



Welcome to [E-XFL.COM](https://www.e-xfl.com)

Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

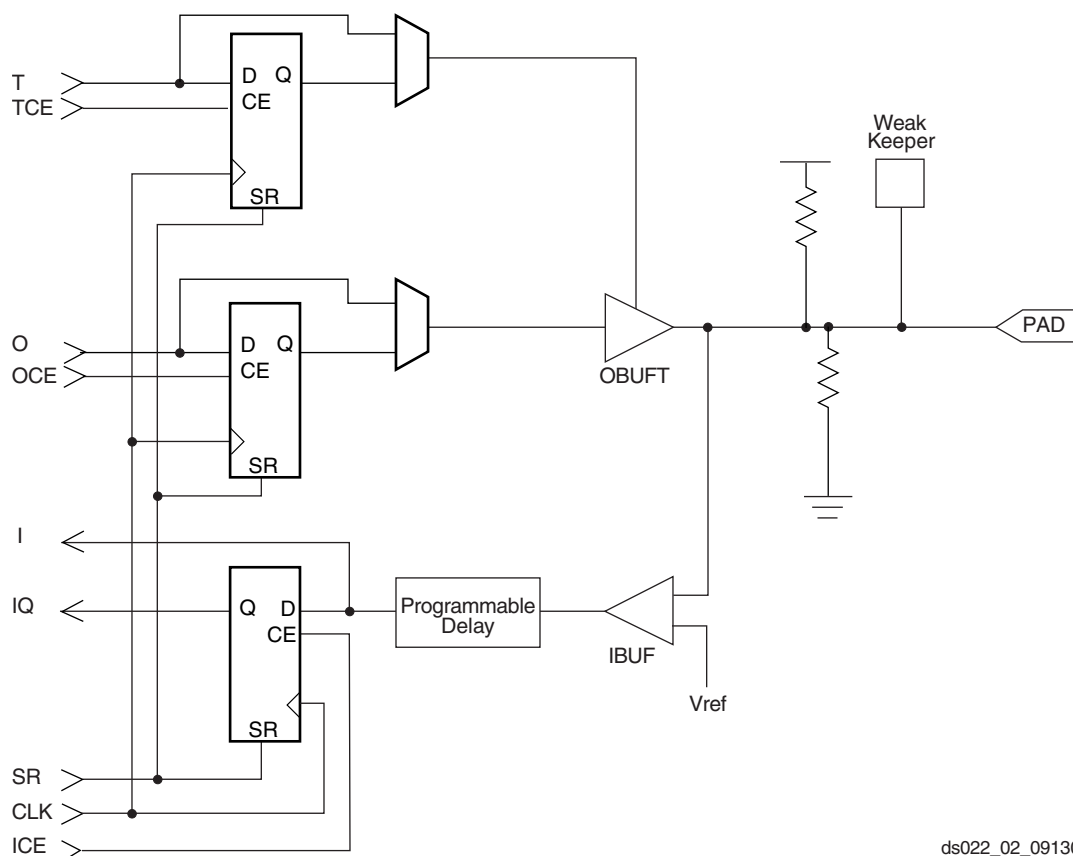
Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	4704
Number of Logic Elements/Cells	21168
Total RAM Bits	114688
Number of I/O	316
Number of Gates	888439
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	432-LBGA Exposed Pad, Metal
Supplier Device Package	432-MBGA (40x40)
Purchase URL	https://www.e-xfl.com/product-detail/xilinx/xcv800-5bg432i



ds022_02_091300

Figure 2: Virtex Input/Output Block (IOB)

Table 1: Supported Select I/O Standards

I/O Standard	Input Reference Voltage (V_{REF})	Output Source Voltage (V_{CCO})	Board Termination Voltage (V_{TT})	5 V Tolerant
LVTTL 2 – 24 mA	N/A	3.3	N/A	Yes
LVC MOS2	N/A	2.5	N/A	Yes
PCI, 5 V	N/A	3.3	N/A	Yes
PCI, 3.3 V	N/A	3.3	N/A	No
GTL	0.8	N/A	1.2	No
GTL+	1.0	N/A	1.5	No
HSTL Class I	0.75	1.5	0.75	No
HSTL Class III	0.9	1.5	1.5	No
HSTL Class IV	0.9	1.5	1.5	No
SSTL3 Class I & II	1.5	3.3	1.5	No
SSTL2 Class I & II	1.25	2.5	1.25	No
CTT	1.5	3.3	1.5	No
AGP	1.32	3.3	N/A	No

more I/O pins convert to V_{REF} pins. Since these are always a superset of the V_{REF} pins used for smaller devices, it is possible to design a PCB that permits migration to a larger device if necessary. All the V_{REF} pins for the largest device anticipated must be connected to the V_{REF} voltage, and not used for I/O.

In smaller devices, some V_{CCO} pins used in larger devices do not connect within the package. These unconnected pins can be left unconnected externally, or can be connected to the V_{CCO} voltage to permit migration to a larger device if necessary.

In TQ144 and PQ/HQ240 packages, all V_{CCO} pins are bonded together internally, and consequently the same V_{CCO} voltage must be connected to all of them. In the CS144 package, bank pairs that share a side are interconnected internally, permitting four choices for V_{CCO} . In both cases, the V_{REF} pins remain internally connected as eight banks, and can be used as described previously.

Configurable Logic Block

The basic building block of the Virtex CLB is the logic cell (LC). An LC includes a 4-input function generator, carry logic, and a storage element. The output from the function generator in each LC drives both the CLB output and the D input of the flip-flop. Each Virtex CLB contains four LCs, organized in two similar slices, as shown in Figure 4.

Figure 5 shows a more detailed view of a single slice.

In addition to the four basic LCs, the Virtex CLB contains logic that combines function generators to provide functions

of five or six inputs. Consequently, when estimating the number of system gates provided by a given device, each CLB counts as 4.5 LCs.

Look-Up Tables

Virtex function generators are implemented as 4-input look-up tables (LUTs). In addition to operating as a function generator, each LUT can provide a 16 x 1-bit synchronous RAM. Furthermore, the two LUTs within a slice can be combined to create a 16 x 2-bit or 32 x 1-bit synchronous RAM, or a 16x1-bit dual-port synchronous RAM.

The Virtex LUT can also provide a 16-bit shift register that is ideal for capturing high-speed or burst-mode data. This mode can also be used to store data in applications such as Digital Signal Processing.

Storage Elements

The storage elements in the Virtex slice can be configured either as edge-triggered D-type flip-flops or as level-sensitive latches. The D inputs can be driven either by the function generators within the slice or directly from slice inputs, bypassing the function generators.

In addition to Clock and Clock Enable signals, each Slice has synchronous set and reset signals (SR and BY). SR forces a storage element into the initialization state specified for it in the configuration. BY forces it into the opposite state. Alternatively, these signals can be configured to operate asynchronously. All of the control signals are independently invertible, and are shared by the two flip-flops within the slice.

