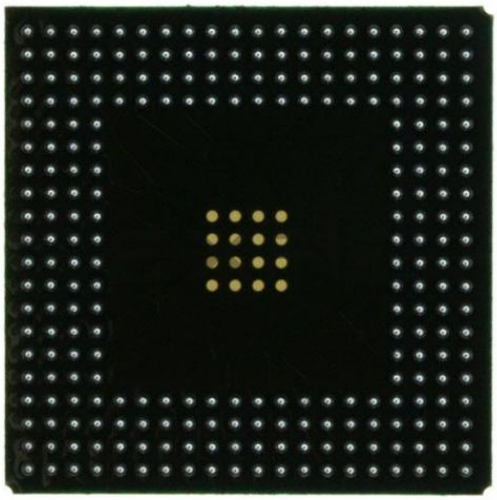


Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	1176
Number of Logic Elements/Cells	5292
Total RAM Bits	57344
Number of I/O	180
Number of Gates	236666
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-BBGA
Supplier Device Package	256-PBGA (27x27)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcv200-6bg256c

Revision History

Date	Version	Revision
11/98	1.0	Initial Xilinx release.
01/99-02/99	1.2-1.3	Both versions updated package drawings and specs.
05/99	1.4	Addition of package drawings and specifications.
05/99	1.5	Replaced FG 676 & FG680 package drawings.
07/99	1.6	Changed Boundary Scan Information and changed Figure 11, Boundary Scan Bit Sequence. Updated IOB Input & Output delays. Added Capacitance info for different I/O Standards. Added 5 V tolerant information. Added DLL Parameters and waveforms and new Pin-to-pin Input and Output Parameter tables for Global Clock Input to Output and Setup and Hold. Changed Configuration Information including Figures 12, 14, 17 & 19. Added device-dependent listings for quiescent currents ICCINTQ and ICCOQ. Updated IOB Input and Output Delays based on default standard of LVTTL, 12 mA, Fast Slew Rate. Added IOB Input Switching Characteristics Standard Adjustments.
09/99	1.7	Speed grade update to preliminary status, Power-on specification and Clock-to-Out Minimums additions, “0” hold time listing explanation, quiescent current listing update, and Figure 6 ADDRA input label correction. Added T_{IJITCC} parameter, changed T_{OJIT} to T_{OPHASE}.
01/00	1.8	Update to speed.txt file 1.96. Corrections for CRs 111036,111137, 112697, 115479, 117153, 117154, and 117612. Modified notes for Recommended Operating Conditions (voltage and temperature). Changed Bank information for V_{CCO} in CS144 package on p.43.
01/00	1.9	Updated DLL Jitter Parameter table and waveforms, added Delay Measurement Methodology table for different I/O standards, changed buffered Hex line info and Input/Output Timing measurement notes.
03/00	2.0	New TBCKO values; corrected FG680 package connection drawing; new note about status of CCLK pin after configuration.
05/00	2.1	Modified “Pins not listed...” statement. Speed grade update to Final status.
05/00	2.2	Modified Table 18.
09/00	2.3	• Added XCV400 values to table under **Minimum Clock-to-Out for Virtex Devices**. • Corrected Units column in table under **IOB Input Switching Characteristics**. • Added values to table under **CLB SelectRAM Switching Characteristics**.
10/00	2.4	• Corrected Pinout information for devices in the BG256, BG432, and BG560 packages in Table 18. • Corrected **BG256 Pin Function Diagram**.
04/01	2.5	• Revised minimums for **Global Clock Set-Up and Hold for LVTTL Standard, with DLL**. • Converted file to modularized format. See **Virtex Data Sheet** section.
03/13	4.0	The products listed in this data sheet are obsolete. See **XCN10016** for further information.

Virtex Data Sheet

The Virtex Data Sheet contains the following modules:

- DS003-1, Virtex 2.5V FPGAs: Introduction and Ordering Information (Module 1)
- DS003-2, Virtex 2.5V FPGAs: Functional Description (Module 2)
- DS003-3, Virtex 2.5V FPGAs: DC and Switching Characteristics (Module 3)
- DS003-4, Virtex 2.5V FPGAs: Pinout Tables (Module 4)

Virtex™ 2.5 V Field Programmable Gate Arrays

DS003-2 (v4.0) March 1, 2013

Product Specification

Architectural Description

Virtex Array

The Virtex user-programmable gate array, shown in Figure 1, comprises two major configurable elements: configurable logic blocks (CLBs) and input/output blocks (IOBs).

- CLBs provide the functional elements for constructing logic
- IOBs provide the interface between the package pins and the CLBs

CLBs interconnect through a general routing matrix (GRM). The GRM comprises an array of routing switches located at the intersections of horizontal and vertical routing channels. Each CLB nests into a VersaBlock™ that also provides local routing resources to connect the CLB to the GRM.

The VersaRing™ I/O interface provides additional routing resources around the periphery of the device. This routing improves I/O routability and facilitates pin locking.

The Virtex architecture also includes the following circuits that connect to the GRM.

- Dedicated block memories of 4096 bits each
- Clock DLLs for clock-distribution delay compensation and clock domain control
- 3-State buffers (BUFTs) associated with each CLB that drive dedicated segmentable horizontal routing resources

Values stored in static memory cells control the configurable logic elements and interconnect resources. These values load into the memory cells on power-up, and can reload if necessary to change the function of the device.

Input/Output Block

The Virtex IOB, Figure 2, features SelectIO™ inputs and outputs that support a wide variety of I/O signalling standards, see Table 1.

The three IOB storage elements function either as edge-triggered D-type flip-flops or as level sensitive latches. Each IOB has a clock signal (CLK) shared by the three flip-flops and independent clock enable signals for each flip-flop.

In addition to the CLK and CE control signals, the three flip-flops share a Set/Reset (SR). For each flip-flop, this signal can be independently configured as a synchronous Set, a synchronous Reset, an asynchronous Preset, or an asynchronous Clear.

The output buffer and all of the IOB control signals have independent polarity controls.

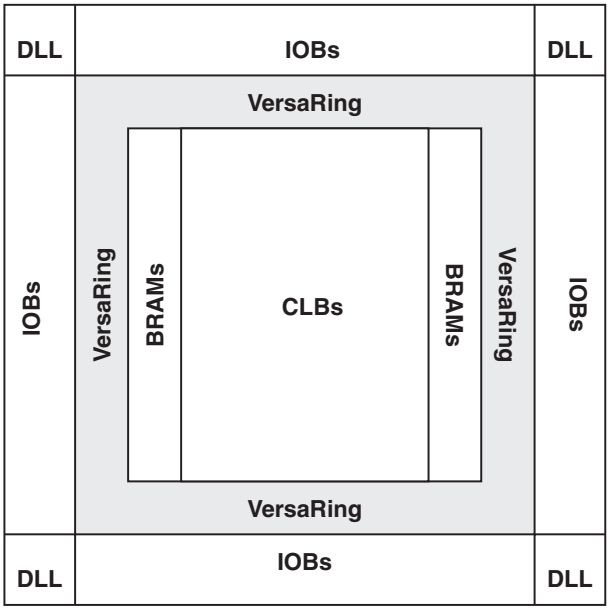


Figure 1: **Virtex Architecture Overview**

All pads are protected against damage from electrostatic discharge (ESD) and from over-voltage transients. Two forms of over-voltage protection are provided, one that permits 5 V compliance, and one that does not. For 5 V compliance, a Zener-like structure connected to ground turns on when the output rises to approximately 6.5 V. When PCI 3.3 V compliance is required, a conventional clamp diode is connected to the output supply voltage, V_{CCO}.

