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Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	864
Number of Logic Elements/Cells	3888
Total RAM Bits	49152
Number of I/O	166
Number of Gates	164674
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	240-BFQFP
Supplier Device Package	240-PQFP (32x32)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcv150-4pq240c

Virtex Architecture

Virtex devices feature a flexible, regular architecture that comprises an array of configurable logic blocks (CLBs) surrounded by programmable input/output blocks (IOBs), all interconnected by a rich hierarchy of fast, versatile routing resources. The abundance of routing resources permits the Virtex family to accommodate even the largest and most complex designs.

Virtex FPGAs are SRAM-based, and are customized by loading configuration data into internal memory cells. In some modes, the FPGA reads its own configuration data from an external PROM (master serial mode). Otherwise, the configuration data is written into the FPGA (SelectMAP™, slave serial, and JTAG modes).

The standard Xilinx Foundation™ and Alliance Series™ Development systems deliver complete design support for Virtex, covering every aspect from behavioral and schematic entry, through simulation, automatic design translation and implementation, to the creation, downloading, and readback of a configuration bit stream.

Higher Performance

Virtex devices provide better performance than previous generations of FPGA. Designs can achieve synchronous system clock rates up to 200 MHz including I/O. Virtex inputs and outputs comply fully with PCI specifications, and interfaces can be implemented that operate at 33 MHz or 66 MHz. Additionally, Virtex supports the hot-swapping requirements of Compact PCI.

Xilinx thoroughly benchmarked the Virtex family. While performance is design-dependent, many designs operated internally at speeds in excess of 100 MHz and can achieve 200 MHz. Table 2 shows performance data for representative circuits, using worst-case timing parameters.

Table 2: Performance for Common Circuit Functions

Function	Bits	Virtex -6
Register-to-Register		
Adder	16	5.0 ns
	64	7.2 ns
Pipelined Multiplier	8 x 8	5.1 ns
	16 x 16	6.0 ns
Address Decoder	16	4.4 ns
	64	6.4 ns
16:1 Multiplexer		5.4 ns
Parity Tree	9	4.1 ns
	18	5.0 ns
	36	6.9 ns
Chip-to-Chip		
HSTL Class IV		200 MHz
LVTTTL, 16mA, fast slew		180 MHz

Virtex Device/Package Combinations and Maximum I/O

Table 3: Virtex Family Maximum User I/O by Device/Package (Excluding Dedicated Clock Pins)

Package	XCV50	XCV100	XCV150	XCV200	XCV300	XCV400	XCV600	XCV800	XCV1000
CS144	94	94							
TQ144	98	98							
PQ240	166	166	166	166	166				
HQ240						166	166	166	
BG256	180	180	180	180					
BG352			260	260	260				
BG432					316	316	316	316	
BG560						404	404	404	404
FG256	176	176	176	176					
FG456			260	284	312				
FG676						404	444	444	
FG680							512	512	512

Virtex Ordering Information

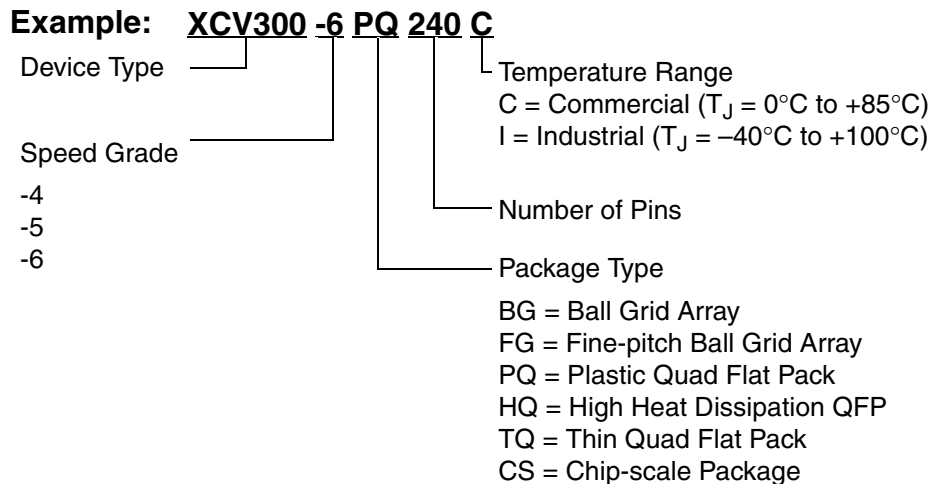


Figure 1: Virtex Ordering Information

Input Path

A buffer in the Virtex IOB input path routes the input signal either directly to internal logic or through an optional input flip-flop.

An optional delay element at the D-input of this flip-flop eliminates pad-to-pad hold time. The delay is matched to the internal clock-distribution delay of the FPGA, and when used, assures that the pad-to-pad hold time is zero.

Each input buffer can be configured to conform to any of the low-voltage signalling standards supported. In some of these standards the input buffer utilizes a user-supplied threshold voltage, V_{REF} . The need to supply V_{REF} imposes constraints on which standards can be used in close proximity to each other. See [I/O Banking, page 3](#).

There are optional pull-up and pull-down resistors at each user I/O input for use after configuration. Their value is in the range 50 k Ω – 100 k Ω .

Output Path

The output path includes a 3-state output buffer that drives the output signal onto the pad. The output signal can be routed to the buffer directly from the internal logic or through an optional IOB output flip-flop.

The 3-state control of the output can also be routed directly from the internal logic or through a flip-flop that provides synchronous enable and disable.

Each output driver can be individually programmed for a wide range of low-voltage signalling standards. Each output buffer can source up to 24 mA and sink up to 48mA. Drive strength and slew rate controls minimize bus transients.

In most signalling standards, the output High voltage depends on an externally supplied V_{CCO} voltage. The need to supply V_{CCO} imposes constraints on which standards can be used in close proximity to each other. See [I/O Banking, page 3](#).

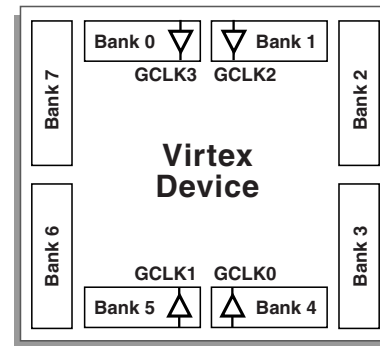
An optional weak-keeper circuit is connected to each output. When selected, the circuit monitors the voltage on the pad and weakly drives the pin High or Low to match the input signal. If the pin is connected to a multiple-source signal, the weak keeper holds the signal in its last state if all drivers are disabled. Maintaining a valid logic level in this way eliminates bus chatter.

Because the weak-keeper circuit uses the IOB input buffer to monitor the input level, an appropriate V_{REF} voltage must be provided if the signalling standard requires one. The provision of this voltage must comply with the I/O banking rules.

I/O Banking

Some of the I/O standards described above require V_{CCO} and/or V_{REF} voltages. These voltages externally and connected to device pins that serve groups of IOBs, called banks. Consequently, restrictions exist about which I/O standards can be combined within a given bank.

Eight I/O banks result from separating each edge of the FPGA into two banks, as shown in [Figure 3](#). Each bank has multiple V_{CCO} pins, all of which must be connected to the same voltage. This voltage is determined by the output standards in use.



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Figure 3: Virtex I/O Banks

Within a bank, output standards can be mixed only if they use the same V_{CCO} . Compatible standards are shown in [Table 2](#). GTL and GTL+ appear under all voltages because their open-drain outputs do not depend on V_{CCO} .

