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### Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

### Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

#### Details

Product Status	Obsolete
Number of LABs/CLBs	6144
Number of Logic Elements/Cells	27648
Total RAM Bits	131072
Number of I/O	404
Number of Gates	1124022
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	560-LBGA Exposed Pad, Metal
Supplier Device Package	560-MBGA (42.5x42.5)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/xilinx/xcv1000-4bg560c">https://www.e-xfl.com/product-detail/xilinx/xcv1000-4bg560c</a>

## Revision History

Date	Version	Revision
11/98	1.0	Initial Xilinx release.
01/99-02/99	1.2-1.3	Both versions updated package drawings and specs.
05/99	1.4	Addition of package drawings and specifications.
05/99	1.5	Replaced FG 676 & FG680 package drawings.
07/99	1.6	Changed Boundary Scan Information and changed Figure 11, Boundary Scan Bit Sequence. Updated IOB Input & Output delays. Added Capacitance info for different I/O Standards. Added 5 V tolerant information. Added DLL Parameters and waveforms and new Pin-to-pin Input and Output Parameter tables for Global Clock Input to Output and Setup and Hold. Changed Configuration Information including Figures 12, 14, 17 & 19. Added device-dependent listings for quiescent currents ICCINTQ and ICCOQ. Updated IOB Input and Output Delays based on default standard of LVTTTL, 12 mA, Fast Slew Rate. Added IOB Input Switching Characteristics Standard Adjustments.
09/99	1.7	Speed grade update to preliminary status, Power-on specification and Clock-to-Out Minimums additions, "0" hold time listing explanation, quiescent current listing update, and Figure 6 ADDRA input label correction. Added T <sub>IJITCC</sub> parameter, changed T <sub>OJIT</sub> to T <sub>OPHASE</sub> .
01/00	1.8	Update to speed.txt file 1.96. Corrections for CRs 111036, 111137, 112697, 115479, 117153, 117154, and 117612. Modified notes for Recommended Operating Conditions (voltage and temperature). Changed Bank information for V <sub>CCO</sub> in CS144 package on p.43.
01/00	1.9	Updated DLL Jitter Parameter table and waveforms, added Delay Measurement Methodology table for different I/O standards, changed buffered Hex line info and Input/Output Timing measurement notes.
03/00	2.0	New TBCKO values; corrected FG680 package connection drawing; new note about status of CCLK pin after configuration.
05/00	2.1	Modified "Pins not listed..." statement. Speed grade update to Final status.
05/00	2.2	Modified Table 18.
09/00	2.3	<ul style="list-style-type: none"> <li>Added XCV400 values to table under <b>Minimum Clock-to-Out for Virtex Devices</b>.</li> <li>Corrected Units column in table under <b>IOB Input Switching Characteristics</b>.</li> <li>Added values to table under <b>CLB SelectRAM Switching Characteristics</b>.</li> </ul>
10/00	2.4	<ul style="list-style-type: none"> <li>Corrected Pinout information for devices in the BG256, BG432, and BG560 packages in Table 18.</li> <li>Corrected <b>BG256 Pin Function Diagram</b>.</li> </ul>
04/01	2.5	<ul style="list-style-type: none"> <li>Revised minimums for <b>Global Clock Set-Up and Hold for LVTTTL Standard, with DLL</b>.</li> <li>Converted file to modularized format. See <b>Virtex Data Sheet</b> section.</li> </ul>
03/13	4.0	The products listed in this data sheet are obsolete. See <a href="#">XCN10016</a> for further information.

## Virtex Data Sheet

The Virtex Data Sheet contains the following modules:

- DS003-1, Virtex 2.5V FPGAs:  
Introduction and Ordering Information (Module 1)
- DS003-2, Virtex 2.5V FPGAs:  
Functional Description (Module 2)
- DS003-3, Virtex 2.5V FPGAs:  
DC and Switching Characteristics (Module 3)
- DS003-4, Virtex 2.5V FPGAs:  
Pinout Tables (Module 4)



# Virtex™ 2.5 V Field Programmable Gate Arrays

DS003-2 (v4.0) March 1, 2013

## Product Specification

## Architectural Description

### Virtex Array

The Virtex user-programmable gate array, shown in [Figure 1](#), comprises two major configurable elements: configurable logic blocks (CLBs) and input/output blocks (IOBs).

- CLBs provide the functional elements for constructing logic
- IOBs provide the interface between the package pins and the CLBs

CLBs interconnect through a general routing matrix (GRM). The GRM comprises an array of routing switches located at the intersections of horizontal and vertical routing channels. Each CLB nests into a VersaBlock™ that also provides local routing resources to connect the CLB to the GRM.

The VersaRing™ I/O interface provides additional routing resources around the periphery of the device. This routing improves I/O routability and facilitates pin locking.

The Virtex architecture also includes the following circuits that connect to the GRM.

- Dedicated block memories of 4096 bits each
- Clock DLLs for clock-distribution delay compensation and clock domain control
- 3-State buffers (BUFTs) associated with each CLB that drive dedicated segmentable horizontal routing resources

Values stored in static memory cells control the configurable logic elements and interconnect resources. These values load into the memory cells on power-up, and can reload if necessary to change the function of the device.

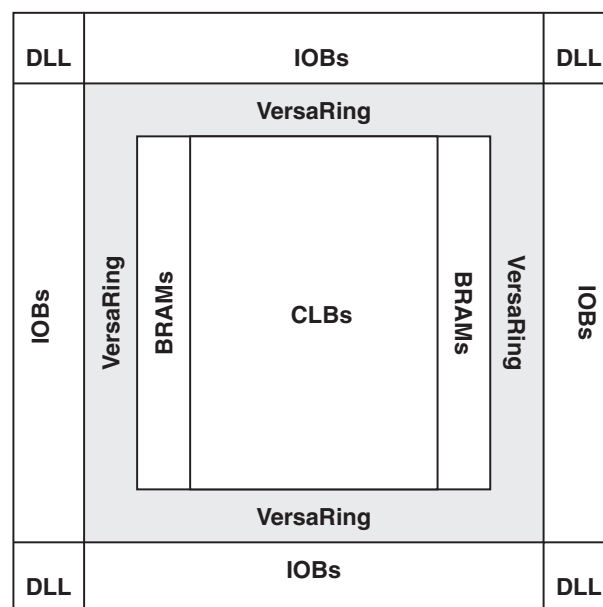
### Input/Output Block

The Virtex IOB, [Figure 2](#), features SelectIO™ inputs and outputs that support a wide variety of I/O signalling standards, see [Table 1](#).

The three IOB storage elements function either as edge-triggered D-type flip-flops or as level sensitive latches. Each IOB has a clock signal (CLK) shared by the three flip-flops and independent clock enable signals for each flip-flop.

In addition to the CLK and CE control signals, the three flip-flops share a Set/Reset (SR). For each flip-flop, this signal can be independently configured as a synchronous Set, a synchronous Reset, an asynchronous Preset, or an asynchronous Clear.

The output buffer and all of the IOB control signals have independent polarity controls.



vao\_b.eps

Figure 1: Virtex Architecture Overview

All pads are protected against damage from electrostatic discharge (ESD) and from over-voltage transients. Two forms of over-voltage protection are provided, one that permits 5 V compliance, and one that does not. For 5 V compliance, a Zener-like structure connected to ground turns on when the output rises to approximately 6.5 V. When PCI 3.3 V compliance is required, a conventional clamp diode is connected to the output supply voltage,  $V_{CCO}$ .

Optional pull-up and pull-down resistors and an optional weak-keeper circuit are attached to each pad. Prior to configuration, all pins not involved in configuration are forced into their high-impedance state. The pull-down resistors and the weak-keeper circuits are inactive, but inputs can optionally be pulled up.

The activation of pull-up resistors prior to configuration is controlled on a global basis by the configuration mode pins. If the pull-up resistors are not activated, all the pins will float. Consequently, external pull-up or pull-down resistors must be provided on pins required to be at a well-defined logic level prior to configuration.

All Virtex IOBs support IEEE 1149.1-compatible boundary scan testing.

### Input Path

A buffer in the Virtex IOB input path routes the input signal either directly to internal logic or through an optional input flip-flop.

An optional delay element at the D-input of this flip-flop eliminates pad-to-pad hold time. The delay is matched to the internal clock-distribution delay of the FPGA, and when used, assures that the pad-to-pad hold time is zero.

Each input buffer can be configured to conform to any of the low-voltage signalling standards supported. In some of these standards the input buffer utilizes a user-supplied threshold voltage,  $V_{REF}$ . The need to supply  $V_{REF}$  imposes constraints on which standards can be used in close proximity to each other. See [I/O Banking, page 3](#).

