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Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	1176
Number of Logic Elements/Cells	5292
Total RAM Bits	57344
Number of I/O	284
Number of Gates	236666
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	456-BBGA
Supplier Device Package	456-FBGA (23x23)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcv200-6fg456c

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong



Virtex Device/Package Combinations and Maximum I/O

Table 3: Virtex Family Maximum User I/O by Device/Package (Excluding Dedicated Clock Pins)

Package	XCV50	XCV100	XCV150	XCV200	XCV300	XCV400	XCV600	XCV800	XCV1000
CS144	94	94							
TQ144	98	98							
PQ240	166	166	166	166	166				
HQ240						166	166	166	
BG256	180	180	180	180					
BG352			260	260	260				
BG432					316	316	316	316	
BG560						404	404	404	404
FG256	176	176	176	176					
FG456			260	284	312				
FG676						404	444	444	
FG680							512	512	512

Virtex Ordering Information

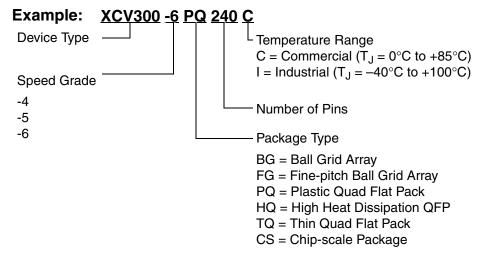


Figure 1: Virtex Ordering Information



Input Path

A buffer In the Virtex IOB input path routes the input signal either directly to internal logic or through an optional input flip-flop.

An optional delay element at the D-input of this flip-flop eliminates pad-to-pad hold time. The delay is matched to the internal clock-distribution delay of the FPGA, and when used, assures that the pad-to-pad hold time is zero.

Each input buffer can be configured to conform to any of the low-voltage signalling standards supported. In some of these standards the input buffer utilizes a user-supplied threshold voltage, V_{REF}. The need to supply V_{REF} imposes constraints on which standards can used in close proximity to each other. See I/O Banking, page 3.

There are optional pull-up and pull-down resistors at each user I/O input for use after configuration. Their value is in the range 50 k Ω – 100 k Ω .

Output Path

The output path includes a 3-state output buffer that drives the output signal onto the pad. The output signal can be routed to the buffer directly from the internal logic or through an optional IOB output flip-flop.

The 3-state control of the output can also be routed directly from the internal logic or through a flip-flip that provides synchronous enable and disable.

Each output driver can be individually programmed for a wide range of low-voltage signalling standards. Each output buffer can source up to 24 mA and sink up to 48mA. Drive strength and slew rate controls minimize bus transients.

In most signalling standards, the output High voltage depends on an externally supplied V_{CCO} voltage. The need to supply V_{CCO} imposes constraints on which standards can be used in close proximity to each other. See **I/O Banking**, page 3.

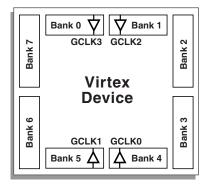
An optional weak-keeper circuit is connected to each output. When selected, the circuit monitors the voltage on the pad and weakly drives the pin High or Low to match the input signal. If the pin is connected to a multiple-source signal, the weak keeper holds the signal in its last state if all drivers are disabled. Maintaining a valid logic level in this way eliminates bus chatter.

Because the weak-keeper circuit uses the IOB input buffer to monitor the input level, an appropriate V_{REF} voltage must be provided if the signalling standard requires one. The provision of this voltage must comply with the I/O banking rules.

I/O Banking

Some of the I/O standards described above require V_{CCO} and/or V_{REF} voltages. These voltages externally and connected to device pins that serve groups of IOBs, called banks. Consequently, restrictions exist about which I/O standards can be combined within a given bank.

Eight I/O banks result from separating each edge of the FPGA into two banks, as shown in Figure 3. Each bank has multiple $V_{\rm CCO}$ pins, all of which must be connected to the same voltage. This voltage is determined by the output standards in use.



X8778_b

Figure 3: Virtex I/O Banks

Within a bank, output standards can be mixed only if they use the same V_{CCO} . Compatible standards are shown in Table 2. GTL and GTL+ appear under all voltages because their open-drain outputs do not depend on V_{CCO} .

Table 2: Compatible Output Standards

V _{CCO}	Compatible Standards
3.3 V	PCI, LVTTL, SSTL3 I, SSTL3 II, CTT, AGP, GTL, GTL+
2.5 V	SSTL2 I, SSTL2 II, LVCMOS2, GTL, GTL+
1.5 V	HSTL I, HSTL III, HSTL IV, GTL, GTL+

Some input standards require a user-supplied threshold voltage, V_{REF} In this case, certain user-I/O pins are automatically configured as inputs for the V_{REF} voltage. Approximately one in six of the I/O pins in the bank assume this role

The V_{REF} pins within a bank are interconnected internally and consequently only one V_{REF} voltage can be used within each bank. All V_{REF} pins in the bank, however, must be connected to the external voltage source for correct operation.

Within a bank, inputs that require V_{REF} can be mixed with those that do not. However, only one V_{REF} voltage can be used within a bank. Input buffers that use V_{REF} are not 5 V tolerant. LVTTL, LVCMOS2, and PCI 33 MHz 5 V, are 5 V tolerant.

The V_{CCO} and V_{REF} pins for each bank appear in the device Pinout tables and diagrams. The diagrams also show the bank affiliation of each I/O.

Within a given package, the number of V_{REF} and V_{CCO} pins can vary depending on the size of device. In larger devices,



more I/O pins convert to V_{REF} pins. Since these are always a superset of the V_{REF} pins used for smaller devices, it is possible to design a PCB that permits migration to a larger device if necessary. All the V_{REF} pins for the largest device anticipated must be connected to the V_{REF} voltage, and not used for I/O.