Table 2: Compatible Output Standards

V_{CCO}	Compatible Standards
3.3 V	PCI, LVTTTL, SSTL3 I, SSTL3 II, CTT, AGP, GTL, GTL+
2.5 V	SSTL2 I, SSTL2 II, LVCMOS2, GTL, GTL+
1.5 V	HSTL I, HSTL III, HSTL IV, GTL, GTL+

Some input standards require a user-supplied threshold voltage, V_{REF} . In this case, certain user-I/O pins are automatically configured as inputs for the V_{REF} voltage. Approximately one in six of the I/O pins in the bank assume this role.

The V_{REF} pins within a bank are interconnected internally and consequently only one V_{REF} voltage can be used within each bank. All V_{REF} pins in the bank, however, must be connected to the external voltage source for correct operation.

Within a bank, inputs that require V_{REF} can be mixed with those that do not. However, only one V_{REF} voltage can be used within a bank. Input buffers that use V_{REF} are not 5 V tolerant. LVTTTL, LVCMOS2, and PCI 33 MHz 5 V, are 5 V tolerant.

The V_{CCO} and V_{REF} pins for each bank appear in the device Pinout tables and diagrams. The diagrams also show the bank affiliation of each I/O.

Within a given package, the number of V_{REF} and V_{CCO} pins can vary depending on the size of device. In larger devices,

General Purpose Routing

Most Virtex signals are routed on the general purpose routing, and consequently, the majority of interconnect resources are associated with this level of the routing hierarchy. The general routing resources are located in horizontal and vertical routing channels associated with the rows and columns CLBs. The general-purpose routing resources are listed below.

- Adjacent to each CLB is a General Routing Matrix (GRM). The GRM is the switch matrix through which horizontal and vertical routing resources connect, and is also the means by which the CLB gains access to the general purpose routing.
- 24 single-length lines route GRM signals to adjacent GRMs in each of the four directions.
- 12 buffered Hex lines route GRM signals to another GRMs six-blocks away in each one of the four directions. Organized in a staggered pattern, Hex lines can be driven only at their endpoints. Hex-line signals can be accessed either at the endpoints or at the midpoint (three blocks from the source). One third of the Hex lines are bidirectional, while the remaining ones are uni-directional.

- 12 Longlines are buffered, bidirectional wires that distribute signals across the device quickly and efficiently. Vertical Longlines span the full height of the device, and horizontal ones span the full width of the device.

I/O Routing

Virtex devices have additional routing resources around their periphery that form an interface between the CLB array and the IOBs. This additional routing, called the VersaRing, facilitates pin-swapping and pin-locking, such that logic redesigns can adapt to existing PCB layouts. Time-to-market is reduced, since PCBs and other system components can be manufactured while the logic design is still in progress.

Dedicated Routing

Some classes of signal require dedicated routing resources to maximize performance. In the Virtex architecture, dedicated routing resources are provided for two classes of signal.

- Horizontal routing resources are provided for on-chip 3-state busses. Four partitionable bus lines are provided per CLB row, permitting multiple busses within a row, as shown in Figure 8.
- Two dedicated nets per CLB propagate carry signals vertically to the adjacent CLB.

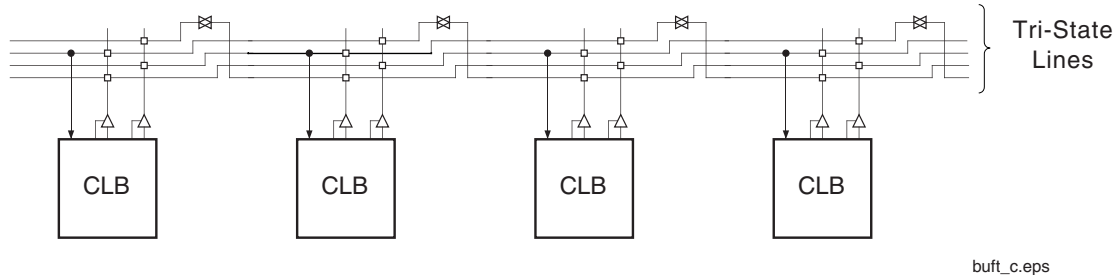


Figure 8: BUFT Connections to Dedicated Horizontal Bus Lines

Global Routing

Global Routing resources distribute clocks and other signals with very high fanout throughout the device. Virtex devices include two tiers of global routing resources referred to as primary global and secondary local clock routing resources.

- The primary global routing resources are four dedicated global nets with dedicated input pins that are designed to distribute high-fanout clock signals with minimal skew. Each global clock net can drive all CLB, IOB, and block RAM clock pins. The primary global nets can only be driven by global buffers. There are four global buffers, one for each global net.

- The secondary local clock routing resources consist of 24 backbone lines, 12 across the top of the chip and 12 across bottom. From these lines, up to 12 unique signals per column can be distributed via the 12 longlines in the column. These secondary resources are more flexible than the primary resources since they are not restricted to routing only to clock pins.

Clock Distribution

Virtex provides high-speed, low-skew clock distribution through the primary global routing resources described above. A typical clock distribution net is shown in Figure 9.

Four global buffers are provided, two at the top center of the device and two at the bottom center. These drive the four primary global nets that in turn drive any clock pin.

In addition to the test instructions outlined above, the boundary-scan circuitry can be used to configure the FPGA, and also to read back the configuration data.

Figure 10 is a diagram of the Virtex Series boundary scan logic. It includes three bits of Data Register per IOB, the IEEE 1149.1 Test Access Port controller, and the Instruction Register with decodes.

Instruction Set

The Virtex Series boundary scan instruction set also includes instructions to configure the device and read back configuration data (CFG_IN, CFG_OUT, and JSTART). The complete instruction set is coded as shown in **Table 5**.

Data Registers

The primary data register is the boundary scan register. For each IOB pin in the FPGA, bonded or not, it includes three bits for In, Out, and 3-State Control. Non-IOB pins have appropriate partial bit population if input-only or output-only. Each EXTEST CAPTURED-OR state captures all In, Out, and 3-state pins.

The other standard data register is the single flip-flop BYPASS register. It synchronizes data being passed through the FPGA to the next downstream boundary scan device.

The FPGA supports up to two additional internal scan chains that can be specified using the BSCAN macro. The macro provides two user pins (SEL1 and SEL2) which are decoded of the USER1 and USER2 instructions respectively. For these instructions, two corresponding pins (TDO1 and TDO2) allow user scan data to be shifted out of TDO.

Likewise, there are individual clock pins (DRCK1 and DRCK2) for each user register. There is a common input pin (TDI) and shared output pins that represent the state of the TAP controller (RESET, SHIFT, and UPDATE).

Bit Sequence

The order within each IOB is: In, Out, 3-State. The input-only pins contribute only the In bit to the boundary scan I/O data register, while the output-only pins contributes all three bits.

From a cavity-up view of the chip (as shown in EPIC), starting in the upper right chip corner, the boundary scan data-register bits are ordered as shown in **Figure 11**.

BSDL (Boundary Scan Description Language) files for Virtex Series devices are available on the Xilinx web site in the File Download area.