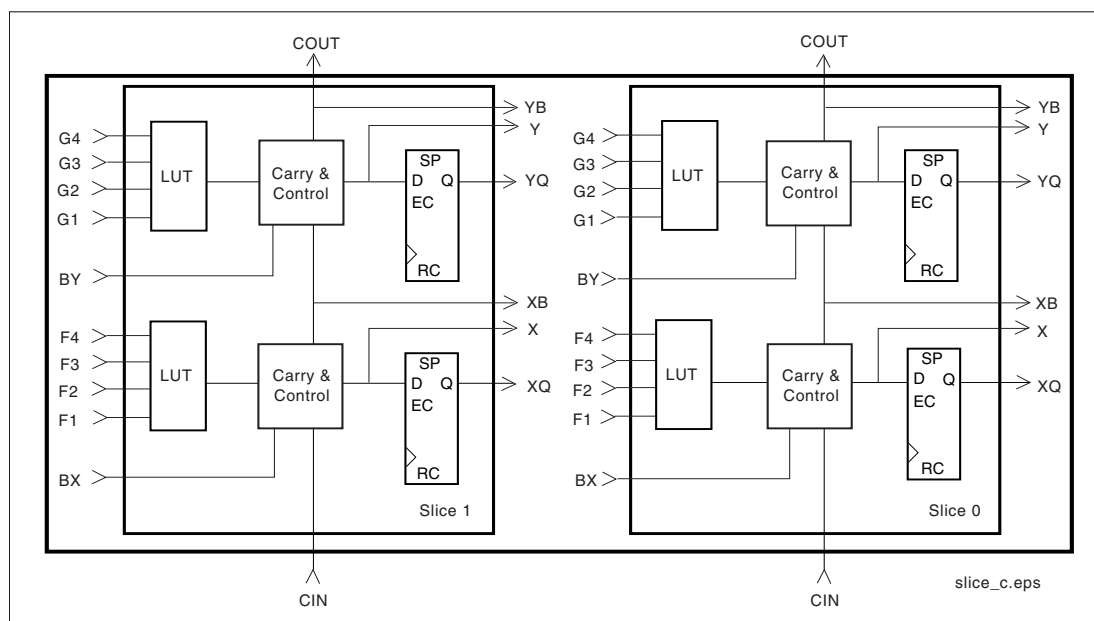


Figure 4: 2-Slice Virtex CLB

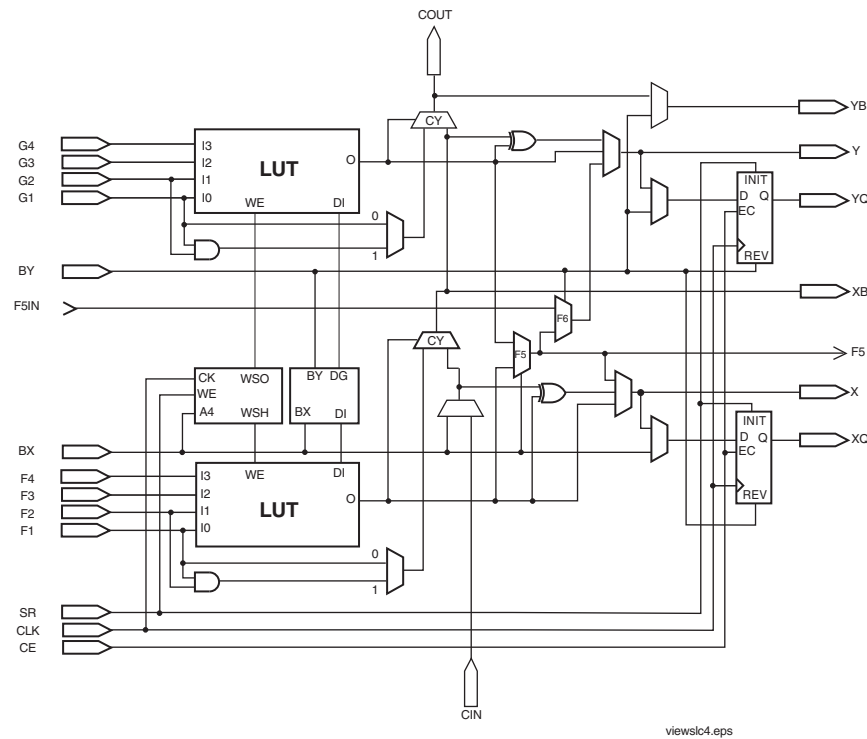


Figure 5: Detailed View of Virtex Slice

Additional Logic

The F5 multiplexer in each slice combines the function generator outputs. This combination provides either a function generator that can implement any 5-input function, a 4:1 multiplexer, or selected functions of up to nine inputs.

Similarly, the F6 multiplexer combines the outputs of all four function generators in the CLB by selecting one of the F5-multiplexer outputs. This permits the implementation of any 6-input function, an 8:1 multiplexer, or selected functions of up to 19 inputs.

Each CLB has four direct feedthrough paths, one per LC. These paths provide extra data input or additional local routing that does not consume logic resources.

Arithmetic Logic

Dedicated carry logic provides fast arithmetic carry capability for high-speed arithmetic functions. The Virtex CLB supports two separate carry chains, one per Slice. The height of the carry chains is two bits per CLB.

The arithmetic logic includes an XOR gate that allows a 1-bit full adder to be implemented within an LC. In addition, a dedicated AND gate improves the efficiency of multiplier implementation.

The dedicated carry path can also be used to cascade function generators for implementing wide logic functions.

BUFTs

Each Virtex CLB contains two 3-state drivers (BUFTs) that can drive on-chip busses. See **Dedicated Routing**, page 7. Each Virtex BUFT has an independent 3-state control pin and an independent input pin.

Block SelectRAM

Virtex FPGAs incorporate several large block SelectRAM memories. These complement the distributed LUT SelectRAMs that provide shallow RAM structures implemented in CLBs.

Block SelectRAM memory blocks are organized in columns. All Virtex devices contain two such columns, one along each vertical edge. These columns extend the full height of the chip. Each memory block is four CLBs high, and consequently, a Virtex device 64 CLBs high contains 16 memory blocks per column, and a total of 32 blocks.

Table 3 shows the amount of block SelectRAM memory that is available in each Virtex device.

Table 3: Virtex Block SelectRAM Amounts

Device	# of Blocks	Total Block SelectRAM Bits
XCV50	8	32,768
XCV100	10	40,960
XCV150	12	49,152
XCV200	14	57,344
XCV300	16	65,536
XCV400	20	81,920
XCV600	24	98,304
XCV800	28	114,688
XCV1000	32	131,072

Each block SelectRAM cell, as illustrated in Figure 6, is a fully synchronous dual-ported 4096-bit RAM with independent control signals for each port. The data widths of the two ports can be configured independently, providing built-in bus-width conversion.

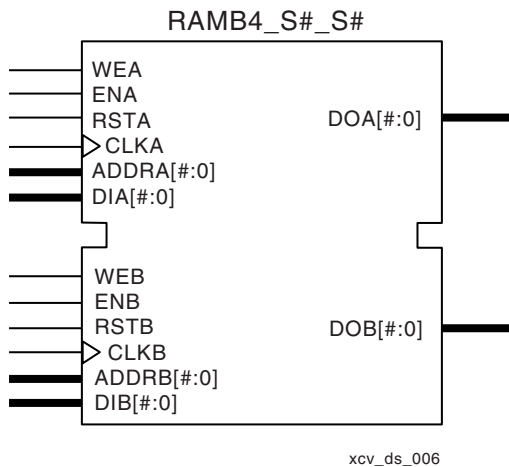


Figure 6: Dual-Port Block SelectRAM

Table 4 shows the depth and width aspect ratios for the block SelectRAM.

Table 4: Block SelectRAM Port Aspect Ratios

Width	Depth	ADDR Bus	Data Bus
1	4096	ADDR<11:0>	DATA<0>
2	2048	ADDR<10:0>	DATA<1:0>
4	1024	ADDR<9:0>	DATA<3:0>
8	512	ADDR<8:0>	DATA<7:0>
16	256	ADDR<7:0>	DATA<15:0>

The Virtex block SelectRAM also includes dedicated routing to provide an efficient interface with both CLBs and other block SelectRAMs. Refer to XAPP130 for block SelectRAM timing waveforms.

Programmable Routing Matrix

It is the longest delay path that limits the speed of any worst-case design. Consequently, the Virtex routing architecture and its place-and-route software were defined in a single optimization process. This joint optimization minimizes long-path delays, and consequently, yields the best system performance.

The joint optimization also reduces design compilation times because the architecture is software-friendly. Design cycles are correspondingly reduced due to shorter design iteration times.

