AMD Xilinx - XCV600-4BG432C Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	3456
Number of Logic Elements/Cells	15552
Total RAM Bits	98304
Number of I/O	316
Number of Gates	661111
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	432-LBGA Exposed Pad, Metal
Supplier Device Package	432-MBGA (40x40)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcv600-4bg432c

Virtex Device/Package Combinations and Maximum I/O

Table 3: **Virtex Family Maximum User I/O by Device/Package (Excluding Dedicated Clock Pins)**

Package	XCV50	XCV100	XCV150	XCV200	XCV300	XCV400	XCV600	XCV800	XCV1000
CS144	94	94							
TQ144	98	98							
PQ240	166	166	166	166	166				
HQ240						166	166	166	
BG256	180	180	180	180					
BG352			260	260	260				
BG432					316	316	316	316	
BG560						404	404	404	404
FG256	176	176	176	176					
FG456			260	284	312				
FG676						404	444	444	
FG680							512	512	512

Virtex Ordering Information

Example: **XCV300 -6 PQ 240 C**

Device Type

Speed Grade
- -4
- -5
- -6

Temperature Range
C = Commercial (T_J = 0°C to +85°C)
I = Industrial (T_J = –40°C to +100°C)

Number of Pins

Package Type
- BG = Ball Grid Array
- FG = Fine-pitch Ball Grid Array
- PQ = Plastic Quad Flat Pack
- HQ = High Heat Dissipation QFP
- TQ = Thin Quad Flat Pack
- CS = Chip-scale Package

Figure 1: **Virtex Ordering Information**

Revision History

Date	Version	Revision
11/98	1.0	Initial Xilinx release.
01/99-02/99	1.2-1.3	Both versions updated package drawings and specs.
05/99	1.4	Addition of package drawings and specifications.
05/99	1.5	Replaced FG 676 & FG680 package drawings.
07/99	1.6	Changed Boundary Scan Information and changed Figure 11, Boundary Scan Bit Sequence. Updated IOB Input & Output delays. Added Capacitance info for different I/O Standards. Added 5 V tolerant information. Added DLL Parameters and waveforms and new Pin-to-pin Input and Output Parameter tables for Global Clock Input to Output and Setup and Hold. Changed Configuration Information including Figures 12, 14, 17 & 19. Added device-dependent listings for quiescent currents ICCINTQ and ICCOQ. Updated IOB Input and Output Delays based on default standard of LVTTL, 12 mA, Fast Slew Rate. Added IOB Input Switching Characteristics Standard Adjustments.
09/99	1.7	Speed grade update to preliminary status, Power-on specification and Clock-to-Out Minimums additions, "0" hold time listing explanation, quiescent current listing update, and Figure 6 ADDRA input label correction. Added T_{IJITCC} parameter, changed T_{OJIT} to T_{OPHASE}.
01/00	1.8	Update to speed.txt file 1.96. Corrections for CRs 111036,111137, 112697, 115479, 117153, 117154, and 117612. Modified notes for Recommended Operating Conditions (voltage and temperature). Changed Bank information for V_{CCO} in CS144 package on p.43.
01/00	1.9	Updated DLL Jitter Parameter table and waveforms, added Delay Measurement Methodology table for different I/O standards, changed buffered Hex line info and Input/Output Timing measurement notes.
03/00	2.0	New TBCKO values; corrected FG680 package connection drawing; new note about status of CCLK pin after configuration.
05/00	2.1	Modified "Pins not listed..." statement. Speed grade update to Final status.
05/00	2.2	Modified Table 18.
09/00	2.3	• Added XCV400 values to table under **Minimum Clock-to-Out for Virtex Devices**. • Corrected Units column in table under **IOB Input Switching Characteristics**. • Added values to table under **CLB SelectRAM Switching Characteristics**.
10/00	2.4	• Corrected Pinout information for devices in the BG256, BG432, and BG560 packages in Table 18. • Corrected **BG256 Pin Function Diagram**.
04/01	2.5	• Revised minimums for **Global Clock Set-Up and Hold for LVTTL Standard, with DLL**. • Converted file to modularized format. See **Virtex Data Sheet** section.
03/13	4.0	The products listed in this data sheet are obsolete. See **XCN10016** for further information.

Virtex Data Sheet

The Virtex Data Sheet contains the following modules:

- DS003-1, Virtex 2.5V FPGAs:
  Introduction and Ordering Information (Module 1)
- DS003-2, Virtex 2.5V FPGAs:
  Functional Description (Module 2)
- DS003-3, Virtex 2.5V FPGAs:
  DC and Switching Characteristics (Module 3)
- DS003-4, Virtex 2.5V FPGAs:
  Pinout Tables (Module 4)

Virtex™ 2.5 V Field Programmable Gate Arrays

DS003-2 (v4.0) March 1, 2013

Product Specification

Architectural Description

Virtex Array

The Virtex user-programmable gate array, shown in Figure 1, comprises two major configurable elements: configurable logic blocks (CLBs) and input/output blocks (IOBs).

- CLBs provide the functional elements for constructing logic
- IOBs provide the interface between the package pins and the CLBs

CLBs interconnect through a general routing matrix (GRM). The GRM comprises an array of routing switches located at the intersections of horizontal and vertical routing channels. Each CLB nests into a VersaBlock™ that also provides local routing resources to connect the CLB to the GRM.

The VersaRing™ I/O interface provides additional routing resources around the periphery of the device. This routing improves I/O routability and facilitates pin locking.

The Virtex architecture also includes the following circuits that connect to the GRM.

- Dedicated block memories of 4096 bits each
- Clock DLLs for clock-distribution delay compensation and clock domain control
- 3-State buffers (BUFTs) associated with each CLB that drive dedicated segmentable horizontal routing resources

Values stored in static memory cells control the configurable logic elements and interconnect resources. These values load into the memory cells on power-up, and can reload if necessary to change the function of the device.

Input/Output Block

The Virtex IOB, Figure 2, features SelectIO™ inputs and outputs that support a wide variety of I/O signalling standards, see Table 1.

The three IOB storage elements function either as edge-triggered D-type flip-flops or as level sensitive latches. Each IOB has a clock signal (CLK) shared by the three flip-flops and independent clock enable signals for each flip-flop.

In addition to the CLK and CE control signals, the three flip-flops share a Set/Reset (SR). For each flip-flop, this signal can be independently configured as a synchronous Set, a synchronous Reset, an asynchronous Preset, or an asynchronous Clear.

The output buffer and all of the IOB control signals have independent polarity controls.

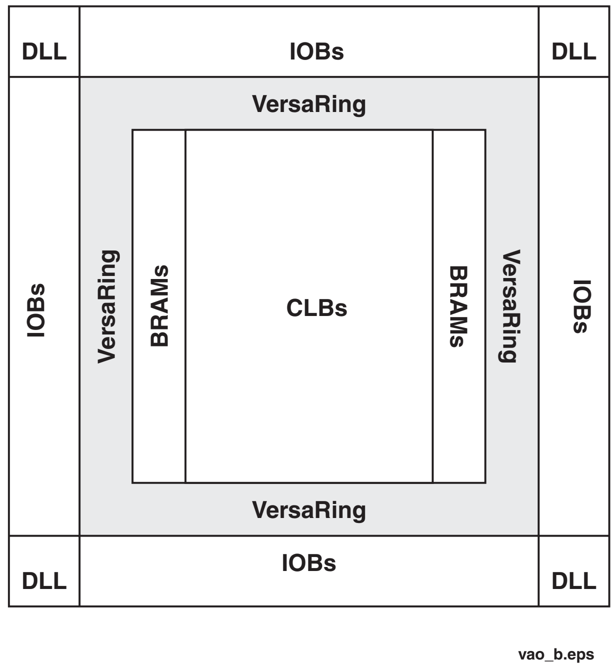

Figure 1: **Virtex Architecture Overview**

All pads are protected against damage from electrostatic discharge (ESD) and from over-voltage transients. Two forms of over-voltage protection are provided, one that permits 5 V compliance, and one that does not. For 5 V compliance, a Zener-like structure connected to ground turns on when the output rises to approximately 6.5 V. When PCI 3.3 V compliance is required, a conventional clamp diode is connected to the output supply voltage, V_{CCO}.

Optional pull-up and pull-down resistors and an optional weak-keeper circuit are attached to each pad. Prior to configuration, all pins not involved in configuration are forced into their high-impedance state. The pull-down resistors and the weak-keeper circuits are inactive, but inputs can optionally be pulled up.

The activation of pull-up resistors prior to configuration is controlled on a global basis by the configuration mode pins. If the pull-up resistors are not activated, all the pins will float. Consequently, external pull-up or pull-down resistors must be provided on pins required to be at a well-defined logic level prior to configuration.