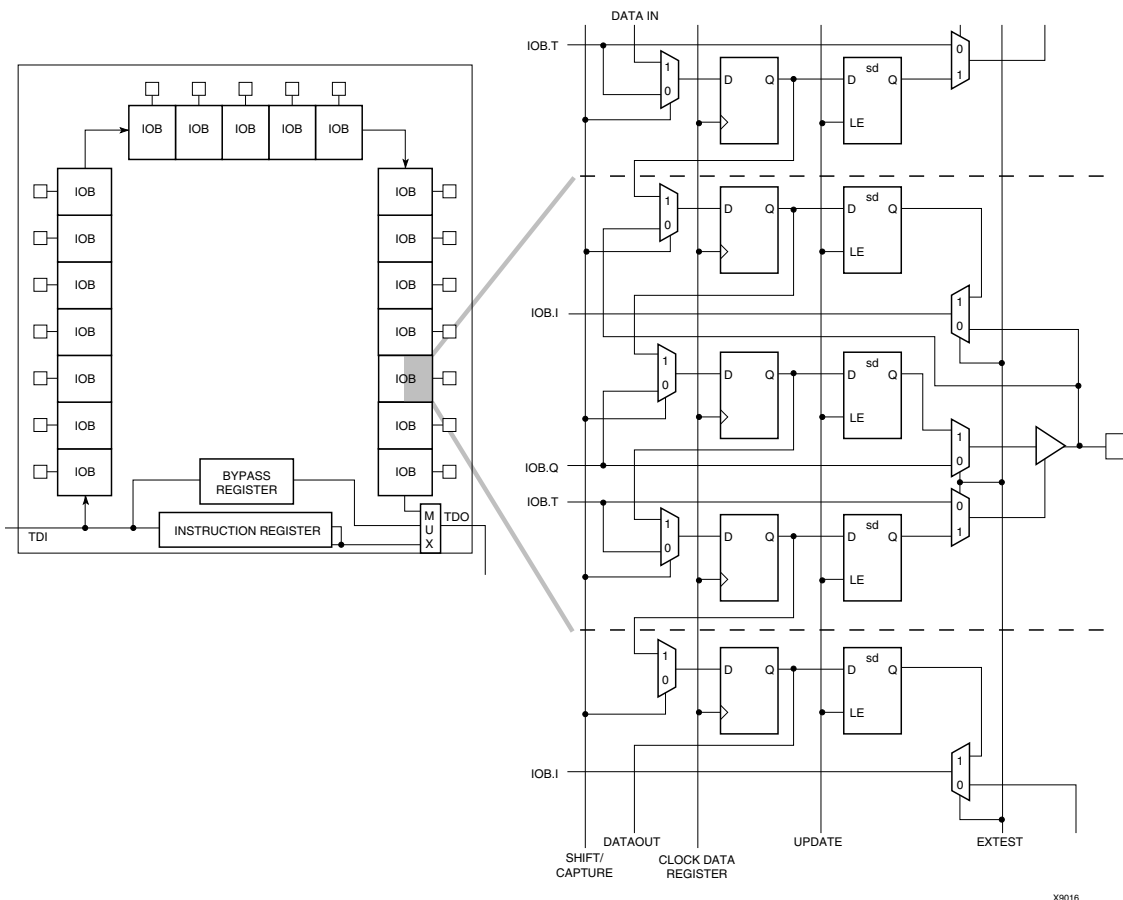


Figure 10: Virtex Series Boundary Scan Logic

ers with a common user interface regardless of their choice of entry and verification tools. The XDM software simplifies the selection of implementation options with pull-down menus and on-line help.

Application programs ranging from schematic capture to Placement and Routing (PAR) can be accessed through the XDM software. The program command sequence is generated prior to execution, and stored for documentation.

Several advanced software features facilitate Virtex design. RPMs, for example, are schematic-based macros with relative location constraints to guide their placement. They help ensure optimal implementation of common functions.

For HDL design entry, the Xilinx FPGA Foundation development system provides interfaces to the following synthesis design environments.

- Synopsys (FPGA Compiler, FPGA Express)
- Exemplar (Spectrum)
- Synplicity (Synplify)

For schematic design entry, the Xilinx FPGA Foundation and alliance development system provides interfaces to the following schematic-capture design environments.

- Mentor Graphics V8 (Design Architect, QuickSim II)
- Viewlogic Systems (Viewdraw)

Third-party vendors support many other environments.

A standard interface-file specification, Electronic Design Interchange Format (EDIF), simplifies file transfers into and out of the development system.

Virtex FPGAs supported by a unified library of standard functions. This library contains over 400 primitives and macros, ranging from 2-input AND gates to 16-bit accumulators, and includes arithmetic functions, comparators, counters, data registers, decoders, encoders, I/O functions, latches, Boolean functions, multiplexers, shift registers, and barrel shifters.

The “soft macro” portion of the library contains detailed descriptions of common logic functions, but does not contain any partitioning or placement information. The performance of these macros depends, therefore, on the partitioning and placement obtained during implementation.

RPMs, on the other hand, do contain predetermined partitioning and placement information that permits optimal implementation of these functions. Users can create their own library of soft macros or RPMs based on the macros and primitives in the standard library.

The design environment supports hierarchical design entry, with high-level schematics that comprise major functional blocks, while lower-level schematics define the logic in these blocks. These hierarchical design elements are automatically combined by the implementation tools. Different design entry tools can be combined within a hierarchical

design, thus allowing the most convenient entry method to be used for each portion of the design.

Design Implementation

The place-and-route tools (PAR) automatically provide the implementation flow described in this section. The partitioner takes the EDIF net list for the design and maps the logic into the architectural resources of the FPGA (CLBs and IOBs, for example). The placer then determines the best locations for these blocks based on their interconnections and the desired performance. Finally, the router interconnects the blocks.

The PAR algorithms support fully automatic implementation of most designs. For demanding applications, however, the user can exercise various degrees of control over the process. User partitioning, placement, and routing information is optionally specified during the design-entry process. The implementation of highly structured designs can benefit greatly from basic floor planning.

The implementation software incorporates Timing Wizard® timing-driven placement and routing. Designers specify timing requirements along entire paths during design entry. The timing path analysis routines in PAR then recognize these user-specified requirements and accommodate them.

Timing requirements are entered on a schematic in a form directly relating to the system requirements, such as the targeted clock frequency, or the maximum allowable delay between two registers. In this way, the overall performance of the system along entire signal paths is automatically tailored to user-generated specifications. Specific timing information for individual nets is unnecessary.

Design Verification

In addition to conventional software simulation, FPGA users can use in-circuit debugging techniques. Because Xilinx devices are infinitely reprogrammable, designs can be verified in real time without the need for extensive sets of software simulation vectors.

The development system supports both software simulation and in-circuit debugging techniques. For simulation, the system extracts the post-layout timing information from the design database, and back-annotates this information into the net list for use by the simulator. Alternatively, the user can verify timing-critical portions of the design using the TRACE® static timing analyzer.

For in-circuit debugging, the development system includes a download and readback cable. This cable connects the FPGA in the target system to a PC or workstation. After downloading the design into the FPGA, the designer can single-step the logic, readback the contents of the flip-flops, and so observe the internal logic state. Simple modifications can be downloaded into the system in a matter of minutes.

Configuration

Virtex devices are configured by loading configuration data into the internal configuration memory. Some of the pins used for this are dedicated configuration pins, while others can be re-used as general purpose inputs and outputs once configuration is complete.

The following are dedicated pins:

- Mode pins (M2, M1, M0)
- Configuration clock pin (CCLK)
- $\overline{\text{PROGRAM}}$ pin
- DONE pin
- Boundary-scan pins (TDI, TDO, TMS, TCK)

Depending on the configuration mode chosen, CCLK can be an output generated by the FPGA, or it can be generated externally and provided to the FPGA as an input. The $\overline{\text{PROGRAM}}$ pin must be pulled High prior to reconfiguration.

Note that some configuration pins can act as outputs. For correct operation, these pins can require a V_{CCO} of 3.3 V to permit LVTTTL operation. All the pins affected are in banks 2 or 3. The configuration pins needed for SelectMap (CS, Write) are located in bank 1.

Table 7: Configuration Codes

Configuration Mode	M2	M1	M0	CCLK Direction	Data Width	Serial D _{out}	Configuration Pull-ups
Master-serial mode	0	0	0	Out	1	Yes	No
Boundary-scan mode	1	0	1	N/A	1	No	No
SelectMAP mode	1	1	0	In	8	No	No
Slave-serial mode	1	1	1	In	1	Yes	No
Master-serial mode	1	0	0	Out	1	Yes	Yes
Boundary-scan mode	0	0	1	N/A	1	No	Yes
SelectMAP mode	0	1	0	In	8	No	Yes
Slave-serial mode	0	1	1	In	1	Yes	Yes

Slave-Serial Mode

In slave-serial mode, the FPGA receives configuration data in bit-serial form from a serial PROM or other source of serial configuration data. The serial bitstream must be setup at the DIN input pin a short time before each rising edge of an externally generated CCLK.

