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#### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

#### Applications of "[Embedded - Microcontrollers](#)"

##### Details

Product Status	Active
Core Processor	ARM® Cortex®-M4
Core Size	32-Bit Single-Core
Speed	168MHz
Connectivity	CANbus, EBI/EMI, I²C, IrDA, LINbus, SPI, UART/USART, USB OTG
Peripherals	Brown-out Detect/Reset, DMA, I²S, LCD, POR, PWM, WDT
Number of I/O	82
Program Memory Size	1MB (1M x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	192K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 16x12b; D/A 2x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP
Supplier Device Package	100-LQFP (14x14)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/stmicroelectronics/stm32f405vgt6w">https://www.e-xfl.com/product-detail/stmicroelectronics/stm32f405vgt6w</a>

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STM32F405xx, STM32F407xx	Description
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## 2.2.9 Flexible static memory controller (FSMC)

The FSMC is embedded in the STM32F405xx and STM32F407xx family. It has four Chip Select outputs supporting the following modes: PCCard/Compact Flash, SRAM, PSRAM, NOR Flash and NAND Flash.

Functionality overview:

- Write FIFO
- Maximum FSMC\_CLK frequency for synchronous accesses is 60 MHz.

### LCD parallel interface

The FSMC can be configured to interface seamlessly with most graphic LCD controllers. It supports the Intel 8080 and Motorola 6800 modes, and is flexible enough to adapt to specific LCD interfaces. This LCD parallel interface capability makes it easy to build cost-effective graphic applications using LCD modules with embedded controllers or high performance solutions using external controllers with dedicated acceleration.

## 2.2.10 Nested vectored interrupt controller (NVIC)

The STM32F405xx and STM32F407xx embed a nested vectored interrupt controller able to manage 16 priority levels, and handle up to 82 maskable interrupt channels plus the 16 interrupt lines of the Cortex®-M4 with FPU core.

- Closely coupled NVIC gives low-latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Allows early processing of interrupts
- Processing of late arriving, higher-priority interrupts
- Support tail chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimum interrupt latency.

## 2.2.11 External interrupt/event controller (EXTI)

The external interrupt/event controller consists of 23 edge-detector lines used to generate interrupt/event requests. Each line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the Internal APB2 clock period. Up to 140 GPIOs can be connected to the 16 external interrupt lines.

## 2.2.12 Clocks and startup

On reset the 16 MHz internal RC oscillator is selected as the default CPU clock. The 16 MHz internal RC oscillator is factory-trimmed to offer 1% accuracy over the full temperature range. The application can then select as system clock either the RC oscillator or an external 4-26 MHz clock source. This clock can be monitored for failure. If a failure is detected, the system automatically switches back to the internal RC oscillator and a software interrupt is generated (if enabled). This clock source is input to a PLL thus allowing to increase the frequency up to 168 MHz. Similarly, full interrupt management of the PLL

STM32F405xx, STM32F407xx	Description
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alternate functions. All GPIOs are high-current-capable and have speed selection to better manage internal noise, power consumption and electromagnetic emission.

The I/O configuration can be locked if needed by following a specific sequence in order to avoid spurious writing to the I/Os registers.

Fast I/O handling allowing maximum I/O toggling up to 84 MHz.

### 2.2.35 Analog-to-digital converters (ADCs)

Three 12-bit analog-to-digital converters are embedded and each ADC shares up to 16 external channels, performing conversions in the single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

Additional logic functions embedded in the ADC interface allow:

- Simultaneous sample and hold
- Interleaved sample and hold

The ADC can be served by the DMA controller. An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

To synchronize A/D conversion and timers, the ADCs could be triggered by any of TIM1, TIM2, TIM3, TIM4, TIM5, or TIM8 timer.

### 2.2.36 Temperature sensor

The temperature sensor has to generate a voltage that varies linearly with temperature. The conversion range is between 1.8 V and 3.6 V. The temperature sensor is internally connected to the ADC1\_IN16 input channel which is used to convert the sensor output voltage into a digital value.

As the offset of the temperature sensor varies from chip to chip due to process variation, the internal temperature sensor is mainly suitable for applications that detect temperature changes instead of absolute temperatures. If an accurate temperature reading is needed, then an external temperature sensor part should be used.