All Virtex IOBs support IEEE 1149.1-compatible boundary scan testing.

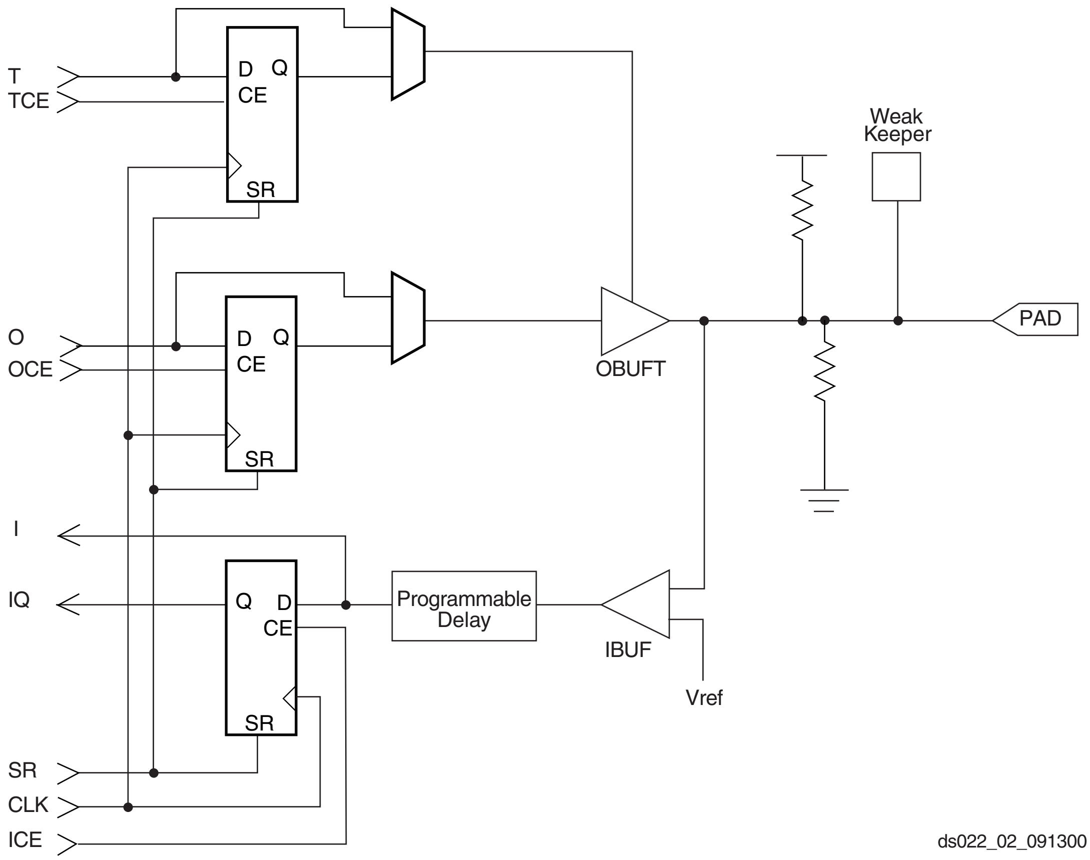

Figure 2: **Virtex Input/Output Block (IOB)**

Table 1: **Supported Select I/O Standards**

I/O Standard	Input Reference Voltage (V_{REF})	Output Source Voltage (V_{CCO})	Board Termination Voltage (V_{TT})	5 V Tolerant
LVTTL 2 – 24 mA	N/A	3.3	N/A	Yes
LVCMOS2	N/A	2.5	N/A	Yes
PCI, 5 V	N/A	3.3	N/A	Yes
PCI, 3.3 V	N/A	3.3	N/A	No
GTL	0.8	N/A	1.2	No
GTL+	1.0	N/A	1.5	No
HSTL Class I	0.75	1.5	0.75	No
HSTL Class III	0.9	1.5	1.5	No
HSTL Class IV	0.9	1.5	1.5	No
SSTL3 Class I &II	1.5	3.3	1.5	No
SSTL2 Class I & II	1.25	2.5	1.25	No
CTT	1.5	3.3	1.5	No
AGP	1.32	3.3	N/A	No

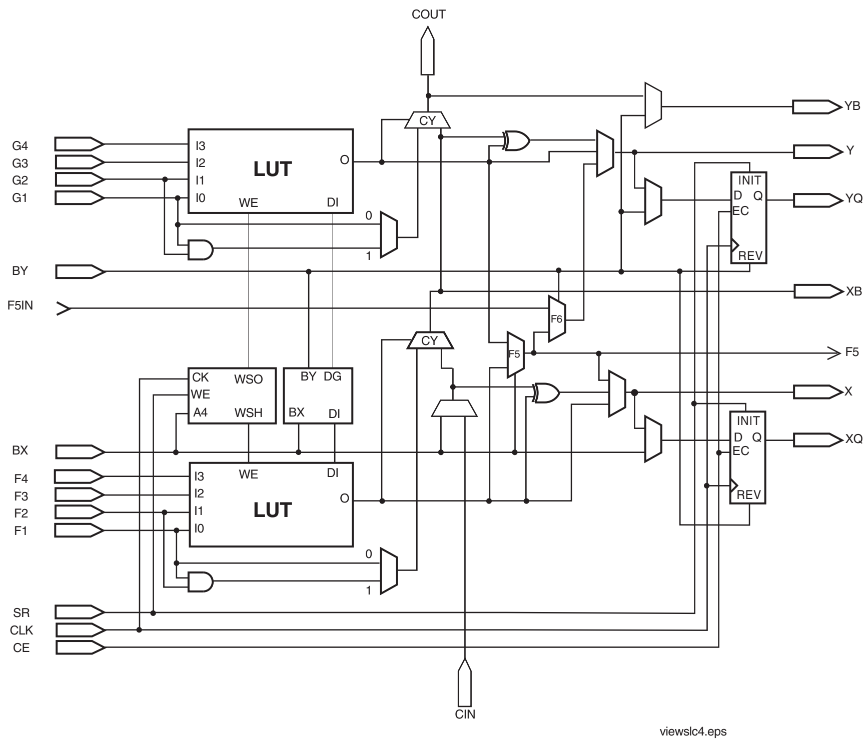

Figure 5: **Detailed View of Virtex Slice**

Additional Logic

The F5 multiplexer in each slice combines the function generator outputs. This combination provides either a function generator that can implement any 5-input function, a 4:1 multiplexer, or selected functions of up to nine inputs.

Similarly, the F6 multiplexer combines the outputs of all four function generators in the CLB by selecting one of the F5-multiplexer outputs. This permits the implementation of any 6-input function, an 8:1 multiplexer, or selected functions of up to 19 inputs.

Each CLB has four direct feedthrough paths, one per LC. These paths provide extra data input lines or additional local routing that does not consume logic resources.

Arithmetic Logic

Dedicated carry logic provides fast arithmetic carry capability for high-speed arithmetic functions. The Virtex CLB supports two separate carry chains, one per Slice. The height of the carry chains is two bits per CLB.

The arithmetic logic includes an XOR gate that allows a 1-bit full adder to be implemented within an LC. In addition, a dedicated AND gate improves the efficiency of multiplier implementation.

The dedicated carry path can also be used to cascade function generators for implementing wide logic functions.

BUFTs

Each Virtex CLB contains two 3-state drivers (BUFTs) that can drive on-chip busses. See **Dedicated Routing**, page 7. Each Virtex BUFT has an independent 3-state control pin and an independent input pin.

Block SelectRAM

Virtex FPGAs incorporate several large block SelectRAM memories. These complement the distributed LUT SelectRAMs that provide shallow RAM structures implemented in CLBs.

Block SelectRAM memory blocks are organized in columns. All Virtex devices contain two such columns, one along each vertical edge. These columns extend the full height of the chip. Each memory block is four CLBs high, and consequently, a Virtex device 64 CLBs high contains 16 memory blocks per column, and a total of 32 blocks.

Table 3 shows the amount of block SelectRAM memory that is available in each Virtex device.

Table 3: **Virtex Block SelectRAM Amounts**

Device	# of Blocks	Total Block SelectRAM Bits
XCV50	8	32,768
XCV100	10	40,960
XCV150	12	49,152
XCV200	14	57,344
XCV300	16	65,536
XCV400	20	81,920
XCV600	24	98,304
XCV800	28	114,688
XCV1000	32	131,072

Each block SelectRAM cell, as illustrated in Figure 6, is a fully synchronous dual-ported 4096-bit RAM with independent control signals for each port. The data widths of the two ports can be configured independently, providing built-in bus-width conversion.