For more information on serial PROMs, see the PROM data sheet at:

<http://www.xilinx.com/bvdocs/publications/ds026.pdf>.

Multiple FPGAs can be daisy-chained for configuration from a single source. After a particular FPGA has been configured, the data for the next device is routed to the DOUT pin. The data on the DOUT pin changes on the rising edge of CCLK.

The change of DOUT on the rising edge of CCLK differs from previous families, but does not cause a problem for

After Virtex devices are configured, unused IOBs function as 3-state OBUFTs with weak pull downs. For a more detailed description than that given below, see the XAPP138, Virtex Configuration and Readback.

Configuration Modes

Virtex supports the following four configuration modes.

- Slave-serial mode
- Master-serial mode
- SelectMAP mode
- Boundary-scan mode

The Configuration mode pins (M2, M1, M0) select among these configuration modes with the option in each case of having the IOB pins either pulled up or left floating prior to configuration. The selection codes are listed in Table 7.

Configuration through the boundary-scan port is always available, independent of the mode selection. Selecting the boundary-scan mode simply turns off the other modes. The three mode pins have internal pull-up resistors, and default to a logic High if left unconnected. However, it is recommended to drive the configuration mode pins externally.

mixed configuration chains. This change was made to improve serial configuration rates for Virtex-only chains.

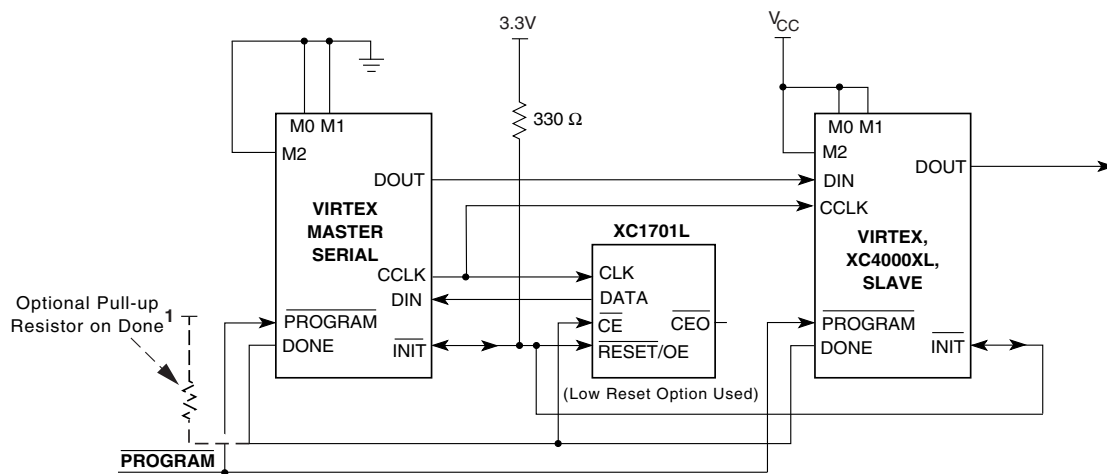
Figure 12 shows a full master/slave system. A Virtex device in slave-serial mode should be connected as shown in the third device from the left.

Slave-serial mode is selected by applying <111> or <011> to the mode pins (M2, M1, M0). A weak pull-up on the mode pins makes slave-serial the default mode if the pins are left unconnected. However, it is recommended to drive the configuration mode pins externally. Figure 13 shows slave-serial mode programming switching characteristics.

Table 8 provides more detail about the characteristics shown in Figure 13. Configuration must be delayed until the $\overline{\text{INIT}}$ pins of all daisy-chained FPGAs are High.

Table 8: Master/Slave Serial Mode Programming Switching

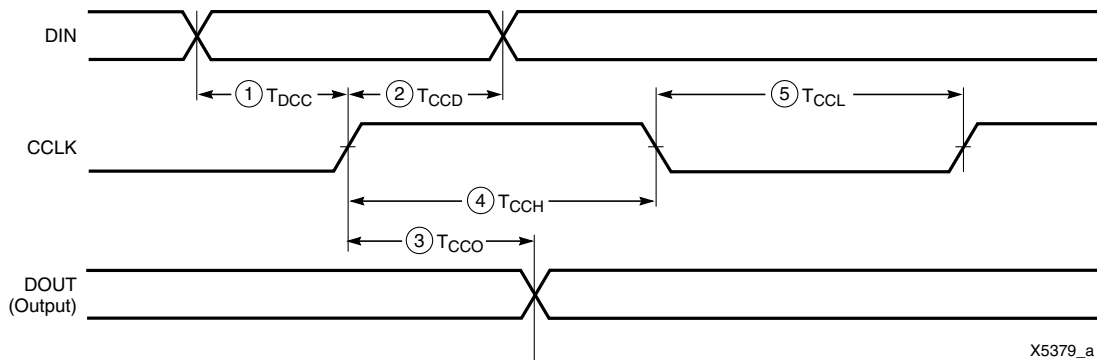
	Description	Figure References	Symbol	Values	Units
CCLK	DIN setup/hold, slave mode	1/2	T_{DCC}/T_{CCD}	5.0 / 0	ns, min
	DIN setup/hold, master mode	1/2	T_{DSCK}/T_{CKDS}	5.0 / 0	ns, min
	DOUT	3	T_{CCO}	12.0	ns, max
	High time	4	T_{CCH}	5.0	ns, min
	Low time	5	T_{CCL}	5.0	ns, min
	Maximum Frequency		F_{CC}	66	MHz, max
	Frequency Tolerance, master mode with respect to nominal			+45% -30%	



Note 1: If none of the Virtex FPGAs have been selected to drive DONE, an external pull-up resistor of 330 Ω should be added to the common DONE line. (For Spartan-XL devices, add a 4.7K Ω pull-up resistor.) This pull-up is not needed if the DriveDONE attribute is set. If used, DriveDONE should be selected only for the last device in the configuration chain.

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Figure 12: Master/Slave Serial Mode Circuit Diagram



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Figure 13: Slave-Serial Mode Programming Switching Characteristics

Master-Serial Mode

In master-serial mode, the CCLK output of the FPGA drives a Xilinx Serial PROM that feeds bit-serial data to the DIN input. The FPGA accepts this data on each rising CCLK edge. After the FPGA has been loaded, the data for the next device in a daisy-chain is presented on the DOUT pin after the rising CCLK edge.

The interface is identical to slave-serial except that an internal oscillator is used to generate the configuration clock (CCLK). A wide range of frequencies can be selected for CCLK which always starts at a slow default frequency. Configuration bits then switch CCLK to a higher frequency for the remainder of the configuration. Switching to a lower frequency is prohibited.

The CCLK frequency is set using the ConfigRate option in the bitstream generation software. The maximum CCLK frequency that can be selected is 60 MHz. When selecting a CCLK frequency, ensure that the serial PROM and any

daisy-chained FPGAs are fast enough to support the clock rate.

On power-up, the CCLK frequency is 2.5 MHz. This frequency is used until the ConfigRate bits have been loaded when the frequency changes to the selected ConfigRate. Unless a different frequency is specified in the design, the default ConfigRate is 4 MHz.

Figure 12 shows a full master/slave system. In this system, the left-most device operates in master-serial mode. The remaining devices operate in slave-serial mode. The SPROM RESET pin is driven by INIT, and the CE input is driven by DONE. There is the potential for contention on the DONE pin, depending on the start-up sequence options chosen.

Figure 14 shows the timing of master-serial configuration. Master-serial mode is selected by a <000> or <100> on the mode pins (M2, M1, M0). Table 8 shows the timing information for Figure 14.