Optional pull-up and pull-down resistors and an optional weak-keeper circuit are attached to each pad. Prior to configuration, all pins not involved in configuration are forced into their high-impedance state. The pull-down resistors and the weak-keeper circuits are inactive, but inputs can optionally be pulled up.

The activation of pull-up resistors prior to configuration is controlled on a global basis by the configuration mode pins. If the pull-up resistors are not activated, all the pins will float. Consequently, external pull-up or pull-down resistors must be provided on pins required to be at a well-defined logic level prior to configuration.

All Virtex IOBs support IEEE 1149.1-compatible boundary scan testing.

Each block SelectRAM cell, as illustrated in Figure 6, is a fully synchronous dual-ported 4096-bit RAM with independent control signals for each port. The data widths of the two ports can be configured independently, providing built-in bus-width conversion.

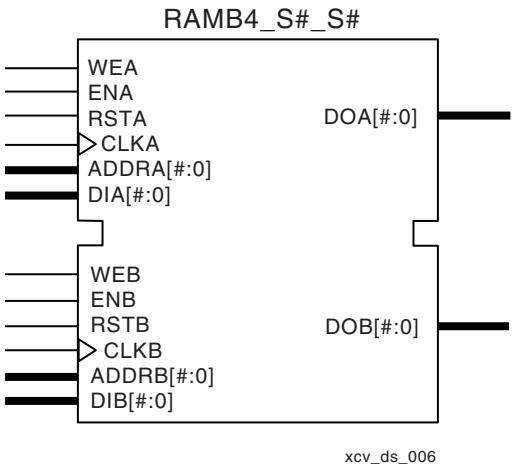


Figure 6: **Dual-Port Block SelectRAM**

Table 4 shows the depth and width aspect ratios for the block SelectRAM.

Table 4: **Block SelectRAM Port Aspect Ratios**

Width	Depth	ADDR Bus	Data Bus
1	4096	ADDR<11:0>	DATA<0>
2	2048	ADDR<10:0>	DATA<1:0>
4	1024	ADDR<9:0>	DATA<3:0>
8	512	ADDR<8:0>	DATA<7:0>
16	256	ADDR<7:0>	DATA<15:0>

The Virtex block SelectRAM also includes dedicated routing to provide an efficient interface with both CLBs and other block SelectRAMs. Refer to XAPP130 for block SelectRAM timing waveforms.

Programmable Routing Matrix

It is the longest delay path that limits the speed of any worst-case design. Consequently, the Virtex routing architecture and its place-and-route software were defined in a single optimization process. This joint optimization minimizes long-path delays, and consequently, yields the best system performance.

The joint optimization also reduces design compilation times because the architecture is software-friendly. Design cycles are correspondingly reduced due to shorter design iteration times.

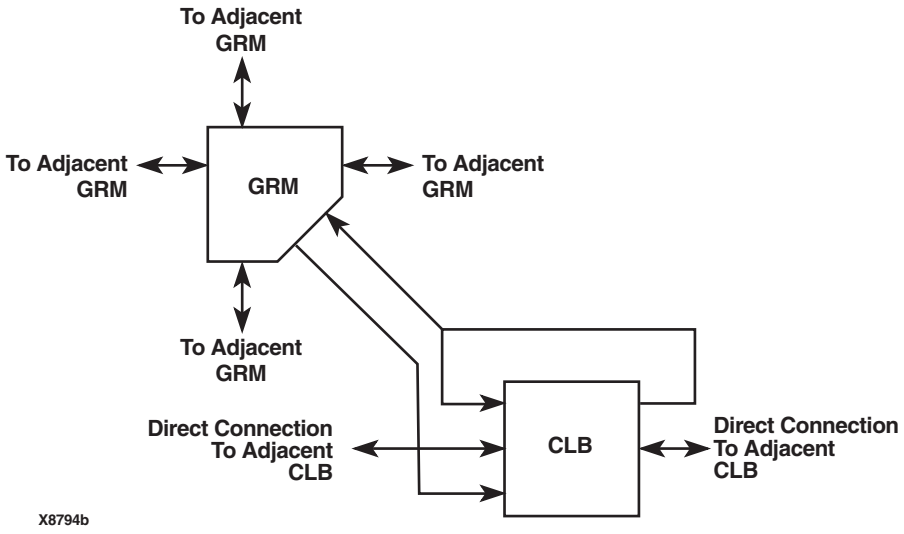


Figure 7: **Virtex Local Routing**

Local Routing

The VersaBlock provides local routing resources, as shown in Figure 7, providing the following three types of connections.

- Interconnections among the LUTs, flip-flops, and GRM
- Internal CLB feedback paths that provide high-speed connections to LUTs within the same CLB, chaining them together with minimal routing delay
- Direct paths that provide high-speed connections between horizontally adjacent CLBs, eliminating the delay of the GRM.

ers with a common user interface regardless of their choice of entry and verification tools. The XDM software simplifies the selection of implementation options with pull-down menus and on-line help.

Application programs ranging from schematic capture to Placement and Routing (PAR) can be accessed through the XDM software. The program command sequence is generated prior to execution, and stored for documentation.

Several advanced software features facilitate Virtex design. RPMs, for example, are schematic-based macros with relative location constraints to guide their placement. They help ensure optimal implementation of common functions.

For HDL design entry, the Xilinx FPGA Foundation development system provides interfaces to the following synthesis design environments.

- Synopsys (FPGA Compiler, FPGA Express)
- Exemplar (Spectrum)
- Synplicity (Synplify)

For schematic design entry, the Xilinx FPGA Foundation and alliance development system provides interfaces to the following schematic-capture design environments.

- Mentor Graphics V8 (Design Architect, QuickSim II)
- Viewlogic Systems (Viewdraw)

Third-party vendors support many other environments.

A standard interface-file specification, Electronic Design Interchange Format (EDIF), simplifies file transfers into and out of the development system.

Virtex FPGAs supported by a unified library of standard functions. This library contains over 400 primitives and macros, ranging from 2-input AND gates to 16-bit accumulators, and includes arithmetic functions, comparators, counters, data registers, decoders, encoders, I/O functions, latches, Boolean functions, multiplexers, shift registers, and barrel shifters.

The “soft macro” portion of the library contains detailed descriptions of common logic functions, but does not contain any partitioning or placement information. The performance of these macros depends, therefore, on the partitioning and placement obtained during implementation.

RPMs, on the other hand, do contain predetermined partitioning and placement information that permits optimal implementation of these functions. Users can create their own library of soft macros or RPMs based on the macros and primitives in the standard library.

The design environment supports hierarchical design entry, with high-level schematics that comprise major functional blocks, while lower-level schematics define the logic in these blocks. These hierarchical design elements are automatically combined by the implementation tools. Different design entry tools can be combined within a hierarchical design, thus allowing the most convenient entry method to be used for each portion of the design.

Design Implementation

The place-and-route tools (PAR) automatically provide the implementation flow described in this section. The partitioner takes the EDIF net list for the design and maps the logic into the architectural resources of the FPGA (CLBs and IOBs, for example). The placer then determines the best locations for these blocks based on their interconnections and the desired performance. Finally, the router interconnects the blocks.

The PAR algorithms support fully automatic implementation of most designs. For demanding applications, however, the user can exercise various degrees of control over the process. User partitioning, placement, and routing information is optionally specified during the design-entry process. The implementation of highly structured designs can benefit greatly from basic floor planning.

The implementation software incorporates Timing Wizard® timing-driven placement and routing. Designers specify timing requirements along entire paths during design entry. The timing path analysis routines in PAR then recognize these user-specified requirements and accommodate them.