There are optional pull-up and pull-down resistors at each user I/O input for use after configuration. Their value is in the range 50 k $\Omega$  – 100 k $\Omega$ .

### Output Path

The output path includes a 3-state output buffer that drives the output signal onto the pad. The output signal can be routed to the buffer directly from the internal logic or through an optional IOB output flip-flop.

The 3-state control of the output can also be routed directly from the internal logic or through a flip-flop that provides synchronous enable and disable.

Each output driver can be individually programmed for a wide range of low-voltage signalling standards. Each output buffer can source up to 24 mA and sink up to 48mA. Drive strength and slew rate controls minimize bus transients.

In most signalling standards, the output High voltage depends on an externally supplied  $V_{CCO}$  voltage. The need to supply  $V_{CCO}$  imposes constraints on which standards can be used in close proximity to each other. See [I/O Banking, page 3](#).

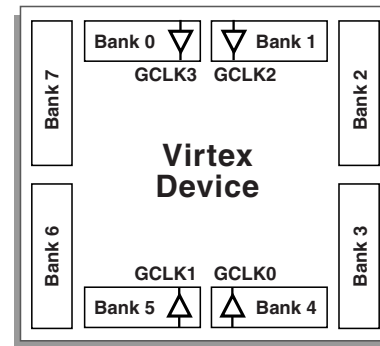
An optional weak-keeper circuit is connected to each output. When selected, the circuit monitors the voltage on the pad and weakly drives the pin High or Low to match the input signal. If the pin is connected to a multiple-source signal, the weak keeper holds the signal in its last state if all drivers are disabled. Maintaining a valid logic level in this way eliminates bus chatter.

Because the weak-keeper circuit uses the IOB input buffer to monitor the input level, an appropriate  $V_{REF}$  voltage must be provided if the signalling standard requires one. The provision of this voltage must comply with the I/O banking rules.

### I/O Banking

Some of the I/O standards described above require  $V_{CCO}$  and/or  $V_{REF}$  voltages. These voltages externally and connected to device pins that serve groups of IOBs, called banks. Consequently, restrictions exist about which I/O standards can be combined within a given bank.

Eight I/O banks result from separating each edge of the FPGA into two banks, as shown in [Figure 3](#). Each bank has multiple  $V_{CCO}$  pins, all of which must be connected to the same voltage. This voltage is determined by the output standards in use.



X8778\_b

Figure 3: Virtex I/O Banks

Within a bank, output standards can be mixed only if they use the same  $V_{CCO}$ . Compatible standards are shown in [Table 2](#). GTL and GTL+ appear under all voltages because their open-drain outputs do not depend on  $V_{CCO}$ .

Table 2: Compatible Output Standards

$V_{CCO}$	Compatible Standards
3.3 V	PCI, LVTTTL, SSTL3 I, SSTL3 II, CTT, AGP, GTL, GTL+
2.5 V	SSTL2 I, SSTL2 II, LVCMOS2, GTL, GTL+
1.5 V	HSTL I, HSTL III, HSTL IV, GTL, GTL+

Some input standards require a user-supplied threshold voltage,  $V_{REF}$ . In this case, certain user-I/O pins are automatically configured as inputs for the  $V_{REF}$  voltage. Approximately one in six of the I/O pins in the bank assume this role.

The  $V_{REF}$  pins within a bank are interconnected internally and consequently only one  $V_{REF}$  voltage can be used within each bank. All  $V_{REF}$  pins in the bank, however, must be connected to the external voltage source for correct operation.

Within a bank, inputs that require  $V_{REF}$  can be mixed with those that do not. However, only one  $V_{REF}$  voltage can be used within a bank. Input buffers that use  $V_{REF}$  are not 5 V tolerant. LVTTTL, LVCMOS2, and PCI 33 MHz 5 V, are 5 V tolerant.

The  $V_{CCO}$  and  $V_{REF}$  pins for each bank appear in the device Pinout tables and diagrams. The diagrams also show the bank affiliation of each I/O.

Within a given package, the number of  $V_{REF}$  and  $V_{CCO}$  pins can vary depending on the size of device. In larger devices,

Each block SelectRAM cell, as illustrated in Figure 6, is a fully synchronous dual-ported 4096-bit RAM with independent control signals for each port. The data widths of the two ports can be configured independently, providing built-in bus-width conversion.

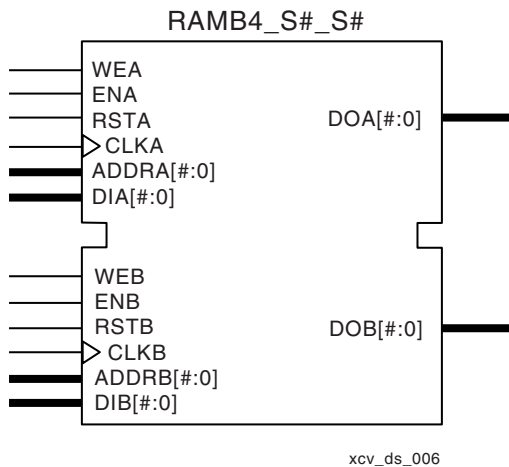


Figure 6: Dual-Port Block SelectRAM

Table 4 shows the depth and width aspect ratios for the block SelectRAM.

Table 4: Block SelectRAM Port Aspect Ratios

Width	Depth	ADDR Bus	Data Bus
1	4096	ADDR<11:0>	DATA<0>
2	2048	ADDR<10:0>	DATA<1:0>
4	1024	ADDR<9:0>	DATA<3:0>
8	512	ADDR<8:0>	DATA<7:0>
16	256	ADDR<7:0>	DATA<15:0>

The Virtex block SelectRAM also includes dedicated routing to provide an efficient interface with both CLBs and other block SelectRAMs. Refer to XAPP130 for block SelectRAM timing waveforms.

## Programmable Routing Matrix

It is the longest delay path that limits the speed of any worst-case design. Consequently, the Virtex routing architecture and its place-and-route software were defined in a single optimization process. This joint optimization minimizes long-path delays, and consequently, yields the best system performance.

The joint optimization also reduces design compilation times because the architecture is software-friendly. Design cycles are correspondingly reduced due to shorter design iteration times.

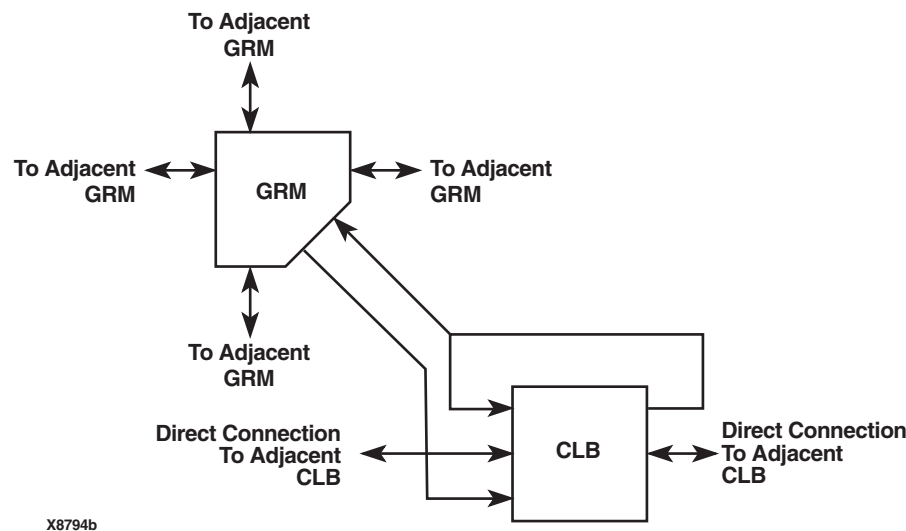


Figure 7: Virtex Local Routing

## Local Routing

The VersaBlock provides local routing resources, as shown in Figure 7, providing the following three types of connections.

- Interconnections among the LUTs, flip-flops, and GRM

- Internal CLB feedback paths that provide high-speed connections to LUTs within the same CLB, chaining them together with minimal routing delay
- Direct paths that provide high-speed connections between horizontally adjacent CLBs, eliminating the delay of the GRM.



ers with a common user interface regardless of their choice of entry and verification tools. The XDM software simplifies the selection of implementation options with pull-down menus and on-line help.

Application programs ranging from schematic capture to Placement and Routing (PAR) can be accessed through the XDM software. The program command sequence is generated prior to execution, and stored for documentation.