In smaller devices, some V_{CCO} pins used in larger devices do not connect within the package. These unconnected pins can be left unconnected externally, or can be connected to the V_{CCO} voltage to permit migration to a larger device if necessary.

In TQ144 and PQ/HQ240 packages, all V_{CCO} pins are bonded together internally, and consequently the same V_{CCO} voltage must be connected to all of them. In the CS144 package, bank pairs that share a side are interconnected internally, permitting four choices for V_{CCO} . In both cases, the V_{REF} pins remain internally connected as eight banks, and can be used as described previously.

Configurable Logic Block

The basic building block of the Virtex CLB is the logic cell (LC). An LC includes a 4-input function generator, carry logic, and a storage element. The output from the function generator in each LC drives both the CLB output and the D input of the flip-flop. Each Virtex CLB contains four LCs, organized in two similar slices, as shown in Figure 4.

Figure 5 shows a more detailed view of a single slice.

In addition to the four basic LCs, the Virtex CLB contains logic that combines function generators to provide functions

of five or six inputs. Consequently, when estimating the number of system gates provided by a given device, each CLB counts as 4.5 LCs.

Look-Up Tables

Virtex function generators are implemented as 4-input look-up tables (LUTs). In addition to operating as a function generator, each LUT can provide a 16 x 1-bit synchronous RAM. Furthermore, the two LUTs within a slice can be combined to create a 16 x 2-bit or 32 x 1-bit synchronous RAM, or a 16x1-bit dual-port synchronous RAM.

The Virtex LUT can also provide a 16-bit shift register that is ideal for capturing high-speed or burst-mode data. This mode can also be used to store data in applications such as Digital Signal Processing.

Storage Elements

The storage elements in the Virtex slice can be configured either as edge-triggered D-type flip-flops or as level-sensitive latches. The D inputs can be driven either by the function generators within the slice or directly from slice inputs, bypassing the function generators.

In addition to Clock and Clock Enable signals, each Slice has synchronous set and reset signals (SR and BY). SR forces a storage element into the initialization state specified for it in the configuration. BY forces it into the opposite state. Alternatively, these signals can be configured to operate asynchronously. All of the control signals are independently invertible, and are shared by the two flip-flops within the slice.

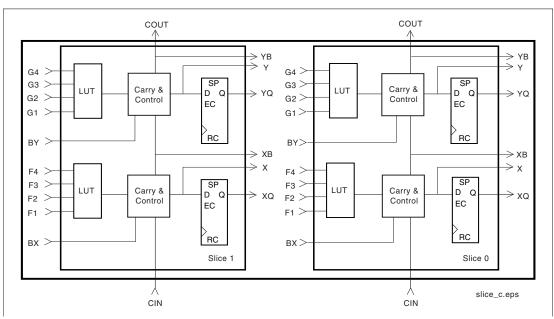


Figure 4: 2-Slice Virtex CLB



General Purpose Routing

Most Virtex signals are routed on the general purpose routing, and consequently, the majority of interconnect resources are associated with this level of the routing hierarchy. The general routing resources are located in horizontal and vertical routing channels associated with the rows and columns CLBs. The general-purpose routing resources are listed below.

- Adjacent to each CLB is a General Routing Matrix (GRM). The GRM is the switch matrix through which horizontal and vertical routing resources connect, and is also the means by which the CLB gains access to the general purpose routing.
- 24 single-length lines route GRM signals to adjacent GRMs in each of the four directions.
- 12 buffered Hex lines route GRM signals to another GRMs six-blocks away in each one of the four directions. Organized in a staggered pattern, Hex lines can be driven only at their endpoints. Hex-line signals can be accessed either at the endpoints or at the midpoint (three blocks from the source). One third of the Hex lines are bidirectional, while the remaining ones are uni-directional.

 12 Longlines are buffered, bidirectional wires that distribute signals across the device quickly and efficiently. Vertical Longlines span the full height of the device, and horizontal ones span the full width of the device.

I/O Routing

Virtex devices have additional routing resources around their periphery that form an interface between the CLB array and the IOBs. This additional routing, called the VersaRing, facilitates pin-swapping and pin-locking, such that logic redesigns can adapt to existing PCB layouts. Time-to-market is reduced, since PCBs and other system components can be manufactured while the logic design is still in progress.

Dedicated Routing

Some classes of signal require dedicated routing resources to maximize performance. In the Virtex architecture, dedicated routing resources are provided for two classes of signal.

- Horizontal routing resources are provided for on-chip 3-state busses. Four partitionable bus lines are provided per CLB row, permitting multiple busses within a row, as shown in Figure 8.
- Two dedicated nets per CLB propagate carry signals vertically to the adjacent CLB.

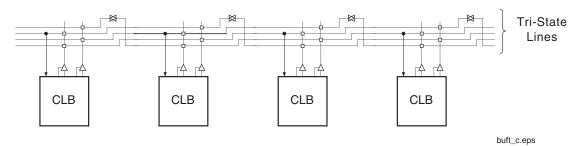


Figure 8: BUFT Connections to Dedicated Horizontal Bus Lines

Global Routing

Global Routing resources distribute clocks and other signals with very high fanout throughout the device. Virtex devices include two tiers of global routing resources referred to as primary global and secondary local clock routing resources.

• The primary global routing resources are four dedicated global nets with dedicated input pins that are designed to distribute high-fanout clock signals with minimal skew. Each global clock net can drive all CLB, IOB, and block RAM clock pins. The primary global nets can only be driven by global buffers. There are four global buffers, one for each global net. The secondary local clock routing resources consist of 24 backbone lines, 12 across the top of the chip and 12 across bottom. From these lines, up to 12 unique signals per column can be distributed via the 12 longlines in the column. These secondary resources are more flexible than the primary resources since they are not restricted to routing only to clock pins.