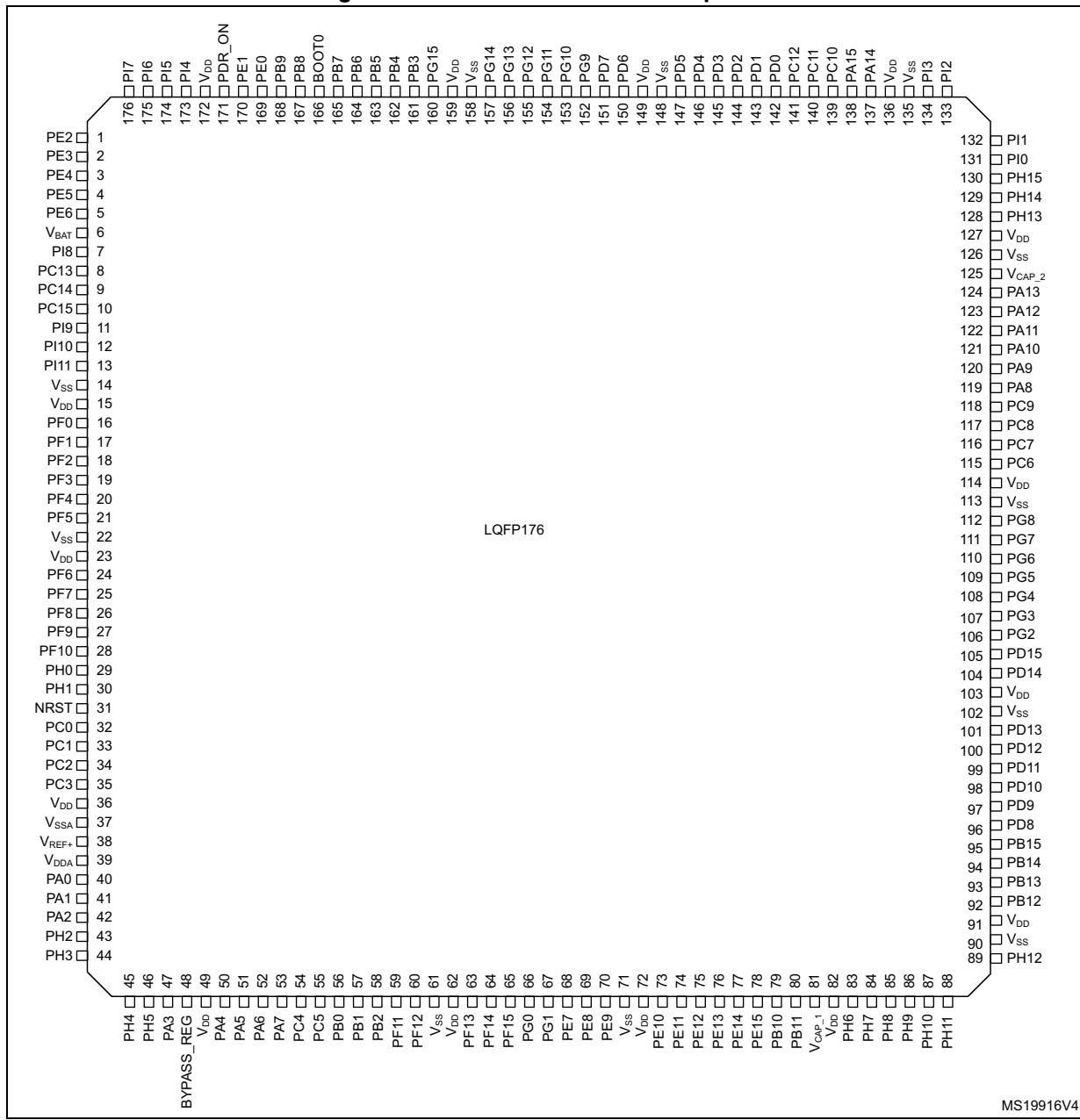
### 2.2.37 Digital-to-analog converter (DAC)

The two 12-bit buffered DAC channels can be used to convert two digital signals into two analog voltage signal outputs.

This dual digital Interface supports the following features:

- two DAC converters: one for each output channel
- 8-bit or 12-bit monotonic output
- left or right data alignment in 12-bit mode
- synchronized update capability
- noise-wave generation
- triangular-wave generation
- dual DAC channel independent or simultaneous conversions
- DMA capability for each channel
- external triggers for conversion
- input voltage reference  $V_{REF+}$

Figure 15. STM32F40xxx LQFP176 pinout



1. The above figure shows the package top view.

Table 7. STM32F40xxx pin and ball definitions

Pin number						Pin name (function after reset) <sup>(1)</sup>	Pin type	I/O structure	Notes	Alternate functions	Additional functions
LQFP64	WL CSP90	LQFP100	LQFP144	UFBGA176	LQFP176						
-	-	1	1	A2	1	PE2	I/O	FT	-	TRACECLK / FSMC_A23 / ETH_MII_RXD3 / EVENTOUT	-
-	-	2	2	A1	2	PE3	I/O	FT	-	TRACED0 / FSMC_A19 / EVENTOUT	-
-	-	3	3	B1	3	PE4	I/O	FT	-	TRACED1 / FSMC_A20 / DCMI_D4 / EVENTOUT	-
-	-	4	4	B2	4	PE5	I/O	FT	-	TRACED2 / FSMC_A21 / TIM9_CH1 / DCMI_D6 / EVENTOUT	-
-	-	5	5	B3	5	PE6	I/O	FT	-	TRACED3 / FSMC_A22 / TIM9_CH2 / DCMI_D7 / EVENTOUT	-
1	A10	6	6	C1	6	V <sub>BAT</sub>	S	-	-	-	-
-	-	-	-	D2	7	PI8	I/O	FT	<sup>(2)(3)</sup>	EVENTOUT	RTC_TAMP1, RTC_TAMP2, RTC_TS
2	A9	7	7	D1	8	PC13	I/O	FT	<sup>(2)(3)</sup>	EVENTOUT	RTC_OUT, RTC_TAMP1, RTC_TS
3	B10	8	8	E1	9	PC14 / OSC32_IN (PC14)	I/O	FT	<sup>(2)(3)</sup>	EVENTOUT	OSC32_IN <sup>(4)</sup>
4	B9	9	9	F1	10	PC15 / OSC32_OUT (PC15)	I/O	FT	<sup>(2)(3)</sup>	EVENTOUT	OSC32_OUT <sup>(4)</sup>
-	-	-	-	D3	11	PI9	I/O	FT	-	CAN1_RX / EVENTOUT	-
-	-	-	-	E3	12	PI10	I/O	FT	-	ETH_MII_RX_ER / EVENTOUT	-
-	-	-	-	E4	13	PI11	I/O	FT	-	OTG_HS_ULPI_DIR / EVENTOUT	-
-	-	-	-	F2	14	V <sub>SS</sub>	S	-	-	-	-
-	-	-	-	F3	15	V <sub>DD</sub>	S	-	-	-	-
-	-	-	10	E2	16	PF0	I/O	FT	-	FSMC_A0 / I2C2_SDA / EVENTOUT	-