Several advanced software features facilitate Virtex design. RPMs, for example, are schematic-based macros with relative location constraints to guide their placement. They help ensure optimal implementation of common functions.

For HDL design entry, the Xilinx FPGA Foundation development system provides interfaces to the following synthesis design environments.

- Synopsys (FPGA Compiler, FPGA Express)
- Exemplar (Spectrum)
- Synplicity (Synplify)

For schematic design entry, the Xilinx FPGA Foundation and alliance development system provides interfaces to the following schematic-capture design environments.

- Mentor Graphics V8 (Design Architect, QuickSim II)
- Viewlogic Systems (Viewdraw)

Third-party vendors support many other environments.

A standard interface-file specification, Electronic Design Interchange Format (EDIF), simplifies file transfers into and out of the development system.

Virtex FPGAs supported by a unified library of standard functions. This library contains over 400 primitives and macros, ranging from 2-input AND gates to 16-bit accumulators, and includes arithmetic functions, comparators, counters, data registers, decoders, encoders, I/O functions, latches, Boolean functions, multiplexers, shift registers, and barrel shifters.

The “soft macro” portion of the library contains detailed descriptions of common logic functions, but does not contain any partitioning or placement information. The performance of these macros depends, therefore, on the partitioning and placement obtained during implementation.

RPMs, on the other hand, do contain predetermined partitioning and placement information that permits optimal implementation of these functions. Users can create their own library of soft macros or RPMs based on the macros and primitives in the standard library.

The design environment supports hierarchical design entry, with high-level schematics that comprise major functional blocks, while lower-level schematics define the logic in these blocks. These hierarchical design elements are automatically combined by the implementation tools. Different design entry tools can be combined within a hierarchical

design, thus allowing the most convenient entry method to be used for each portion of the design.

## Design Implementation

The place-and-route tools (PAR) automatically provide the implementation flow described in this section. The partitioner takes the EDIF net list for the design and maps the logic into the architectural resources of the FPGA (CLBs and IOBs, for example). The placer then determines the best locations for these blocks based on their interconnections and the desired performance. Finally, the router interconnects the blocks.

The PAR algorithms support fully automatic implementation of most designs. For demanding applications, however, the user can exercise various degrees of control over the process. User partitioning, placement, and routing information is optionally specified during the design-entry process. The implementation of highly structured designs can benefit greatly from basic floor planning.

The implementation software incorporates Timing Wizard® timing-driven placement and routing. Designers specify timing requirements along entire paths during design entry. The timing path analysis routines in PAR then recognize these user-specified requirements and accommodate them.

Timing requirements are entered on a schematic in a form directly relating to the system requirements, such as the targeted clock frequency, or the maximum allowable delay between two registers. In this way, the overall performance of the system along entire signal paths is automatically tailored to user-generated specifications. Specific timing information for individual nets is unnecessary.

## Design Verification

In addition to conventional software simulation, FPGA users can use in-circuit debugging techniques. Because Xilinx devices are infinitely reprogrammable, designs can be verified in real time without the need for extensive sets of software simulation vectors.

The development system supports both software simulation and in-circuit debugging techniques. For simulation, the system extracts the post-layout timing information from the design database, and back-annotates this information into the net list for use by the simulator. Alternatively, the user can verify timing-critical portions of the design using the TRACE® static timing analyzer.

For in-circuit debugging, the development system includes a download and readback cable. This cable connects the FPGA in the target system to a PC or workstation. After downloading the design into the FPGA, the designer can single-step the logic, readback the contents of the flip-flops, and so observe the internal logic state. Simple modifications can be downloaded into the system in a matter of minutes.

## Master-Serial Mode

In master-serial mode, the CCLK output of the FPGA drives a Xilinx Serial PROM that feeds bit-serial data to the DIN input. The FPGA accepts this data on each rising CCLK edge. After the FPGA has been loaded, the data for the next device in a daisy-chain is presented on the DOUT pin after the rising CCLK edge.

The interface is identical to slave-serial except that an internal oscillator is used to generate the configuration clock (CCLK). A wide range of frequencies can be selected for CCLK which always starts at a slow default frequency. Configuration bits then switch CCLK to a higher frequency for the remainder of the configuration. Switching to a lower frequency is prohibited.

The CCLK frequency is set using the ConfigRate option in the bitstream generation software. The maximum CCLK frequency that can be selected is 60 MHz. When selecting a CCLK frequency, ensure that the serial PROM and any

daisy-chained FPGAs are fast enough to support the clock rate.

On power-up, the CCLK frequency is 2.5 MHz. This frequency is used until the ConfigRate bits have been loaded when the frequency changes to the selected ConfigRate. Unless a different frequency is specified in the design, the default ConfigRate is 4 MHz.

Figure 12 shows a full master/slave system. In this system, the left-most device operates in master-serial mode. The remaining devices operate in slave-serial mode. The SPROM  $\overline{\text{RESET}}$  pin is driven by  $\overline{\text{INIT}}$ , and the  $\overline{\text{CE}}$  input is driven by DONE. There is the potential for contention on the DONE pin, depending on the start-up sequence options chosen.

Figure 14 shows the timing of master-serial configuration. Master-serial mode is selected by a <000> or <100> on the mode pins (M2, M1, M0). Table 8 shows the timing information for Figure 14.

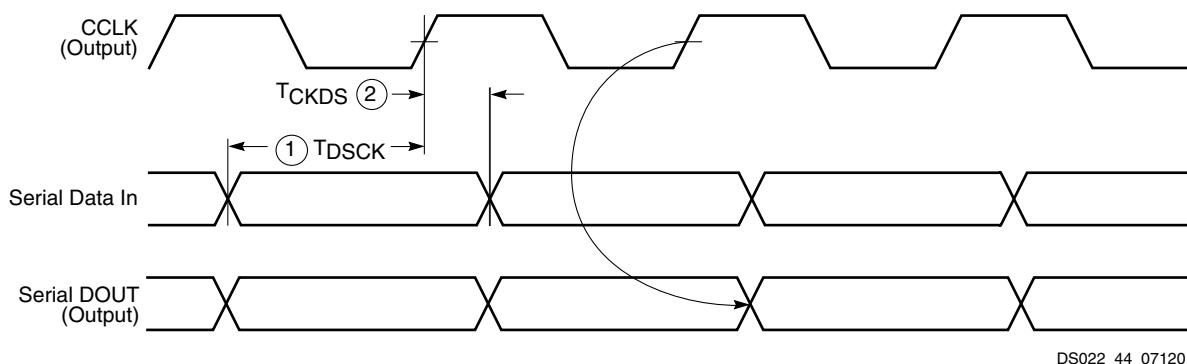


Figure 14: Master-Serial Mode Programming Switching Characteristics

At power-up,  $V_{CC}$  must rise from 1.0 V to  $V_{CC}$  min in less than 50 ms, otherwise delay configuration by pulling  $\overline{\text{PROGRAM}}$  Low until  $V_{CC}$  is valid.

The sequence of operations necessary to configure a Virtex FPGA serially appears in Figure 15.

## SelectMAP Mode

The SelectMAP mode is the fastest configuration option. Byte-wide data is written into the FPGA with a  $\overline{\text{BUSY}}$  flag controlling the flow of data.

An external data source provides a byte stream, CCLK, a Chip Select ( $\overline{\text{CS}}$ ) signal and a Write signal ( $\overline{\text{WRITE}}$ ). If  $\overline{\text{BUSY}}$  is asserted (High) by the FPGA, the data must be held until  $\overline{\text{BUSY}}$  goes Low.

Data can also be read using the SelectMAP mode. If  $\overline{\text{WRITE}}$  is not asserted, configuration data is read out of the FPGA as part of a readback operation.

In the SelectMAP mode, multiple Virtex devices can be chained in parallel. DATA pins (D7:D0), CCLK,  $\overline{\text{WRITE}}$ ,  $\overline{\text{BUSY}}$ ,  $\overline{\text{PROGRAM}}$ , DONE, and  $\overline{\text{INIT}}$  can be connected in parallel between all the FPGAs. Note that the data is organized with the MSB of each byte on pin D0 and the LSB of each byte on D7. The  $\overline{\text{CS}}$  pins are kept separate, insuring that each FPGA can be selected individually.  $\overline{\text{WRITE}}$  should be Low before loading the first bitstream and returned High after the last device has been programmed. Use  $\overline{\text{CS}}$  to select the appropriate FPGA for loading the bitstream and sending the configuration data. At the end of the bitstream, deselect the loaded device and select the next target FPGA by setting its  $\overline{\text{CS}}$  pin High. A free-running oscillator or other externally generated signal can be used for CCLK. The  $\overline{\text{BUSY}}$  signal can be ignored for frequencies below 50 MHz. For details about frequencies above 50 MHz, see XAPP138, Virtex Configuration and Readback. Once all the devices have been programmed, the DONE pin goes High.