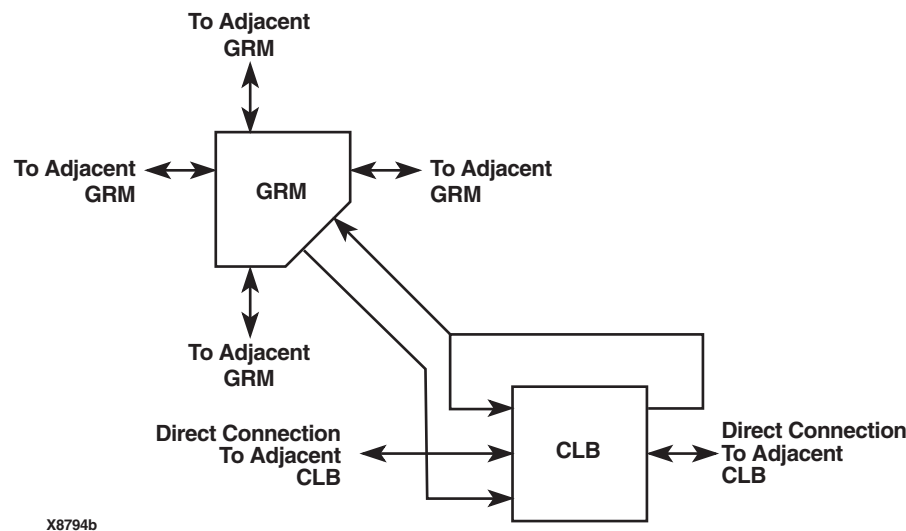


Figure 7: Virtex Local Routing

Local Routing

The VersaBlock provides local routing resources, as shown in Figure 7, providing the following three types of connections.

- Interconnections among the LUTs, flip-flops, and GRM

- Internal CLB feedback paths that provide high-speed connections to LUTs within the same CLB, chaining them together with minimal routing delay
- Direct paths that provide high-speed connections between horizontally adjacent CLBs, eliminating the delay of the GRM.

Four dedicated clock pads are provided, one adjacent to each of the global buffers. The input to the global buffer is

selected either from these pads or from signals in the general purpose routing.

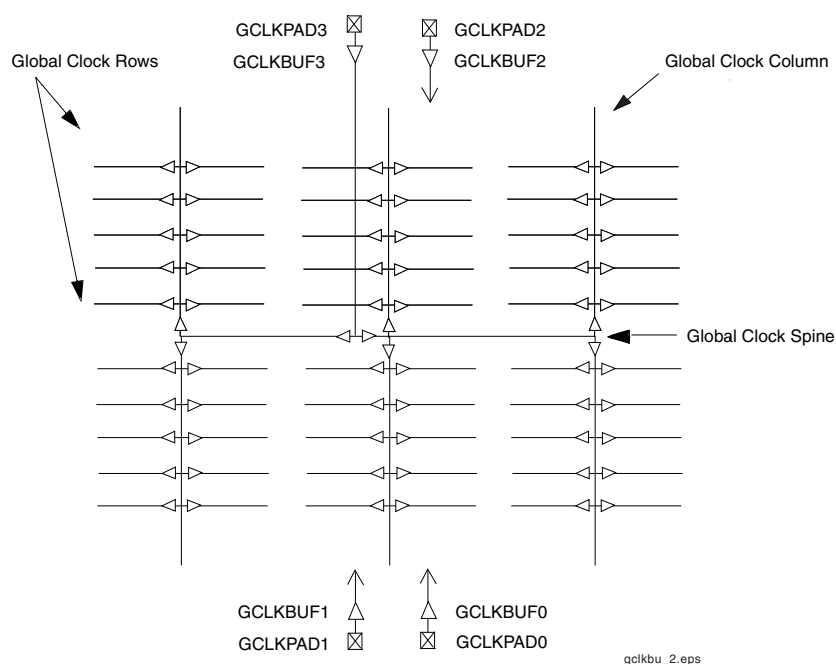


Figure 9: Global Clock Distribution Network

Delay-Locked Loop (DLL)

Associated with each global clock input buffer is a fully digital Delay-Locked Loop (DLL) that can eliminate skew between the clock input pad and internal clock-input pins throughout the device. Each DLL can drive two global clock networks. The DLL monitors the input clock and the distributed clock, and automatically adjusts a clock delay element. Clock edges reach internal flip-flops one to four clock periods after they arrive at the input. This closed-loop system effectively eliminates clock-distribution delay by ensuring that clock edges arrive at internal flip-flops in synchronism with clock edges arriving at the input.

In addition to eliminating clock-distribution delay, the DLL provides advanced control of multiple clock domains. The DLL provides four quadrature phases of the source clock, can double the clock, or divide the clock by 1.5, 2, 2.5, 3, 4, 5, 8, or 16.

The DLL also operates as a clock mirror. By driving the output from a DLL off-chip and then back on again, the DLL can be used to de-skew a board level clock among multiple Virtex devices.

In order to guarantee that the system clock is operating correctly prior to the FPGA starting up after configuration, the DLL can delay the completion of the configuration process until after it has achieved lock.

See **DLL Timing Parameters**, page 21 of Module 3, for frequency range information.

Boundary Scan

Virtex devices support all the mandatory boundary-scan instructions specified in the IEEE standard 1149.1. A Test Access Port (TAP) and registers are provided that implement the EXTEST, INTEST, SAMPLE/PRELOAD, BYPASS, IDCODE, USERCODE, and HIGHZ instructions. The TAP also supports two internal scan chains and configuration/readback of the device. The TAP uses dedicated package pins that always operate using LVTTTL. For TDO to operate using LVTTTL, the V_{CCO} for Bank 2 should be 3.3 V. Otherwise, TDO switches rail-to-rail between ground and V_{CCO} .

Boundary-scan operation is independent of individual IOB configurations, and unaffected by package type. All IOBs, including un-bonded ones, are treated as independent 3-state bidirectional pins in a single scan chain. Retention of the bidirectional test capability after configuration facilitates the testing of external interconnections, provided the user design or application is turned off.

Table 5 lists the boundary-scan instructions supported in Virtex FPGAs. Internal signals can be captured during EXTEST by connecting them to un-bonded or unused IOBs. They can also be connected to the unused outputs of IOBs defined as unidirectional input pins.

Before the device is configured, all instructions except USER1 and USER2 are available. After configuration, all instructions are available. During configuration, it is recommended that those operations using the boundary-scan register (SAMPLE/PRELOAD, INTEST, EXTEST) not be performed.

ers with a common user interface regardless of their choice of entry and verification tools. The XDM software simplifies the selection of implementation options with pull-down menus and on-line help.

Application programs ranging from schematic capture to Placement and Routing (PAR) can be accessed through the XDM software. The program command sequence is generated prior to execution, and stored for documentation.

Several advanced software features facilitate Virtex design. RPMs, for example, are schematic-based macros with relative location constraints to guide their placement. They help ensure optimal implementation of common functions.

For HDL design entry, the Xilinx FPGA Foundation development system provides interfaces to the following synthesis design environments.

- Synopsys (FPGA Compiler, FPGA Express)
- Exemplar (Spectrum)
- Synplicity (Synplify)

For schematic design entry, the Xilinx FPGA Foundation and alliance development system provides interfaces to the following schematic-capture design environments.

- Mentor Graphics V8 (Design Architect, QuickSim II)
- Viewlogic Systems (Viewdraw)

Third-party vendors support many other environments.

A standard interface-file specification, Electronic Design Interchange Format (EDIF), simplifies file transfers into and out of the development system.

Virtex FPGAs supported by a unified library of standard functions. This library contains over 400 primitives and macros, ranging from 2-input AND gates to 16-bit accumulators, and includes arithmetic functions, comparators, counters, data registers, decoders, encoders, I/O functions, latches, Boolean functions, multiplexers, shift registers, and barrel shifters.