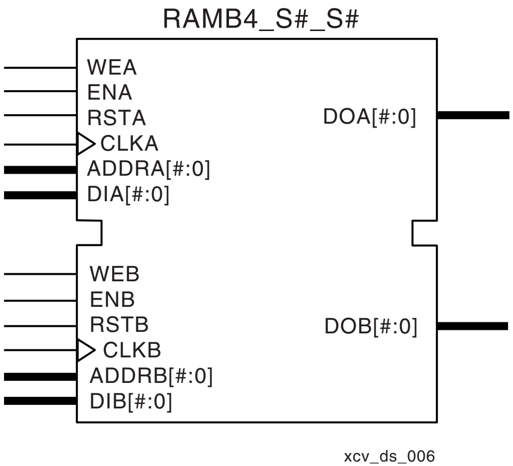

Figure 6: **Dual-Port Block SelectRAM**

Table 4 shows the depth and width aspect ratios for the block SelectRAM.

Table 4: **Block SelectRAM Port Aspect Ratios**

Width	Depth	ADDR Bus	Data Bus
1	4096	ADDR<11:0>	DATA<0>
2	2048	ADDR<10:0>	DATA<1:0>
4	1024	ADDR<9:0>	DATA<3:0>
8	512	ADDR<8:0>	DATA<7:0>
16	256	ADDR<7:0>	DATA<15:0>

The Virtex block SelectRAM also includes dedicated routing to provide an efficient interface with both CLBs and other block SelectRAMs. Refer to XAPP130 for block SelectRAM timing waveforms.

Programmable Routing Matrix

It is the longest delay path that limits the speed of any worst-case design. Consequently, the Virtex routing architecture and its place-and-route software were defined in a single optimization process. This joint optimization minimizes long-path delays, and consequently, yields the best system performance.

The joint optimization also reduces design compilation times because the architecture is software-friendly. Design cycles are correspondingly reduced due to shorter design iteration times.

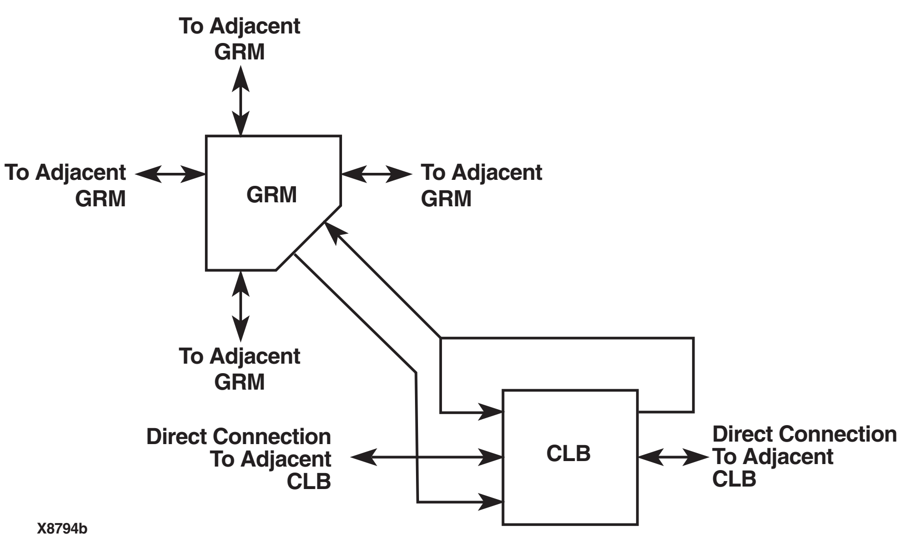

Figure 7: **Virtex Local Routing**

Local Routing

The VersaBlock provides local routing resources, as shown in Figure 7, providing the following three types of connections.

- Interconnections among the LUTs, flip-flops, and GRM
- Internal CLB feedback paths that provide high-speed connections to LUTs within the same CLB, chaining them together with minimal routing delay
- Direct paths that provide high-speed connections between horizontally adjacent CLBs, eliminating the delay of the GRM.

General Purpose Routing

Most Virtex signals are routed on the general purpose routing, and consequently, the majority of interconnect resources are associated with this level of the routing hierarchy. The general routing resources are located in horizontal and vertical routing channels associated with the rows and columns CLBs. The general-purpose routing resources are listed below.

- Adjacent to each CLB is a General Routing Matrix (GRM). The GRM is the switch matrix through which horizontal and vertical routing resources connect, and is also the means by which the CLB gains access to the general purpose routing.
- 24 single-length lines route GRM signals to adjacent GRMs in each of the four directions.
- 12 buffered Hex lines route GRM signals to another GRMs six-blocks away in each one of the four directions. Organized in a staggered pattern, Hex lines can be driven only at their endpoints. Hex-line signals can be accessed either at the endpoints or at the midpoint (three blocks from the source). One third of the Hex lines are bidirectional, while the remaining ones are uni-directional.
- 12 Longlines are buffered, bidirectional wires that distribute signals across the device quickly and efficiently. Vertical Longlines span the full height of the device, and horizontal ones span the full width of the device.

I/O Routing

Virtex devices have additional routing resources around their periphery that form an interface between the CLB array and the IOBs. This additional routing, called the VersaRing, facilitates pin-swapping and pin-locking, such that logic redesigns can adapt to existing PCB layouts. Time-to-market is reduced, since PCBs and other system components can be manufactured while the logic design is still in progress.

Dedicated Routing

Some classes of signal require dedicated routing resources to maximize performance. In the Virtex architecture, dedicated routing resources are provided for two classes of signal.

- Horizontal routing resources are provided for on-chip 3-state busses. Four partitionable bus lines are provided per CLB row, permitting multiple busses within a row, as shown in Figure 8.
- Two dedicated nets per CLB propagate carry signals vertically to the adjacent CLB.

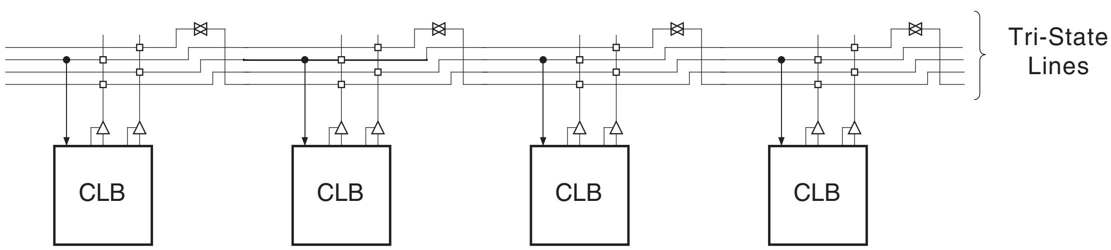

Figure 8: **BUFT Connections to Dedicated Horizontal Bus Lines**

Global Routing

Global Routing resources distribute clocks and other signals with very high fanout throughout the device. Virtex devices include two tiers of global routing resources referred to as primary global and secondary local clock routing resources.

- The primary global routing resources are four dedicated global nets with dedicated input pins that are designed to distribute high-fanout clock signals with minimal skew. Each global clock net can drive all CLB, IOB, and block RAM clock pins. The primary global nets can only be driven by global buffers. There are four global buffers, one for each global net.
- The secondary local clock routing resources consist of 24 backbone lines, 12 across the top of the chip and 12 across bottom. From these lines, up to 12 unique signals per column can be distributed via the 12 longlines in the column. These secondary resources are more flexible than the primary resources since they are not restricted to routing only to clock pins.

Clock Distribution

Virtex provides high-speed, low-skew clock distribution through the primary global routing resources described above. A typical clock distribution net is shown in Figure 9.

Four global buffers are provided, two at the top center of the device and two at the bottom center. These drive the four primary global nets that in turn drive any clock pin.

In addition to the test instructions outlined above, the boundary-scan circuitry can be used to configure the FPGA, and also to read back the configuration data.

Figure 10 is a diagram of the Virtex Series boundary scan logic. It includes three bits of Data Register per IOB, the IEEE 1149.1 Test Access Port controller, and the Instruction Register with decodes.

Instruction Set

The Virtex Series boundary scan instruction set also includes instructions to configure the device and read back configuration data (CFG_IN, CFG_OUT, and JSTART). The complete instruction set is coded as shown in Table 5.

Data Registers

The primary data register is the boundary scan register. For each IOB pin in the FPGA, bonded or not, it includes three bits for In, Out, and 3-State Control. Non-IOB pins have appropriate partial bit population if input-only or output-only. Each EXTEST CAPTURED-OR state captures all In, Out, and 3-state pins.

The other standard data register is the single flip-flop BYPASS register. It synchronizes data being passed through the FPGA to the next downstream boundary scan device.

The FPGA supports up to two additional internal scan chains that can be specified using the BSCAN macro. The macro provides two user pins (SEL1 and SEL2) which are decodes of the USER1 and USER2 instructions respectively. For these instructions, two corresponding pins (TDO1 and TDO2) allow user scan data to be shifted out of TDO.