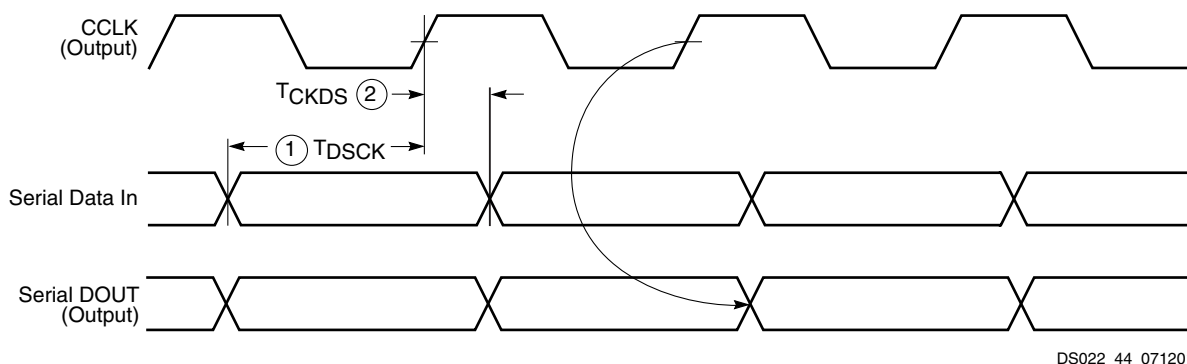


Figure 14: Master-Serial Mode Programming Switching Characteristics

At power-up, V_{CC} must rise from 1.0 V to V_{CC} min in less than 50 ms, otherwise delay configuration by pulling PROGRAM Low until V_{CC} is valid.

The sequence of operations necessary to configure a Virtex FPGA serially appears in Figure 15.

SelectMAP Mode

The SelectMAP mode is the fastest configuration option. Byte-wide data is written into the FPGA with a BUSY flag controlling the flow of data.

An external data source provides a byte stream, CCLK, a Chip Select (\overline{CS}) signal and a Write signal (\overline{WRITE}). If BUSY is asserted (High) by the FPGA, the data must be held until BUSY goes Low.

Data can also be read using the SelectMAP mode. If \overline{WRITE} is not asserted, configuration data is read out of the FPGA as part of a readback operation.

In the SelectMAP mode, multiple Virtex devices can be chained in parallel. DATA pins (D7:D0), CCLK, \overline{WRITE} , \overline{BUSY} , $\overline{PROGRAM}$, \overline{DONE} , and \overline{INIT} can be connected in parallel between all the FPGAs. Note that the data is organized with the MSB of each byte on pin D0 and the LSB of each byte on D7. The \overline{CS} pins are kept separate, insuring that each FPGA can be selected individually. \overline{WRITE} should be Low before loading the first bitstream and returned High after the last device has been programmed. Use \overline{CS} to select the appropriate FPGA for loading the bitstream and sending the configuration data. At the end of the bitstream, deselect the loaded device and select the next target FPGA by setting its \overline{CS} pin High. A free-running oscillator or other externally generated signal can be used for CCLK. The \overline{BUSY} signal can be ignored for frequencies below 50 MHz. For details about frequencies above 50 MHz, see XAPP138, Virtex Configuration and Readback. Once all the devices have been programmed, the \overline{DONE} pin goes High.

IOB Input Switching Characteristics Standard Adjustments

Description	Symbol	Standard ⁽¹⁾	Speed Grade				Units
			Min	-6	-5	-4	
Data Input Delay Adjustments							
Standard-specific data input delay adjustments	T _{ILVTTL}	LVTTL	0	0	0	0	ns
	T _{ILVCMOS2}	LVCMSO2	−0.02	−0.04	−0.04	−0.05	ns
	T _{IPCI33_3}	PCI, 33 MHz, 3.3 V	−0.05	−0.11	−0.12	−0.14	ns
	T _{IPCI33_5}	PCI, 33 MHz, 5.0 V	0.13	0.25	0.28	0.33	ns
	T _{IPCI66_3}	PCI, 66 MHz, 3.3 V	−0.05	−0.11	−0.12	−0.14	ns
	T _{IGTL}	GTL	0.10	0.20	0.23	0.26	ns
	T _{IGTLP}	GTL+	0.06	0.11	0.12	0.14	ns
	T _{IHSTL}	HSTL	0.02	0.03	0.03	0.04	ns
	T _{ISSTL2}	SSTL2	−0.04	−0.08	−0.09	−0.10	ns
	T _{ISSTL3}	SSTL3	−0.02	−0.04	−0.05	−0.06	ns
	T _{ICTT}	CTT	0.01	0.02	0.02	0.02	ns
	T _{IAGP}	AGP	−0.03	−0.06	−0.07	−0.08	ns

Notes:

- Input timing for LVTTL is measured at 1.4 V. For other I/O standards, see [Table 3](#).

IOB Output Switching Characteristics

Output delays terminating at a pad are specified for LVTTL with 12 mA drive and fast slew rate. For other standards, adjust the delays with the values shown in [IOB Output Switching Characteristics Standard Adjustments, page 9](#).

Description	Symbol	Speed Grade				Units
		Min	-6	-5	-4	
Propagation Delays						
O input to Pad	T _{IOOP}	1.2	2.9	3.2	3.5	ns, max
O input to Pad via transparent latch	T _{IOOLP}	1.4	3.4	3.7	4.0	ns, max
3-State Delays						
T input to Pad high-impedance ⁽¹⁾	T _{IOTHZ}	1.0	2.0	2.2	2.4	ns, max
T input to valid data on Pad	T _{IOTON}	1.4	3.1	3.3	3.7	ns, max
T input to Pad high-impedance via transparent latch ⁽¹⁾	T _{IOTLPHZ}	1.2	2.4	2.6	3.0	ns, max
T input to valid data on Pad via transparent latch	T _{IOTLPON}	1.6	3.5	3.8	4.2	ns, max
GTS to Pad high impedance ⁽¹⁾	T _{GTS}	2.5	4.9	5.5	6.3	ns, max
Sequential Delays						
Clock CLK						
Minimum Pulse Width, High	T _{CH}	0.8	1.5	1.7	2.0	ns, min
Minimum Pulse Width, Low	T _{CL}	0.8	1.5	1.7	2.0	ns, min

I/O Standard Global Clock Input Adjustments

Description	Symbol	Standard ⁽¹⁾	Speed Grade				Units
			Min	-6	-5	-4	
Data Input Delay Adjustments							
Standard-specific global clock input delay adjustments	T _{GPLVTTL}	LVTTL	0	0	0	0	ns, max
	T _{GPLVCMOS2}	LVC MOS2	−0.02	−0.04	−0.04	−0.05	ns, max
	T _{GP PCI33_3}	PCI, 33 MHz, 3.3 V	−0.05	−0.11	−0.12	−0.14	ns, max
	T _{GP PCI33_5}	PCI, 33 MHz, 5.0 V	0.13	0.25	0.28	0.33	ns, max
	T _{GP PCI66_3}	PCI, 66 MHz, 3.3 V	−0.05	−0.11	−0.12	−0.14	ns, max
	T _{GPGTL}	GTL	0.7	0.8	0.9	0.9	ns, max
	T _{GPGTLP}	GTL+	0.7	0.8	0.8	0.8	ns, max
	T _{GPHSTL}	HSTL	0.7	0.7	0.7	0.7	ns, max
	T _{GPSSTL2}	SSTL2	0.6	0.52	0.51	0.50	ns, max
	T _{GPSSTL3}	SSTL3	0.6	0.6	0.55	0.54	ns, max
	T _{GPCTT}	CTT	0.7	0.7	0.7	0.7	ns, max
	T _{GPAGP}	AGP	0.6	0.54	0.53	0.52	ns, max

Notes:

1. Input timing for GPLVTTL is measured at 1.4 V. For other I/O standards, see [Table 3](#).

Block RAM Switching Characteristics

Description	Symbol	Speed Grade				Units
		Min	-6	-5	-4	
Sequential Delays						
Clock CLK to DOUT output	T _{BCKO}	1.7	3.4	3.8	4.3	ns, max
Setup and Hold Times before/after Clock CLK ⁽¹⁾	Setup Time / Hold Time					
ADDR inputs	T _{BACK} /T _{BCKA}	0.6 / 0	1.2 / 0	1.3 / 0	1.5 / 0	ns, min
DIN inputs	T _{BDCK} /T _{BCKD}	0.6 / 0	1.2 / 0	1.3 / 0	1.5 / 0	ns, min
EN input	T _{BECK} /T _{BCKE}	1.3 / 0	2.6 / 0	3.0 / 0	3.4 / 0	ns, min
RST input	T _{BRCK} /T _{BCKR}	1.3 / 0	2.5 / 0	2.7 / 0	3.2 / 0	ns, min
WEN input	T _{BWCK} /T _{BCKW}	1.2 / 0	2.3 / 0	2.6 / 0	3.0 / 0	ns, min
Clock CLK						
Minimum Pulse Width, High	T _{BPWH}	0.8	1.5	1.7	2.0	ns, min
Minimum Pulse Width, Low	T _{BPWL}	0.8	1.5	1.7	2.0	ns, min
CLKA -> CLKB setup time for different ports	T _{BCCS}		3.0	3.5	4.0	ns, min

Notes:

1. A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.

TBUF Switching Characteristics

Description	Symbol	Speed Grade				Units
		Min	-6	-5	-4	
Combinatorial Delays						
IN input to OUT output	T _{IO}	0	0	0	0	ns, max
TRI input to OUT output high-impedance	T _{OFF}	0.05	0.09	0.10	0.11	ns, max
TRI input to valid data on OUT output	T _{ON}	0.05	0.09	0.10	0.11	ns, max

JTAG Test Access Port Switching Characteristics

Description	Symbol	Speed Grade			Units
		-6	-5	-4	
TMS and TDI Setup times before TCK	T_{TAPTCK}	4.0	4.0	4.0	ns, min
TMS and TDI Hold times after TCK	T_{TCKTAP}	2.0	2.0	2.0	ns, min
Output delay from clock TCK to output TDO	T_{TCKTDO}	11.0	11.0	11.0	ns, max
Maximum TCK clock frequency	F_{TCK}	33	33	33	MHz, max

Virtex Pin-to-Pin Output Parameter Guidelines

All devices are 100% functionally tested. Listed below are representative values for typical pin locations and normal clock loading. Values are expressed in nanoseconds unless otherwise noted.

Global Clock Input to Output Delay for LVTTL, 12 mA, Fast Slew Rate, *with* DLL

Description	Symbol	Device	Speed Grade				Units
			Min	-6	-5	-4	
LVTTL Global Clock Input to Output Delay using Output Flip-flop, 12 mA, Fast Slew Rate, <i>with</i> DLL. For data <i>output</i> with different standards, adjust delays with the values shown in Output Delay Adjustments.	T _{ICKOFDLL}	XCV50	1.0	3.1	3.3	3.6	ns, max
		XCV100	1.0	3.1	3.3	3.6	ns, max
		XCV150	1.0	3.1	3.3	3.6	ns, max
		XCV200	1.0	3.1	3.3	3.6	ns, max
		XCV300	1.0	3.1	3.3	3.6	ns, max
		XCV400	1.0	3.1	3.3	3.6	ns, max
		XCV600	1.0	3.1	3.3	3.6	ns, max
		XCV800	1.0	3.1	3.3	3.6	ns, max
		XCV1000	1.0	3.1	3.3	3.6	ns, max

Notes:

1. Listed above are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.
2. Output timing is measured at 1.4 V with 35 pF external capacitive load for LVTTL. The 35 pF load does not apply to the Min values. For other I/O standards and different loads, see [Table 2](#) and [Table 3](#).
3. DLL output jitter is already included in the timing calculation.

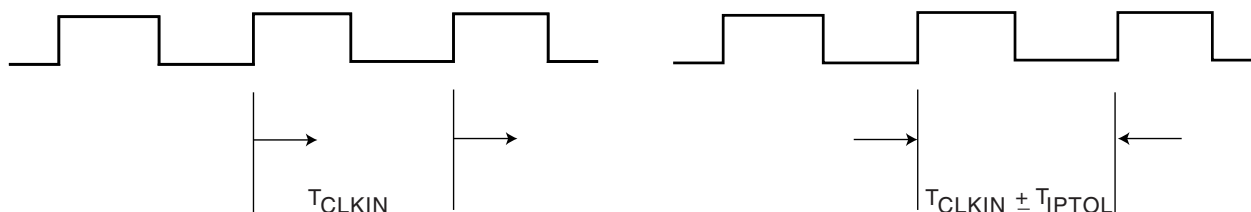
Global Clock Input-to-Output Delay for LVTTL, 12 mA, Fast Slew Rate, *without* DLL

Description	Symbol	Device	Speed Grade				Units
			Min	-6	-5	-4	
LVTTL Global Clock Input to Output Delay using Output Flip-flop, 12 mA, Fast Slew Rate, <i>without</i> DLL. For data <i>output</i> with different standards, adjust delays with the values shown in Input and Output Delay Adjustments. For I/O standards requiring V _{REF} such as GTL, GTL+, SSTL, HSTL, CTT, and AGO, an additional 600 ps must be added.	T _{ICKOF}	XCV50	1.5	4.6	5.1	5.7	ns, max
		XCV100	1.5	4.6	5.1	5.7	ns, max
		XCV150	1.5	4.7	5.2	5.8	ns, max
		XCV200	1.5	4.7	5.2	5.8	ns, max
		XCV300	1.5	4.7	5.2	5.9	ns, max
		XCV400	1.5	4.8	5.3	6.0	ns, max
		XCV600	1.6	4.9	5.4	6.0	ns, max
		XCV800	1.6	4.9	5.5	6.2	ns, max
		XCV1000	1.7	5.0	5.6	6.3	ns, max

Notes:

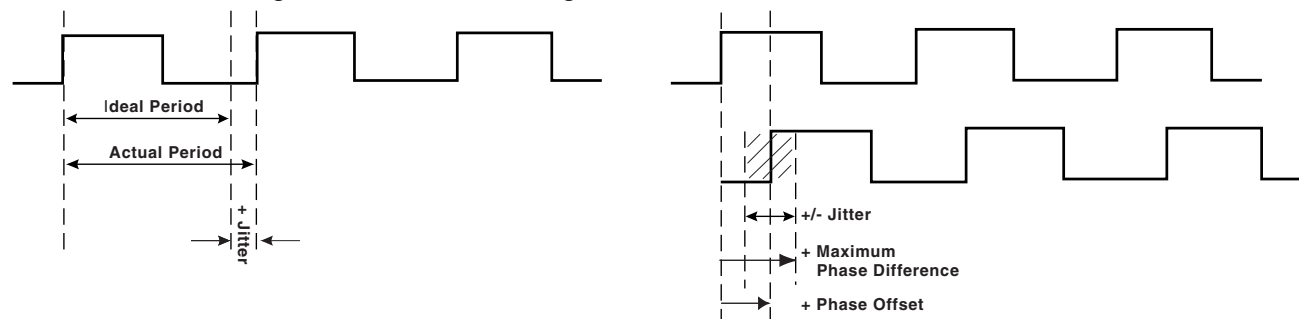
1. Listed above are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.
2. Output timing is measured at 1.4 V with 35 pF external capacitive load for LVTTL. The 35 pF load does not apply to the Min values. For other I/O standards and different loads, see [Table 2](#) and [Table 3](#).