The “soft macro” portion of the library contains detailed descriptions of common logic functions, but does not contain any partitioning or placement information. The performance of these macros depends, therefore, on the partitioning and placement obtained during implementation.

RPMs, on the other hand, do contain predetermined partitioning and placement information that permits optimal implementation of these functions. Users can create their own library of soft macros or RPMs based on the macros and primitives in the standard library.

The design environment supports hierarchical design entry, with high-level schematics that comprise major functional blocks, while lower-level schematics define the logic in these blocks. These hierarchical design elements are automatically combined by the implementation tools. Different design entry tools can be combined within a hierarchical

design, thus allowing the most convenient entry method to be used for each portion of the design.

Design Implementation

The place-and-route tools (PAR) automatically provide the implementation flow described in this section. The partitioner takes the EDIF net list for the design and maps the logic into the architectural resources of the FPGA (CLBs and IOBs, for example). The placer then determines the best locations for these blocks based on their interconnections and the desired performance. Finally, the router interconnects the blocks.

The PAR algorithms support fully automatic implementation of most designs. For demanding applications, however, the user can exercise various degrees of control over the process. User partitioning, placement, and routing information is optionally specified during the design-entry process. The implementation of highly structured designs can benefit greatly from basic floor planning.

The implementation software incorporates Timing Wizard® timing-driven placement and routing. Designers specify timing requirements along entire paths during design entry. The timing path analysis routines in PAR then recognize these user-specified requirements and accommodate them.

Timing requirements are entered on a schematic in a form directly relating to the system requirements, such as the targeted clock frequency, or the maximum allowable delay between two registers. In this way, the overall performance of the system along entire signal paths is automatically tailored to user-generated specifications. Specific timing information for individual nets is unnecessary.

Design Verification

In addition to conventional software simulation, FPGA users can use in-circuit debugging techniques. Because Xilinx devices are infinitely reprogrammable, designs can be verified in real time without the need for extensive sets of software simulation vectors.

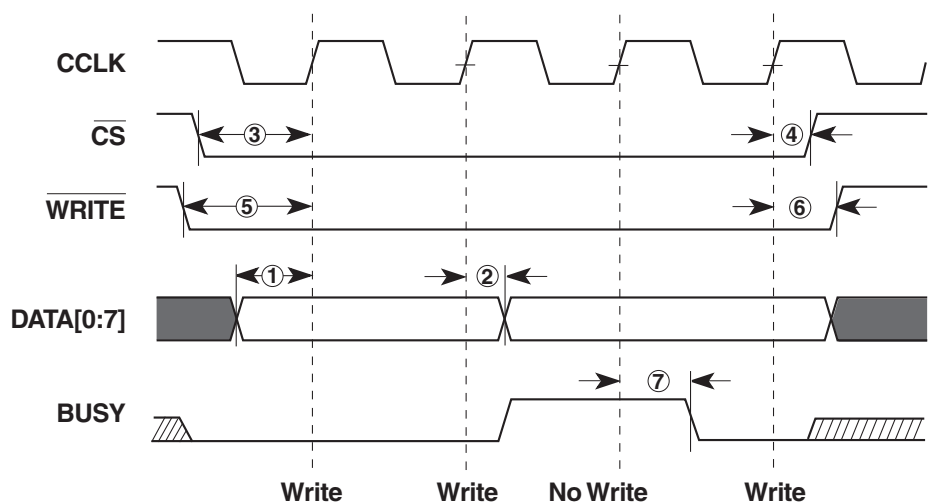
The development system supports both software simulation and in-circuit debugging techniques. For simulation, the system extracts the post-layout timing information from the design database, and back-annotates this information into the net list for use by the simulator. Alternatively, the user can verify timing-critical portions of the design using the TRACE® static timing analyzer.

For in-circuit debugging, the development system includes a download and readback cable. This cable connects the FPGA in the target system to a PC or workstation. After downloading the design into the FPGA, the designer can single-step the logic, readback the contents of the flip-flops, and so observe the internal logic state. Simple modifications can be downloaded into the system in a matter of minutes.

3. At the rising edge of CCLK: If BUSY is Low, the data is accepted on this clock. If BUSY is High (from a previous write), the data is not accepted. Acceptance will instead occur on the first clock after BUSY goes Low, and the data must be held until this has happened.
4. Repeat steps 2 and 3 until all the data has been sent.

5. De-assert $\overline{\text{CS}}$ and $\overline{\text{WRITE}}$.

A flowchart for the write operation appears in [Figure 17](#). Note that if CCLK is slower than f_{CCNH} , the FPGA never asserts BUSY. In this case, the above handshake is unnecessary, and data can simply be entered into the FPGA every CCLK cycle.



ds003_16_071902

Figure 16: Write Operations

IOB Input Switching Characteristics Standard Adjustments

Description	Symbol	Standard ⁽¹⁾	Speed Grade				Units
			Min	-6	-5	-4	
Data Input Delay Adjustments							
Standard-specific data input delay adjustments	T _{ILVTTL}	LVTTL	0	0	0	0	ns
	T _{ILVCMOS2}	LVC MOS2	−0.02	−0.04	−0.04	−0.05	ns
	T _{IPCI33_3}	PCI, 33 MHz, 3.3 V	−0.05	−0.11	−0.12	−0.14	ns
	T _{IPCI33_5}	PCI, 33 MHz, 5.0 V	0.13	0.25	0.28	0.33	ns
	T _{IPCI66_3}	PCI, 66 MHz, 3.3 V	−0.05	−0.11	−0.12	−0.14	ns
	T _{IGTL}	GTL	0.10	0.20	0.23	0.26	ns
	T _{IGTLP}	GTL+	0.06	0.11	0.12	0.14	ns
	T _{IHSTL}	HSTL	0.02	0.03	0.03	0.04	ns
	T _{ISSTL2}	SSTL2	−0.04	−0.08	−0.09	−0.10	ns
	T _{ISSTL3}	SSTL3	−0.02	−0.04	−0.05	−0.06	ns
	T _{ICTT}	CTT	0.01	0.02	0.02	0.02	ns
	T _{IAGP}	AGP	−0.03	−0.06	−0.07	−0.08	ns

Notes:

- Input timing for LVTTL is measured at 1.4 V. For other I/O standards, see [Table 3](#).

IOB Output Switching Characteristics

Output delays terminating at a pad are specified for LVTTL with 12 mA drive and fast slew rate. For other standards, adjust the delays with the values shown in [IOB Output Switching Characteristics Standard Adjustments, page 9](#).

Description	Symbol	Speed Grade				Units
		Min	-6	-5	-4	
Propagation Delays						
O input to Pad	T _{IOOP}	1.2	2.9	3.2	3.5	ns, max
O input to Pad via transparent latch	T _{IOOLP}	1.4	3.4	3.7	4.0	ns, max
3-State Delays						
T input to Pad high-impedance ⁽¹⁾	T _{IOTHZ}	1.0	2.0	2.2	2.4	ns, max
T input to valid data on Pad	T _{IOTON}	1.4	3.1	3.3	3.7	ns, max
T input to Pad high-impedance via transparent latch ⁽¹⁾	T _{IOTLPHZ}	1.2	2.4	2.6	3.0	ns, max
T input to valid data on Pad via transparent latch	T _{IOTLPON}	1.6	3.5	3.8	4.2	ns, max
GTS to Pad high impedance ⁽¹⁾	T _{GTS}	2.5	4.9	5.5	6.3	ns, max
Sequential Delays						
Clock CLK						
Minimum Pulse Width, High	T _{CH}	0.8	1.5	1.7	2.0	ns, min
Minimum Pulse Width, Low	T _{CL}	0.8	1.5	1.7	2.0	ns, min

Virtex Pin-to-Pin Output Parameter Guidelines

All devices are 100% functionally tested. Listed below are representative values for typical pin locations and normal clock loading. Values are expressed in nanoseconds unless otherwise noted.