Timing requirements are entered on a schematic in a form directly relating to the system requirements, such as the targeted clock frequency, or the maximum allowable delay between two registers. In this way, the overall performance of the system along entire signal paths is automatically tailored to user-generated specifications. Specific timing information for individual nets is unnecessary.

Design Verification

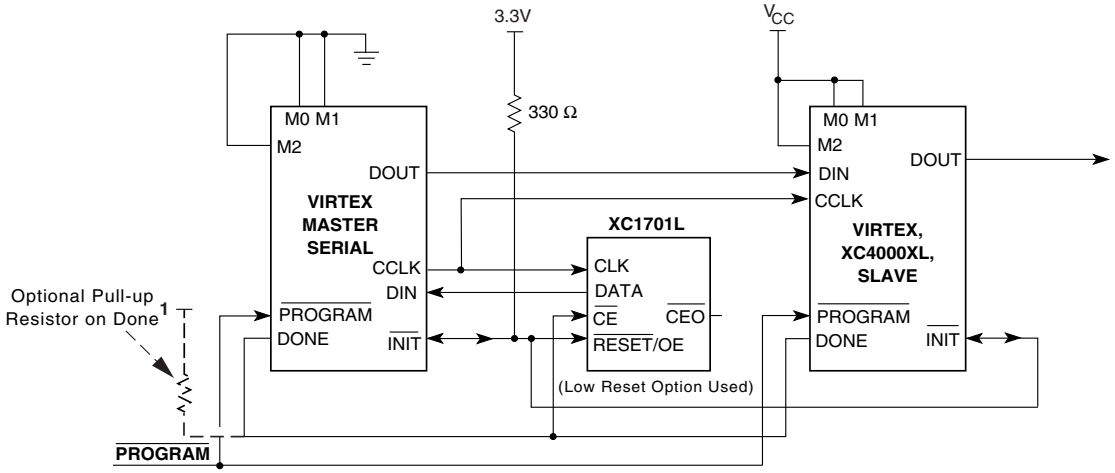
In addition to conventional software simulation, FPGA users can use in-circuit debugging techniques. Because Xilinx devices are infinitely reprogrammable, designs can be verified in real time without the need for extensive sets of software simulation vectors.

The development system supports both software simulation and in-circuit debugging techniques. For simulation, the system extracts the post-layout timing information from the design database, and back-annotates this information into the net list for use by the simulator. Alternatively, the user can verify timing-critical portions of the design using the TRACE® static timing analyzer.

For in-circuit debugging, the development system includes a download and readback cable. This cable connects the FPGA in the target system to a PC or workstation. After downloading the design into the FPGA, the designer can single-step the logic, readback the contents of the flip-flops, and so observe the internal logic state. Simple modifications can be downloaded into the system in a matter of minutes.

Table 8: **Master/Slave Serial Mode Programming Switching**

	Description	Figure References	Symbol	Values	Units
CCLK	DIN setup/hold, slave mode	1/2	T_{DCC}/T_{CCD}	5.0 / 0	ns, min
	DIN setup/hold, master mode	1/2	T_{DSCK}/T_{CKDS}	5.0 / 0	ns, min
	DOUT	3	T_{CCO}	12.0	ns, max
	High time	4	T_{CCH}	5.0	ns, min
	Low time	5	T_{CCL}	5.0	ns, min
	Maximum Frequency		F_{CC}	66	MHz, max
	Frequency Tolerance, master mode with respect to nominal			+45% –30%	



Note 1: If none of the Virtex FPGAs have been selected to drive DONE, an external pull-up resistor of 330 Ω should be added to the common DONE line. (For Spartan-XL devices, add a 4.7K Ω pull-up resistor.) This pull-up is not needed if the DriveDONE attribute is set. If used, DriveDONE should be selected only for the last device in the configuration chain.

xcv_12_050103

Figure 12: **Master/Slave Serial Mode Circuit Diagram**

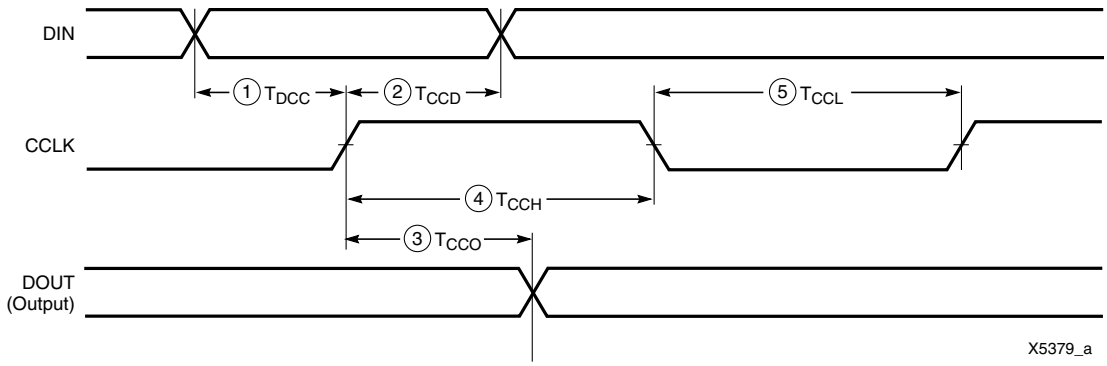


Figure 13: **Slave-Serial Mode Programming Switching Characteristics**

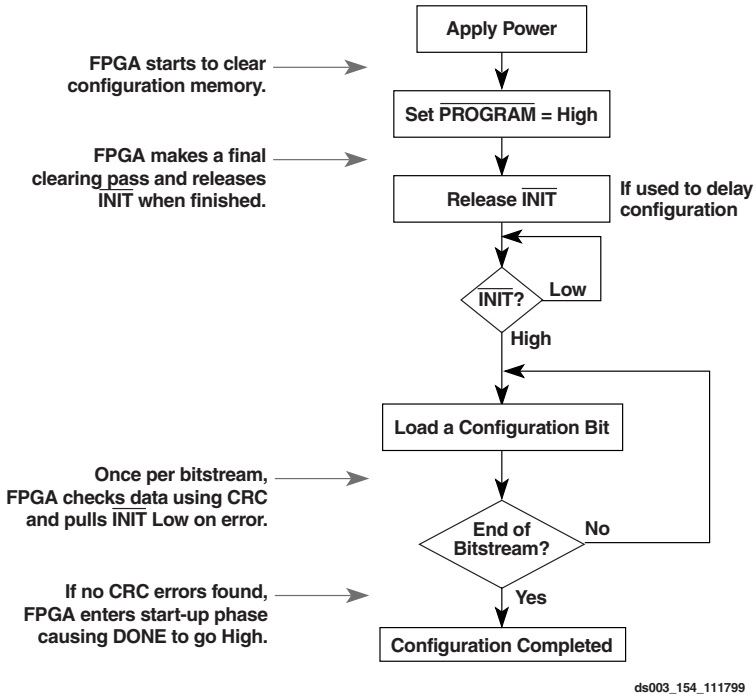


Figure 15: **Serial Configuration Flowchart**

After configuration, the pins of the SelectMAP port can be used as additional user I/O. Alternatively, the port can be retained to permit high-speed 8-bit readback.

Retention of the SelectMAP port is selectable on a design-by-design basis when the bitstream is generated. If retention is selected, PROHIBIT constraints are required to prevent the SelectMAP-port pins from being used as user I/O.