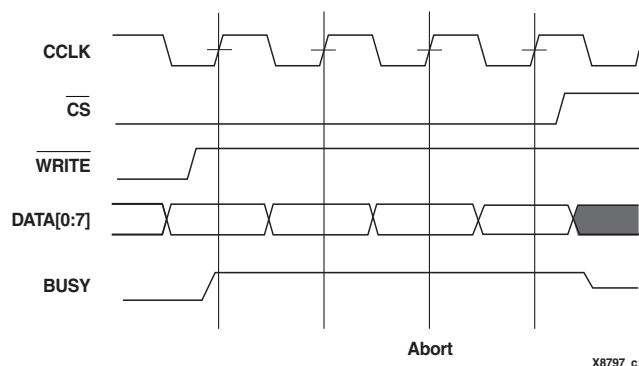


Figure 18: SelectMAP Write Abort Waveforms

## Boundary-Scan Mode

In the boundary-scan mode, configuration is done through the IEEE 1149.1 Test Access Port. Note that the **PROGRAM** pin must be pulled High prior to reconfiguration. A Low on the **PROGRAM** pin resets the TAP controller and no JTAG operations can be performed.

Configuration through the TAP uses the **CFG\_IN** instruction. This instruction allows data input on TDI to be converted into data packets for the internal configuration bus.

The following steps are required to configure the FPGA through the boundary-scan port (when using TCK as a start-up clock).

1. Load the **CFG\_IN** instruction into the boundary-scan instruction register (IR)
2. Enter the Shift-DR (SDR) state
3. Shift a configuration bitstream into TDI
4. Return to Run-Test-Idle (RTI)
5. Load the **JSTART** instruction into IR
6. Enter the SDR state
7. Clock TCK through the startup sequence
8. Return to RTI

Configuration and readback via the TAP is always available. The boundary-scan mode is selected by a <101> or 001> on the mode pins (M2, M1, M0). For details on TAP characteristics, refer to XAPP139.

## Configuration Sequence

The configuration of Virtex devices is a three-phase process. First, the configuration memory is cleared. Next, configuration data is loaded into the memory, and finally, the logic is activated by a start-up process.

Configuration is automatically initiated on power-up unless it is delayed by the user, as described below. The configuration process can also be initiated by asserting **PROGRAM**.

The end of the memory-clearing phase is signalled by **INIT** going High, and the completion of the entire process is signalled by **DONE** going High.

The power-up timing of configuration signals is shown in Figure 19. The corresponding timing characteristics are listed in Table 10.

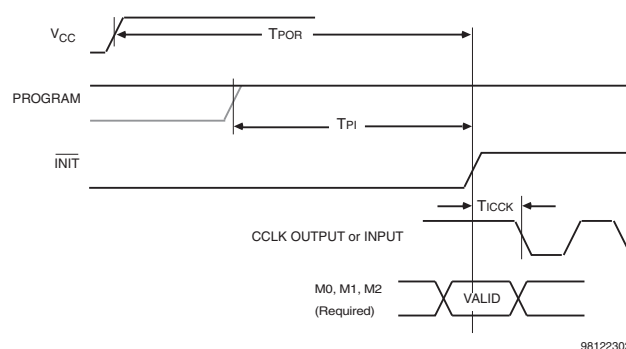


Figure 19: Power-Up Timing Configuration Signals

Table 10: Power-up Timing Characteristics

Description	Symbol	Value	Units
Power-on Reset	T <sub>POR</sub>	2.0	ms, max
Program Latency	T <sub>PL</sub>	100.0	μs, max
CCLK (output) Delay	T <sub>ICCK</sub>	0.5	μs, min
		4.0	μs, max
Program Pulse Width	T <sub>PROGRAM</sub>	300	ns, min

## Delaying Configuration

**INIT** can be held Low using an open-drain driver. An open-drain is required since **INIT** is a bidirectional open-drain pin that is held Low by the FPGA while the configuration memory is being cleared. Extending the time that the pin is Low causes the configuration sequencer to wait. Thus, configuration is delayed by preventing entry into the phase where data is loaded.

## Start-Up Sequence

The default Start-up sequence is that one CCLK cycle after **DONE** goes High, the global 3-state signal (GTS) is released. This permits device outputs to turn on as necessary.

One CCLK cycle later, the Global Set/Reset (GSR) and Global Write Enable (GWE) signals are released. This permits the internal storage elements to begin changing state in response to the logic and the user clock.

The relative timing of these events can be changed. In addition, the GTS, GSR, and GWE events can be made dependent on the **DONE** pins of multiple devices all going High, forcing the devices to start in synchronism. The sequence can also be paused at any stage until lock has been achieved on any or all DLLs.





## Virtex™ 2.5 V Field Programmable Gate Arrays

DS003-3 (v4.0) March 1, 2013

Production Product Specification

### Virtex Electrical Characteristics

#### Definition of Terms

Electrical and switching characteristics are specified on a per-speed-grade basis and can be designated as Advance, Preliminary, or Production. Each designation is defined as follows:

**Advance:** These speed files are based on simulations only and are typically available soon after device design specifications are frozen. Although speed grades with this designation are considered relatively stable and conservative, some under-reporting might still occur.

**Preliminary:** These speed files are based on complete ES (engineering sample) silicon characterization. Devices and speed grades with this designation are intended to give a better indication of the expected performance of production silicon. The probability of under-reporting delays is greatly reduced as compared to Advance data.

**Production:** These speed files are released once enough production silicon of a particular device family member has been characterized to provide full correlation between speed files and devices over numerous production lots. There is no under-reporting of delays, and customers receive formal notification of any subsequent changes. Typically, the slowest speed grades transition to Production before faster speed grades.

All specifications are representative of worst-case supply voltage and junction temperature conditions. The parameters included are common to popular designs and typical applications. Contact the factory for design considerations requiring more detailed information.

**Table 1** correlates the current status of each Virtex device with a corresponding speed file designation.

*Table 1: Virtex Device Speed Grade Designations*

Device	Speed Grade Designations		
	Advance	Preliminary	Production
XCV50			–6, –5, –4
XCV100			–6, –5, –4
XCV150			–6, –5, –4
XCV200			–6, –5, –4
XCV300			–6, –5, –4
XCV400			–6, –5, –4
XCV600			–6, –5, –4
XCV800			–6, –5, –4
XCV1000			–6, –5, –4

All specifications are subject to change without notice.

## Power-On Power Supply Requirements

Xilinx FPGAs require a certain amount of supply current during power-on to insure proper device operation. The actual current consumed depends on the power-on ramp rate of the power supply. This is the time required to reach the nominal power supply voltage of the device<sup>(1)</sup> from 0 V. The current is highest at the fastest suggested ramp rate (0 V to nominal voltage in 2 ms) and is lowest at the slowest allowed ramp rate (0 V to nominal voltage in 50 ms). For more details on power supply requirements, see Application Note XAPP158 on [www.xilinx.com](http://www.xilinx.com).

Product	Description <sup>(2)</sup>	Current Requirement <sup>(1,3)</sup>
Virtex Family, Commercial Grade	Minimum required current supply	500 mA
Virtex Family, Industrial Grade	Minimum required current supply	2 A

### Notes:

- Ramp rate used for this specification is from 0 - 2.7 VDC. Peak current occurs on or near the internal power-on reset threshold of 1.0V and lasts for less than 3 ms.
- Devices are guaranteed to initialize properly with the minimum current available from the power supply as noted above.
- Larger currents can result if ramp rates are forced to be faster.

## DC Input and Output Levels

Values for  $V_{IL}$  and  $V_{IH}$  are recommended input voltages. Values for  $I_{OL}$  and  $I_{OH}$  are guaranteed output currents over the recommended operating conditions at the  $V_{OL}$  and  $V_{OH}$  test points. Only selected standards are tested. These are chosen to ensure that all standards meet their specifications. The selected standards are tested at minimum  $V_{CCO}$  for each standard with the respective  $V_{OL}$  and  $V_{OH}$  voltage levels shown. Other standards are sample tested.