Likewise, there are individual clock pins (DRCK1 and DRCK2) for each user register. There is a common input pin (TDI) and shared output pins that represent the state of the TAP controller (RESET, SHIFT, and UPDATE).

Bit Sequence

The order within each IOB is: In, Out, 3-State. The input-only pins contribute only the In bit to the boundary scan I/O data register, while the output-only pins contributes all three bits.

From a cavity-up view of the chip (as shown in EPIC), starting in the upper right chip corner, the boundary scan data-register bits are ordered as shown in Figure 11.

BSDL (Boundary Scan Description Language) files for Virtex Series devices are available on the Xilinx web site in the File Download area.

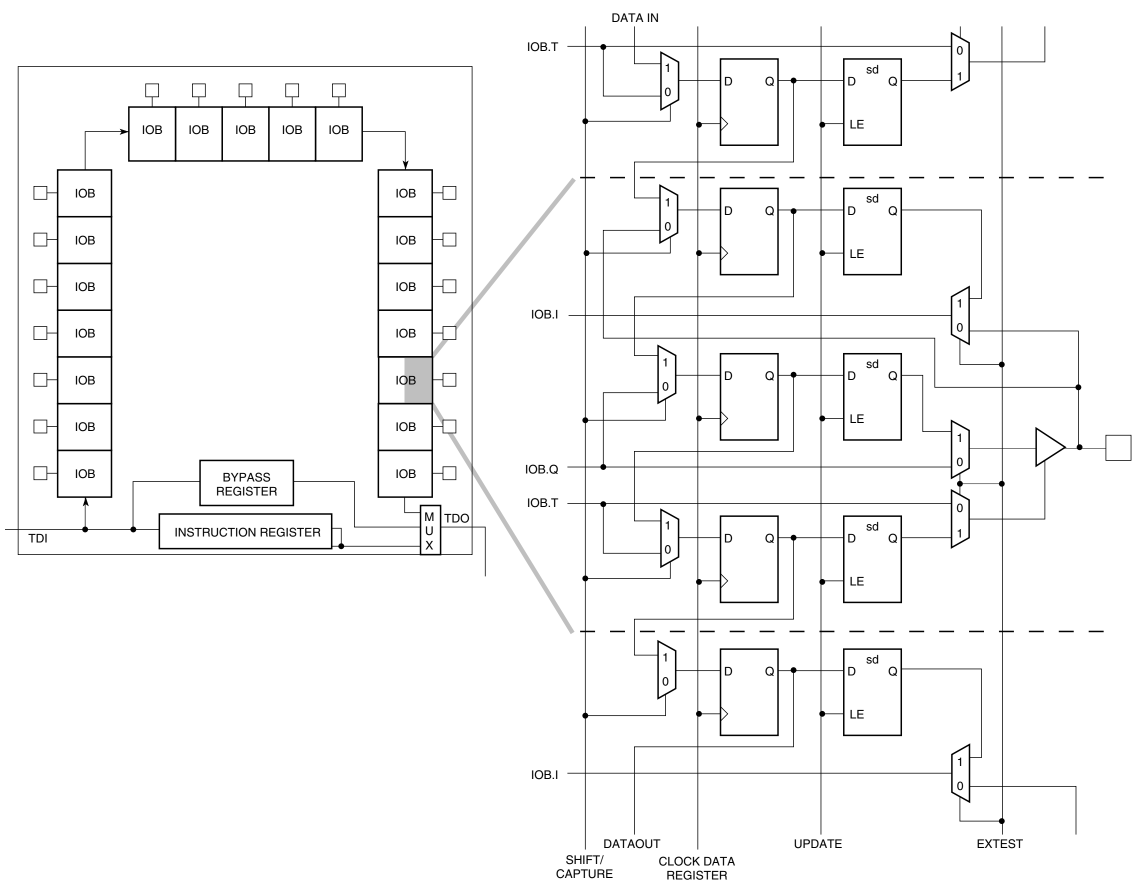

Figure 10: **Virtex Series Boundary Scan Logic**

ers with a common user interface regardless of their choice of entry and verification tools. The XDM software simplifies the selection of implementation options with pull-down menus and on-line help.

Application programs ranging from schematic capture to Placement and Routing (PAR) can be accessed through the XDM software. The program command sequence is generated prior to execution, and stored for documentation.

Several advanced software features facilitate Virtex design. RPMs, for example, are schematic-based macros with relative location constraints to guide their placement. They help ensure optimal implementation of common functions.

For HDL design entry, the Xilinx FPGA Foundation development system provides interfaces to the following synthesis design environments.

- Synopsys (FPGA Compiler, FPGA Express)
- Exemplar (Spectrum)
- Synplicity (Synplify)

For schematic design entry, the Xilinx FPGA Foundation and alliance development system provides interfaces to the following schematic-capture design environments.

- Mentor Graphics V8 (Design Architect, QuickSim II)
- Viewlogic Systems (Viewdraw)

Third-party vendors support many other environments.

A standard interface-file specification, Electronic Design Interchange Format (EDIF), simplifies file transfers into and out of the development system.

Virtex FPGAs supported by a unified library of standard functions. This library contains over 400 primitives and macros, ranging from 2-input AND gates to 16-bit accumulators, and includes arithmetic functions, comparators, counters, data registers, decoders, encoders, I/O functions, latches, Boolean functions, multiplexers, shift registers, and barrel shifters.

The "soft macro" portion of the library contains detailed descriptions of common logic functions, but does not contain any partitioning or placement information. The performance of these macros depends, therefore, on the partitioning and placement obtained during implementation.

RPMs, on the other hand, do contain predetermined partitioning and placement information that permits optimal implementation of these functions. Users can create their own library of soft macros or RPMs based on the macros and primitives in the standard library.

The design environment supports hierarchical design entry, with high-level schematics that comprise major functional blocks, while lower-level schematics define the logic in these blocks. These hierarchical design elements are automatically combined by the implementation tools. Different design entry tools can be combined within a hierarchical design, thus allowing the most convenient entry method to be used for each portion of the design.

Design Implementation

The place-and-route tools (PAR) automatically provide the implementation flow described in this section. The partitioner takes the EDIF net list for the design and maps the logic into the architectural resources of the FPGA (CLBs and IOBs, for example). The placer then determines the best locations for these blocks based on their interconnections and the desired performance. Finally, the router interconnects the blocks.

The PAR algorithms support fully automatic implementation of most designs. For demanding applications, however, the user can exercise various degrees of control over the process. User partitioning, placement, and routing information is optionally specified during the design-entry process. The implementation of highly structured designs can benefit greatly from basic floor planning.

The implementation software incorporates Timing Wizard® timing-driven placement and routing. Designers specify timing requirements along entire paths during design entry. The timing path analysis routines in PAR then recognize these user-specified requirements and accommodate them.

Timing requirements are entered on a schematic in a form directly relating to the system requirements, such as the targeted clock frequency, or the maximum allowable delay between two registers. In this way, the overall performance of the system along entire signal paths is automatically tailored to user-generated specifications. Specific timing information for individual nets is unnecessary.

Design Verification

In addition to conventional software simulation, FPGA users can use in-circuit debugging techniques. Because Xilinx devices are infinitely reprogrammable, designs can be verified in real time without the need for extensive sets of software simulation vectors.

The development system supports both software simulation and in-circuit debugging techniques. For simulation, the system extracts the post-layout timing information from the design database, and back-annotates this information into the net list for use by the simulator. Alternatively, the user can verify timing-critical portions of the design using the TRACE® static timing analyzer.

For in-circuit debugging, the development system includes a download and readback cable. This cable connects the FPGA in the target system to a PC or workstation. After downloading the design into the FPGA, the designer can single-step the logic, readback the contents of the flip-flops, and so observe the internal logic state. Simple modifications can be downloaded into the system in a matter of minutes.