Clock Distribution

Virtex provides high-speed, low-skew clock distribution through the primary global routing resources described above. A typical clock distribution net is shown in Figure 9.

Four global buffers are provided, two at the top center of the device and two at the bottom center. These drive the four primary global nets that in turn drive any clock pin.

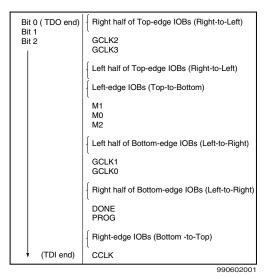


Figure 11: Boundary Scan Bit Sequence

Table 5: Boundary Scan Instructions

Boundary-Scan Command	Binary Code(4:0)	Description
EXTEST	00000	Enables boundary-scan EXTEST operation
SAMPLE/PRELOAD	00001	Enables boundary-scan SAMPLE/PRELOAD operation
USER 1	00010	Access user-defined register 1
USER 2	00011	Access user-defined register 2
CFG_OUT	00100	Access the configuration bus for read operations.
CFG_IN	00101	Access the configuration bus for write operations.
INTEST	00111	Enables boundary-scan INTEST operation
USERCODE	01000	Enables shifting out USER code
IDCODE	01001	Enables shifting out of ID Code
HIGHZ	01010	3-states output pins while enabling the Bypass Register
JSTART	01100	Clock the start-up sequence when StartupClk is TCK
BYPASS	11111	Enables BYPASS
RESERVED	All other codes	Xilinx reserved instructions

Identification Registers

The IDCODE register is supported. By using the IDCODE, the device connected to the JTAG port can be determined.

The IDCODE register has the following binary format:

vvvv:ffff:fffa:aaaa:aaaa:cccc:cccc1

where

v = the die version number

f = the family code (03h for Virtex family)

a = the number of CLB rows (ranges from 010h for XCV50 to 040h for XCV1000)

c = the company code (49h for Xilinx)

The USERCODE register is supported. By using the USER-CODE, a user-programmable identification code can be loaded and shifted out for examination. The identification code is embedded in the bitstream during bitstream generation and is valid only after configuration.

Table 6: IDCODEs Assigned to Virtex FPGAs

FPGA	IDCODE
XCV50	v0610093h
XCV100	v0614093h
XCV150	v0618093h
XCV200	v061C093h
XCV300	v0620093h
XCV400	v0628093h
XCV600	v0630093h
XCV800	v0638093h
XCV1000	v0640093h

Including Boundary Scan in a Design

Since the boundary scan pins are dedicated, no special element needs to be added to the design unless an internal data register (USER1 or USER2) is desired.

If an internal data register is used, insert the boundary scan symbol and connect the necessary pins as appropriate.

Development System

Virtex FPGAs are supported by the Xilinx Foundation and Alliance CAE tools. The basic methodology for Virtex design consists of three interrelated steps: design entry, implementation, and verification. Industry-standard tools are used for design entry and simulation (for example, Synopsys FPGA Express), while Xilinx provides proprietary architecture-specific tools for implementation.

The Xilinx development system is integrated under the Xilinx Design Manager (XDM™) software, providing design-

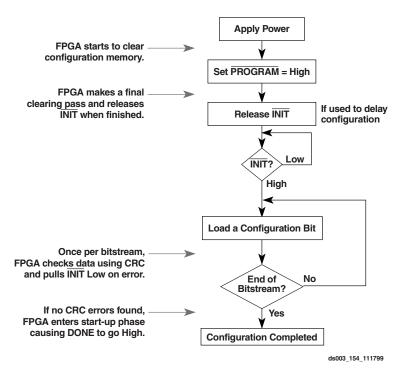


Figure 15: Serial Configuration Flowchart

After configuration, the pins of the SelectMAP port can be used as additional user I/O. Alternatively, the port can be retained to permit high-speed 8-bit readback.

Retention of the SelectMAP port is selectable on a design-by-design basis when the bitstream is generated. If retention is selected, PROHIBIT constraints are required to prevent the SelectMAP-port pins from being used as user I/O.

Multiple Virtex FPGAs can be configured using the Select-MAP mode, and be made to start-up simultaneously. To configure multiple devices in this way, wire the individual CCLK, Data, $\overline{\text{WRITE}}$, and BUSY pins of all the devices in parallel. The individual devices are loaded separately by asserting the $\overline{\text{CS}}$ pin of each device in turn and writing the appropriate data. see Table 9 for SelectMAP Write Timing Characteristics.

Table 9: SelectMAP Write Timing Characteristics

	Description		Symbol		Units
	D ₀₋₇ Setup/Hold	1/2	T _{SMDCC} /T _{SMCCD}	5.0 / 1.7	ns, min
	CS Setup/Hold	3/4	T _{SMCSCC} /T _{SMCCCS}	7.0 / 1.7	ns, min
CCLK	WRITE Setup/Hold	5/6	T _{SMCCW} /T _{SMWCC}	7.0 / 1.7	ns, min
COLK	BUSY Propagation Delay	7	T _{SMCKBY}	12.0	ns, max
	Maximum Frequency		F _{CC}	66	MHz, max
	Maximum Frequency with no handshake		F _{CCNH}	50	MHz, max

Write

Write operations send packets of configuration data into the FPGA. The sequence of operations for a multi-cycle write operation is shown below. Note that a configuration packet can be split into many such sequences. The packet does not have to complete within one assertion of \overline{CS} , illustrated in Figure 16.

- 1. Assert WRITE and CS Low. Note that when CS is asserted on successive CCLKs, WRITE must remain either asserted or de-asserted. Otherwise an abort will be initiated, as described below.
- 2. Drive data onto D[7:0]. Note that to avoid contention, the data source should not be enabled while \overline{CS} is Low and \overline{WRITE} is High. Similarly, while \overline{WRITE} is High, no more that one \overline{CS} should be asserted.