Input/Output Standard	$V_{IL}$		$V_{IH}$		$V_{OL}$	$V_{OH}$	$I_{OL}$	$I_{OH}$
	V, min	V, max	V, min	V, max	V, Max	V, Min	mA	mA
LVTTL <sup>(1)</sup>	-0.5	0.8	2.0	5.5	0.4	2.4	24	-24
LVC MOS2	-0.5	.7	1.7	5.5	0.4	1.9	12	-12
PCI, 3.3 V	-0.5	44% $V_{CCINT}$	60% $V_{CCINT}$	$V_{CCO} + 0.5$	10% $V_{CCO}$	90% $V_{CCO}$	Note 2	Note 2
PCI, 5.0 V	-0.5	0.8	2.0	5.5	0.55	2.4	Note 2	Note 2
GTL	-0.5	$V_{REF} - 0.05$	$V_{REF} + 0.05$	3.6	0.4	n/a	40	n/a
GTL+	-0.5	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.6	n/a	36	n/a
HSTL I <sup>(3)</sup>	-0.5	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.4	$V_{CCO} - 0.4$	8	-8
HSTL III	-0.5	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.4	$V_{CCO} - 0.4$	24	-8
HSTL IV	-0.5	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.4	$V_{CCO} - 0.4$	48	-8
SSTL3 I	-0.5	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	$V_{REF} - 0.6$	$V_{REF} + 0.6$	8	-8
SSTL3 II	-0.5	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	$V_{REF} - 0.8$	$V_{REF} + 0.8$	16	-16
SSTL2 I	-0.5	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	$V_{REF} - 0.61$	$V_{REF} + 0.61$	7.6	-7.6
SSTL2 II	-0.5	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	$V_{REF} - 0.80$	$V_{REF} + 0.80$	15.2	-15.2
CTT	-0.5	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	$V_{REF} - 0.4$	$V_{REF} + 0.4$	8	-8
AGP	-0.5	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	10% $V_{CCO}$	90% $V_{CCO}$	Note 2	Note 2

### Notes:

- $V_{OL}$  and  $V_{OH}$  for lower drive currents are sample tested.
- Tested according to the relevant specifications.
- DC input and output levels for HSTL18 (HSTL I/O standard with  $V_{CCO}$  of 1.8 V) are provided in an HSTL white paper on [www.xilinx.com](http://www.xilinx.com).

### Virtex Switching Characteristics

All devices are 100% functionally tested. Internal timing parameters are derived from measuring internal test patterns. Listed below are representative values. For more specific, more precise, and worst-case guaranteed data, use the values reported by the static timing analyzer (TRCE

in the Xilinx Development System) and back-annotated to the simulation net list. All timing parameters assume worst-case operating conditions (supply voltage and junction temperature). Values apply to all Virtex devices unless otherwise noted.

### IOB Input Switching Characteristics

Input delays associated with the pad are specified for LVTTTL levels. For other standards, adjust the delays with the values shown in , page 6.

Description	Device	Symbol	Speed Grade				Units
			Min	-6	-5	-4	
Propagation Delays							
Pad to I output, no delay	All	T <sub>IOPI</sub>	0.39	0.8	0.9	1.0	ns, max
Pad to I output, with delay	XCV50	T <sub>IOPID</sub>	0.8	1.5	1.7	1.9	ns, max
	XCV100		0.8	1.5	1.7	1.9	ns, max
	XCV150		0.8	1.5	1.7	1.9	ns, max
	XCV200		0.8	1.5	1.7	1.9	ns, max
	XCV300		0.8	1.5	1.7	1.9	ns, max
	XCV400		0.9	1.8	2.0	2.3	ns, max
	XCV600		0.9	1.8	2.0	2.3	ns, max
	XCV800		1.1	2.1	2.4	2.7	ns, max
	XCV1000		1.1	2.1	2.4	2.7	ns, max
Pad to output IQ via transparent latch, no delay	All	T <sub>IOPLI</sub>	0.8	1.6	1.8	2.0	ns, max
Pad to output IQ via transparent latch, with delay	XCV50	T <sub>IOPLID</sub>	1.9	3.7	4.2	4.8	ns, max
	XCV100		1.9	3.7	4.2	4.8	ns, max
	XCV150		2.0	3.9	4.3	4.9	ns, max
	XCV200		2.0	4.0	4.4	5.1	ns, max
	XCV300		2.0	4.0	4.4	5.1	ns, max
	XCV400		2.1	4.1	4.6	5.3	ns, max
	XCV600		2.1	4.2	4.7	5.4	ns, max
	XCV800		2.2	4.4	4.9	5.6	ns, max
	XCV1000		2.3	4.5	5.1	5.8	ns, max
Sequential Delays							
Clock CLK	All						
Minimum Pulse Width, High		T <sub>CH</sub>	0.8	1.5	1.7	2.0	ns, min
Minimum Pulse Width, Low		T <sub>CL</sub>	0.8	1.5	1.7	2.0	ns, min
Clock CLK to output IQ		T <sub>IOCKIQ</sub>	0.2	0.7	0.7	0.8	ns, max

### IOB Output Switching Characteristics Standard Adjustments

Output delays terminating at a pad are specified for LVTTTL with 12 mA drive and fast slew rate. For other standards, adjust the delays by the values shown.

Description	Symbol	Standard <sup>(1)</sup>	Speed Grade				Unit s
			Min	-6	-5	-4	
Output Delay Adjustments							
Standard-specific adjustments for output delays terminating at pads (based on standard capacitive load, Csl)	T <sub>OLVTTTL_S2</sub>	LVTTL, Slow, 2 mA	4.2	14.7	15.8	17.0	ns
	T <sub>OLVTTTL_S4</sub>	4 mA	2.5	7.5	8.0	8.6	ns
	T <sub>OLVTTTL_S6</sub>	6 mA	1.8	4.8	5.1	5.6	ns
	T <sub>OLVTTTL_S8</sub>	8 mA	1.2	3.0	3.3	3.5	ns
	T <sub>OLVTTTL_S12</sub>	12 mA	1.0	1.9	2.1	2.2	ns
	T <sub>OLVTTTL_S16</sub>	16 mA	0.9	1.7	1.9	2.0	ns
	T <sub>OLVTTTL_S24</sub>	24 mA	0.8	1.3	1.4	1.6	ns
	T <sub>OLVTTTL_F2</sub>	LVTTL, Fast, 2mA	1.9	13.1	14.0	15.1	ns
	T <sub>OLVTTTL_F4</sub>	4 mA	0.7	5.3	5.7	6.1	ns
	T <sub>OLVTTTL_F6</sub>	6 mA	0.2	3.1	3.3	3.6	ns
	T <sub>OLVTTTL_F8</sub>	8 mA	0.1	1.0	1.1	1.2	ns
	T <sub>OLVTTTL_F12</sub>	12 mA	0	0	0	0	ns
	T <sub>OLVTTTL_F16</sub>	16 mA	−0.10	−0.05	−0.05	−0.05	ns
	T <sub>OLVTTTL_F24</sub>	24 mA	−0.10	−0.20	−0.21	−0.23	ns
	T <sub>OLVCMOS2</sub>	LVC MOS2	0.10	0.10	0.11	0.12	ns
	T <sub>OPCI33_3</sub>	PCI, 33 MHz, 3.3 V	0.50	2.3	2.5	2.7	ns
	T <sub>OPCI33_5</sub>	PCI, 33 MHz, 5.0 V	0.40	2.8	3.0	3.3	ns
	T <sub>OPCI66_3</sub>	PCI, 66 MHz, 3.3 V	0.10	−0.40	−0.42	−0.46	ns
	T <sub>OGTL</sub>	GTL	0.6	0.50	0.54	0.6	ns
	T <sub>OGTLP</sub>	GTL+	0.7	0.8	0.9	1.0	ns
	T <sub>OHSTL_I</sub>	HSTL I	0.10	−0.50	−0.53	−0.5	ns
	T <sub>OHSTL_III</sub>	HSTL III	−0.10	−0.9	−0.9	−1.0	ns
	T <sub>OHSTL_IV</sub>	HSTL IV	−0.20	−1.0	−1.0	−1.1	ns
	T <sub>OSSTL2_I</sub>	SSTL2 I	−0.10	−0.50	−0.53	−0.5	ns
	T <sub>OSSTL2_II</sub>	SSTL2 II	−0.20	−0.9	−0.9	−1.0	ns
	T <sub>OSSTL3_I</sub>	SSTL3 I	−0.20	−0.50	−0.53	−0.5	ns
	T <sub>OSSTL3_II</sub>	SSTL3 II	−0.30	−1.0	−1.0	−1.1	ns
	T <sub>OCTT</sub>	CTT	0	−0.6	−0.6	−0.6	ns
	T <sub>OAGP</sub>	AGP	0	−0.9	−0.9	−1.0	ns

#### Notes:

- Output timing is measured at 1.4 V with 35 pF external capacitive load for LVTTTL. For other I/O standards and different loads, see [Table 2](#) and [Table 3](#).