Table 8: **Master/Slave Serial Mode Programming Switching**

	Description	Figure References	Symbol	Values	Units
CCLK	DIN setup/hold, slave mode	1/2	T_{DCC}/T_{CCD}	5.0 / 0	ns, min
	DIN setup/hold, master mode	1/2	T_{DSCK}/T_{CKDS}	5.0 / 0	ns, min
	DOUT	3	T_{CCO}	12.0	ns, max
	High time	4	T_{CCH}	5.0	ns, min
	Low time	5	T_{CCL}	5.0	ns, min
	Maximum Frequency		F_{CC}	66	MHz, max
	Frequency Tolerance, master mode with respect to nominal			+45% –30%	

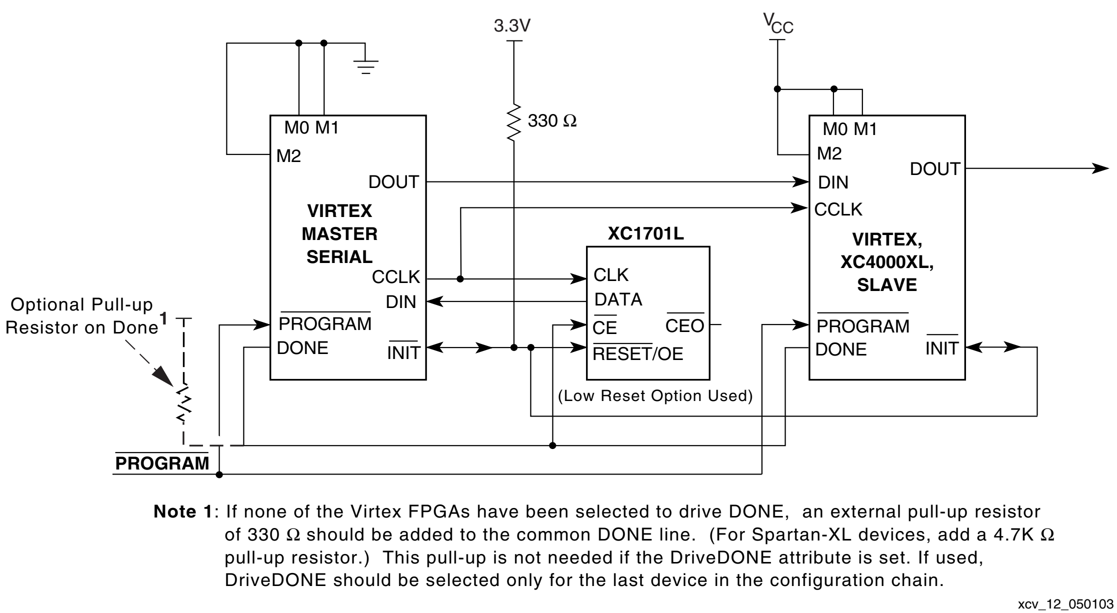

Figure 12: **Master/Slave Serial Mode Circuit Diagram**

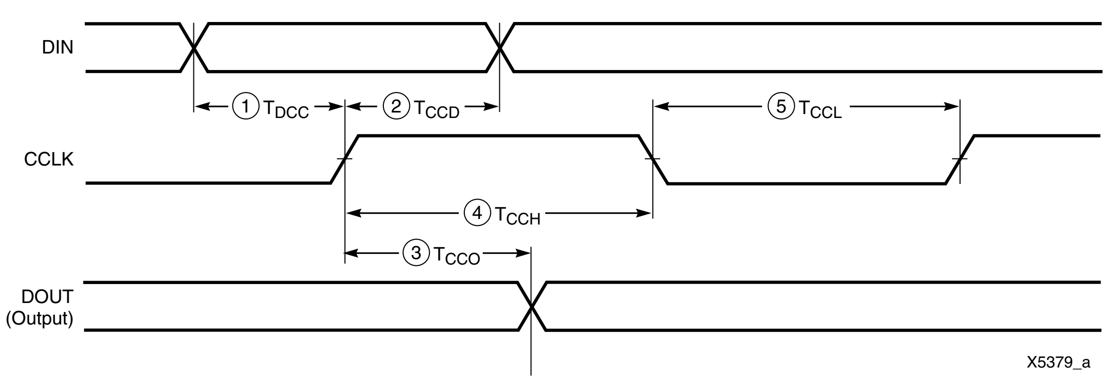

Figure 13: **Slave-Serial Mode Programming Switching Characteristics**

Master-Serial Mode

In master-serial mode, the CCLK output of the FPGA drives a Xilinx Serial PROM that feeds bit-serial data to the DIN input. The FPGA accepts this data on each rising CCLK edge. After the FPGA has been loaded, the data for the next device in a daisy-chain is presented on the DOUT pin after the rising CCLK edge.

The interface is identical to slave-serial except that an internal oscillator is used to generate the configuration clock (CCLK). A wide range of frequencies can be selected for CCLK which always starts at a slow default frequency. Configuration bits then switch CCLK to a higher frequency for the remainder of the configuration. Switching to a lower frequency is prohibited.

The CCLK frequency is set using the ConfigRate option in the bitstream generation software. The maximum CCLK frequency that can be selected is 60 MHz. When selecting a CCLK frequency, ensure that the serial PROM and any daisy-chained FPGAs are fast enough to support the clock rate.

On power-up, the CCLK frequency is 2.5 MHz. This frequency is used until the ConfigRate bits have been loaded when the frequency changes to the selected ConfigRate. Unless a different frequency is specified in the design, the default ConfigRate is 4 MHz.

Figure 12 shows a full master/slave system. In this system, the left-most device operates in master-serial mode. The remaining devices operate in slave-serial mode. The SPROM RESET pin is driven by INIT, and the CE input is driven by DONE. There is the potential for contention on the DONE pin, depending on the start-up sequence options chosen.

Figure 14 shows the timing of master-serial configuration. Master-serial mode is selected by a <000> or <100> on the mode pins (M2, M1, M0). Table 8 shows the timing information for Figure 14.

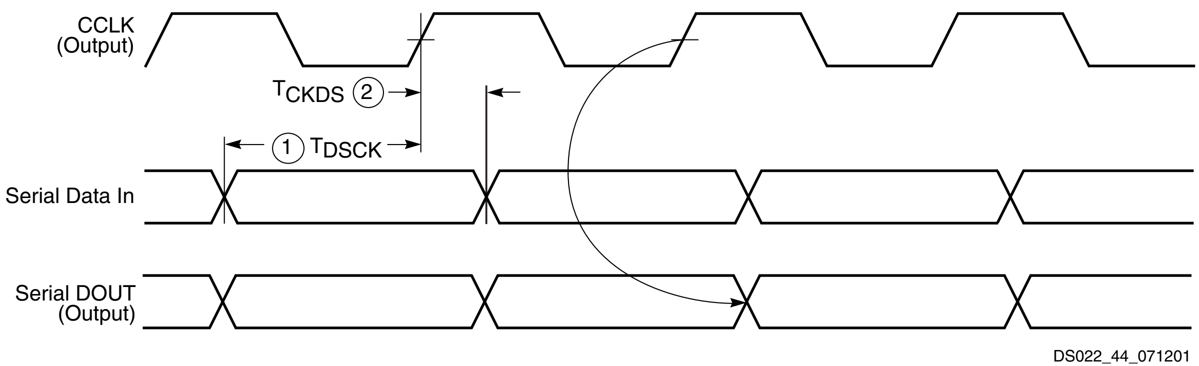

Figure 14: **Master-Serial Mode Programming Switching Characteristics**

At power-up, V_{CC} must rise from 1.0 V to V_{CC} min in less than 50 ms, otherwise delay configuration by pulling PROGRAM Low until V_{CC} is valid.

The sequence of operations necessary to configure a Virtex FPGA serially appears in Figure 15.

SelectMAP Mode

The SelectMAP mode is the fastest configuration option. Byte-wide data is written into the FPGA with a BUSY flag controlling the flow of data.

An external data source provides a byte stream, CCLK, a Chip Select (CS) signal and a Write signal (WRITE). If BUSY is asserted (High) by the FPGA, the data must be held until BUSY goes Low.

Data can also be read using the SelectMAP mode. If WRITE is not asserted, configuration data is read out of the FPGA as part of a readback operation.

In the SelectMAP mode, multiple Virtex devices can be chained in parallel. DATA pins (D7:D0), CCLK, WRITE, BUSY, PROGRAM, DONE, and INIT can be connected in parallel between all the FPGAs. Note that the data is organized with the MSB of each byte on pin DO and the LSB of each byte on D7. The CS pins are kept separate, insuring that each FPGA can be selected individually. WRITE should be Low before loading the first bitstream and returned High after the last device has been programmed. Use CS to select the appropriate FPGA for loading the bitstream and sending the configuration data. at the end of the bitstream, deselect the loaded device and select the next target FPGA by setting its CS pin High. A free-running oscillator or other externally generated signal can be used for CCLK. The BUSY signal can be ignored for frequencies below 50 MHz. For details about frequencies above 50 MHz, see XAPP138, Virtex Configuration and Readback. Once all the devices have been programmed, the DONE pin goes High.