Multiple Virtex FPGAs can be configured using the Select-MAP mode, and be made to start-up simultaneously. To configure multiple devices in this way, wire the individual CCLK, Data, $\overline{\text{WRITE}}$, and BUSY pins of all the devices in parallel. The individual devices are loaded separately by asserting the $\overline{\text{CS}}$ pin of each device in turn and writing the appropriate data. see Table 9 for SelectMAP Write Timing Characteristics.

Table 9: **SelectMAP Write Timing Characteristics**

	Description		Symbol		Units
CCLK	D_{0-7} Setup/Hold	1/2	T_{SMDCC}/T_{SMCCD}	5.0 / 1.7	ns, min
	$\overline{\text{CS}}$ Setup/Hold	3/4	T_{SMCSCC}/T_{SMCCCS}	7.0 / 1.7	ns, min
	$\overline{\text{WRITE}}$ Setup/Hold	5/6	T_{SMCCW}/T_{SMWCC}	7.0 / 1.7	ns, min
	BUSY Propagation Delay	7	T_{SMCKBY}	12.0	ns, max
	Maximum Frequency		F_{CC}	66	MHz, max
	Maximum Frequency with no handshake		F_{CCNH}	50	MHz, max

Write

Write operations send packets of configuration data into the FPGA. The sequence of operations for a multi-cycle write operation is shown below. Note that a configuration packet can be split into many such sequences. The packet does not have to complete within one assertion of $\overline{\text{CS}}$, illustrated in Figure 16.

1. Assert $\overline{\text{WRITE}}$ and $\overline{\text{CS}}$ Low. Note that when $\overline{\text{CS}}$ is asserted on successive CCLKs, $\overline{\text{WRITE}}$ must remain either asserted or de-asserted. Otherwise an abort will be initiated, as described below.
2. Drive data onto D[7:0]. Note that to avoid contention, the data source should not be enabled while $\overline{\text{CS}}$ is Low and $\overline{\text{WRITE}}$ is High. Similarly, while $\overline{\text{WRITE}}$ is High, no more that one $\overline{\text{CS}}$ should be asserted.

3. At the rising edge of CCLK: If BUSY is Low, the data is accepted on this clock. If BUSY is High (from a previous write), the data is not accepted. Acceptance will instead occur on the first clock after BUSY goes Low, and the data must be held until this has happened.
4. Repeat steps 2 and 3 until all the data has been sent.
5. De-assert $\overline{CS}$ and $\overline{WRITE}$.

A flowchart for the write operation appears in Figure 17. Note that if CCLK is slower than f_{CCNH}, the FPGA never asserts BUSY. In this case, the above handshake is unnecessary, and data can simply be entered into the FPGA every CCLK cycle.

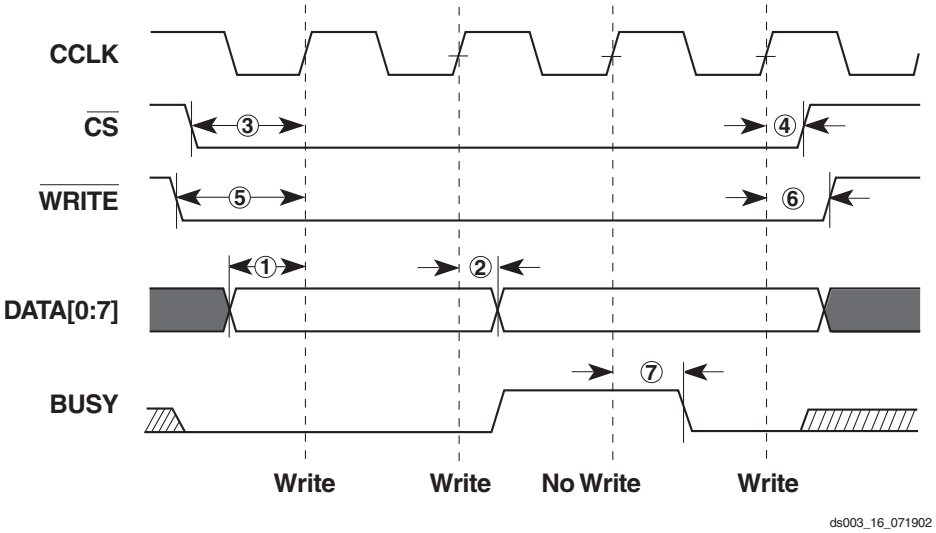


Figure 16: **Write Operations**

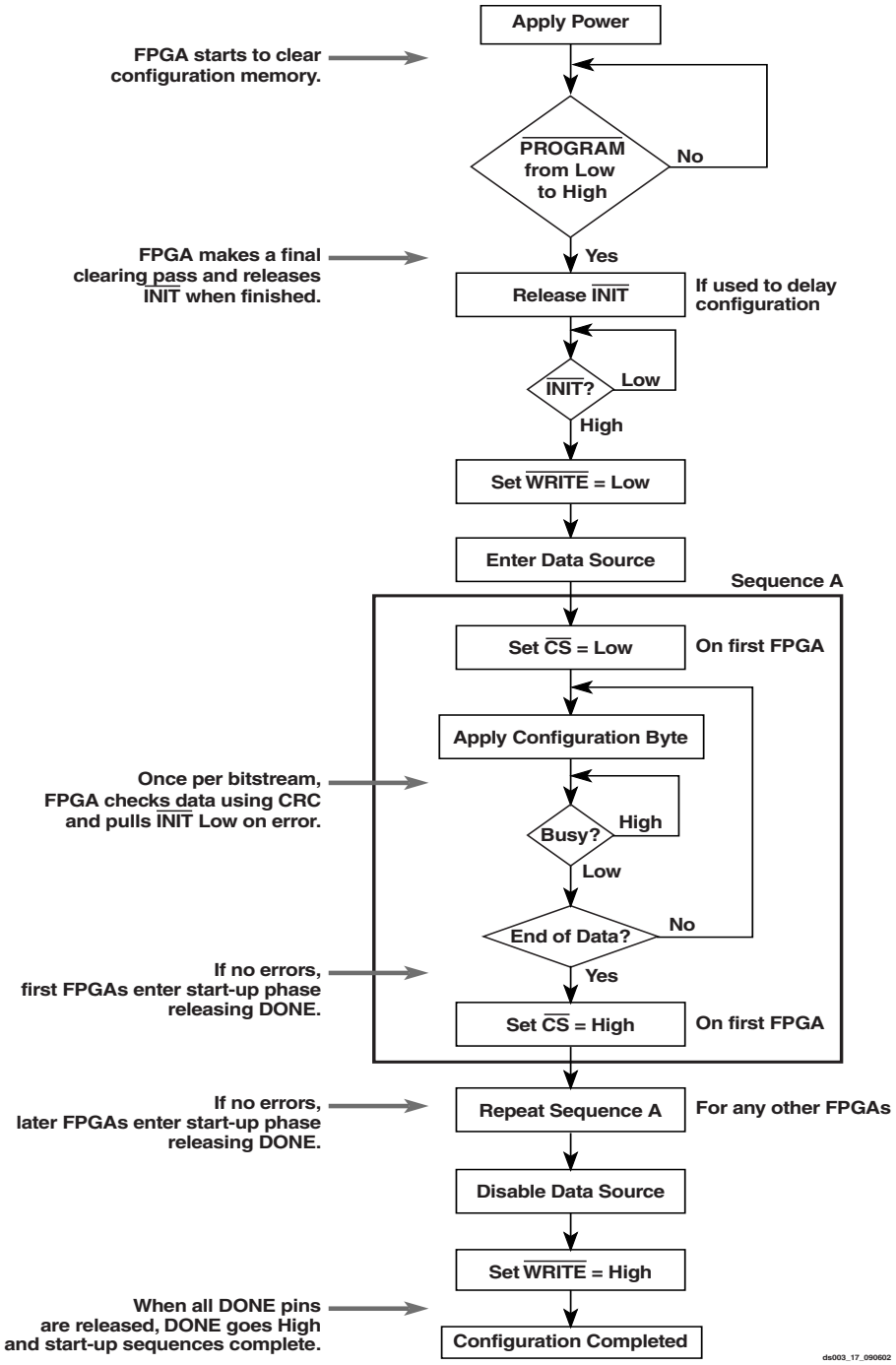


Figure 17: **SelectMAP Flowchart for Write Operation**

Abort

During a given assertion of $\overline{CS}$, the user cannot switch from a write to a read, or vice-versa. This action causes the current packet command to be aborted. The device will remain BUSY until the aborted operation has completed. Following an abort, data is assumed to be unaligned to word boundaries, and the FPGA requires a new synchronization word prior to accepting any new packets.