- At the rising edge of CCLK: If BUSY is Low, the data is accepted on this clock. If BUSY is High (from a previous write), the data is not accepted. Acceptance will instead occur on the first clock after BUSY goes Low, and the data must be held until this has happened.
- 4. Repeat steps 2 and 3 until all the data has been sent.
- 5. De-assert $\overline{\text{CS}}$ and $\overline{\text{WRITE}}$.

A flowchart for the write operation appears in Figure 17. Note that if CCLK is slower than f_{CCNH} , the FPGA never asserts BUSY. In this case, the above handshake is unnecessary, and data can simply be entered into the FPGA every CCLK cycle.

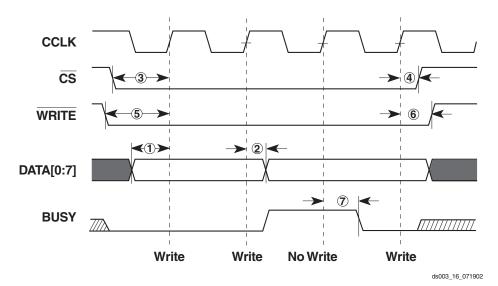


Figure 16: Write Operations



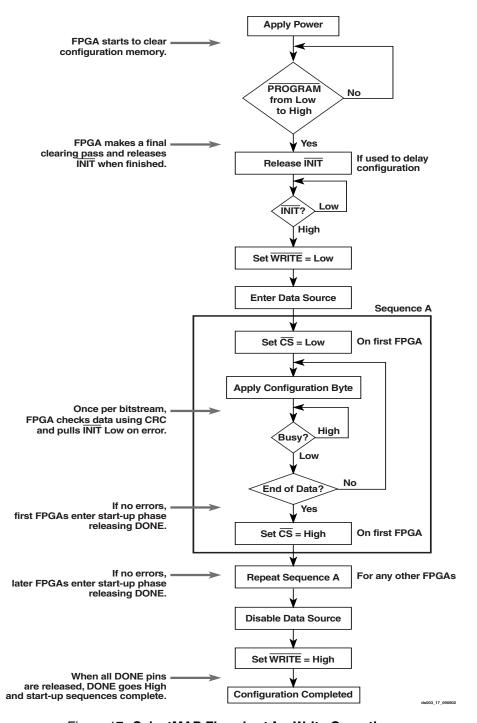


Figure 17: SelectMAP Flowchart for Write Operation

Abort

During a given assertion of $\overline{\text{CS}}$, the user cannot switch from a write to a read, or vice-versa. This action causes the current packet command to be aborted. The device will remain BUSY until the aborted operation has completed. Following an abort, data is assumed to be unaligned to word boundar-

ies, and the FPGA requires a new synchronization word prior to accepting any new packets.

To initiate an abort during a write operation, de-assert WRITE. At the rising edge of CCLK, an abort is initiated, as shown in Figure 18.



Power-On Power Supply Requirements

Xilinx FPGAs require a certain amount of supply current during power-on to insure proper device operation. The actual current consumed depends on the power-on ramp rate of the power supply. This is the time required to reach the nominal power supply voltage of the device⁽¹⁾ from 0 V. The current is highest at the fastest suggested ramp rate (0 V to nominal voltage in 2 ms) and is lowest at the slowest allowed ramp rate (0 V to nominal voltage in 50 ms). For more details on power supply requirements, see Application Note XAPP158 on www.xilinx.com.

Product	Description ⁽²⁾	Current Requirement ^(1,3)		
Virtex Family, Commercial Grade	Minimum required current supply	500 mA		
Virtex Family, Industrial Grade	Minimum required current supply	2 A		

Notes:

- Ramp rate used for this specification is from 0 2.7 VDC. Peak current occurs on or near the internal power-on reset threshold of 1.0V and lasts for less than 3 ms.
- Devices are guaranteed to initialize properly with the minimum current available from the power supply as noted above.
- 3. Larger currents can result if ramp rates are forced to be faster.

DC Input and Output Levels

Values for V_{IL} and V_{IH} are recommended input voltages. Values for I_{OL} and I_{OH} are guaranteed output currents over the recommended operating conditions at the V_{OL} and V_{OH} test points. Only selected standards are tested. These are chosen to ensure that all standards meet their specifications. The selected standards are tested at minimum V_{CCO} for each standard with the respective V_{OL} and V_{OH} voltage levels shown. Other standards are sample tested.

Input/Output		V _{IL}	VI	Н	V _{OL}	V _{OH}	I _{OL}	I _{OH}
Standard	V, min	V, max	V, min	V, max	V, Max	V, Min	mA	mA
LVTTL ⁽¹⁾	- 0.5	0.8	2.0	5.5	0.4	2.4	24	-24
LVCMOS2	- 0.5	.7	1.7	5.5	0.4	1.9	12	-12
PCI, 3.3 V	- 0.5	44% V _{CCINT}	60% V _{CCINT}	V _{CCO} + 0.5	10% V _{CCO}	90% V _{CCO}	Note 2	Note 2
PCI, 5.0 V	- 0.5	0.8	2.0	5.5	0.55	2.4	Note 2	Note 2
GTL	- 0.5	V _{REF} - 0.05	V _{REF} + 0.05	3.6	0.4	n/a	40	n/a
GTL+	- 0.5	V _{REF} – 0.1	V _{REF} + 0.1	3.6	0.6	n/a	36	n/a
HSTL I ⁽³⁾	- 0.5	V _{REF} – 0.1	V _{REF} + 0.1	3.6	0.4	V _{CCO} - 0.4	8	-8
HSTL III	- 0.5	V _{REF} – 0.1	V _{REF} + 0.1	3.6	0.4	V _{CCO} - 0.4	24	-8
HSTL IV	- 0.5	V _{REF} – 0.1	V _{REF} + 0.1	3.6	0.4	V _{CCO} - 0.4	48	-8
SSTL3 I	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	V _{REF} - 0.6	V _{REF} + 0.6	8	-8
SSTL3 II	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	V _{REF} - 0.8	V _{REF} + 0.8	16	-16
SSTL2 I	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	V _{REF} - 0.61	V _{REF} + 0.61	7.6	-7.6
SSTL2 II	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	V _{REF} - 0.80	V _{REF} + 0.80	15.2	-15.2
CTT	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	V _{REF} - 0.4	V _{REF} + 0.4	8	-8
AGP	- 0.5	V _{REF} - 0.2	V _{REF} + 0.2	3.6	10% V _{CCO}	90% V _{CCO}	Note 2	Note 2