## Calculation of $T_{i\text{oop}}$ as a Function of Capacitance

$T_{i\text{oop}}$  is the propagation delay from the O Input of the IOB to the pad. The values for  $T_{i\text{oop}}$  were based on the standard capacitive load ( $C_{sl}$ ) for each I/O standard as listed in Table 2.

Table 2: Constants for Calculating  $T_{i\text{oop}}$

Standard	Csl (pF)	fl (ns/pF)
LVTTL Fast Slew Rate, 2mA drive	35	0.41
LVTTL Fast Slew Rate, 4mA drive	35	0.20
LVTTL Fast Slew Rate, 6mA drive	35	0.13
LVTTL Fast Slew Rate, 8mA drive	35	0.079
LVTTL Fast Slew Rate, 12mA drive	35	0.044
LVTTL Fast Slew Rate, 16mA drive	35	0.043
LVTTL Fast Slew Rate, 24mA drive	35	0.033
LVTTL Slow Slew Rate, 2mA drive	35	0.41
LVTTL Slow Slew Rate, 4mA drive	35	0.20
LVTTL Slow Slew Rate, 6mA drive	35	0.100
LVTTL Slow Slew Rate, 8mA drive	35	0.086
LVTTL Slow Slew Rate, 12mA drive	35	0.058
LVTTL Slow Slew Rate, 16mA drive	35	0.050
LVTTL Slow Slew Rate, 24mA drive	35	0.048
LVCMS2	35	0.041
PCI 33MHz 5V	50	0.050
PCI 33MHZ 3.3 V	10	0.050
PCI 66 MHz 3.3 V	10	0.033
GTL	0	0.014
GTL+	0	0.017
HSTL Class I	20	0.022
HSTL Class III	20	0.016
HSTL Class IV	20	0.014
SSTL2 Class I	30	0.028
SSTL2 Class II	30	0.016
SSTL3 Class I	30	0.029
SSTL3 Class II	30	0.016
CTT	20	0.035
AGP	10	0.037

### Notes:

1. I/O parameter measurements are made with the capacitance values shown above. See Application Note XAPP133 on [www.xilinx.com](http://www.xilinx.com) for appropriate terminations.
2. I/O standard measurements are reflected in the IBIS model information except where the IBIS format precludes it.

For other capacitive loads, use the formulas below to calculate the corresponding  $T_{i\text{oop}}$ .

$$T_{i\text{oop}} = T_{i\text{oop}} + T_{\text{opadjust}} + (C_{\text{load}} - C_{sl}) * fl$$

Where:

$T_{\text{opadjust}}$  is reported above in the Output Delay Adjustment section.

$C_{\text{load}}$  is the capacitive load for the design.

Table 3: Delay Measurement Methodology

Standard	$V_L$ (1)	$V_H$ (1)	Meas. Point	$V_{REF}$ Typ (2)
LVTTL	0	3	1.4	-
LVCMS2	0	2.5	1.125	-
PCI33_5	Per PCI Spec			-
PCI33_3	Per PCI Spec			-
PCI66_3	Per PCI Spec			-
GTL	$V_{REF} - 0.2$	$V_{REF} + 0.2$	$V_{REF}$	0.80
GTL+	$V_{REF} - 0.2$	$V_{REF} + 0.2$	$V_{REF}$	1.0
HSTL Class I	$V_{REF} - 0.5$	$V_{REF} + 0.5$	$V_{REF}$	0.75
HSTL Class III	$V_{REF} - 0.5$	$V_{REF} + 0.5$	$V_{REF}$	0.90
HSTL Class IV	$V_{REF} - 0.5$	$V_{REF} + 0.5$	$V_{REF}$	0.90
SSTL3 I & II	$V_{REF} - 1.0$	$V_{REF} + 1.0$	$V_{REF}$	1.5
SSTL2 I & II	$V_{REF} - 0.75$	$V_{REF} + 0.75$	$V_{REF}$	1.25
CTT	$V_{REF} - 0.2$	$V_{REF} + 0.2$	$V_{REF}$	1.5
AGP	$V_{REF} - (0.2 \times V_{CCO})$	$V_{REF} + (0.2 \times V_{CCO})$	$V_{REF}$	Per AGP Spec

### Notes:

1. Input waveform switches between  $V_L$  and  $V_H$ .
2. Measurements are made at  $V_{REF}$  (Typ), Maximum, and Minimum. Worst-case values are reported.
3. I/O parameter measurements are made with the capacitance values shown in Table 2. See Application Note XAPP133 on [www.xilinx.com](http://www.xilinx.com) for appropriate terminations.
4. I/O standard measurements are reflected in the IBIS model information except where the IBIS format precludes it.



Date	Version	Revision
01/00	1.9	Updated DLL Jitter Parameter table and waveforms, added Delay Measurement Methodology table for different I/O standards, changed buffered Hex line info and Input/Output Timing measurement notes.
03/00	2.0	New TBCKO values; corrected FG680 package connection drawing; new note about status of CCLK pin after configuration.
05/00	2.1	Modified "Pins not listed..." statement. Speed grade update to Final status.
05/00	2.2	Modified Table 18.
09/00	2.3	<ul style="list-style-type: none"> <li>Added XCV400 values to table under <b>Minimum Clock-to-Out for Virtex Devices</b>.</li> <li>Corrected Units column in table under <b>IOB Input Switching Characteristics</b>.</li> <li>Added values to table under <b>CLB SelectRAM Switching Characteristics</b>.</li> </ul>
10/00	2.4	<ul style="list-style-type: none"> <li>Corrected Pinout information for devices in the BG256, BG432, and BG560 packages in Table 18.</li> <li>Corrected <b>BG256 Pin Function Diagram</b>.</li> </ul>
04/02/01	2.5	<ul style="list-style-type: none"> <li>Revised minimums for <b>Global Clock Set-Up and Hold for LVTTTL Standard, with DLL</b>.</li> <li>Converted file to modularized format. See the <b>Virtex Data Sheet</b> section.</li> </ul>
04/19/01	2.6	<ul style="list-style-type: none"> <li>Clarified TIOCKP and TIOCKON <b>IOB Output Switching Characteristics</b> descriptors.</li> </ul>
07/19/01	2.7	<ul style="list-style-type: none"> <li>Under <b>Absolute Maximum Ratings</b>, changed (<math>T_{SOL}</math>) to 220 °C.</li> </ul>
07/26/01	2.8	<ul style="list-style-type: none"> <li>Removed <math>T_{SOL}</math> parameter and added footnote to <b>Absolute Maximum Ratings</b> table.</li> </ul>
10/29/01	2.9	<ul style="list-style-type: none"> <li>Updated the speed grade designations used in data sheets, and added <b>Table 1</b>, which shows the current speed grade designation for each device.</li> </ul>
02/01/02	3.0	<ul style="list-style-type: none"> <li>Added footnote to <b>DC Input and Output Levels</b> table.</li> </ul>
07/19/02	3.1	<ul style="list-style-type: none"> <li>Removed mention of MIL-M-38510/605 specification.</li> <li>Added link to xapp158 from the <b>Power-On Power Supply Requirements</b> section.</li> </ul>
09/10/02	3.2	<ul style="list-style-type: none"> <li>Added Clock CLK to <b>IOB Input Switching Characteristics</b> and <b>IOB Output Switching Characteristics</b>.</li> </ul>
03/01/13	4.0	The products listed in this data sheet are obsolete. See <a href="#">XCN10016</a> for further information.

## Virtex Data Sheet

The Virtex Data Sheet contains the following modules:

- DS003-1, Virtex 2.5V FPGAs:  
Introduction and Ordering Information (Module 1)
- DS003-2, Virtex 2.5V FPGAs:  
Functional Description (Module 2)
- DS003-3, Virtex 2.5V FPGAs:  
DC and Switching Characteristics (Module 3)
- DS003-4, Virtex 2.5V FPGAs:  
Pinout Tables (Module 4)

## Virtex Pinout Information

### Pinout Tables

See [www.xilinx.com](http://www.xilinx.com) for updates or additional pinout information. For convenience, [Table 2](#), [Table 3](#) and [Table 4](#) list the locations of special-purpose and power-supply pins. Pins not listed are either user I/Os or not connected, depending on the device/package combination. See the Pinout Diagrams starting on [page 17](#) for any pins not listed for a particular part/package combination.