**Table 9. Alternate function mapping (continued)**

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	TIM1/2	TIM3/4/5	TIM8/9/10 /11	I2C1/2/3	SPI1/SPI2/ I2S2/I2S2ext	SPI3/I2Sext /I2S3	USART1/2/3/ I2S3ext	UART4/5/ USART6	CAN1/2 TIM12/13/ 14	OTG_FS/ OTG_HS	ETH	FSMC/SDIO /OTG_FS	DCMI		
Port G	PG0	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A10	-	-	EVENTOUT
	PG1	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A11	-	-	EVENTOUT
	PG2	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A12	-	-	EVENTOUT
	PG3	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A13	-	-	EVENTOUT
	PG4	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A14	-	-	EVENTOUT
	PG5	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_A15	-	-	EVENTOUT
	PG6	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_INT2	-	-	EVENTOUT
	PG7	-	-	-	-	-	-	-	-	USART6_CK	-	-	-	FSMC_INT3	-	-	EVENTOUT
	PG8	-	-	-	-	-	-	-	-	USART6_RTS	-	-	ETH_PPS_OUT	-	-	-	EVENTOUT
	PG9	-	-	-	-	-	-	-	-	USART6_RX	-	-	-	FSMC_NE2/ FSMC_NCE3	-	-	EVENTOUT
	PG10	-	-	-	-	-	-	-	-	-	-	-	-	FSMC_NCE4_1/ FSMC_NE3	-	-	EVENTOUT
	PG11	-	-	-	-	-	-	-	-	-	-	-	ETH_MII_TX_EN ETH_RMII_TX_EN	FSMC_NCE4_2	-	-	EVENTOUT
	PG12	-	-	-	-	-	-	-	-	USART6_RTS	-	-	-	FSMC_NE4	-	-	EVENTOUT
	PG13	-	-	-	-	-	-	-	-	USART6_CTS	-	-	ETH_MII_TXD0 ETH_RMII_TXD0	FSMC_A24	-	-	EVENTOUT
	PG14	-	-	-	-	-	-	-	-	USART6_TX	-	-	ETH_MII_TXD1 ETH_RMII_TXD1	FSMC_A25	-	-	EVENTOUT
	PG15	-	-	-	-	-	-	-	-	USART6_CTS	-	-	-	DCMI_D13	-	-	EVENTOUT

**Table 10. register boundary addresses (continued)**

<b>Bus</b>	<b>Boundary address</b>	<b>Peripheral</b>
APB2	0x4001 4C00 - 0x4001 57FF	Reserved
	0x4001 4800 - 0x4001 4BFF	TIM11
	0x4001 4400 - 0x4001 47FF	TIM10
	0x4001 4000 - 0x4001 43FF	TIM9
	0x4001 3C00 - 0x4001 3FFF	EXTI
	0x4001 3800 - 0x4001 3BFF	SYSCFG
	0x4001 3400 - 0x4001 37FF	Reserved
	0x4001 3000 - 0x4001 33FF	SPI1
	0x4001 2C00 - 0x4001 2FFF	SDIO
	0x4001 2400 - 0x4001 2BFF	Reserved
	0x4001 2000 - 0x4001 23FF	ADC1 - ADC2 - ADC3
	0x4001 1800 - 0x4001 1FFF	Reserved
	0x4001 1400 - 0x4001 17FF	USART6
	0x4001 1000 - 0x4001 13FF	USART1
	0x4001 0800 - 0x4001 0FFF	Reserved
	0x4001 0400 - 0x4001 07FF	TIM8
	0x4001 0000 - 0x4001 03FF	TIM1
	0x4000 7800- 0x4000 FFFF	Reserved

**Table 21. Typical and maximum current consumption in Run mode, code with data processing running from Flash memory (ART accelerator disabled)**