Notes:

- V_{OL} and V_{OH} for lower drive currents are sample tested.
- 2. Tested according to the relevant specifications.
- DC input and output levels for HSTL18 (HSTL I/O standard with V_{CCO} of 1.8 V) are provided in an HSTL white paper on www.xilinx.com.



Block RAM Switching Characteristics

	Speed Grade					
Description	Symbol	Min	-6	-5	-4	Units
Sequential Delays						
Clock CLK to DOUT output	T _{BCKO}	1.7	3.4	3.8	4.3	ns, max
Setup and Hold Times before/after Clock CLK ⁽¹⁾		Setu	p Time / H	old Time		
ADDR inputs	T _{BACK} /T _{BCKA}	0.6 / 0	1.2 / 0	1.3 / 0	1.5 / 0	ns, min
DIN inputs	T _{BDCK} /T _{BCKD}	0.6 / 0	1.2 / 0	1.3 / 0	1.5 / 0	ns, min
EN input	T _{BECK} /T _{BCKE}	1.3 / 0	2.6 / 0	3.0 / 0	3.4 / 0	ns, min
RST input	T _{BRCK} /T _{BCKR}	1.3 / 0	2.5 / 0	2.7 / 0	3.2 / 0	ns, min
WEN input	T _{BWCK} /T _{BCKW}	1.2 / 0	2.3 / 0	2.6 / 0	3.0 / 0	ns, min
Clock CLK						
Minimum Pulse Width, High	T _{BPWH}	0.8	1.5	1.7	2.0	ns, min
Minimum Pulse Width, Low	T _{BPWL}	0.8	1.5	1.7	2.0	ns, min
CLKA -> CLKB setup time for different ports	T _{BCCS}		3.0	3.5	4.0	ns, min

Notes:

TBUF Switching Characteristics

		Speed Grade				
Description	Symbol	Min	-6	-5	-4	Units
Combinatorial Delays						
IN input to OUT output	T _{IO}	0	0	0	0	ns, max
TRI input to OUT output high-impedance	T _{OFF}	0.05	0.09	0.10	0.11	ns, max
TRI input to valid data on OUT output	T _{ON}	0.05	0.09	0.10	0.11	ns, max

JTAG Test Access Port Switching Characteristics

Description	Symbol	-6	-5	-4	Units
TMS and TDI Setup times before TCK	T _{TAPTCK}	4.0	4.0	4.0	ns, min
TMS and TDI Hold times after TCK	T _{TCKTAP}	2.0	2.0	2.0	ns, min
Output delay from clock TCK to output TDO	T _{TCKTDO}	11.0	11.0	11.0	ns, max
Maximum TCK clock frequency	F _{TCK}	33	33	33	MHz, max

^{1.} A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.



Global Clock Set-Up and Hold for LVTTL Standard, without DLL

Description	Symbol	Device	Min	-6	-5	-4	Units				
Input Setup and Hold Time Relat standards, adjust the setup time of					For data inp	ut with diffe	rent				
Full Delay Global Clock and IFF, without	T _{PSFD} /T _{PHFD}	XCV50	0.6 / 0	2.3 / 0	2.6 / 0	2.9 / 0	ns, min				
DLL		XCV100	0.6 / 0	2.3 / 0	2.6 / 0	3.0 / 0	ns, min				
		XCV150	0.6 / 0	2.4 / 0	2.7 / 0	3.1 / 0	ns, min				
		XCV200	0.7 / 0	2.5 / 0	2.8 / 0	3.2 / 0	ns, min				
		XCV300	0.7 / 0	2.5 / 0	2.8 / 0	3.2 / 0	ns, min				
		XCV400	0.7 / 0	2.6 / 0	2.9 / 0	3.3 / 0	ns, min				
		XCV600	0.7 / 0	2.6 / 0	2.9 / 0	3.3 / 0	ns, min				
								XCV800	0.7 / 0	2.7 / 0	3.1 / 0
		XCV1000	0.7 / 0	2.8 / 0	3.1 / 0	3.6 / 0	ns, min				

IFF = Input Flip-Flop or Latch

Notes: Notes:

- 1. Set-up time is measured relative to the Global Clock input signal with the fastest route and the lightest load. Hold time is measured relative to the Global Clock input signal with the slowest route and heaviest load.
- 2. A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.



DLL Timing Parameters

All devices are 100 percent functionally tested. Because of the difficulty in directly measuring many internal timing parameters, those parameters are derived from benchmark timing patterns. The following guidelines reflect worst-case values across the recommended operating conditions.

		Speed Grade						
		-6		-5		-4		
Description	Symbol	Min	Max	Min	Max	Min	Max	Units
Input Clock Frequency (CLKDLLHF)	FCLKINHF	60	200	60	180	60	180	MHz
Input Clock Frequency (CLKDLL)	FCLKINLF	25	100	25	90	25	90	MHz
Input Clock Pulse Width (CLKDLLHF)	T _{DLLPWHF}	2.0	-	2.4	-	2.4	-	ns
Input Clock Pulse Width (CLKDLL)	T _{DLLPWLF}	2.5	-	3.0		3.0	-	ns

Notes:

DLL Clock Tolerance, Jitter, and Phase Information

All DLL output jitter and phase specifications determined through statistical measurement at the package pins using a clock mirror configuration and matched drivers.