Global Clock Input to Output Delay for LVTTL, 12 mA, Fast Slew Rate, *with* DLL

Description	Symbol	Device	Speed Grade				Units
			Min	-6	-5	-4	
LVTTL Global Clock Input to Output Delay using Output Flip-flop, 12 mA, Fast Slew Rate, <i>with</i> DLL. For data <i>output</i> with different standards, adjust delays with the values shown in Output Delay Adjustments.	T _{ICKOFDLL}	XCV50	1.0	3.1	3.3	3.6	ns, max
		XCV100	1.0	3.1	3.3	3.6	ns, max
		XCV150	1.0	3.1	3.3	3.6	ns, max
		XCV200	1.0	3.1	3.3	3.6	ns, max
		XCV300	1.0	3.1	3.3	3.6	ns, max
		XCV400	1.0	3.1	3.3	3.6	ns, max
		XCV600	1.0	3.1	3.3	3.6	ns, max
		XCV800	1.0	3.1	3.3	3.6	ns, max
		XCV1000	1.0	3.1	3.3	3.6	ns, max

Notes:

1. Listed above are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.
2. Output timing is measured at 1.4 V with 35 pF external capacitive load for LVTTL. The 35 pF load does not apply to the Min values. For other I/O standards and different loads, see [Table 2](#) and [Table 3](#).
3. DLL output jitter is already included in the timing calculation.

Global Clock Input-to-Output Delay for LVTTL, 12 mA, Fast Slew Rate, *without* DLL

Description	Symbol	Device	Speed Grade				Units
			Min	-6	-5	-4	
LVTTL Global Clock Input to Output Delay using Output Flip-flop, 12 mA, Fast Slew Rate, <i>without</i> DLL. For data <i>output</i> with different standards, adjust delays with the values shown in Input and Output Delay Adjustments. For I/O standards requiring V _{REF} such as GTL, GTL+, SSTL, HSTL, CTT, and AGO, an additional 600 ps must be added.	T _{ICKOF}	XCV50	1.5	4.6	5.1	5.7	ns, max
		XCV100	1.5	4.6	5.1	5.7	ns, max
		XCV150	1.5	4.7	5.2	5.8	ns, max
		XCV200	1.5	4.7	5.2	5.8	ns, max
		XCV300	1.5	4.7	5.2	5.9	ns, max
		XCV400	1.5	4.8	5.3	6.0	ns, max
		XCV600	1.6	4.9	5.4	6.0	ns, max
		XCV800	1.6	4.9	5.5	6.2	ns, max
		XCV1000	1.7	5.0	5.6	6.3	ns, max

Notes:

1. Listed above are representative values where one global clock input drives one vertical clock line in each accessible column, and where all accessible IOB and CLB flip-flops are clocked by the global clock net.
2. Output timing is measured at 1.4 V with 35 pF external capacitive load for LVTTL. The 35 pF load does not apply to the Min values. For other I/O standards and different loads, see [Table 2](#) and [Table 3](#).

Virtex Pin-to-Pin Input Parameter Guidelines

All devices are 100% functionally tested. Listed below are representative values for typical pin locations and normal clock loading. Values are expressed in nanoseconds unless otherwise noted

Global Clock Set-Up and Hold for LVTTL Standard, *with DLL*

Description	Symbol	Device	Speed Grade				Units
			Min	-6	-5	-4	
Input Setup and Hold Time Relative to Global Clock Input Signal for LVTTL Standard. For data input with different standards, adjust the setup time delay by the values shown in Input Delay Adjustments.							
No Delay Global Clock and IFF, with DLL	T_{PSDLL}/T_{PHDLL}	XCV50	0.40 / -0.4	1.7 / -0.4	1.8 / -0.4	2.1 / -0.4	ns, min
		XCV100	0.40 / -0.4	1.7 / -0.4	1.9 / -0.4	2.1 / -0.4	ns, min
		XCV150	0.40 / -0.4	1.7 / -0.4	1.9 / -0.4	2.1 / -0.4	ns, min
		XCV200	0.40 / -0.4	1.7 / -0.4	1.9 / -0.4	2.1 / -0.4	ns, min
		XCV300	0.40 / -0.4	1.7 / -0.4	1.9 / -0.4	2.1 / -0.4	ns, min
		XCV400	0.40 / -0.4	1.7 / -0.4	1.9 / -0.4	2.1 / -0.4	ns, min
		XCV600	0.40 / -0.4	1.7 / -0.4	1.9 / -0.4	2.1 / -0.4	ns, min
		XCV800	0.40 / -0.4	1.7 / -0.4	1.9 / -0.4	2.1 / -0.4	ns, min
		XCV1000	0.40 / -0.4	1.7 / -0.4	1.9 / -0.4	2.1 / -0.4	ns, min

IFF = Input Flip-Flop or Latch

Notes:

1. Set-up time is measured relative to the Global Clock input signal with the fastest route and the lightest load. Hold time is measured relative to the Global Clock input signal with the slowest route and heaviest load.
2. DLL output jitter is already included in the timing calculation.
3. A Zero "0" Hold Time listing indicates no hold time or a negative hold time. Negative values can not be guaranteed "best-case", but if a "0" is listed, there is no positive hold time.

DLL Timing Parameters

All devices are 100 percent functionally tested. Because of the difficulty in directly measuring many internal timing parameters, those parameters are derived from benchmark timing patterns. The following guidelines reflect worst-case values across the recommended operating conditions.

Description	Symbol	Speed Grade						Units
		-6		-5		-4		
		Min	Max	Min	Max	Min	Max	
Input Clock Frequency (CLKDLLHF)	FCLKINHF	60	200	60	180	60	180	MHz
Input Clock Frequency (CLKDLL)	FCLKINLF	25	100	25	90	25	90	MHz
Input Clock Pulse Width (CLKDLLHF)	T _{DLLPWHF}	2.0	-	2.4	-	2.4	-	ns
Input Clock Pulse Width (CLKDLL)	T _{DLLPWL}	2.5	-	3.0		3.0	-	ns

Notes:

1. All specifications correspond to Commercial Operating Temperatures (0°C to +85°C).

DLL Clock Tolerance, Jitter, and Phase Information

All DLL output jitter and phase specifications determined through statistical measurement at the package pins using a clock mirror configuration and matched drivers.

Description	Symbol	F _{CLKIN}	CLKDLLHF		CLKDLL		Units
			Min	Max	Min	Max	
Input Clock Period Tolerance	T _{IP} TOL		-	1.0	-	1.0	ns
Input Clock Jitter Tolerance (Cycle to Cycle)	T _{IJ} TCC		-	± 150	-	± 300	ps
Time Required for DLL to Acquire Lock	T _{LOCK}	> 60 MHz	-	20	-	20	μs
		50 - 60 MHz	-	-	-	25	μs
		40 - 50 MHz	-	-	-	50	μs
		30 - 40 MHz	-	-	-	90	μs
		25 - 30 MHz	-	-	-	120	μs
Output Jitter (cycle-to-cycle) for any DLL Clock Output ⁽¹⁾	T _{OJ} TCC			± 60		± 60	ps
Phase Offset between CLKIN and CLKO ⁽²⁾	T _{PHIO}			± 100		± 100	ps
Phase Offset between Clock Outputs on the DLL ⁽³⁾	T _{PHOO}			± 140		± 140	ps
Maximum Phase Difference between CLKIN and CLKO ⁽⁴⁾	T _{PHIOM}			± 160		± 160	ps
Maximum Phase Difference between Clock Outputs on the DLL ⁽⁵⁾	T _{PHOOM}			± 200		± 200	ps

Notes:

1. **Output Jitter** is cycle-to-cycle jitter measured on the DLL output clock, *excluding* input clock jitter.
2. **Phase Offset between CLKIN and CLKO** is the worst-case fixed time difference between rising edges of CLKIN and CLKO, *excluding* Output Jitter and input clock jitter.
3. **Phase Offset between Clock Outputs on the DLL** is the worst-case fixed time difference between rising edges of any two DLL outputs, *excluding* Output Jitter and input clock jitter.
4. **Maximum Phase Difference between CLKIN and CLKO** is the sum of Output Jitter and Phase Offset between CLKIN and CLKO, or the greatest difference between CLKIN and CLKO rising edges due to DLL alone (*excluding* input clock jitter).
5. **Maximum Phase Difference between Clock Outputs on the DLL** is the sum of Output Jitter and Phase Offset between any two DLL clock outputs, or the greatest difference between any two DLL output rising edges due to DLL alone (*excluding* input clock jitter).
6. All specifications correspond to Commercial Operating Temperatures (0°C to +85°C).