Block RAM Switching Characteristics

Description	Symbol	Speed Grade				Units
		Min	-6	-5	-4	
Sequential Delays						
Clock CLK to DOUT output	T_{BCKO}	1.7	3.4	3.8	4.3	ns, max
Setup and Hold Times before/after Clock CLK(1)	Setup Time / Hold Time					
ADDR inputs	T_{BACK}/T_{BCKA}	0.6 / 0	1.2 / 0	1.3 / 0	1.5 / 0	ns, min
DIN inputs	T_{BDCK}/T_{BCKD}	0.6 / 0	1.2 / 0	1.3 / 0	1.5 / 0	ns, min
EN input	T_{BECK}/T_{BCKE}	1.3 / 0	2.6 / 0	3.0 / 0	3.4 / 0	ns, min
RST input	T_{BRCK}/T_{BCKR}	1.3 / 0	2.5 / 0	2.7 / 0	3.2 / 0	ns, min
WEN input	T_{BWCK}/T_{BCKW}	1.2 / 0	2.3 / 0	2.6 / 0	3.0 / 0	ns, min
Clock CLK						
Minimum Pulse Width, High	T_{BPWH}	0.8	1.5	1.7	2.0	ns, min
Minimum Pulse Width, Low	T_{BPWL}	0.8	1.5	1.7	2.0	ns, min
CLKA -> CLKB setup time for different ports	T_{BCCS}		3.0	3.5	4.0	ns, min

Notes:
1. A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.

TBUF Switching Characteristics

Description	Symbol	Speed Grade				Units
		Min	-6	-5	-4	
Combinatorial Delays						
IN input to OUT output	T_{IO}	0	0	0	0	ns, max
TRI input to OUT output high-impedance	T_{OFF}	0.05	0.09	0.10	0.11	ns, max
TRI input to valid data on OUT output	T_{ON}	0.05	0.09	0.10	0.11	ns, max

JTAG Test Access Port Switching Characteristics

Description	Symbol	Speed Grade			Units
		-6	-5	-4	
TMS and TDI Setup times before TCK	T_{TAPTCK}	4.0	4.0	4.0	ns, min
TMS and TDI Hold times after TCK	T_{TCKTAP}	2.0	2.0	2.0	ns, min
Output delay from clock TCK to output TDO	T_{TCKTDO}	11.0	11.0	11.0	ns, max
Maximum TCK clock frequency	F_{TCK}	33	33	33	MHz, max

Minimum Clock-to-Out for Virtex Devices

I/O Standard	With DLL	Without DLL									
	All Devices	V50	V100	V150	V200	V300	V400	V600	V800	V1000	Units
*LVTTL_S2	5.2	6.0	6.0	6.0	6.0	6.1	6.1	6.1	6.1	6.1	ns
*LVTTL_S4	3.5	4.3	4.3	4.3	4.3	4.4	4.4	4.4	4.4	4.4	ns
*LVTTL_S6	2.8	3.6	3.6	3.6	3.6	3.7	3.7	3.7	3.7	3.7	ns
*LVTTL_S8	2.2	3.1	3.1	3.1	3.1	3.1	3.1	3.2	3.2	3.2	ns
*LVTTL_S12	2.0	2.9	2.9	2.9	2.9	2.9	2.9	3.0	3.0	3.0	ns
*LVTTL_S16	1.9	2.8	2.8	2.8	2.8	2.8	2.8	2.9	2.9	2.9	ns
*LVTTL_S24	1.8	2.6	2.6	2.7	2.7	2.7	2.7	2.7	2.7	2.8	ns
*LVTTL_F2	2.9	3.8	3.8	3.8	3.8	3.8	3.8	3.9	3.9	3.9	ns
*LVTTL_F4	1.7	2.6	2.6	2.6	2.6	2.6	2.6	2.7	2.7	2.7	ns
*LVTTL_F6	1.2	2.0	2.0	2.0	2.1	2.1	2.1	2.1	2.1	2.2	ns
*LVTTL_F8	1.1	1.9	1.9	1.9	1.9	2.0	2.0	2.0	2.0	2.0	ns
*LVTTL_F12	1.0	1.8	1.8	1.8	1.8	1.9	1.9	1.9	1.9	1.9	ns
*LVTTL_F16	0.9	1.7	1.8	1.8	1.8	1.8	1.8	1.8	1.9	1.9	ns
*LVTTL_F24	0.9	1.7	1.7	1.7	1.8	1.8	1.8	1.8	1.8	1.9	ns
LVCMOS2	1.1	1.9	1.9	1.9	2.0	2.0	2.0	2.0	2.0	2.1	ns
PCI33_3	1.5	2.4	2.4	2.4	2.4	2.4	2.4	2.5	2.5	2.5	ns
PCI33_5	1.4	2.2	2.2	2.3	2.3	2.3	2.3	2.3	2.3	2.4	ns
PCI66_3	1.1	1.9	1.9	2.0	2.0	2.0	2.0	2.0	2.1	2.1	ns
GTL	1.6	2.5	2.5	2.5	2.5	2.5	2.5	2.6	2.6	2.6	ns
GTL+	1.7	2.5	2.5	2.6	2.6	2.6	2.6	2.6	2.6	2.7	ns
HSTL I	1.1	1.9	1.9	1.9	1.9	2.0	2.0	2.0	2.0	2.0	ns
HSTL III	0.9	1.7	1.7	1.8	1.8	1.8	1.8	1.8	1.8	1.9	ns
HSTL IV	0.8	1.6	1.6	1.6	1.7	1.7	1.7	1.7	1.7	1.8	ns
SSTL2 I	0.9	1.7	1.7	1.7	1.7	1.8	1.8	1.8	1.8	1.8	ns
SSTL2 II	0.8	1.6	1.6	1.6	1.6	1.7	1.7	1.7	1.7	1.7	ns
SSTL3 I	0.8	1.6	1.7	1.7	1.7	1.7	1.7	1.7	1.8	1.8	ns
SSTL3 II	0.7	1.5	1.5	1.6	1.6	1.6	1.6	1.6	1.6	1.7	ns
CTT	1.0	1.8	1.8	1.8	1.9	1.9	1.9	1.9	1.9	2.0	ns
AGP	1.0	1.8	1.8	1.9	1.9	1.9	1.9	1.9	1.9	2.0	ns

*S = Slow Slew Rate, F = Fast Slew Rate

Notes:
1. Listed above are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.
2. Input and output timing is measured at 1.4 V for LVTTL. For other I/O standards, see Table 3. In all cases, an 8 pF external capacitive load is used.