Symbol	Parameter	Conditions	$f_{HCLK}$	Typ	Max <sup>(1)</sup>		Unit
				$T_A = 25^\circ\text{C}$	$T_A = 85^\circ\text{C}$	$T_A = 105^\circ\text{C}$	
$I_{DD}$	Supply current in Run mode	External clock <sup>(2)</sup> , all peripherals enabled <sup>(3)(4)</sup>	168 MHz	93	109	117	mA
			144 MHz	76	89	96	
			120 MHz	67	79	86	
			90 MHz	53	65	73	
			60 MHz	37	49	56	
			30 MHz	20	32	39	
			25 MHz	16	27	35	
			16 MHz	11	23	30	
			8 MHz	6	18	25	
			4 MHz	4	16	23	
			2 MHz	3	15	22	
		External clock <sup>(2)</sup> , all peripherals disabled <sup>(3)(4)</sup>	168 MHz	46	61	69	
			144 MHz	40	52	60	
			120 MHz	37	48	56	
			90 MHz	30	42	50	
			60 MHz	22	33	41	
			30 MHz	12	24	31	
			25 MHz	10	21	29	
			16 MHz	7	19	26	
			8 MHz	4	16	23	
			4 MHz	3	15	22	
			2 MHz	2	14	21	

1. Guaranteed by characterization, tested in production at  $V_{DD}$  max and  $f_{HCLK}$  max with peripherals enabled.
2. External clock is 4 MHz and PLL is on when  $f_{HCLK} > 25$  MHz.
3. When analog peripheral blocks such as (ADCs, DACs, HSE, LSE, HSI, LSI) are on, an additional power consumption should be considered.
4. When the ADC is ON (ADON bit set in the ADC\_CR2 register), add an additional power consumption of 1.6 mA per ADC for the analog part.

**Table 23. Typical and maximum current consumptions in Stop mode**

Symbol	Parameter	Conditions	Typ	Max			Unit
			T <sub>A</sub> = 25 °C	T <sub>A</sub> = 25 °C	T <sub>A</sub> = 85 °C	T <sub>A</sub> = 105 °C	
I <sub>DD_STOP</sub>	Supply current in Stop mode with main regulator in Run mode	Flash in Stop mode, low-speed and high-speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	0.45	1.5	11.00	20.00	mA
		Flash in Deep power-down mode, low-speed and high-speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	0.40	1.5	11.00	20.00	
	Supply current in Stop mode with main regulator in Low-power mode	Flash in Stop mode, low-speed and high-speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	0.31	1.1	8.00	15.00	
		Flash in Deep power-down mode, low-speed and high-speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	0.28	1.1	8.00	15.00	

**Table 24. Typical and maximum current consumptions in Standby mode**

Symbol	Parameter	Conditions	Typ			Max <sup>(1)</sup>		Unit
			T <sub>A</sub> = 25 °C			T <sub>A</sub> = 85 °C	T <sub>A</sub> = 105 °C	
			V <sub>DD</sub> = 1.8 V	V <sub>DD</sub> = 2.4 V	V <sub>DD</sub> = 3.3 V	V <sub>DD</sub> = 3.6 V		
I <sub>DD_STBY</sub>	Supply current in Standby mode	Backup SRAM ON, low-speed oscillator and RTC ON	3.0	3.4	4.0	20	36	µA
		Backup SRAM OFF, low-speed oscillator and RTC ON	2.4	2.7	3.3	16	32	
		Backup SRAM ON, RTC OFF	2.4	2.6	3.0	12.5	24.8	
		Backup SRAM OFF, RTC OFF	1.7	1.9	2.2	9.8	19.2	

1. Guaranteed by characterization.

### 5.3.18 TIM timer characteristics

The parameters given in [Table 52](#) and [Table 53](#) are guaranteed by design.

Refer to [Section 5.3.16: I/O port characteristics](#) for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

**Table 52. Characteristics of TIMx connected to the APB1 domain<sup>(1)</sup>**

Symbol	Parameter	Conditions	Min	Max	Unit	
$t_{\text{res}(\text{TIM})}$	Timer resolution time	AHB/APB1 prescaler distinct from 1, $f_{\text{TIMxCLK}} = 84 \text{ MHz}$	1	-	$t_{\text{TIMxCLK}}$	
			11.9	-	ns	
		AHB/APB1 prescaler = 1, $f_{\text{TIMxCLK}} = 42 \text{ MHz}$	1	-	$t_{\text{TIMxCLK}}$	
			23.8	-	ns	
$f_{\text{EXT}}$	Timer external clock frequency on CH1 to CH4	$f_{\text{TIMxCLK}} = 84 \text{ MHz}$ $\text{APB1} = 42 \text{ MHz}$	0	$f_{\text{TIMxCLK}}/2$	MHz	
$\text{Res}_{\text{TIM}}$	Timer resolution		0	42	MHz	
$t_{\text{COUNTER}}$	16-bit counter clock period when internal clock is selected		-	16/32	bit	
	32-bit counter clock period when internal clock is selected		1	65536	$t_{\text{TIMxCLK}}$	
			0.0119	780	μs	
			1	-	$t_{\text{TIMxCLK}}$	
$t_{\text{MAX\_COUNT}}$	Maximum possible count		0.0119	51130563	μs	
			-	$65536 \times 65536$	$t_{\text{TIMxCLK}}$	
			-	51.1	s	