Virtex Pinout Information

Pinout Tables

See www.xilinx.com for updates or additional pinout information. For convenience, [Table 2](#), [Table 3](#) and [Table 4](#) list the locations of special-purpose and power-supply pins. Pins not listed are either user I/Os or not connected, depending on the device/package combination. See the Pinout Diagrams starting on [page 17](#) for any pins not listed for a particular part/package combination.

Table 2: Virtex Pinout Tables (Chip-Scale and QFP Packages)

Pin Name	Device	CS144	TQ144	PQ/HQ240
GCK0	All	K7	90	92
GCK1	All	M7	93	89
GCK2	All	A7	19	210
GCK3	All	A6	16	213
M0	All	M1	110	60
M1	All	L2	112	58
M2	All	N2	108	62
CCLK	All	B13	38	179
PROGRAM	All	L12	72	122
DONE	All	M12	74	120
INIT	All	L13	71	123
BUSY/DOUT	All	C11	39	178
D0/DIN	All	C12	40	177
D1	All	E10	45	167
D2	All	E12	47	163
D3	All	F11	51	156
D4	All	H12	59	145
D5	All	J13	63	138
D6	All	J11	65	134
D7	All	K10	70	124
WRITE	All	C10	32	185
CS	All	D10	33	184
TDI	All	A11	34	183
TDO	All	A12	36	181
TMS	All	B1	143	2
TCK	All	C3	2	239
V _{CCINT}	All	A9, B6, C5, G3, G12, M5, M9, N6	10, 15, 25, 57, 84, 94, 99, 126	16, 32, 43, 77, 88, 104, 137, 148, 164, 198, 214, 225

Table 2: Virtex Pinout Tables (Chip-Scale and QFP Packages) (Continued)

Pin Name	Device	CS144	TQ144	PQ/HQ240
V_{CCO}	All	Banks 0 and 1: A2, A13, D7 Banks 2 and 3: B12, G11, M13 Banks 4 and 5: N1, N7, N13 Banks 6 and 7: B2, G2, M2	No I/O Banks in this package: 1, 17, 37, 55, 73, 92, 109, 128	No I/O Banks in this package: 15, 30, 44, 61, 76, 90, 105, 121, 136, 150, 165, 180, 197, 212, 226, 240
V_{REF} Bank 0 (V_{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	C4, D6	5, 13	218, 232
	XCV100/150	... + B4	... + 7	... + 229
	XCV200/300	N/A	N/A	... + 236
	XCV400	N/A	N/A	... + 215
	XCV600	N/A	N/A	... + 230
	XCV800	N/A	N/A	... + 222
V_{REF} Bank 1 (V_{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	A10, B8	22, 30	191, 205
	XCV100/150	... + D9	... + 28	... + 194
	XCV200/300	N/A	N/A	... + 187
	XCV400	N/A	N/A	... + 208
	XCV600	N/A	N/A	... + 193
	XCV800	N/A	N/A	... + 201
V_{REF} Bank 2 (V_{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V_{REF} pins are general I/O.	XCV50	D11, F10	42, 50	157, 171
	XCV100/150	... + D13	... + 44	... + 168
	XCV200/300	N/A	N/A	... + 175
	XCV400	N/A	N/A	... + 154
	XCV600	N/A	N/A	... + 169
	XCV800	N/A	N/A	... + 161

Table 2: Virtex Pinout Tables (Chip-Scale and QFP Packages) (Continued)

Pin Name	Device	CS144	TQ144	PQ/HQ240
V_{REF} Bank 6 (V _{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V _{REF} pins are general I/O.	XCV50	H2, K1	116, 123	36, 50
	XCV100/150	... + J3	... + 118	... + 47
	XCV200/300	N/A	N/A	... + 54
	XCV400	N/A	N/A	... + 33
	XCV600	N/A	N/A	... + 48
	XCV800	N/A	N/A	... + 40
V_{REF} Bank 7 (V _{REF} pins are listed incrementally. Connect all pins listed for both the required device and all smaller devices listed in the same package.) Within each bank, if input reference voltage is not required, all V _{REF} pins are general I/O.	XCV50	D4, E1	133, 140	9, 23
	XCV100/150	... + D2	... + 138	... + 12
	XCV200/300	N/A	N/A	... + 5
	XCV400	N/A	N/A	... + 26
	XCV600	N/A	N/A	... + 11
	XCV800	N/A	N/A	... + 19
GND	All	A1, B9, B11, C7, D5, E4, E11, F1, G10, J1, J12, L3, L5, L7, L9, N12	9, 18, 26, 35, 46, 54, 64, 75, 83, 91, 100, 111, 120, 129, 136, 144,	1, 8, 14, 22, 29, 37, 45, 51, 59, 69, 75, 83, 91, 98, 106, 112, 119, 129, 135, 143, 151, 158, 166, 172, 182, 190, 196, 204, 211, 219, 227, 233

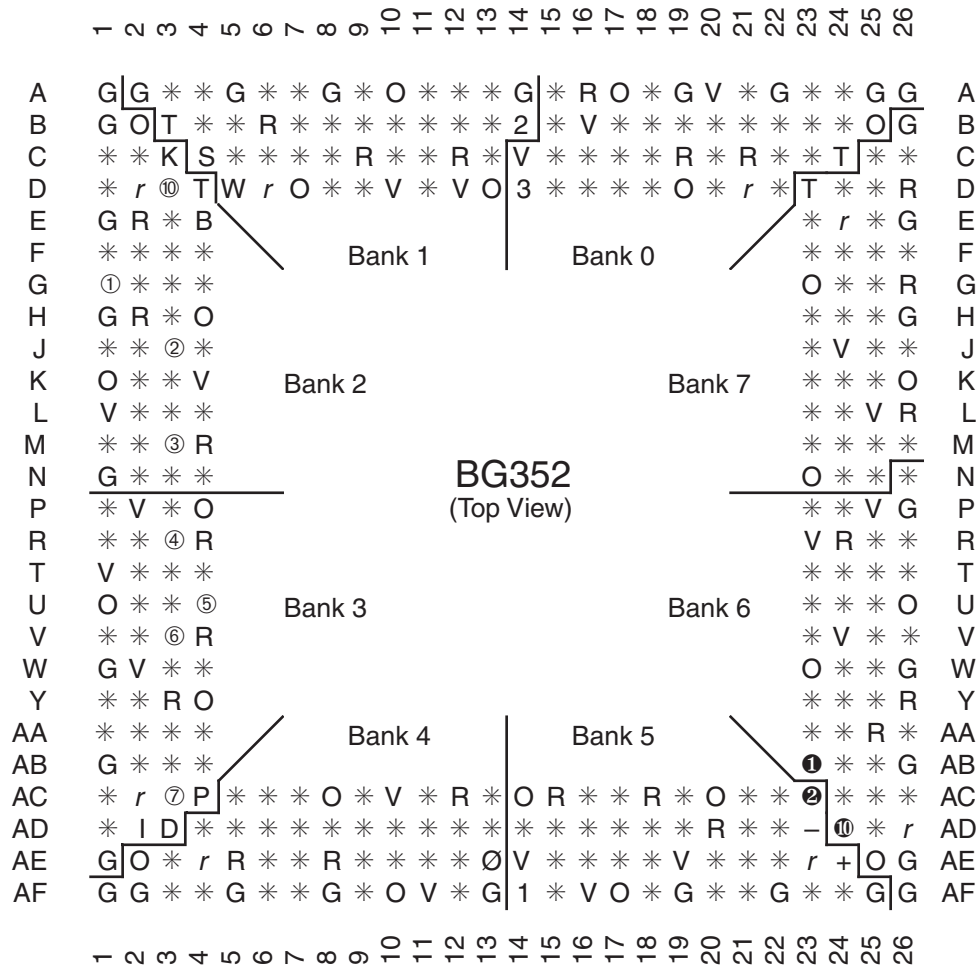
Table 4: Virtex Pinout Tables (Fine-Pitch BGA)