*Table 2: Virtex Pinout Tables (Chip-Scale and QFP Packages)*

Pin Name	Device	CS144	TQ144	PQ/HQ240
GCK0	All	K7	90	92
GCK1	All	M7	93	89
GCK2	All	A7	19	210
GCK3	All	A6	16	213
M0	All	M1	110	60
M1	All	L2	112	58
M2	All	N2	108	62
CCLK	All	B13	38	179
PROGRAM	All	L12	72	122
DONE	All	M12	74	120
INIT	All	L13	71	123
BUSY/DOUT	All	C11	39	178
D0/DIN	All	C12	40	177
D1	All	E10	45	167
D2	All	E12	47	163
D3	All	F11	51	156
D4	All	H12	59	145
D5	All	J13	63	138
D6	All	J11	65	134
D7	All	K10	70	124
WRITE	All	C10	32	185
CS	All	D10	33	184
TDI	All	A11	34	183
TDO	All	A12	36	181
TMS	All	B1	143	2
TCK	All	C3	2	239
V <sub>CCINT</sub>	All	A9, B6, C5, G3, G12, M5, M9, N6	10, 15, 25, 57, 84, 94, 99, 126	16, 32, 43, 77, 88, 104, 137, 148, 164, 198, 214, 225

Table 3: Virtex Pinout Tables (BGA) (Continued)

Pin Name	Device	BG256	BG352	BG432	BG560
V <sub>CCO</sub> , Bank 7	All	G4, H4	G23, K26, N23	A31, L28, L31	C32, D33, K33, N32, T33
V <sub>REF</sub> Bank 0 (VREF pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all VREF pins are general I/O.	XCV50	A8, B4	N/A	N/A	N/A
	XCV100/150	... + A4	A16, C19, C21	N/A	N/A
	XCV200/300	... + A2	... + D21	B19, D22, D24, D26	N/A
	XCV400	N/A	N/A	... + C18	A19, D20, D26, E23, E27
	XCV600	N/A	N/A	... + C24	... + E24
	XCV800	N/A	N/A	... + B21	... + E21
	XCV1000	N/A	N/A	N/A	... + D29
V <sub>REF</sub> Bank 1 (VREF pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all VREF pins are general I/O.	XCV50	A17, B12	N/A	N/A	N/A
	XCV100/150	... + B15	B6, C9, C12	N/A	N/A
	XCV200/300	... + B17	... + D6	A13, B7, C6, C10	N/A
	XCV400	N/A	N/A	... + B15	A6, D7, D11, D16, E15
	XCV600	N/A	N/A	... + D10	... + D10
	XCV800	N/A	N/A	... + B12	... + D13
	XCV1000	N/A	N/A	N/A	... + E7
V <sub>REF</sub> Bank 2 (VREF pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all VREF pins are general I/O.	XCV50	C20, J18	N/A	N/A	N/A
	XCV100/150	... + F19	E2, H2, M4	N/A	N/A
	XCV200/300	... + G18	... + D2	E2, G3, J2, N1	N/A
	XCV400	N/A	N/A	... + R3	G5, H4, L5, P4, R1
	XCV600	N/A	N/A	... + H1	... + K5
	XCV800	N/A	N/A	... + M3	... + N5
	XCV1000	N/A	N/A	N/A	... + B3

Table 4: Virtex Pinout Tables (Fine-Pitch BGA) (Continued)

Pin Name	Device	FG256	FG456	FG676	FG680
<b>V<sub>REF</sub> Bank 1</b> (V <sub>REF</sub> pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V <sub>REF</sub> pins are general I/O.	XCV50	B9, C11	N/A	N/A	N/A
	XCV100/150	... + E11	A18, B13, E14	N/A	N/A
	XCV200/300	... + A14	... + A19	N/A	N/A
	XCV400	N/A	N/A	A14, C20, C21, D15, G16	N/A
	XCV600	N/A	N/A	... + B19	B6, B8, B18, D11, D13, D17
	XCV800	N/A	N/A	... + A17	... + B14
	XCV1000	N/A	N/A	N/A	... + B5
<b>V<sub>REF</sub> Bank 2</b> (V <sub>REF</sub> pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V <sub>REF</sub> pins are general I/O.	XCV50	F13, H13	N/A	N/A	N/A
	XCV100/150	... + F14	F21, H18, K21	N/A	N/A
	XCV200/300	... + E13	... + D22	N/A	N/A
	XCV400	N/A	N/A	F24, H23, K20, M23, M26	N/A
	XCV600	N/A	N/A	... + G26	G1, H4, J1, L2, V5, W3
	XCV800	N/A	N/A	... + K25	... + N1
	XCV1000	N/A	N/A	N/A	... + D2
<b>V<sub>REF</sub> Bank 3</b> (V <sub>REF</sub> pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V <sub>REF</sub> pins are general I/O.	XCV50	K16, L14	N/A	N/A	N/A
	XCV100/150	... + L13	N21, R19, U21	N/A	N/A
	XCV200/300	... + M13	... + U20	N/A	N/A
	XCV400	N/A	N/A	R23, R25, U21, W22, W23	N/A
	XCV600	N/A	N/A	... + W26	AC1, AJ2, AK3, AL4, AR1, Y1
	XCV800	N/A	N/A	... + U25	... + AF3
	XCV1000	N/A	N/A	N/A	... + AP4

## Pinout Diagrams

The following diagrams, **CS144 Pin Function Diagram**, page 17 through **FG680 Pin Function Diagram**, page 27, illustrate the locations of special-purpose pins on Virtex FPGAs. Table 5 lists the symbols used in these diagrams. The diagrams also show I/O-bank boundaries.

Table 5: Pinout Diagram Symbols

Symbol	Pin Function
*	General I/O
*	Device-dependent general I/O, n/c on smaller devices
V	V <sub>CCINT</sub>
v	Device-dependent V <sub>CCINT</sub> , n/c on smaller devices
O	V <sub>CCO</sub>
R	V <sub>REF</sub>
r	Device-dependent V <sub>REF</sub> , remains I/O on smaller devices
G	Ground
Ø, 1, 2, 3	Global Clocks

Table 5: Pinout Diagram Symbols (Continued)

Symbol	Pin Function
⑩, ①, ②	M0, M1, M2
⑩, ①, ②, ③, ④, ⑤, ⑥, ⑦	D0/DIN, D1, D2, D3, D4, D5, D6, D7
B	DOUT/BUSY
D	DONE
P	PROGRAM
I	INIT
K	CCLK
W	WRITE
S	CS
T	Boundary-scan Test Access Port
+	Temperature diode, anode
–	Temperature diode, cathode
n	No connect

## CS144 Pin Function Diagram

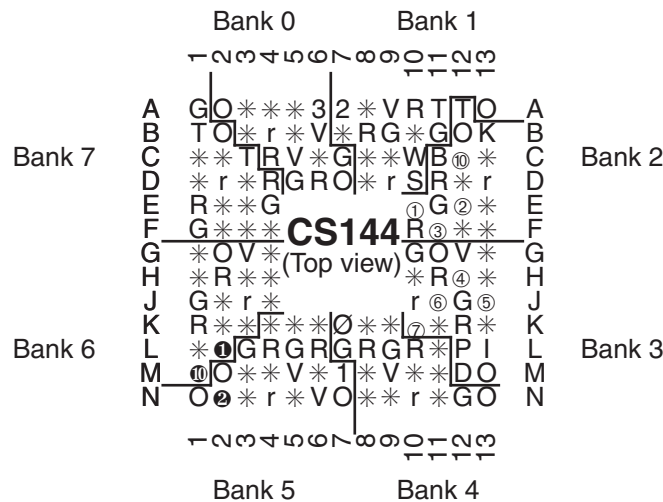
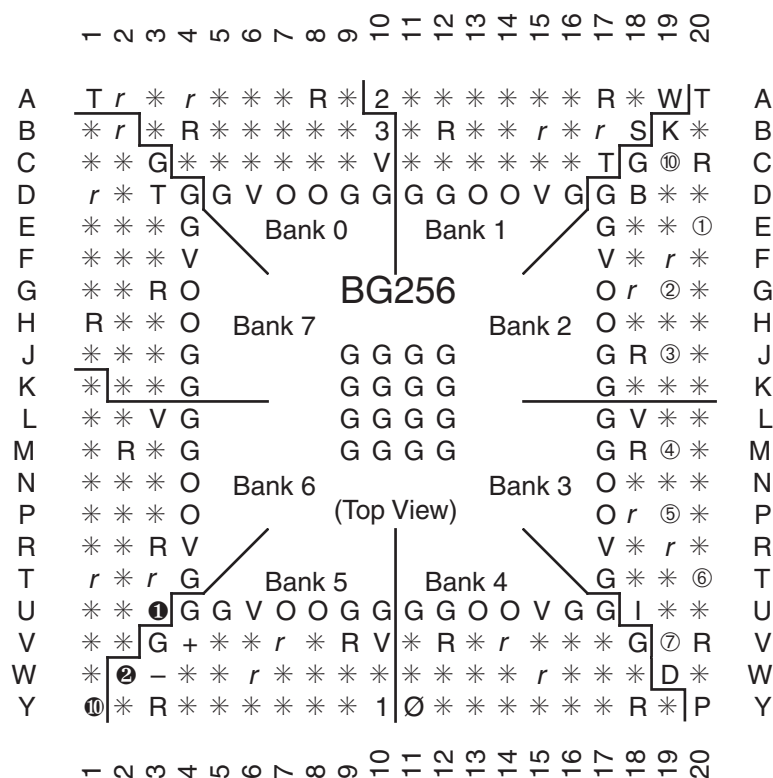