1. TIMx is used as a general term to refer to the TIM2, TIM3, TIM4, TIM5, TIM6, TIM7, and TIM12 timers.

Table 67. ADC characteristics (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{lat}^{(4)}$	Injection trigger conversion latency	$f_{ADC} = 30 \text{ MHz}$	-	-	0.100	$\mu\text{s}$
			-	-	$3^{(7)}$	$1/f_{ADC}$
$t_{latr}^{(4)}$	Regular trigger conversion latency	$f_{ADC} = 30 \text{ MHz}$	-	-	0.067	$\mu\text{s}$
			-	-	$2^{(7)}$	$1/f_{ADC}$
$t_S^{(4)}$	Sampling time	$f_{ADC} = 30 \text{ MHz}$	0.100	-	16	$\mu\text{s}$
		-	3	-	480	$1/f_{ADC}$
$t_{STAB}^{(4)}$	Power-up time	-	-	2	3	$\mu\text{s}$
$t_{CONV}^{(4)}$	Total conversion time (including sampling time)	$f_{ADC} = 30 \text{ MHz}$ 12-bit resolution	0.50	-	16.40	$\mu\text{s}$
		$f_{ADC} = 30 \text{ MHz}$ 10-bit resolution	0.43	-	16.34	$\mu\text{s}$
		$f_{ADC} = 30 \text{ MHz}$ 8-bit resolution	0.37	-	16.27	$\mu\text{s}$
		$f_{ADC} = 30 \text{ MHz}$ 6-bit resolution	0.30	-	16.20	$\mu\text{s}$
		9 to 492 ( $t_S$ for sampling +n-bit resolution for successive approximation)				$1/f_{ADC}$
$f_S^{(4)}$	Sampling rate ( $f_{ADC} = 30 \text{ MHz}$ , and $t_S = 3 \text{ ADC cycles}$ )	12-bit resolution Single ADC	-	-	2	Msps
		12-bit resolution Interleave Dual ADC mode	-	-	3.75	Msps
		12-bit resolution Interleave Triple ADC mode	-	-	6	Msps
$I_{VREF+}^{(4)}$	ADC $V_{REF}$ DC current consumption in conversion mode	-	-	300	500	$\mu\text{A}$
$I_{VDDA}^{(4)}$	ADC $V_{DDA}$ DC current consumption in conversion mode	-	-	1.6	1.8	mA

1.  $V_{DD}/V_{DDA}$  minimum value of 1.7 V is obtained when the device operates in reduced temperature range, and with the use of an external power supply supervisor (refer to [Section : Internal reset OFF](#)).
2. It is recommended to maintain the voltage difference between  $V_{REF+}$  and  $V_{DDA}$  below 1.8 V.
3.  $V_{DDA} - V_{REF+} < 1.2 \text{ V}$ .
4. Guaranteed by characterization.
5.  $V_{REF+}$  is internally connected to  $V_{DDA}$  and  $V_{REF-}$  is internally connected to  $V_{SSA}$ .
6.  $R_{ADC}$  maximum value is given for  $V_{DD}=1.8 \text{ V}$ , and minimum value for  $V_{DD}=3.3 \text{ V}$ .
7. For external triggers, a delay of  $1/f_{PCLK2}$  must be added to the latency specified in [Table 67](#).

**Equation 1:  $R_{AIN}$  max formula**

$$R_{AIN} = \frac{(k - 0.5)}{f_{ADC} \times C_{ADC} \times \ln(2^{N+2})} - R_{ADC}$$

The formula above ([Equation 1](#)) is used to determine the maximum external impedance allowed for an error below 1/4 of LSB. N = 12 (from 12-bit resolution) and k is the number of sampling periods defined in the ADC\_SMPR1 register.

**Table 68. ADC accuracy at  $f_{ADC} = 30$  MHz**

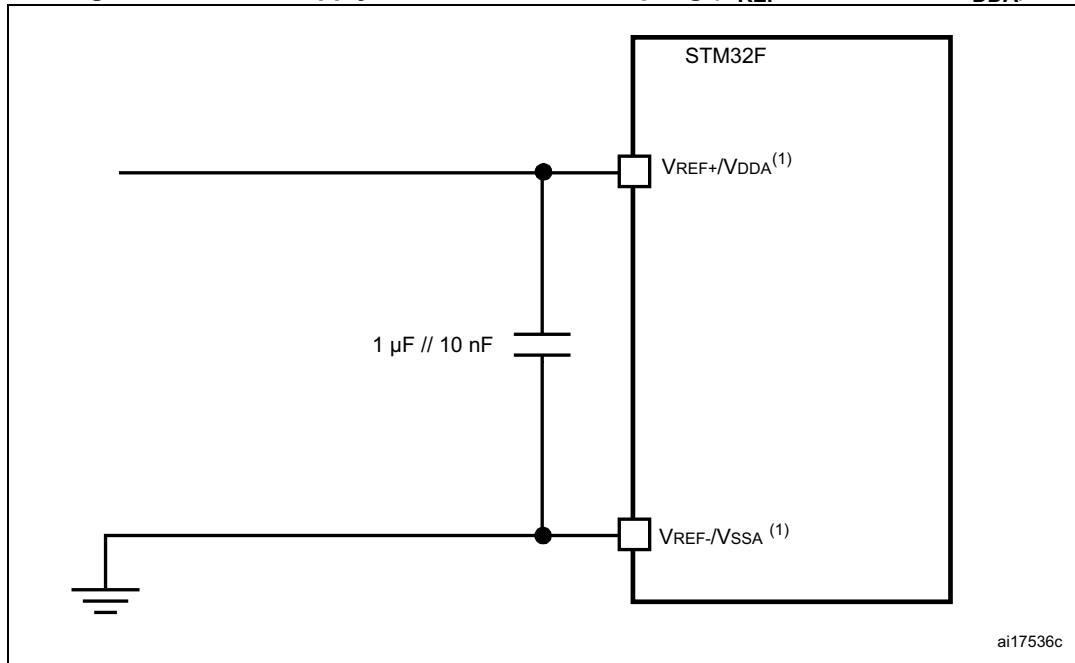
Symbol	Parameter	Test conditions	Typ	Max <sup>(1)</sup>	Unit
ET	Total unadjusted error	$f_{PCLK2} = 60$ MHz, $f_{ADC} = 30$ MHz, $R_{AIN} < 10$ k $\Omega$ , $V_{DDA} = 1.8^{(2)}$ to 3.6 V	$\pm 2$	$\pm 5$	LSB
EO	Offset error		$\pm 1.5$	$\pm 2.5$	
EG	Gain error		$\pm 1.5$	$\pm 3$	
ED	Differential linearity error		$\pm 1$	$\pm 2$	
EL	Integral linearity error		$\pm 1.5$	$\pm 3$	