Pin Name	Device	FG256	FG456	FG676	FG680
GCK0	All	N8	W12	AA14	AW19
GCK1	All	R8	Y11	AB13	AU22
GCK2	All	C9	A11	C13	D21
GCK3	All	B8	C11	E13	A20
M0	All	N3	AB2	AD4	AT37
M1	All	P2	U5	W7	AU38
M2	All	R3	Y4	AB6	AT35
CCLK	All	D15	B22	D24	E4
PROGRAM	All	P15	W20	AA22	AT5
DONE	All	R14	Y19	AB21	AU5
INIT	All	N15	V19	Y21	AU2
BUSY/DOUT	All	C15	C21	E23	E3
D0/DIN	All	D14	D20	F22	C2
D1	All	E16	H22	K24	P4
D2	All	F15	H20	K22	P3
D3	All	G16	K20	M22	R1
D4	All	J16	N22	R24	AD3
D5	All	M16	R21	U23	AG2
D6	All	N16	T22	V24	AH1
D7	All	N14	Y21	AB23	AR4
WRITE	All	C13	A20	C22	B4
CS	All	B13	C19	E21	D5
TDI	All	A15	B20	D22	B3
TDO	All	B14	A21	C23	C4
TMS	All	D3	D3	F5	E36
TCK	All	C4	C4	E6	C36
DXN	All	R4	Y5	AB7	AV37
DXP	All	P4	V6	Y8	AU35

Table 4: Virtex Pinout Tables (Fine-Pitch BGA) (Continued)

Pin Name	Device	FG256	FG456	FG676	FG680
No Connect (No-connect pins are listed incrementally. All pins listed for both the required device and all larger devices listed in the same package are no connects.)	XCV800	N/A	N/A	A2, A3, A15, A25, B1, B6, B11, B16, B21, B24, B26, C1, C2, C25, C26, F2, F6, F21, F25, L2, L25, N25, P2, T2, T25, AA2, AA6, AA21, AA25, AD1, AD2, AD25, AE1, AE3, AE6, AE11, AE14, AE16, AE21, AE24, AE26, AF2, AF24, AF25	N/A
	XCV600	N/A	N/A	same as above	N/A
	XCV400	N/A	N/A	... + A9, A10, A13, A16, A24, AC1, AC25, AE12, AE15, AF3, AF10, AF11, AF13, AF14, AF16, AF18, AF23, B4, B12, B13, B15, B17, D1, D25, H26, J1, K26, L1, M1, M25, N1, N26, P1, P26, R2, R26, T1, T26, U26, V1	N/A
	XCV300	N/A	D4, D19, W4, W19	N/A	N/A
	XCV200	N/A	... + A2, A6, A12, B11, B16, C2, D1, D18, E17, E19, G2, G22, L2, L19, M2, M21, R3, R20, U3, U18, Y22, AA1, AA3, AA11, AA16, AB7, AB12, AB21,	N/A	N/A
	XCV150	N/A	... + A13, A14, C8, C9, E13, F11, H21, J1, J4, K2, K18, K19, M17, N1, P1, P5, P22, R22, W13, W15, AA9, AA10, AB8, AB14	N/A	N/A