			CLKDLLHF		CLKDLL		
Description	Symbol	F _{CLKIN}	Min	Max	Min	Max	Units
Input Clock Period Tolerance	T _{IPTOL}		-	1.0	-	1.0	ns
Input Clock Jitter Tolerance (Cycle to Cycle)	T _{IJITCC}		-	± 150	-	± 300	ps
Time Required for DLL to Acquire Lock	T _{LOCK}	> 60 MHz	ı	20	-	20	μs
		50 - 60 MHz	ı	-	-	25	μs
		40 - 50 MHz	ı	-	-	50	μs
		30 - 40 MHz	ı	-	-	90	μs
		25 - 30 MHz	ı	-	-	120	μs
Output Jitter (cycle-to-cycle) for any DLL Clock Output (1)	T _{OJITCC}			± 60		± 60	ps
Phase Offset between CLKIN and CLKO ⁽²⁾	T _{PHIO}			± 100		± 100	ps
Phase Offset between Clock Outputs on the DLL ⁽³⁾	T _{PHOO}			± 140		± 140	ps
Maximum Phase Difference between CLKIN and CLKO ⁽⁴⁾	T _{PHIOM}			± 160		± 160	ps
Maximum Phase Difference between Clock Outputs on the DLL (5)	T _{PHOOM}			± 200		± 200	ps

Notes:

- 1. Output Jitter is cycle-to-cycle jitter measured on the DLL output clock, excluding input clock jitter.
- Phase Offset between CLKIN and CLKO is the worst-case fixed time difference between rising edges of CLKIN and CLKO, excluding Output Jitter and input clock jitter.
- Phase Offset between Clock Outputs on the DLL is the worst-case fixed time difference between rising edges of any two DLL outputs, excluding Output Jitter and input clock jitter.
- 4. Maximum Phase Difference between CLKIN an CLKO is the sum of Output Jitter and Phase Offset between CLKIN and CLKO, or the greatest difference between CLKIN and CLKO rising edges due to DLL alone (excluding input clock jitter).
- Maximum Phase Difference between Clock Outputs on the DLL is the sum of Output Jitter and Phase Offset between any DLL
 clock outputs, or the greatest difference between any two DLL output rising edges sue to DLL alone (excluding input clock jitter).
- 6. All specifications correspond to Commercial Operating Temperatures (0°C to +85°C).

^{1.} All specifications correspond to Commercial Operating Temperatures (0°C to + 85°C).



Date	Version	Revision				
01/00	1.9	Updated DLL Jitter Parameter table and waveforms, added Delay Measurement Methodology table for different I/O standards, changed buffered Hex line info and Input/Output Timing measurement notes.				
03/00	2.0	New TBCKO values; corrected FG680 package connection drawing; new note about status of CCLK pin after configuration.				
05/00	2.1	Modified "Pins not listed" statement. Speed grade update to Final status.				
05/00	2.2	Modified Table 18.				
09/00	2.3	 Added XCV400 values to table under Minimum Clock-to-Out for Virtex Devices. Corrected Units column in table under IOB Input Switching Characteristics. Added values to table under CLB SelectRAM Switching Characteristics. 				
10/00	2.4	 Corrected Pinout information for devices in the BG256, BG432, and BG560 packages in Table 18. Corrected BG256 Pin Function Diagram. 				
04/02/01	2.5	 Revised minimums for Global Clock Set-Up and Hold for LVTTL Standard, with DLL. Converted file to modularized format. See the Virtex Data Sheet section. 				
04/19/01	2.6	Clarified TIOCKP and TIOCKON IOB Output Switching Characteristics descriptors.				
07/19/01	2.7	Under Absolute Maximum Ratings, changed (T _{SOL}) to 220 °C.				
07/26/01	2.8	Removed T _{SOL} parameter and added footnote to Absolute Maximum Ratings table.				
10/29/01	2.9	 Updated the speed grade designations used in data sheets, and added Table 1, which shows the current speed grade designation for each device. 				
02/01/02	3.0	Added footnote to DC Input and Output Levels table.				
07/19/02	3.1	 Removed mention of MIL-M-38510/605 specification. Added link to xapp158 from the Power-On Power Supply Requirements section. 				
09/10/02	3.2	 Added Clock CLK to IOB Input Switching Characteristics and IOB Output Switching Characteristics. 				
03/01/13	4.0	The products listed in this data sheet are obsolete. See XCN10016 for further information.				

Virtex Data Sheet

The Virtex Data Sheet contains the following modules:

- DS003-1, Virtex 2.5V FPGAs: Introduction and Ordering Information (Module 1)
- DS003-2, Virtex 2.5V FPGAs: Functional Description (Module 2)

- DS003-3, Virtex 2.5V FPGAs:
 DC and Switching Characteristics (Module 3)
- DS003-4, Virtex 2.5V FPGAs: Pinout Tables (Module 4)



Table 4: Virtex Pinout Tables (Fine-Pitch BGA)