1. Guaranteed by characterization.
2.  $V_{DD}/V_{DDA}$  minimum value of 1.7 V is obtained when the device operates in reduced temperature range, and with the use of an external power supply supervisor (refer to [Section : Internal reset OFF](#)).

**Note:**

*ADC accuracy vs. negative injection current: injecting a negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.*

*Any positive injection current within the limits specified for  $I_{INJ(PIN)}$  and  $S/I_{INJ(PIN)}$  in [Section 5.3.16](#) does not affect the ADC accuracy.*

**Figure 52. Power supply and reference decoupling ( $V_{REF+}$  connected to  $V_{DDA}$ )**

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- $V_{REF+}$  and  $V_{REF-}$  inputs are both available on UFBGA176.  $V_{REF+}$  is also available on LQFP100, LQFP144, and LQFP176. When  $V_{REF+}$  and  $V_{REF-}$  are not available, they are internally connected to  $V_{DDA}$  and  $V_{SSA}$ .

### 5.3.22 Temperature sensor characteristics

**Table 69. Temperature sensor characteristics**

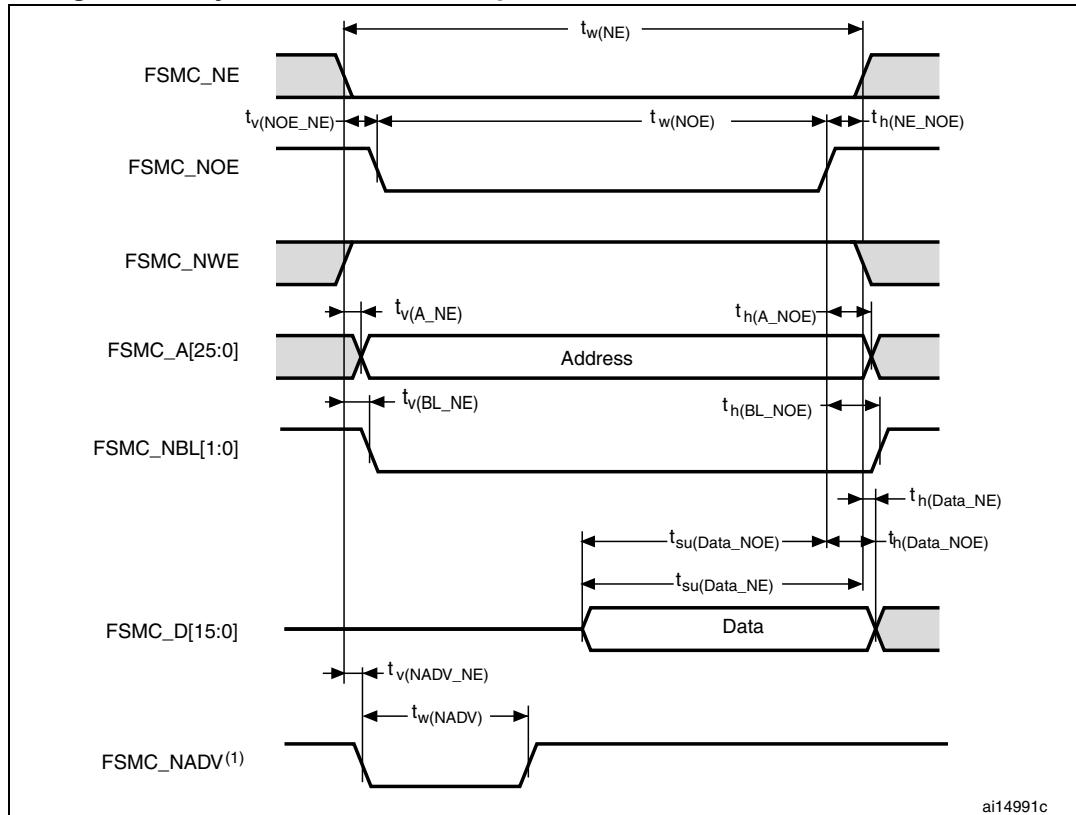
Symbol	Parameter	Min	Typ	Max	Unit
$T_L^{(1)}$	$V_{SENSE}$ linearity with temperature	-	$\pm 1$	$\pm 2$	°C
Avg_Slope <sup>(1)</sup>	Average slope	-	2.5		mV/°C
$V_{25}^{(1)}$	Voltage at 25 °C	-	0.76		V
$t_{START}^{(2)}$	Startup time	-	6	10	μs
$T_{S\_temp}^{(2)}$	ADC sampling time when reading the temperature (1 °C accuracy)	10	-	-	μs