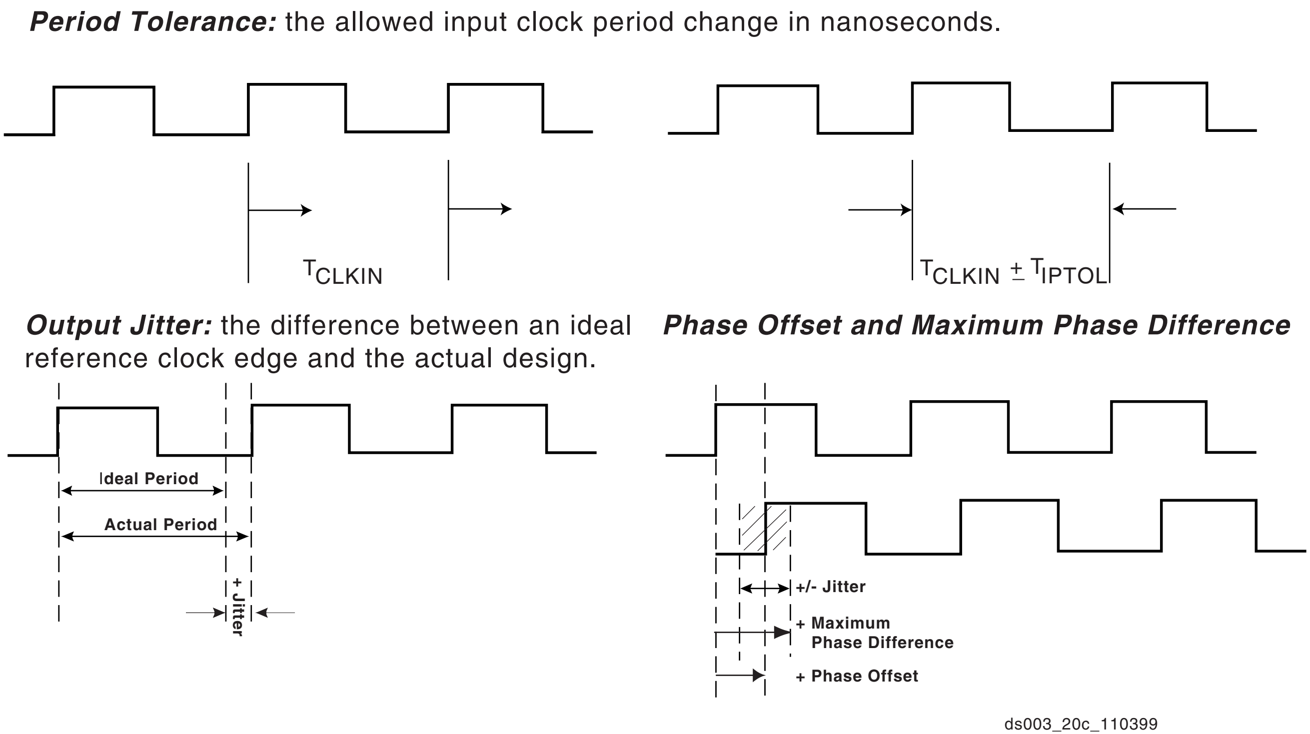

Figure 1: **Frequency Tolerance and Clock Jitter**

Revision History

Date	Version	Revision
11/98	1.0	Initial Xilinx release.
01/99	1.2	Updated package drawings and specs.
02/99	1.3	Update of package drawings, updated specifications.
05/99	1.4	Addition of package drawings and specifications.
05/99	1.5	Replaced FG 676 & FG680 package drawings.
07/99	1.6	Changed Boundary Scan Information and changed Figure 11, Boundary Scan Bit Sequence. Updated IOB Input & Output delays. Added Capacitance info for different I/O Standards. Added 5 V tolerant information. Added DLL Parameters and waveforms and new Pin-to-pin Input and Output Parameter tables for Global Clock Input to Output and Setup and Hold. Changed Configuration Information including Figures 12, 14, 17 & 19. Added device-dependent listings for quiescent currents ICCINTQ and ICCOQ. Updated IOB Input and Output Delays based on default standard of LVTTL, 12 mA, Fast Slew Rate. Added IOB Input Switching Characteristics Standard Adjustments.
09/99	1.7	Speed grade update to preliminary status, Power-on specification and Clock-to-Out Minimums additions, "0" hold time listing explanation, quiescent current listing update, and Figure 6 ADDRA input label correction. Added T_{IJITCC} parameter, changed T_{OJIT} to T_{OPHASE}.
01/00	1.8	Update to speed.txt file 1.96. Corrections for CRs 111036,111137, 112697, 115479, 117153, 117154, and 117612. Modified notes for Recommended Operating Conditions (voltage and temperature). Changed Bank information for V_{CCO} in CS144 package on p.43.

Virtex™ 2.5 V Field Programmable Gate Arrays

DS003-4 (v4.0) March 1, 2013 **Production Product Specification**

Virtex Pin Definitions

Table 1: **Special Purpose Pins**

Pin Name	Dedicated Pin	Direction	Description
GCK0, GCK1, GCK2, GCK3	Yes	Input	Clock input pins that connect to Global Clock Buffers. These pins become user inputs when not needed for clocks.
M0, M1, M2	Yes	Input	Mode pins are used to specify the configuration mode.
CCLK	Yes	Input or Output	The configuration Clock I/O pin: it is an input for SelectMAP and slave-serial modes, and output in master-serial mode. After configuration, it is input only, logic level = Don't Care.
PROGRAM	Yes	Input	Initiates a configuration sequence when asserted Low.
DONE	Yes	Bidirectional	Indicates that configuration loading is complete, and that the start-up sequence is in progress. The output can be open drain.
INIT	No	Bidirectional (Open-drain)	When Low, indicates that the configuration memory is being cleared. The pin becomes a user I/O after configuration.
BUSY/ DOUT	No	Output	In SelectMAP mode, BUSY controls the rate at which configuration data is loaded. The pin becomes a user I/O after configuration unless the SelectMAP port is retained. In bit-serial modes, DOUT provides header information to downstream devices in a daisy-chain. The pin becomes a user I/O after configuration.
D0/DIN, D1, D2, D3, D4, D5, D6, D7	No	Input or Output	In SelectMAP mode, D0 - D7 are configuration data pins. These pins become user I/Os after configuration unless the SelectMAP port is retained. In bit-serial modes, DIN is the single data input. This pin becomes a user I/O after configuration.
WRITE	No	Input	In SelectMAP mode, the active-low Write Enable signal. The pin becomes a user I/O after configuration unless the SelectMAP port is retained.
CS	No	Input	In SelectMAP mode, the active-low Chip Select signal. The pin becomes a user I/O after configuration unless the SelectMAP port is retained.
TDI, TDO, TMS, TCK	Yes	Mixed	Boundary-scan Test-Access-Port pins, as defined in IEEE 1149.1.
DXN, DXP	Yes	N/A	Temperature-sensing diode pins. (Anode: DXP, cathode: DXN)
V_{CCINT}	Yes	Input	Power-supply pins for the internal core logic.
V_{CCO}	Yes	Input	Power-supply pins for the output drivers (subject to banking rules)
V_{REF}	No	Input	Input threshold voltage pins. Become user I/Os when an external threshold voltage is not needed (subject to banking rules).
GND	Yes	Input	Ground

Table 3: **Virtex Pinout Tables (BGA)** ***(Continued)***

Pin Name	Device	BG256	BG352	BG432	BG560
V_{CCO}, Bank 7	All	G4, H4	G23, K26, N23	A31, L28, L31	C32, D33, K33, N32, T33
V_{REF}, Bank 0 (VREF pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	A8, B4	N/A	N/A	N/A
	XCV100/150	... + A4	A16,C19, C21	N/A	N/A
	XCV200/300	... + A2	... + D21	B19, D22, D24, D26	N/A
	XCV400	N/A	N/A	... + C18	A19, D20, D26, E23, E27
	XCV600	N/A	N/A	... + C24	... + E24
	XCV800	N/A	N/A	... + B21	... + E21
	XCV1000	N/A	N/A	N/A	... + D29
V_{REF}, Bank 1 (VREF pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	A17, B12	N/A	N/A	N/A
	XCV100/150	... + B15	B6, C9, C12	N/A	N/A
	XCV200/300	... + B17	... + D6	A13, B7, C6, C10	N/A
	XCV400	N/A	N/A	... + B15	A6, D7, D11, D16, E15
	XCV600	N/A	N/A	... + D10	... + D10
	XCV800	N/A	N/A	... + B12	... + D13
	XCV1000	N/A	N/A	N/A	... + E7
V_{REF}, Bank 2 (V_{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	C20, J18	N/A	N/A	N/A
	XCV100/150	... + F19	E2, H2, M4	N/A	N/A
	XCV200/300	... + G18	... + D2	E2, G3, J2, N1	N/A
	XCV400	N/A	N/A	... + R3	G5, H4, L5, P4, R1
	XCV600	N/A	N/A	... + H1	... + K5
	XCV800	N/A	N/A	... + M3	... + N5
	XCV1000	N/A	N/A	N/A	... + B3

Table 4: **Virtex Pinout Tables (Fine-Pitch BGA)** ***(Continued)***

Pin Name	Device	FG256	FG456	FG676	FG680
V_{REF}, Bank 1 (VREF pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	B9, C11	N/A	N/A	N/A
	XCV100/150	... + E11	A18, B13, E14	N/A	N/A
	XCV200/300	... + A14	... + A19	N/A	N/A
	XCV400	N/A	N/A	A14, C20, C21, D15, G16	N/A
	XCV600	N/A	N/A	... + B19	B6, B8, B18, D11, D13, D17
	XCV800	N/A	N/A	... + A17	... + B14
	XCV1000	N/A	N/A	N/A	... + B5
V_{REF}, Bank 2 (V_{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	F13, H13	N/A	N/A	N/A
	XCV100/150	... + F14	F21, H18, K21	N/A	N/A
	XCV200/300	... + E13	... + D22	N/A	N/A
	XCV400	N/A	N/A	F24, H23, K20, M23, M26	N/A
	XCV600	N/A	N/A	... + G26	G1, H4, J1, L2, V5, W3
	XCV800	N/A	N/A	... + K25	... + N1
	XCV1000	N/A	N/A	N/A	... + D2
V_{REF}, Bank 3 (V_{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	K16, L14	N/A	N/A	N/A
	XCV100/150	... + L13	N21, R19, U21	N/A	N/A
	XCV200/300	... + M13	... + U20	N/A	N/A
	XCV400	N/A	N/A	R23, R25, U21, W22, W23	N/A
	XCV600	N/A	N/A	... + W26	AC1, AJ2, AK3, AL4, AR1, Y1
	XCV800	N/A	N/A	... + U25	... + AF3
	XCV1000	N/A	N/A	N/A	... + AP4