Pin Name	Device	FG256	FG456	FG676	FG680
GCK0	All	N8	W12	AA14	AW19
GCK1	All	R8	Y11	AB13	AU22
GCK2	All	C9	A11	C13	D21
GCK3	All	B8	C11	E13	A20
M0	All	N3	AB2	AD4	AT37
M1	All	P2	U5	W7	AU38
M2	All	R3	Y4	AB6	AT35
CCLK	All	D15	B22	D24	E4
PROGRAM	All	P15	W20	AA22	AT5
DONE	All	R14	Y19	AB21	AU5
INIT	All	N15	V19	Y21	AU2
BUSY/DOUT	All	C15	C21	E23	E3
D0/DIN	All	D14	D20	F22	C2
D1	All	E16	H22	K24	P4
D2	All	F15	H20	K22	P3
D3	All	G16	K20	M22	R1
D4	All	J16	N22	R24	AD3
D5	All	M16	R21	U23	AG2
D6	All	N16	T22	V24	AH1
D7	All	N14	Y21	AB23	AR4
WRITE	All	C13	A20	C22	B4
CS	All	B13	C19	E21	D5
TDI	All	A15	B20	D22	В3
TDO	All	B14	A21	C23	C4
TMS	All	D3	D3	F5	E36
TCK	All	C4	C4	E6	C36
DXN	All	R4	Y5	AB7	AV37
DXP	All	P4	V6	Y8	AU35



Table 4: Virtex Pinout Tables (Fine-Pitch BGA) (Continued)

Pin Name	Device	FG256	FG456	FG676	FG680
V _{REF} Bank 1	XCV50	B9, C11	N/A	N/A	N/A
(VREF pins are listed	XCV100/150	+ E11	A18, B13, E14	N/A	N/A
incrementally. Connect all pins listed for both	XCV200/300	+ A14	+ A19	N/A	N/A
the required device and all smaller devices	XCV400	N/A	N/A	A14, C20, C21, D15, G16	N/A
listed in the same package.) Within each bank, if	XCV600	N/A	N/A	+ B19	B6, B8, B18, D11, D13, D17
input reference voltage	XCV800	N/A	N/A	+ A17	+ B14
is not required, all V _{REF} pins are general I/O.	XCV1000	N/A	N/A	N/A	+ B5
V _{REF} , Bank 2	XCV50	F13, H13	N/A	N/A	N/A
(V _{REF} pins are listed	XCV100/150	+ F14	F21, H18, K21	N/A	N/A
incrementally. Connect all pins listed for both	XCV200/300	+ E13	+ D22	N/A	N/A
the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V _{REF} pins are general I/O.	XCV400	N/A	N/A	F24, H23, K20, M23, M26	N/A
	XCV600	N/A	N/A	+ G26	G1, H4, J1, L2, V5, W3
	XCV800	N/A	N/A	+ K25	+ N1
	XCV1000	N/A	N/A	N/A	+ D2
V _{REF} , Bank 3	XCV50	K16, L14	N/A	N/A	N/A
(V _{REF} pins are listed	XCV100/150	+ L13	N21, R19, U21	N/A	N/A
incrementally. Connect all pins listed for both	XCV200/300	+ M13	+ U20	N/A	N/A
the required device and all smaller devices listed in the same package.) Within each bank, if	XCV400	N/A	N/A	R23, R25, U21, W22, W23	N/A
	XCV600	N/A	N/A	+ W26	AC1, AJ2, AK3, AL4, AR1, Y1
input reference voltage	XCV800	N/A	N/A	+ U25	+ AF3
is not required, all V _{REF} pins are general I/O.	XCV1000	N/A	N/A	N/A	+ AP4

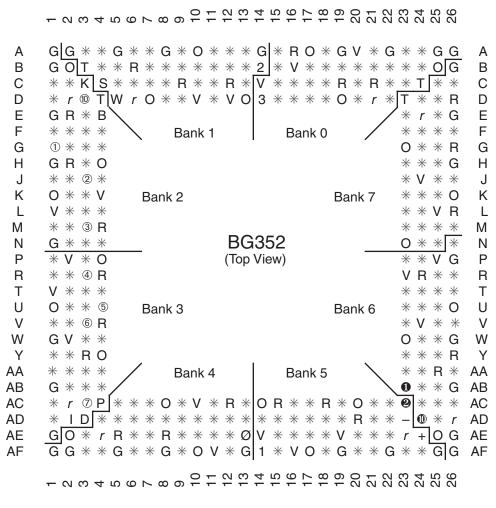


Table 4: Virtex Pinout Tables (Fine-Pitch BGA) (Continued)

Pin Name	Device	FG256	FG456	FG676	FG680
No Connect (No-connect pins are listed incrementally. All pins listed for both the required device and all larger devices listed in the same package are no connects.)	XCV800	N/A	N/A	A2, A3, A15, A25, B1, B6, B11, B16, B21, B24, B26, C1, C2, C25, C26, F2, F6, F21, F25, L2, L25, N25, P2, T2, T25, AA2, AA6, AA21, AA25, AD1, AD2, AD25, AE1, AE3, AE6, AE11, AE14, AE16, AE21, AE24, AE26, AF2, AF24, AF25	N/A
	XCV600	N/A	N/A	same as above	N/A
	XCV400	N/A	N/A	+ A9, A10, A13, A16, A24, AC1, AC25, AE12, AE15, AF3, AF10, AF11, AF13, AF14, AF16, AF18, AF23, B4, B12, B13, B15, B17, D1, D25, H26, J1, K26, L1, M1, M25, N1, N26, P1, P26, R2, R26, T1, T26, U26, V1	N/A
	XCV300	N/A	D4, D19, W4, W19	N/A	N/A
	XCV200	N/A	+ A2, A6, A12, B11, B16, C2, D1, D18, E17, E19, G2, G22, L2, L19, M2, M21, R3, R20, U3, U18, Y22, AA1, AA3, AA11, AA16, AB7, AB12, AB21,	N/A	N/A
	XCV150	N/A	+ A13, A14, C8, C9, E13, F11, H21, J1, J4, K2, K18, K19, M17, N1, P1, P5, P22, R22, W13, W15, AA9, AA10, AB8, AB14	N/A	N/A