Period Tolerance: the allowed input clock period change in nanoseconds.



Output Jitter: the difference between an ideal reference clock edge and the actual design.

Phase Offset and Maximum Phase Difference



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Figure 1: Frequency Tolerance and Clock Jitter

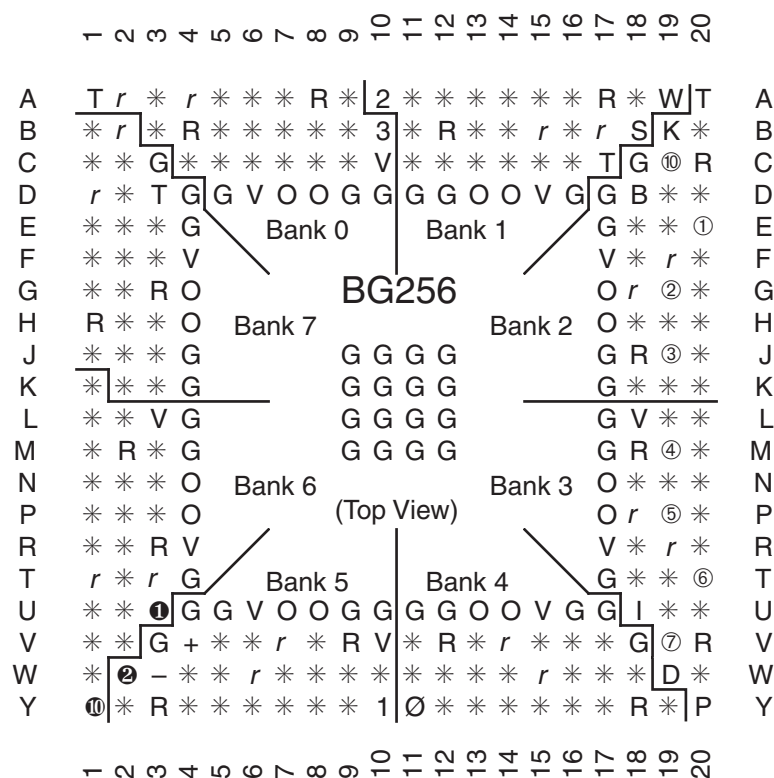
Revision History

Date	Version	Revision
11/98	1.0	Initial Xilinx release.
01/99	1.2	Updated package drawings and specs.
02/99	1.3	Update of package drawings, updated specifications.
05/99	1.4	Addition of package drawings and specifications.
05/99	1.5	Replaced FG 676 & FG680 package drawings.
07/99	1.6	Changed Boundary Scan Information and changed Figure 11, Boundary Scan Bit Sequence. Updated IOB Input & Output delays. Added Capacitance info for different I/O Standards. Added 5 V tolerant information. Added DLL Parameters and waveforms and new Pin-to-pin Input and Output Parameter tables for Global Clock Input to Output and Setup and Hold. Changed Configuration Information including Figures 12, 14, 17 & 19. Added device-dependent listings for quiescent currents ICCINTQ and ICCOQ. Updated IOB Input and Output Delays based on default standard of LVTTTL, 12 mA, Fast Slew Rate. Added IOB Input Switching Characteristics Standard Adjustments.
09/99	1.7	Speed grade update to preliminary status, Power-on specification and Clock-to-Out Minimums additions, "0" hold time listing explanation, quiescent current listing update, and Figure 6 ADDRA input label correction. Added T _{IJITCC} parameter, changed T _{OJIT} to T _{OPHASE} .
01/00	1.8	Update to speed.txt file 1.96. Corrections for CRs 111036, 111137, 112697, 115479, 117153, 117154, and 117612. Modified notes for Recommended Operating Conditions (voltage and temperature). Changed Bank information for V _{CCO} in CS144 package on p.43.

Table 4: Virtex Pinout Tables (Fine-Pitch BGA)

Pin Name	Device	FG256	FG456	FG676	FG680
GCK0	All	N8	W12	AA14	AW19
GCK1	All	R8	Y11	AB13	AU22
GCK2	All	C9	A11	C13	D21
GCK3	All	B8	C11	E13	A20
M0	All	N3	AB2	AD4	AT37
M1	All	P2	U5	W7	AU38
M2	All	R3	Y4	AB6	AT35
CCLK	All	D15	B22	D24	E4
PROGRAM	All	P15	W20	AA22	AT5
DONE	All	R14	Y19	AB21	AU5
INIT	All	N15	V19	Y21	AU2
BUSY/DOUT	All	C15	C21	E23	E3
D0/DIN	All	D14	D20	F22	C2
D1	All	E16	H22	K24	P4
D2	All	F15	H20	K22	P3
D3	All	G16	K20	M22	R1
D4	All	J16	N22	R24	AD3
D5	All	M16	R21	U23	AG2
D6	All	N16	T22	V24	AH1
D7	All	N14	Y21	AB23	AR4
WRITE	All	C13	A20	C22	B4
CS	All	B13	C19	E21	D5
TDI	All	A15	B20	D22	B3
TDO	All	B14	A21	C23	C4
TMS	All	D3	D3	F5	E36
TCK	All	C4	C4	E6	C36
DXN	All	R4	Y5	AB7	AV37
DXP	All	P4	V6	Y8	AU35

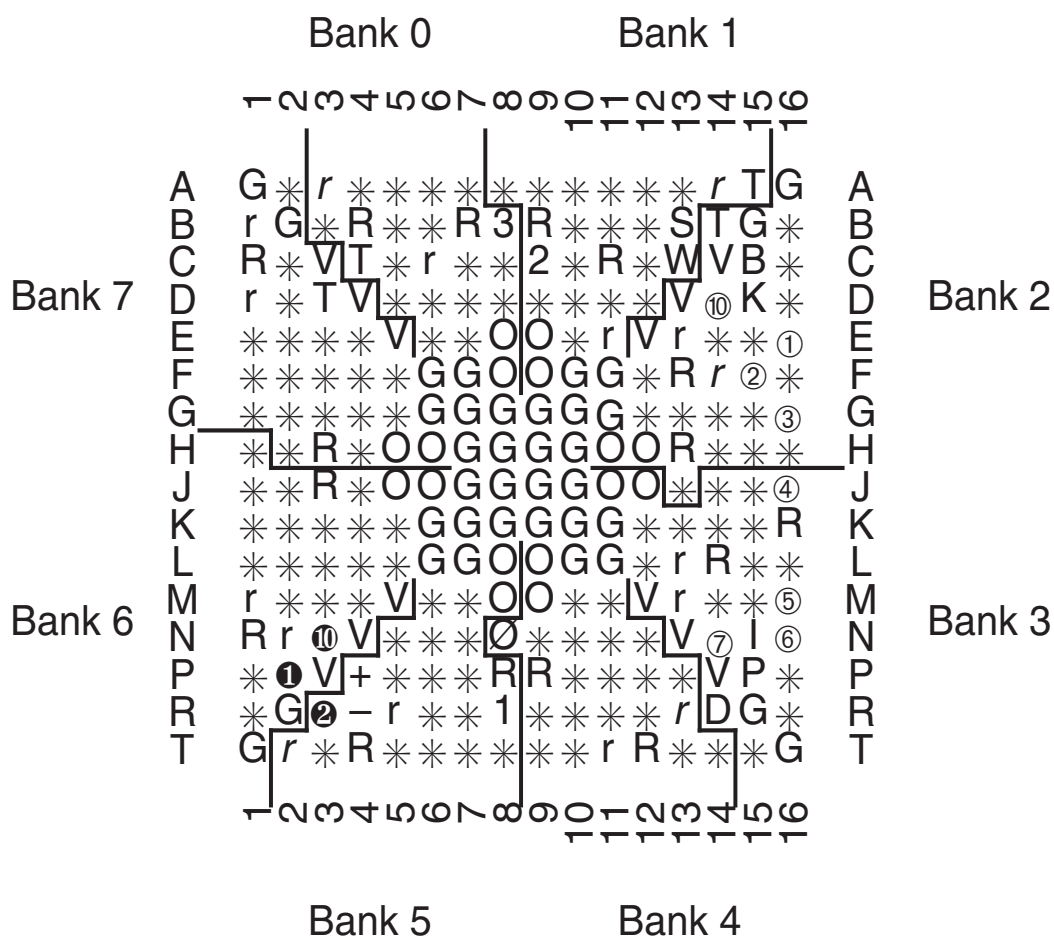
BG256 Pin Function Diagram



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Figure 4: BG256 Pin Function Diagram