Table 4: **Virtex Pinout Tables (Fine-Pitch BGA) *(Continued)***

Pin Name	Device	FG256	FG456	FG676	FG680
V_{REF}, Bank 4 (V_{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	P9, T12	N/A	N/A	N/A
	XCV100/150	... + T11	AA13, AB16, AB19	N/A	N/A
	XCV200/300	... + R13	... + AB20	N/A	N/A
	XCV400	N/A	N/A	AC15, AD18, AD21, AD22, AF15	N/A
	XCV600	N/A	N/A	... + AF20	AT19, AU7, AU17, AV8, AV10, AW11
	XCV800	N/A	N/A	... + AF17	... + AV14
	XCV1000	N/A	N/A	N/A	... + AU6
V_{REF}, Bank 5 (V_{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	T4, P8	N/A	N/A	N/A
	XCV100/150	... + R5	W8, Y10, AA5	N/A	N/A
	XCV200/300	... + T2	... + Y6	N/A	N/A
	XCV400	N/A	N/A	AA10, AB8, AB12, AC7, AF12	N/A
	XCV600	N/A	N/A	... + AF8	AT27, AU29, AU31, AV35, AW21, AW23
	XCV800	N/A	N/A	... + AE10	... + AT25
	XCV1000	N/A	N/A	N/A	... + AV36
V_{REF}, Bank 6 (V_{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	J3, N1	N/A	N/A	N/A
	XCV100/150	... + M1	N2, R4, T3	N/A	N/A
	XCV200/300	... + N2	... + Y1	N/A	N/A
	XCV400	N/A	N/A	AB3, R1, R4, U6, V5	N/A
	XCV600	N/A	N/A	... + Y1	AB35, AD37, AH39, AK39, AM39, AN36
	XCV800	N/A	N/A	... + U2	... + AE39
	XCV1000	N/A	N/A	N/A	... + AT39

TQ144 Pin Function Diagram

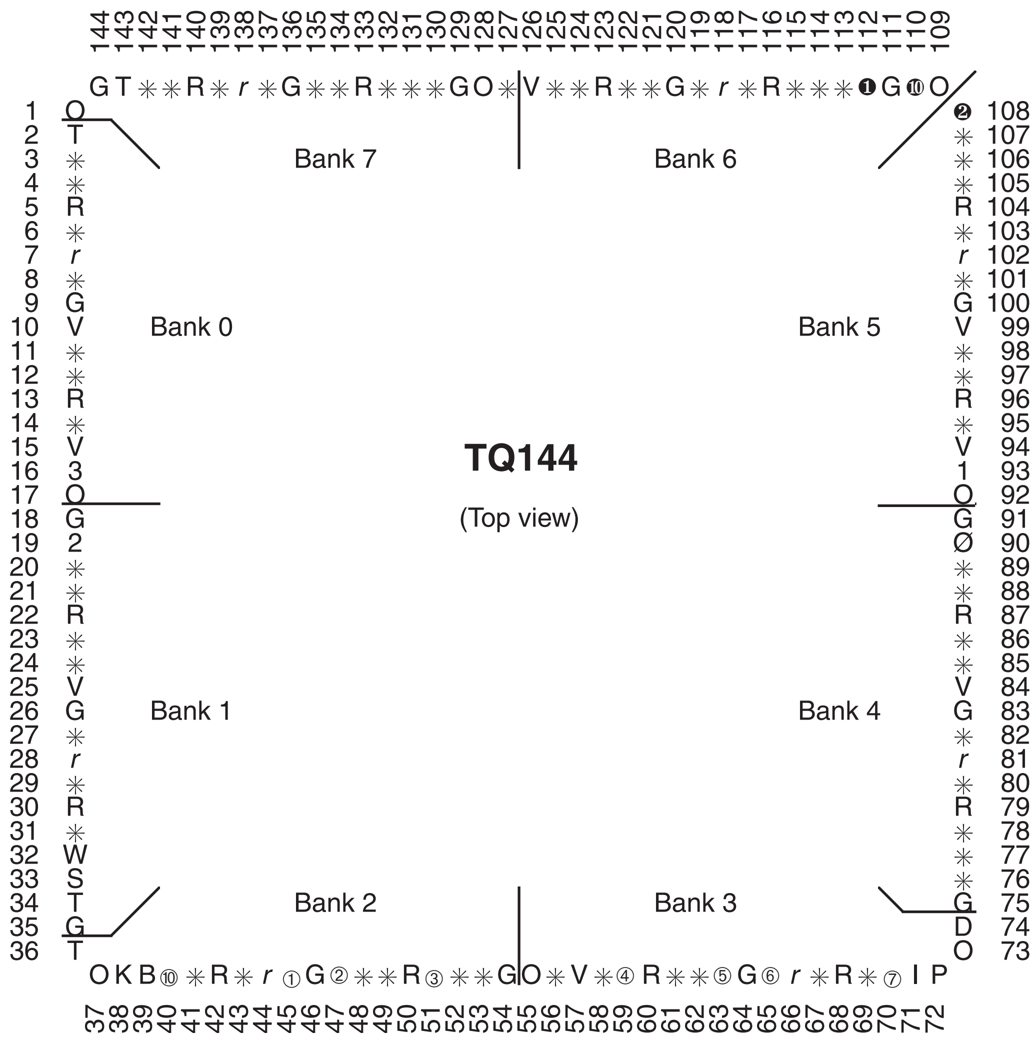

Figure 2: **TQ144 Pin Function Diagram**

BG432 Pin Function Diagram

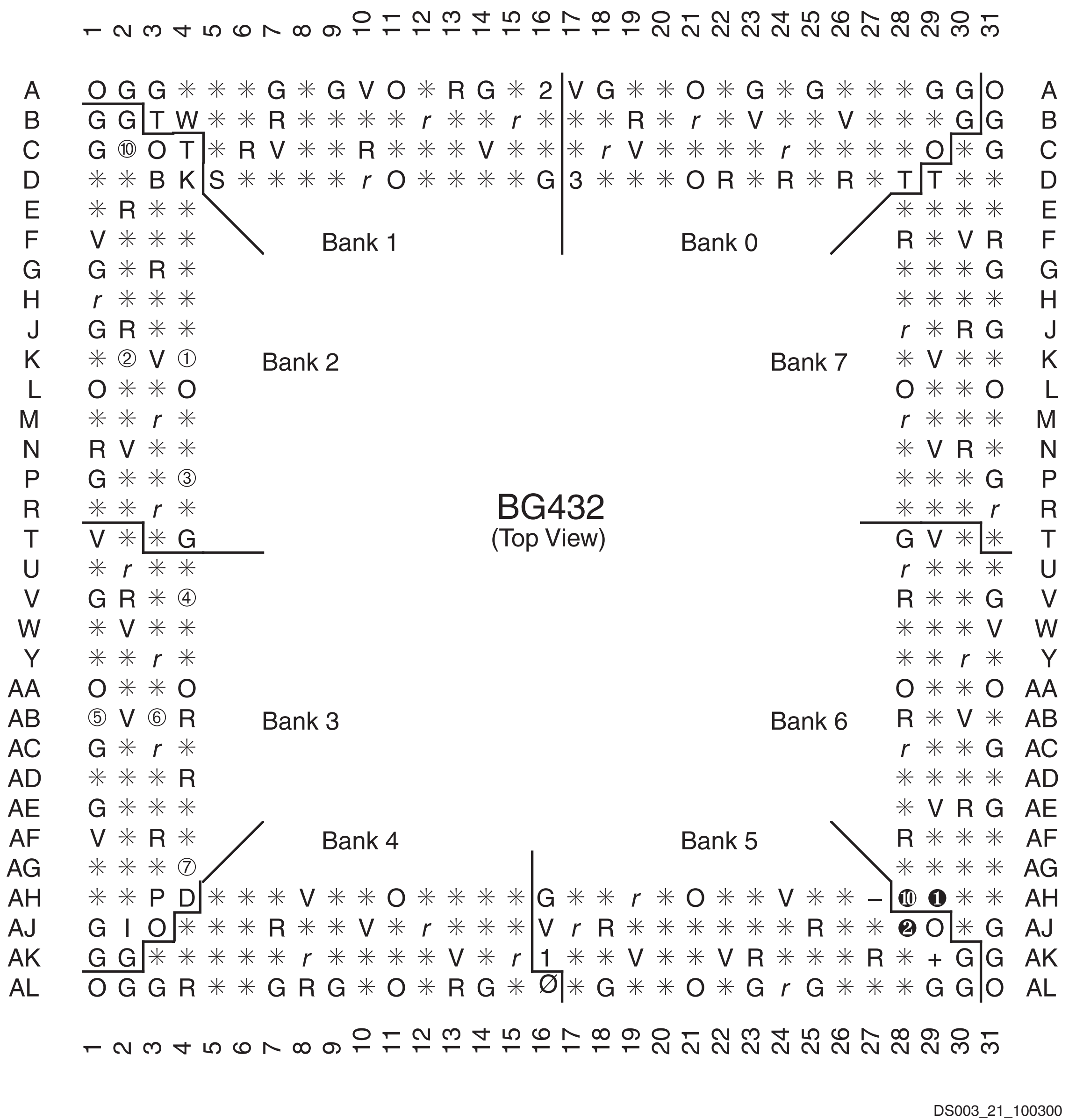

Figure 6: **BG432 Pin Function Diagram**