1. Guaranteed by characterization.

2. Guaranteed by design.

**Table 70. Temperature sensor calibration values**

Symbol	Parameter	Memory address
TS_CAL1	TS ADC raw data acquired at temperature of 30 °C, $V_{DDA}=3.3$ V	0x1FFF 7A2C - 0x1FFF 7A2D
TS_CAL2	TS ADC raw data acquired at temperature of 110 °C, $V_{DDA}=3.3$ V	0x1FFF 7A2E - 0x1FFF 7A2F

**Figure 54. Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms**

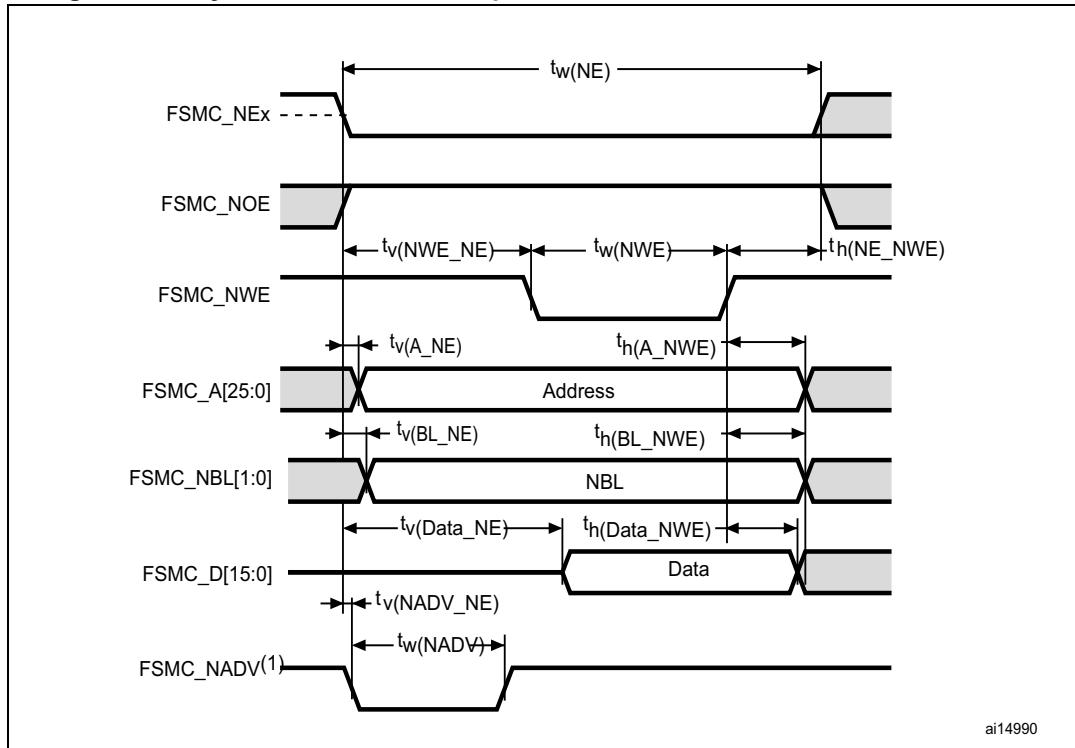
1. Mode 2/B, C and D only. In Mode 1,  $\text{FSMC\_NADV}$  is not used.

**Table 75. Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_w(NE)$	FSMC_NE low time	$2T_{HCLK}-0.5$	$2T_{HCLK}+1$	ns
$t_v(NOE_NE)$	FSMC_NEx low to FSMC_NOE low	0.5	3	ns
$t_w(NOE)$	FSMC_NOE low time	$2T_{HCLK}-2$	$2T_{HCLK}+2$	ns
$t_h(NE\_NOE)$	FSMC_NOE high to FSMC_NE high hold time	0	-	ns
$t_v(A\_NE)$	FSMC_NEx low to FSMC_A valid	-	4.5	ns
$t_h(A\_NOE)$	Address hold time after FSMC_NOE high	4	-	ns
$t_v(BL\_NE)$	FSMC_NEx low to FSMC_BL valid	-	1.5	ns
$t_h(BL\_NOE)$	FSMC_BL hold time after FSMC_NOE high	0	-	ns
$t_{su}(Data\_NE)$	Data to FSMC_NEx high setup time	$T_{HCLK}+4$	-	ns
$t_{su}(Data\_NOE)$	Data to FSMC_NOEx high setup time	$T_{HCLK}+4$	-	ns
$t_h(Data\_NOE)$	Data hold time after FSMC_NOE high	0	-	ns
$t_h(Data\_NE)$	Data hold time after FSMC_NEx high	0	-	ns
$t_v(NADV\_NE)$	FSMC_NEx low to FSMC_NADV low	-	2	ns
$t_w(NADV)$	FSMC_NADV low time	-	$T_{HCLK}$	ns

1.  $C_L = 30 \text{ pF}$ .