To initiate an abort during a write operation, de-assert $\overline{WRITE}$. At the rising edge of CCLK, an abort is initiated, as shown in Figure 18.

Description	Symbol	Speed Grade				Units
		Min	-6	-5	-4	
Clock CLK to Pad delay with OBUFT enabled (non-3-state)	T_{IOCKP}	1.0	2.9	3.2	3.5	ns, max
Clock CLK to Pad high-impedance (synchronous)(1)	T_{IOCKHZ}	1.1	2.3	2.5	2.9	ns, max
Clock CLK to valid data on Pad delay, plus enable delay for OBUFT	T_{IOCKON}	1.5	3.4	3.7	4.1	ns, max
Setup and Hold Times before/after Clock CLK(2)		**Setup Time / Hold Time**				
O input	T_{IOOCK}/T_{IOCKO}	0.51 / 0	1.1 / 0	1.2 / 0	1.3 / 0	ns, min
OCE input	$T_{IOOCECK}/T_{IOCKOCE}$	0.37 / 0	0.8 / 0	0.9 / 0	1.0 / 0	ns, min
SR input (OFF)	$T_{IOSRCKO}/T_{IOCKOSR}$	0.52 / 0	1.1 / 0	1.2 / 0	1.4 / 0	ns, min
3-State Setup Times, T input	T_{IOTCK}/T_{IOCKT}	0.34 / 0	0.7 / 0	0.8 / 0	0.9 / 0	ns, min
3-State Setup Times, TCE input	$T_{IOTCECK}/T_{IOCKTCE}$	0.41 / 0	0.9 / 0	0.9 / 0	1.1 / 0	ns, min
3-State Setup Times, SR input (TFF)	$T_{IOSRCKT}/T_{IOCKTSR}$	0.49 / 0	1.0 / 0	1.1 / 0	1.3 / 0	ns, min
Set/Reset Delays						
SR input to Pad (asynchronous)	T_{IOSRP}	1.6	3.8	4.1	4.6	ns, max
SR input to Pad high-impedance (asynchronous)(1)	T_{IOSRHZ}	1.6	3.1	3.4	3.9	ns, max
SR input to valid data on Pad (asynchronous)	T_{IOSRON}	2.0	4.2	4.6	5.1	ns, max
GSR to Pad	T_{IOGSRQ}	4.9	9.7	10.9	12.5	ns, max

Notes:
1. 3-state turn-off delays should not be adjusted.
2. A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.

Calculation of T_{ioop} as a Function of Capacitance

T_{ioop} is the propagation delay from the O Input of the IOB to the pad. The values for T_{ioop} were based on the standard capacitive load (Csl) for each I/O standard as listed in Table 2.

Table 2: **Constants for Calculating T_{ioop}**

Standard	Csl (pF)	fl (ns/pF)
LVTTL Fast Slew Rate, 2mA drive	35	0.41
LVTTL Fast Slew Rate, 4mA drive	35	0.20
LVTTL Fast Slew Rate, 6mA drive	35	0.13
LVTTL Fast Slew Rate, 8mA drive	35	0.079
LVTTL Fast Slew Rate, 12mA drive	35	0.044
LVTTL Fast Slew Rate, 16mA drive	35	0.043
LVTTL Fast Slew Rate, 24mA drive	35	0.033
LVTTL Slow Slew Rate, 2mA drive	35	0.41
LVTTL Slow Slew Rate, 4mA drive	35	0.20
LVTTL Slow Slew Rate, 6mA drive	35	0.100
LVTTL Slow Slew Rate, 8mA drive	35	0.086
LVTTL Slow Slew Rate, 12mA drive	35	0.058
LVTTL Slow Slew Rate, 16mA drive	35	0.050
LVTTL Slow Slew Rate, 24mA drive	35	0.048
LVCMOS2	35	0.041
PCI 33MHz 5V	50	0.050
PCI 33MHZ 3.3 V	10	0.050
PCI 66 MHz 3.3 V	10	0.033
GTL	0	0.014
GTL+	0	0.017
HSTL Class I	20	0.022
HSTL Class III	20	0.016
HSTL Class IV	20	0.014
SSTL2 Class I	30	0.028
SSTL2 Class II	30	0.016
SSTL3 Class I	30	0.029
SSTL3 Class II	30	0.016
CTT	20	0.035
AGP	10	0.037

Notes:
1. I/O parameter measurements are made with the capacitance values shown above. See Application Note XAPP133 on www.xilinx.com for appropriate terminations.
2. I/O standard measurements are reflected in the IBIS model information except where the IBIS format precludes it.

For other capacitive loads, use the formulas below to calculate the corresponding T_{ioop}.

$$T_{ioop} = T_{ioop} + T_{opadjust} + (C_{load} - C_{sl}) * fl$$

Where:

$T_{opadjust}$ is reported above in the Output Delay Adjustment section.

C_{load} is the capacitive load for the design.

Table 3: **Delay Measurement Methodology**

Standard	V_L (1)	V_H (1)	Meas. Point	V_{REF} Typ (2)
LVTTL	0	3	1.4	-
LVCMOS2	0	2.5	1.125	-
PCI33_5	Per PCI Spec			-
PCI33_3	Per PCI Spec			-
PCI66_3	Per PCI Spec			-
GTL	V_{REF} –0.2	V_{REF} +0.2	V_{REF}	0.80
GTL+	V_{REF} –0.2	V_{REF} +0.2	V_{REF}	1.0
HSTL Class I	V_{REF} –0.5	V_{REF} +0.5	V_{REF}	0.75
HSTL Class III	V_{REF} –0.5	V_{REF} +0.5	V_{REF}	0.90
HSTL Class IV	V_{REF} –0.5	V_{REF} +0.5	V_{REF}	0.90
SSTL3 I & II	V_{REF} –1.0	V_{REF} +1.0	V_{REF}	1.5
SSTL2 I & II	V_{REF} –0.75	V_{REF} +0.75	V_{REF}	1.25
CTT	V_{REF} –0.2	V_{REF} +0.2	V_{REF}	1.5
AGP	V_{REF} – (0.2xV_{CCO})	V_{REF} + (0.2xV_{CCO})	V_{REF}	Per AGP Spec

Notes:
1. Input waveform switches between V_L and V_H.
2. Measurements are made at VREF (Typ), Maximum, and Minimum. Worst-case values are reported.
3. I/O parameter measurements are made with the capacitance values shown in Table 2. See Application Note XAPP133 on www.xilinx.com for appropriate terminations.
4. I/O standard measurements are reflected in the IBIS model information except where the IBIS format precludes it.