BG352 Pin Function Diagram



DS003_19_100600

Figure 5: BG352 Pin Function Diagram



FG256 Pin Function Diagram

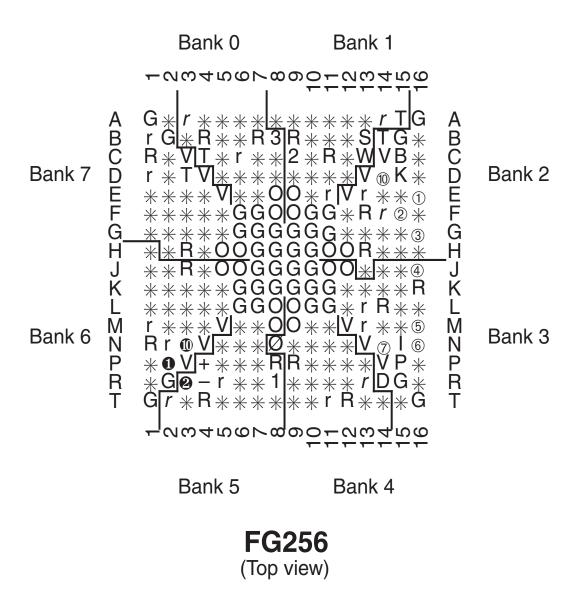


Figure 8: FG256 Pin Function Diagram



FG680 Pin Function Diagram

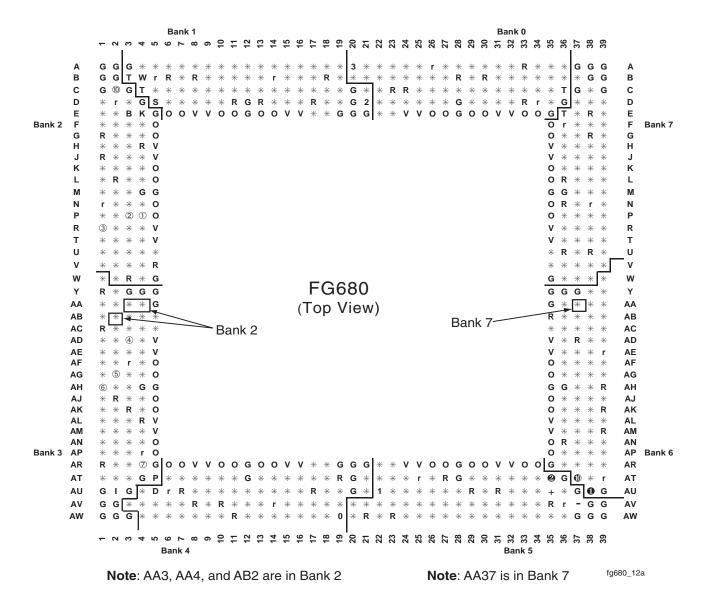


Figure 11: FG680 Pin Function Diagram



Revision History

Date	Version	Revision
11/98	1.0	Initial Xilinx release.
01/99-02/99	1.2-1.3	Both versions updated package drawings and specs.
05/99	1.4	Addition of package drawings and specifications.
05/99	1.5	Replaced FG 676 & FG680 package drawings.
07/99	1.6	Changed Boundary Scan Information and changed Figure 11, Boundary Scan Bit Sequence. Updated IOB Input & Output delays. Added Capacitance info for different I/O Standards. Added 5 V tolerant information. Added DLL Parameters and waveforms and new Pin-to-pin Input and Output Parameter tables for Global Clock Input to Output and Setup and Hold. Changed Configuration Information including Figures 12, 14, 17 & 19. Added device-dependent listings for quiescent currents ICCINTQ and ICCOQ. Updated IOB Input and Output Delays based on default standard of LVTTL, 12 mA, Fast Slew Rate. Added IOB Input Switching Characteristics Standard Adjustments.
09/99	1.7	Speed grade update to preliminary status, Power-on specification and Clock-to-Out Minimums additions, "0" hold time listing explanation, quiescent current listing update, and Figure 6 ADDRA input label correction. Added T _{IJITCC} parameter, changed T _{OJIT} to T _{OPHASE} .
01/00	1.8	Update to speed.txt file 1.96. Corrections for CRs 111036,111137, 112697, 115479, 117153, 117154, and 117612. Modified notes for Recommended Operating Conditions (voltage and temperature). Changed Bank information for V _{CCO} in CS144 package on p.43.
01/00	1.9	Updated DLL Jitter Parameter table and waveforms, added Delay Measurement Methodology table for different I/O standards, changed buffered Hex line info and Input/Output Timing measurement notes.
03/00	2.0	New TBCKO values; corrected FG680 package connection drawing; new note about status of CCLK pin after configuration.
05/00	2.1	Modified "Pins not listed" statement. Speed grade update to Final status.
05/00	2.2	Modified Table 18.
09/00	2.3	 Added XCV400 values to table under Minimum Clock-to-Out for Virtex Devices. Corrected Units column in table under IOB Input Switching Characteristics. Added values to table under CLB SelectRAM Switching Characteristics.
10/00	2.4	 Corrected pinout info for devices in the BG256, BG432, and BG560 pkgs in Table 18. Corrected BG256 Pin Function Diagram.
04/02/01	2.5	 Revised minimums for Global Clock Set-Up and Hold for LVTTL Standard, with DLL. Converted file to modularized format. See section Virtex Data Sheet, below.
04/19/01	2.6	Corrected pinout information for FG676 device in Table 4. (Added AB22 pin.)
07/19/01	2.7	 Clarified V_{CCINT} pinout information and added AE19 pin for BG352 devices in Table 3. Changed pinouts listed for BG352 XCV400 devices in banks 0 thru 7.
07/19/02	2.8	Changed pinouts listed for GND in TQ144 devices (see Table 2).
03/01/13	4.0	The products listed in this data sheet are obsolete. See XCN10016 for further information.

Virtex Data Sheet

The Virtex Data Sheet contains the following modules:

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 DC and Switching Characteristics (Module 3)
- DS003-4, Virtex 2.5V FPGAs: Pinout Tables (Module 4)