Figure 1: CS144 Pin Function Diagram



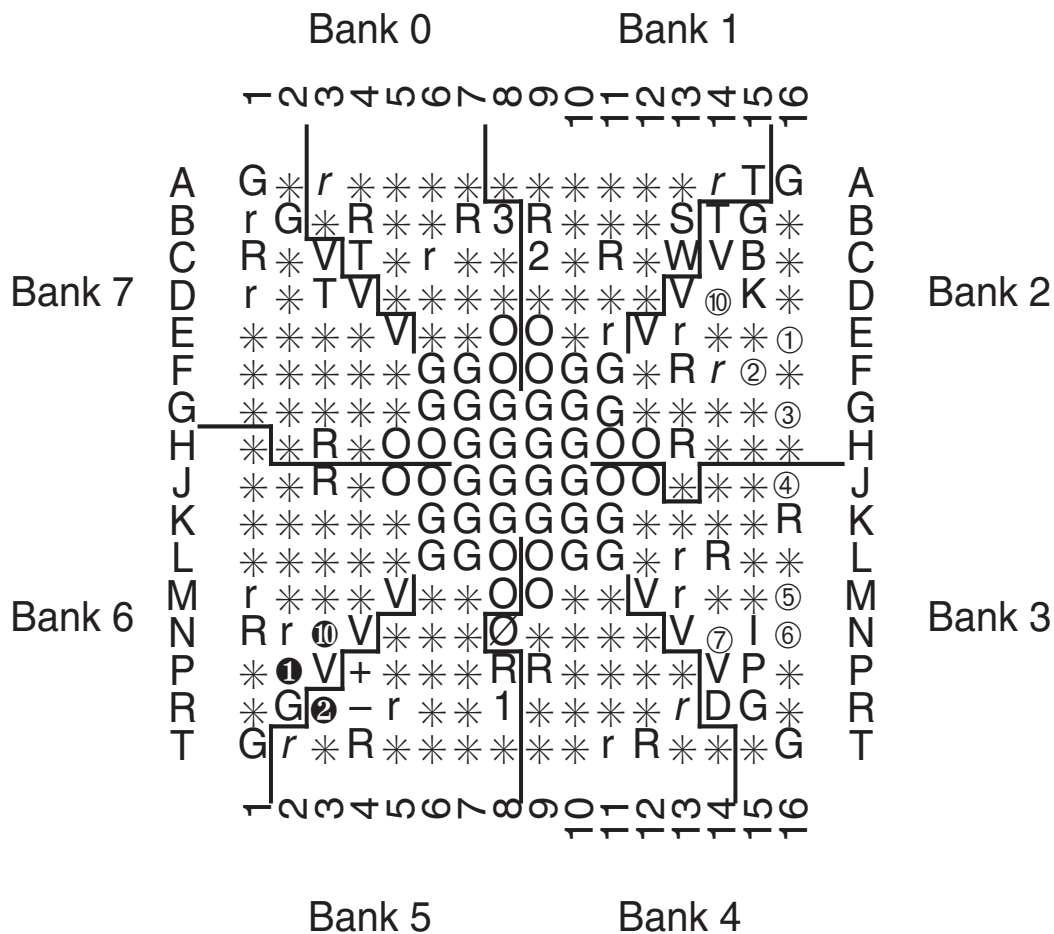
## BG256 Pin Function Diagram



DS003\_18\_100300

Figure 4: BG256 Pin Function Diagram

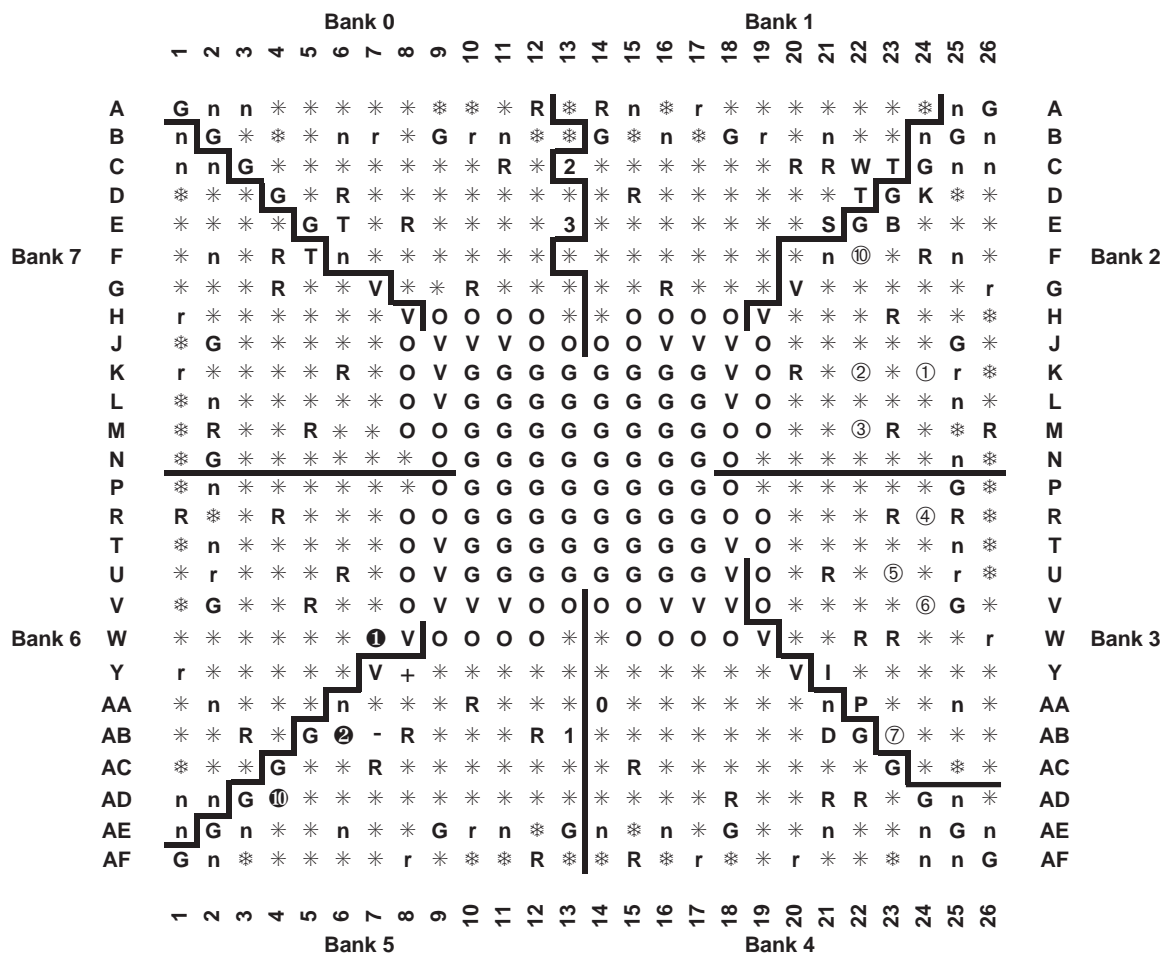
## FG256 Pin Function Diagram



**FG256**  
(Top view)

**Figure 8: FG256 Pin Function Diagram**

## FG676 Pin Function Diagram

FG676  
(Top view)

fg676a

Figure 10: FG676 Pin Function Diagram

## Notes:

Packages FG456 and FG676 are layout compatible.