Clock Distribution Guidelines

Description	Device	Symbol	Speed Grade			Units
			-6	-5	-4	
Global Clock Skew(1)						
Global Clock Skew between IOB Flip-flops	XCV50	$T_{GSKEWIOB}$	0.10	0.12	0.14	ns, max
	XCV100		0.12	0.13	0.15	ns, max
	XCV150		0.12	0.13	0.15	ns, max
	XCV200		0.13	0.14	0.16	ns, max
	XCV300		0.14	0.16	0.18	ns, max
	XCV400		0.13	0.13	0.14	ns, max
	XCV600		0.14	0.15	0.17	ns, max
	XCV800		0.16	0.17	0.20	ns, max
	XCV1000		0.20	0.23	0.25	ns, max

Notes:
1. These clock-skew delays are provided for guidance only. They reflect the delays encountered in a typical design under worst-case conditions. Precise values for a particular design are provided by the timing analyzer.

Clock Distribution Switching Characteristics

Description	Symbol	Speed Grade				Units
		Min	-6	-5	-4	
GCLK IOB and Buffer						
Global Clock PAD to output.	T_{GPIO}	0.33	0.7	0.8	0.9	ns, max
Global Clock Buffer I input to O output	T_{GIO}	0.34	0.7	0.8	0.9	ns, max

CLB SelectRAM Switching Characteristics

Description	Symbol	Speed Grade				Units
		Min	-6	-5	-4	
Sequential Delays						
Clock CLK to X/Y outputs (WE active) 16 x 1 mode	$T_{SHCKO16}$	1.2	2.3	2.6	3.0	ns, max
Clock CLK to X/Y outputs (WE active) 32 x 1 mode	$T_{SHCKO32}$	1.2	2.7	3.1	3.5	ns, max
Shift-Register Mode						
Clock CLK to X/Y outputs	T_{REG}	1.2	3.7	4.1	4.7	ns, max
Setup and Hold Times before/after Clock CLK(1)	Setup Time / Hold Time					
F/G address inputs	T_{AS}/T_{AH}	0.25 / 0	0.5 / 0	0.6 / 0	0.7 / 0	ns, min
BX/BY data inputs (DIN)	T_{DS}/T_{DH}	0.34 / 0	0.7 / 0	0.8 / 0	0.9 / 0	ns, min
CE input (WE)	T_{WS}/T_{WH}	0.38 / 0	0.8 / 0	0.9 / 0	1.0 / 0	ns, min
Shift-Register Mode						
BX/BY data inputs (DIN)	T_{SHDICK}	0.34	0.7	0.8	0.9	ns, min
CE input (WS)	T_{SHCECK}	0.38	0.8	0.9	1.0	ns, min
Clock CLK						
Minimum Pulse Width, High	T_{WPH}	1.2	2.4	2.7	3.1	ns, min
Minimum Pulse Width, Low	T_{WPL}	1.2	2.4	2.7	3.1	ns, min
Minimum clock period to meet address write cycle time	T_{WC}	2.4	4.8	5.4	6.2	ns, min
Shift-Register Mode						
Minimum Pulse Width, High	T_{SRPH}	1.2	2.4	2.7	3.1	ns, min
Minimum Pulse Width, Low	T_{SRPL}	1.2	2.4	2.7	3.1	ns, min

Notes:
1. A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.

Virtex Pin-to-Pin Input Parameter Guidelines

All devices are 100% functionally tested. Listed below are representative values for typical pin locations and normal clock loading. Values are expressed in nanoseconds unless otherwise noted

Global Clock Set-Up and Hold for LVTTL Standard, *with* DLL

Description	Symbol	Device	Speed Grade				Units
			Min	-6	-5	-4	
Input Setup and Hold Time Relative to Global Clock Input Signal for LVTTL Standard. For data input with different standards, adjust the setup time delay by the values shown in Input Delay Adjustments.							
No Delay Global Clock and IFF, with DLL	T_{PSDLL}/T_{PHDLL}	XCV50	0.40 / –0.4	1.7 /–0.4	1.8 /–0.4	2.1 /–0.4	ns, min
		XCV100	0.40 /–0.4	1.7 /–0.4	1.9 /–0.4	2.1 /–0.4	ns, min
		XCV150	0.40 /–0.4	1.7 /–0.4	1.9 /–0.4	2.1 /–0.4	ns, min
		XCV200	0.40 /–0.4	1.7 /–0.4	1.9 /–0.4	2.1 /–0.4	ns, min
		XCV300	0.40 /–0.4	1.7 /–0.4	1.9 /–0.4	2.1 /–0.4	ns, min
		XCV400	0.40 /–0.4	1.7 /–0.4	1.9 /–0.4	2.1 /–0.4	ns, min
		XCV600	0.40 /–0.4	1.7 /–0.4	1.9 /–0.4	2.1 /–0.4	ns, min
		XCV800	0.40 /–0.4	1.7 /–0.4	1.9 /–0.4	2.1 /–0.4	ns, min
		XCV1000	0.40 /–0.4	1.7 /–0.4	1.9 /–0.4	2.1 /–0.4	ns, min

IFF = Input Flip-Flop or Latch

Notes:

1. Set-up time is measured relative to the Global Clock input signal with the fastest route and the lightest load. Hold time is measured relative to the Global Clock input signal with the slowest route and heaviest load.
2. DLL output jitter is already included in the timing calculation.
3. A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.

Global Clock Set-Up and Hold for LVTTL Standard, *without* DLL

Description	Symbol	Device	Speed Grade				Units
			Min	-6	-5	-4	
Input Setup and Hold Time Relative to Global Clock Input Signal for LVTTL Standard.(2) For data input with different standards, adjust the setup time delay by the values shown in Input Delay Adjustments.							
Full Delay Global Clock and IFF, without DLL	T_{PSFD}/T_{PHFD}	XCV50	0.6 / 0	2.3 / 0	2.6 / 0	2.9 / 0	ns, min
		XCV100	0.6 / 0	2.3 / 0	2.6 / 0	3.0 / 0	ns, min
		XCV150	0.6 / 0	2.4 / 0	2.7 / 0	3.1 / 0	ns, min
		XCV200	0.7 / 0	2.5 / 0	2.8 / 0	3.2 / 0	ns, min
		XCV300	0.7 / 0	2.5 / 0	2.8 / 0	3.2 / 0	ns, min
		XCV400	0.7 / 0	2.6 / 0	2.9 / 0	3.3 / 0	ns, min
		XCV600	0.7 / 0	2.6 / 0	2.9 / 0	3.3 / 0	ns, min
		XCV800	0.7 / 0	2.7 / 0	3.1 / 0	3.5 / 0	ns, min
		XCV1000	0.7 / 0	2.8 / 0	3.1 / 0	3.6 / 0	ns, min

IFF = Input Flip-Flop or Latch

Notes: Notes:

1. Set-up time is measured relative to the Global Clock input signal with the fastest route and the lightest load. Hold time is measured relative to the Global Clock input signal with the slowest route and heaviest load.
2. A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.

DLL Timing Parameters

All devices are 100 percent functionally tested. Because of the difficulty in directly measuring many internal timing parameters, those parameters are derived from benchmark timing patterns. The following guidelines reflect worst-case values across the recommended operating conditions.

Description	Symbol	Speed Grade						Units
		-6		-5		-4		
		Min	Max	Min	Max	Min	Max	
Input Clock Frequency (CLKDLLHF)	FCLKINHF	60	200	60	180	60	180	MHz
Input Clock Frequency (CLKDLL)	FCLKINLF	25	100	25	90	25	90	MHz
Input Clock Pulse Width (CLKDLLHF)	$T_{DLLPWHF}$	2.0	-	2.4	-	2.4	-	ns
Input Clock Pulse Width (CLKDLL)	$T_{DLLPWLF}$	2.5	-	3.0		3.0	-	ns

Notes:
1. All specifications correspond to Commercial Operating Temperatures (0°C to + 85°C).