FG256 Pin Function Diagram

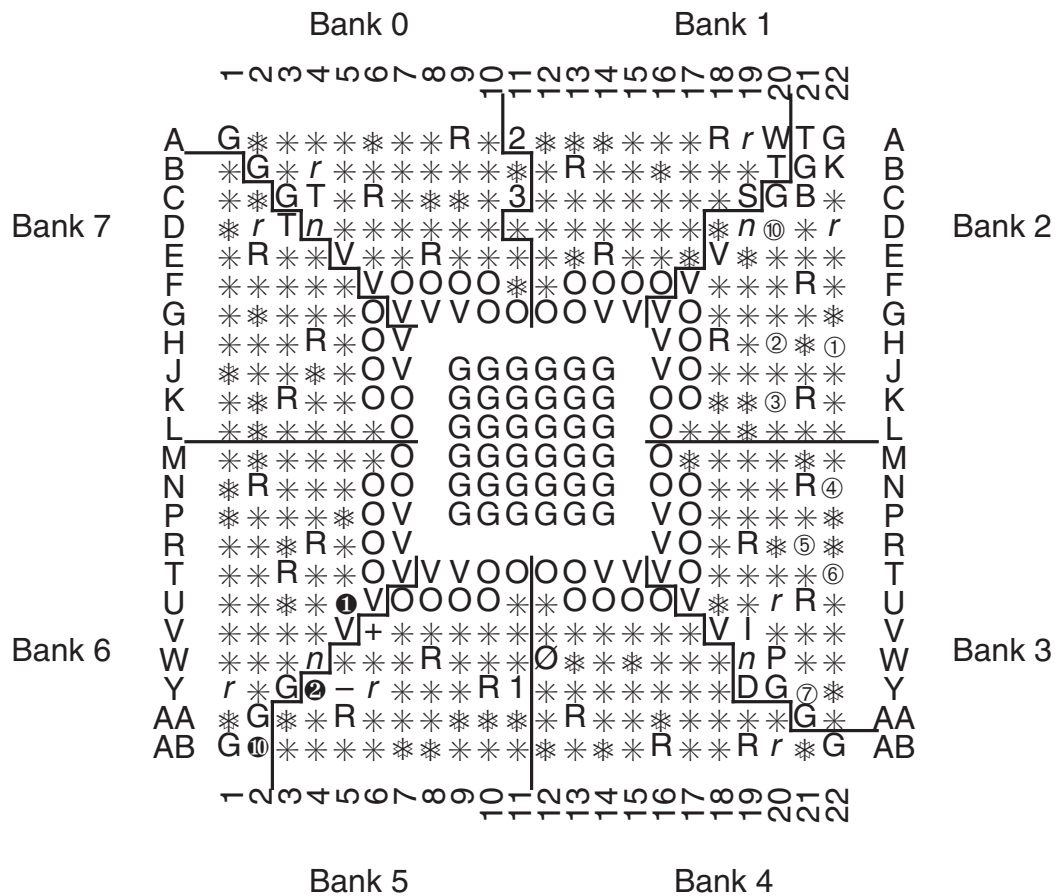


FG256

(Top view)

Figure 8: FG256 Pin Function Diagram

FG456 Pin Function Diagram



FG456 (Top view)

Figure 9: FG456 Pin Function Diagram

Notes:

Packages FG456 and FG676 are layout compatible.

FG676 Pin Function Diagram

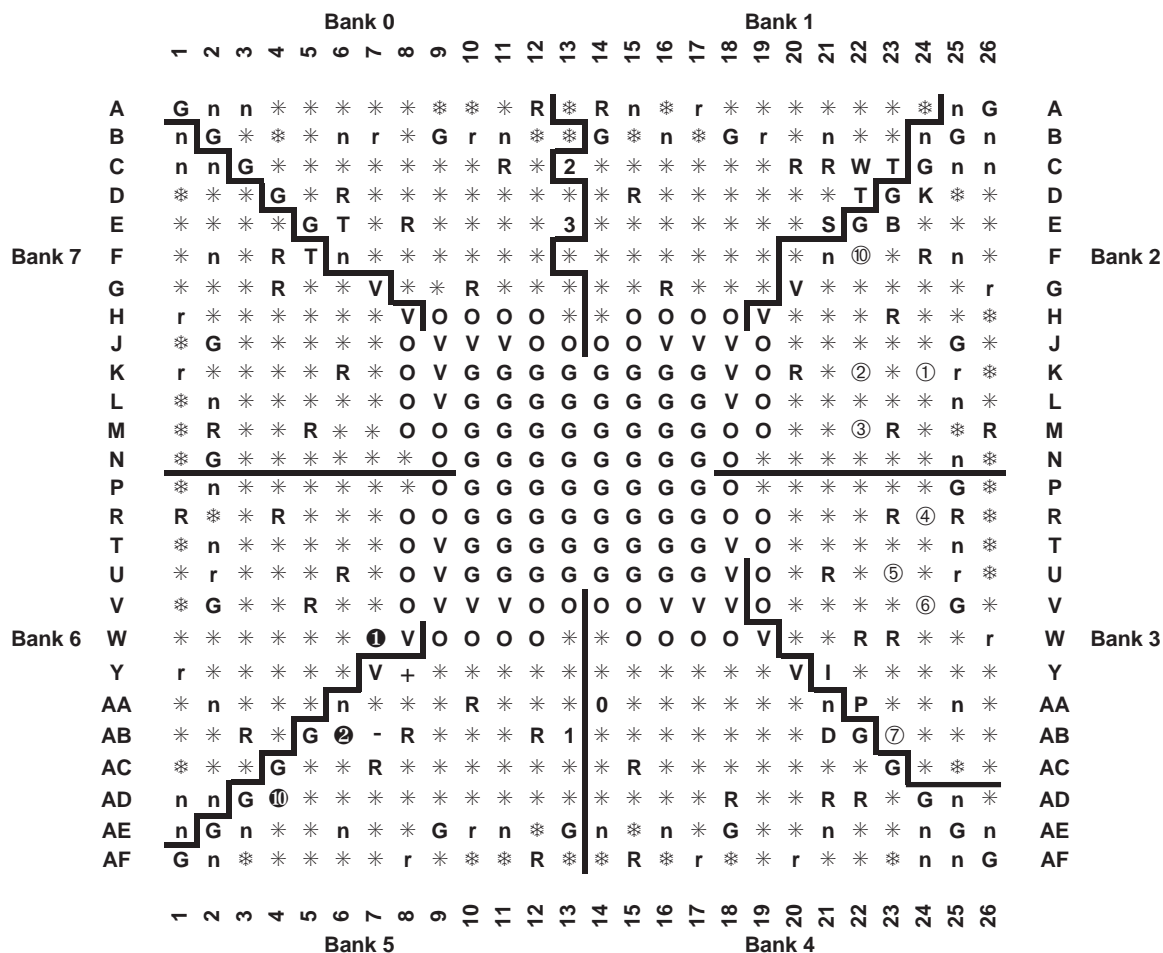


Figure 10: FG676 Pin Function Diagram

Notes:

Packages FG456 and FG676 are layout compatible.