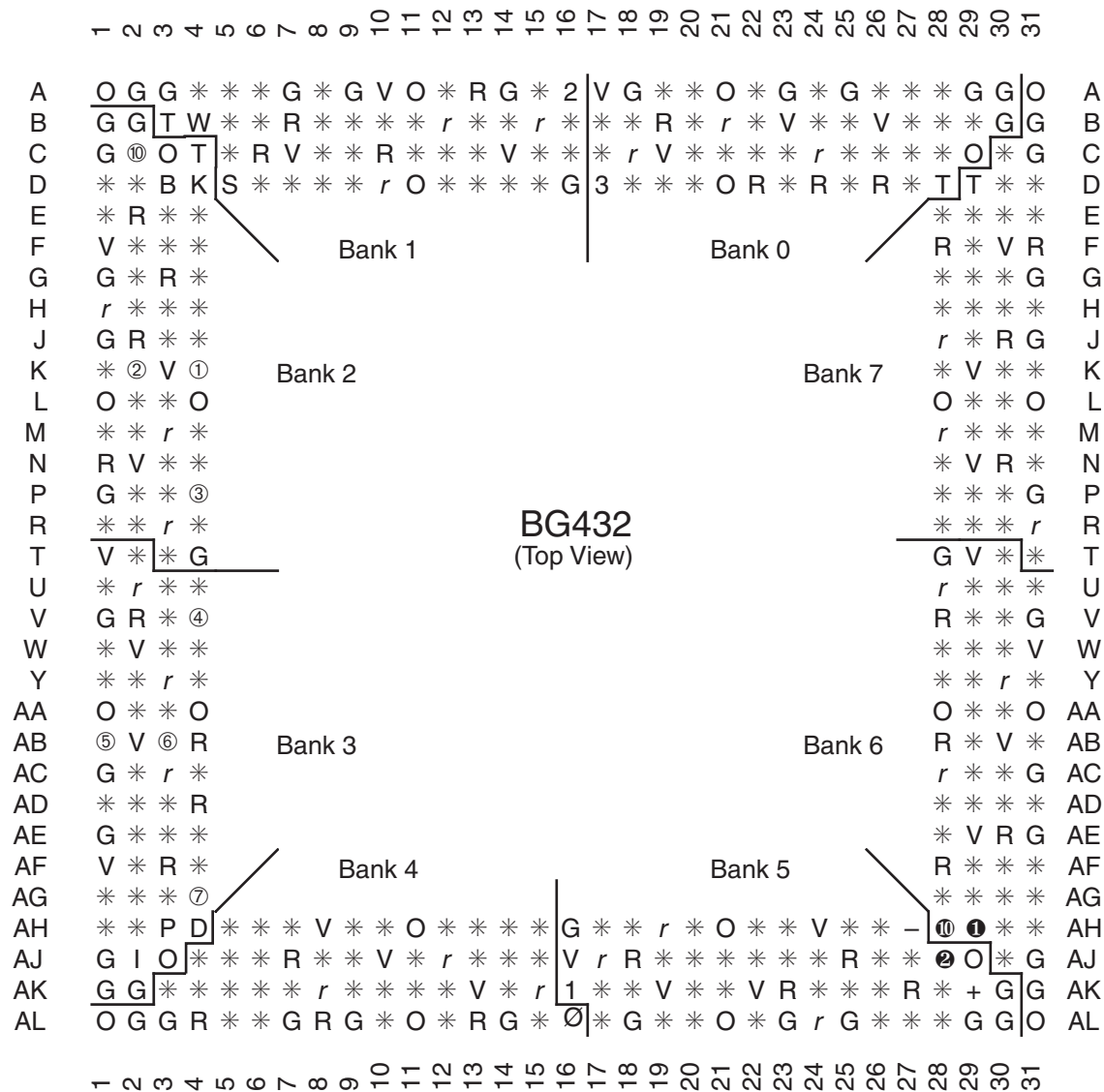
BG352 Pin Function Diagram



DS003_19_100600

Figure 5: BG352 Pin Function Diagram

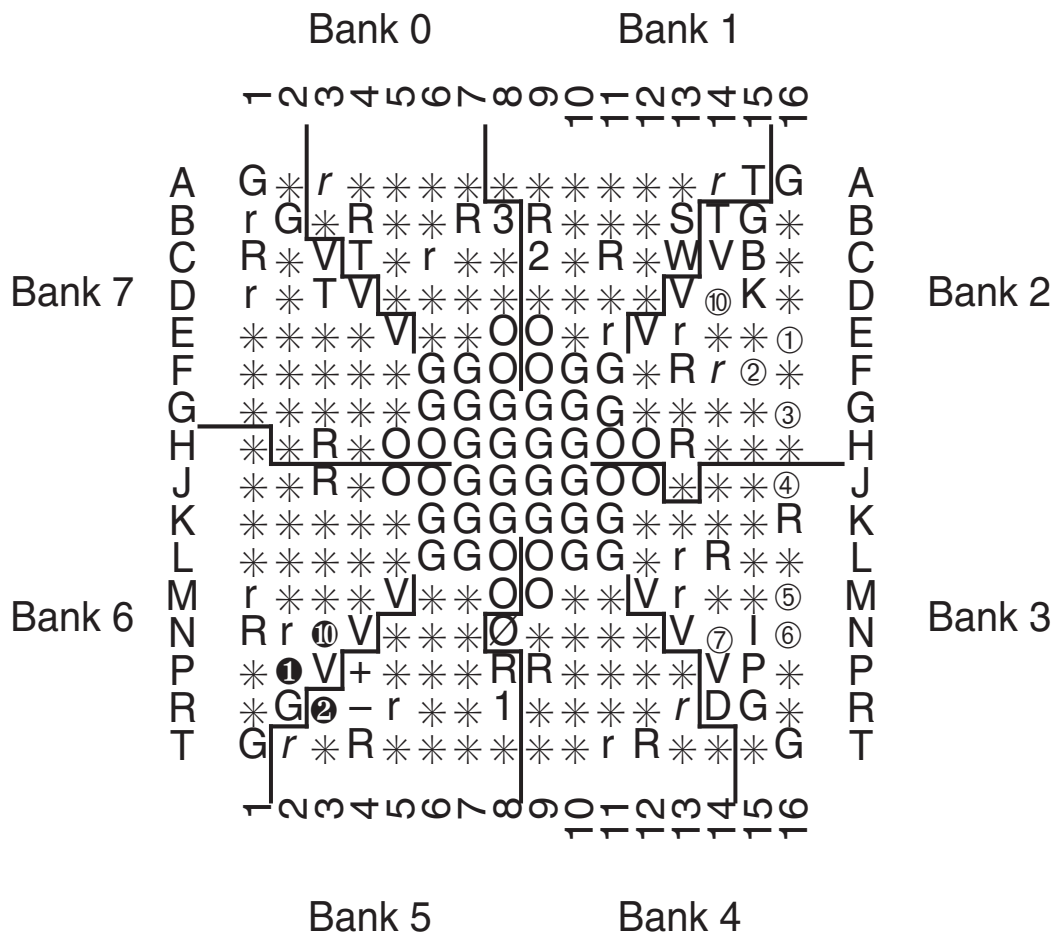
BG432 Pin Function Diagram



DS003_21_100300

Figure 6: BG432 Pin Function Diagram

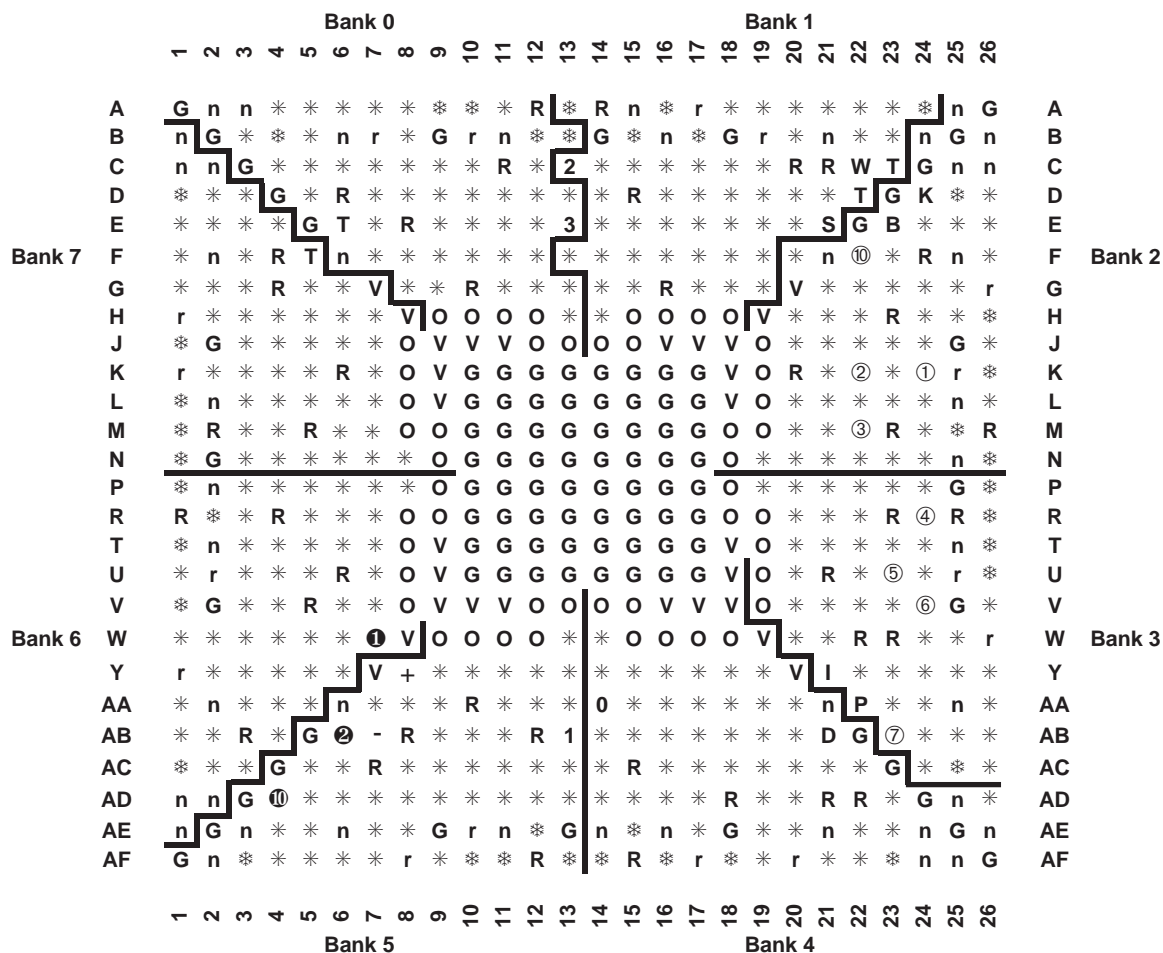
FG256 Pin Function Diagram



FG256
(Top view)

Figure 8: FG256 Pin Function Diagram

FG676 Pin Function Diagram



FG676
(Top view)

fg676a

Figure 10: FG676 Pin Function Diagram

Notes:

Packages FG456 and FG676 are layout compatible.