DLL Clock Tolerance, Jitter, and Phase Information

All DLL output jitter and phase specifications determined through statistical measurement at the package pins using a clock mirror configuration and matched drivers.

Description	Symbol	F_{CLKIN}	CLKDLLHF		CLKDLL		Units
			Min	Max	Min	Max	
Input Clock Period Tolerance	T_{IPTOL}		-	1.0	-	1.0	ns
Input Clock Jitter Tolerance (Cycle to Cycle)	T_{IJITCC}		-	± 150	-	± 300	ps
Time Required for DLL to Acquire Lock	T_{LOCK}	> 60 MHz	-	20	-	20	µs
		50 - 60 MHz	-	-	-	25	µs
		40 - 50 MHz	-	-	-	50	µs
		30 - 40 MHz	-	-	-	90	µs
		25 - 30 MHz	-	-	-	120	µs
Output Jitter (cycle-to-cycle) for any DLL Clock Output(1)	T_{OJITCC}			± 60		± 60	ps
Phase Offset between CLKIN and CLKO(2)	T_{PHIO}			± 100		± 100	ps
Phase Offset between Clock Outputs on the DLL(3)	T_{PHOO}			± 140		± 140	ps
Maximum Phase Difference between CLKIN and CLKO(4)	T_{PHIOM}			± 160		± 160	ps
Maximum Phase Difference between Clock Outputs on the DLL(5)	T_{PHOOM}			± 200		± 200	ps

Notes:
1. **Output Jitter** is cycle-to-cycle jitter measured on the DLL output clock, *excluding* input clock jitter.
2. **Phase Offset between CLKIN and CLKO** is the worst-case fixed time difference between rising edges of CLKIN and CLKO, *excluding* Output Jitter and input clock jitter.
3. **Phase Offset between Clock Outputs on the DLL** is the worst-case fixed time difference between rising edges of any two DLL outputs, *excluding* Output Jitter and input clock jitter.
4. **Maximum Phase Difference between CLKIN an CLKO** is the sum of Output Jitter and Phase Offset between CLKIN and CLKO, or the greatest difference between CLKIN and CLKO rising edges due to DLL alone (*excluding* input clock jitter).
5. **Maximum Phase DIfference between Clock Outputs on the DLL** is the sum of Output JItter and Phase Offset between any DLL clock outputs, or the greatest difference between any two DLL output rising edges sue to DLL alone (*excluding* input clock jitter).
6. All specifications correspond to Commercial Operating Temperatures (0°C to +85°C).

Virtex Pin Definitions

Table 1: **Special Purpose Pins**

Pin Name	Dedicated Pin	Direction	Description
GCK0, GCK1, GCK2, GCK3	Yes	Input	Clock input pins that connect to Global Clock Buffers. These pins become user inputs when not needed for clocks.
M0, M1, M2	Yes	Input	Mode pins are used to specify the configuration mode.
CCLK	Yes	Input or Output	The configuration Clock I/O pin: it is an input for SelectMAP and slave-serial modes, and output in master-serial mode. After configuration, it is input only, logic level = Don't Care.
PROGRAM	Yes	Input	Initiates a configuration sequence when asserted Low.
DONE	Yes	Bidirectional	Indicates that configuration loading is complete, and that the start-up sequence is in progress. The output can be open drain.
INIT	No	Bidirectional (Open-drain)	When Low, indicates that the configuration memory is being cleared. The pin becomes a user I/O after configuration.
BUSY/ DOUT	No	Output	In SelectMAP mode, BUSY controls the rate at which configuration data is loaded. The pin becomes a user I/O after configuration unless the SelectMAP port is retained. In bit-serial modes, DOUT provides header information to downstream devices in a daisy-chain. The pin becomes a user I/O after configuration.
D0/DIN, D1, D2, D3, D4, D5, D6, D7	No	Input or Output	In SelectMAP mode, D0 - D7 are configuration data pins. These pins become user I/Os after configuration unless the SelectMAP port is retained. In bit-serial modes, DIN is the single data input. This pin becomes a user I/O after configuration.
WRITE	No	Input	In SelectMAP mode, the active-low Write Enable signal. The pin becomes a user I/O after configuration unless the SelectMAP port is retained.
CS	No	Input	In SelectMAP mode, the active-low Chip Select signal. The pin becomes a user I/O after configuration unless the SelectMAP port is retained.
TDI, TDO, TMS, TCK	Yes	Mixed	Boundary-scan Test-Access-Port pins, as defined in IEEE 1149.1.
DXN, DXP	Yes	N/A	Temperature-sensing diode pins. (Anode: DXP, cathode: DXN)
V_{CCINT}	Yes	Input	Power-supply pins for the internal core logic.
V_{CCO}	Yes	Input	Power-supply pins for the output drivers (subject to banking rules)
V_{REF}	No	Input	Input threshold voltage pins. Become user I/Os when an external threshold voltage is not needed (subject to banking rules).
GND	Yes	Input	Ground

Table 3: **Virtex Pinout Tables (BGA)**

Pin Name	Device	BG256	BG352	BG432	BG560
GCK0	All	Y11	AE13	AL16	AL17
GCK1	All	Y10	AF14	AK16	AJ17
GCK2	All	A10	B14	A16	D17
GCK3	All	B10	D14	D17	A17
M0	All	Y1	AD24	AH28	AJ29
M1	All	U3	AB23	AH29	AK30
M2	All	W2	AC23	AJ28	AN32
CCLK	All	B19	C3	D4	C4
PROGRAM	All	Y20	AC4	AH3	AM1
DONE	All	W19	AD3	AH4	AJ5
INIT	All	U18	AD2	AJ2	AH5
BUSY/DOUT	All	D18	E4	D3	D4
D0/DIN	All	C19	D3	C2	E4
D1	All	E20	G1	K4	K3
D2	All	G19	J3	K2	L4
D3	All	J19	M3	P4	P3
D4	All	M19	R3	V4	W4
D5	All	P19	U4	AB1	AB5
D6	All	T20	V3	AB3	AC4
D7	All	V19	AC3	AG4	AJ4
WRITE	All	A19	D5	B4	D6
CS	All	B18	C4	D5	A2
TDI	All	C17	B3	B3	D5
TDO	All	A20	D4	C4	E6
TMS	All	D3	D23	D29	B33
TCK	All	A1	C24	D28	E29
DXN	All	W3	AD23	AH27	AK29
DXP	All	V4	AE24	AK29	AJ28

Table 3: **Virtex Pinout Tables (BGA) *(Continued)***

Pin Name	Device	BG256	BG352	BG432	BG560
V_{REF}, Bank 7 (V_{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	G3, H1	N/A	N/A	N/A
	XCV100/150	... + D1	D26, G26, L26	N/A	N/A
	XCV200/300	... + B2	... + E24	F28, F31, J30, N30	N/A
	XCV400	N/A	N/A	... + R31	E31, G31, K31, P31, T31
	XCV600	N/A	N/A	... + J28	... + H32
	XCV800	N/A	N/A	... + M28	... + L33
	XCV1000	N/A	N/A	N/A	... + D31
GND	All	C3, C18, D4, D5, D9, D10, D11, D12, D16, D17, E4, E17, J4, J17, K4, K17, L4, L17, M4, M17, T4, T17, U4, U5, U9, U10, U11, U12, U16, U17, V3, V18	A1, A2, A5, A8, A14, A19, A22, A25, A26, B1, B26, E1, E26, H1, H26, N1, P26, W1, W26, AB1, AB26, AE1, AE26, AF1, AF2, AF5, AF8, AF13, AF19, AF22, AF25, AF26	A2, A3, A7, A9, A14, A18, A23, A25, A29, A30, B1, B2, B30, B31, C1, C31, D16, G1, G31, J1, J31, P1, P31, T4, T28, V1, V31, AC1, AC31, AE1, AE31, AH16, AJ1, AJ31, AK1, AK2, AK30, AK31, AL2, AL3, AL7, AL9 AL14, AL18 AL23, AL25, AL29, AL30	A1, A7, A12, A14, A18, A20, A24, A29, A32, A33, B1, B6, B9, B15, B23, B27, B31, C2, E1, F32, G2, G33, J32, K1, L2, M33, P1, P33, R32, T1, V33, W2, Y1, Y33, AB1, AC32, AD33, AE2, AG1, AG32, AH2, AJ33, AL32, AM3, AM7, AM11, AM19, AM25, AM28, AM33, AN1, AN2, AN5, AN10, AN14, AN16, AN20, AN22, AN27, AN33
GND(1)	All	J9, J10, J11, J12, K9, K10, K11, K12, L9, L10, L11, L12, M9, M10, M11, M12	N/A	N/A	N/A
No Connect	All	N/A	N/A	N/A	C31, AC2, AK4, AL3

Notes:
1. 16 extra balls (grounded) at package center.

PQ240/HQ240 Pin Function Diagram

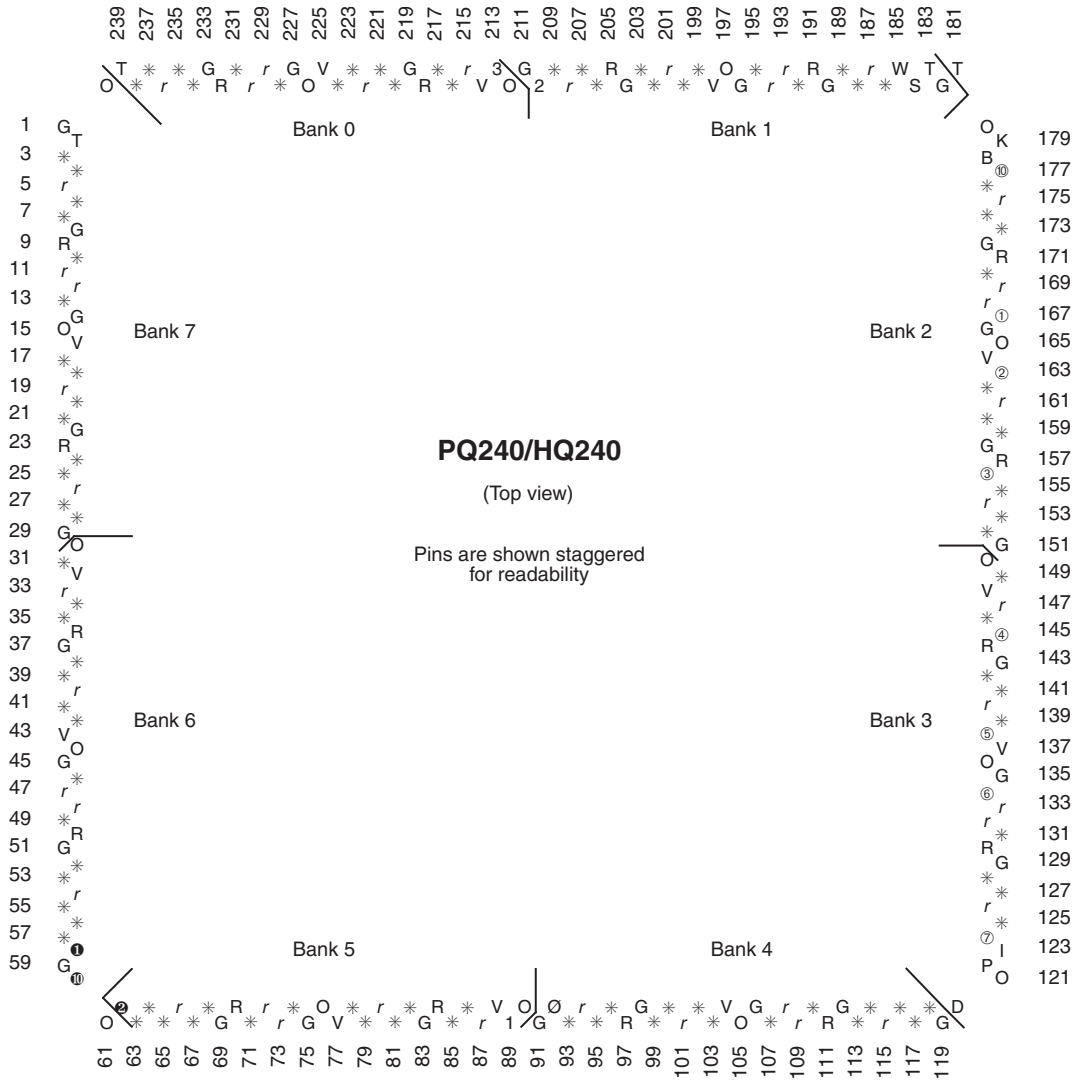


Figure 3: **PQ240/HQ240 Pin Function Diagram**

BG256 Pin Function Diagram

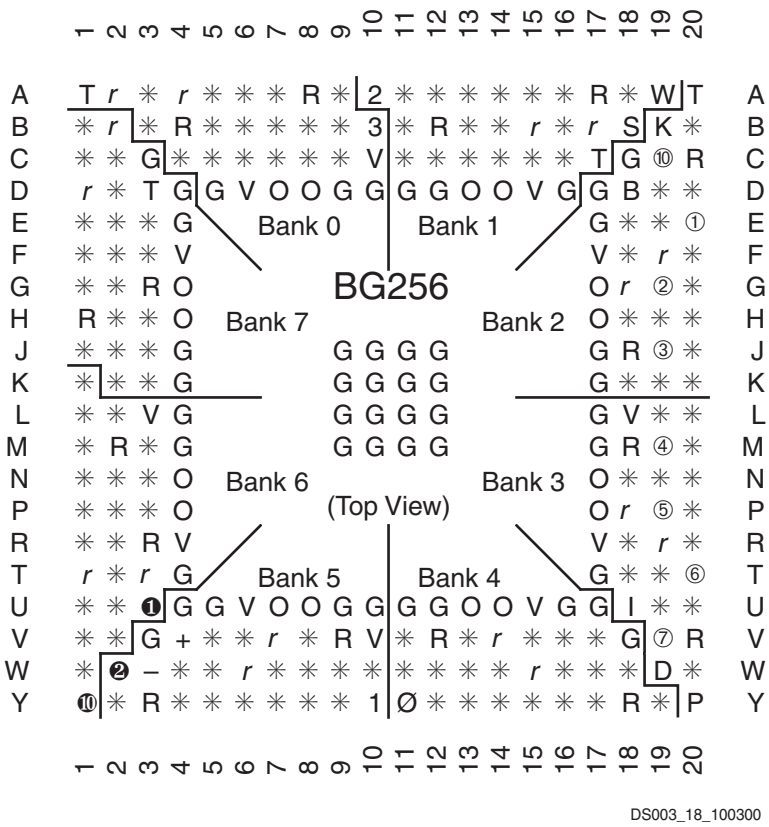


Figure 4: **BG256 Pin Function Diagram**