2. Guaranteed by characterization.

**Figure 55. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms**

1. Mode 2/B, C and D only. In Mode 1, FSMC\_NADV is not used.

**Table 76. Asynchronous non-multiplexed SRAM/PSRAM/NOR write timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FSMC_NE low time	$3T_{HCLK}$	$3T_{HCLK}+4$	ns
$t_{v(NWE\_NE)}$	FSMC_NEx low to FSMC_NWE low	$T_{HCLK}-0.5$	$T_{HCLK}+0.5$	ns
$t_{w(NWE)}$	FSMC_NWE low time	$T_{HCLK}-1$	$T_{HCLK}+2$	ns
$t_{h(NE\_NWE)}$	FSMC_NWE high to FSMC_NE high hold time	$T_{HCLK}-1$	-	ns
$t_{v(A\_NE)}$	FSMC_NEx low to FSMC_A valid	-	0	ns
$t_{h(A\_NWE)}$	Address hold time after FSMC_NWE high	$T_{HCLK}-2$	-	ns
$t_{v(BL\_NE)}$	FSMC_NEx low to FSMC_BL valid	-	1.5	ns
$t_{h(BL\_NWE)}$	FSMC_BL hold time after FSMC_NWE high	$T_{HCLK}-1$	-	ns
$t_{v(Data\_NE)}$	Data to FSMC_NEx low to Data valid	-	$T_{HCLK}+3$	ns
$t_{h(Data\_NWE)}$	Data hold time after FSMC_NWE high	$T_{HCLK}-1$	-	ns
$t_{v(NADV\_NE)}$	FSMC_NEx low to FSMC_NADV low	-	2	ns
$t_{w(NADV)}$	FSMC_NADV low time	-	$T_{HCLK}+0.5$	ns

1.  $C_L = 30 \text{ pF}$ .
2. Guaranteed by characterization.

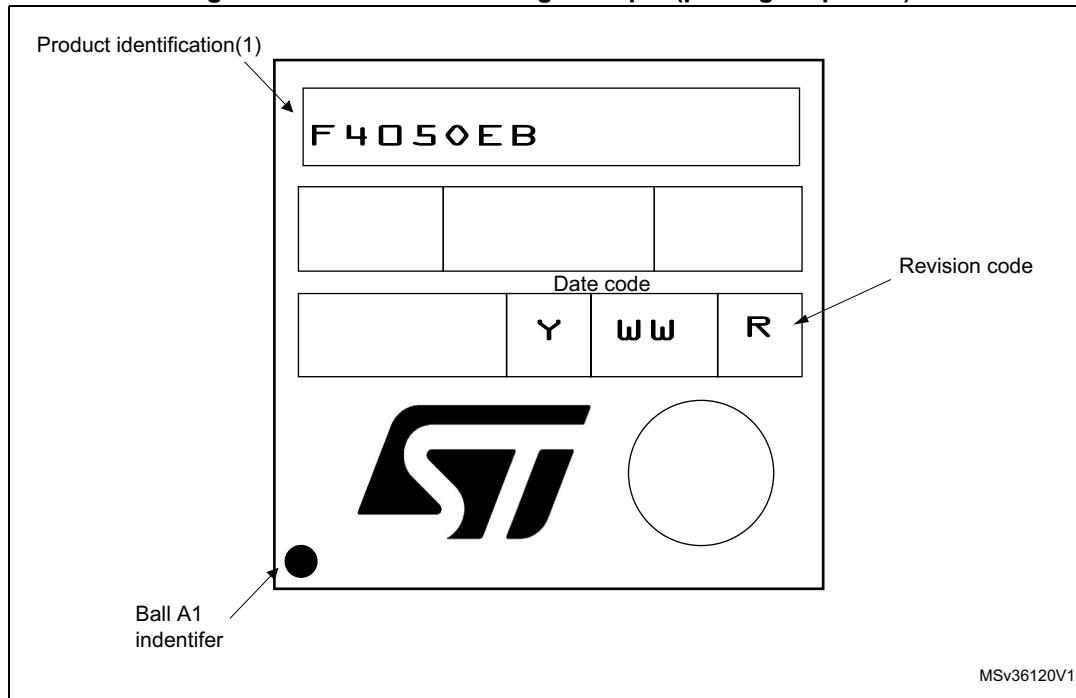
**Table 91. WLCSP90 recommended PCB design rules**

Dimension	Recommended values
Pitch	0.4 mm
Dpad	260 µm max. (circular) 220 µm recommended
Dsm	300 µm min. (for 260 µm diameter pad)
PCB pad design	Non-solder mask defined via underbump allowed

**Device marking for WLCSP90**

The following figure gives an example of topside marking and ball A1 position identifier location.

Other optional marking or inset/upset marks, which depend on supply chain operations, are not indicated below.

**Figure 77. WLCSP90 marking example (package top view)**

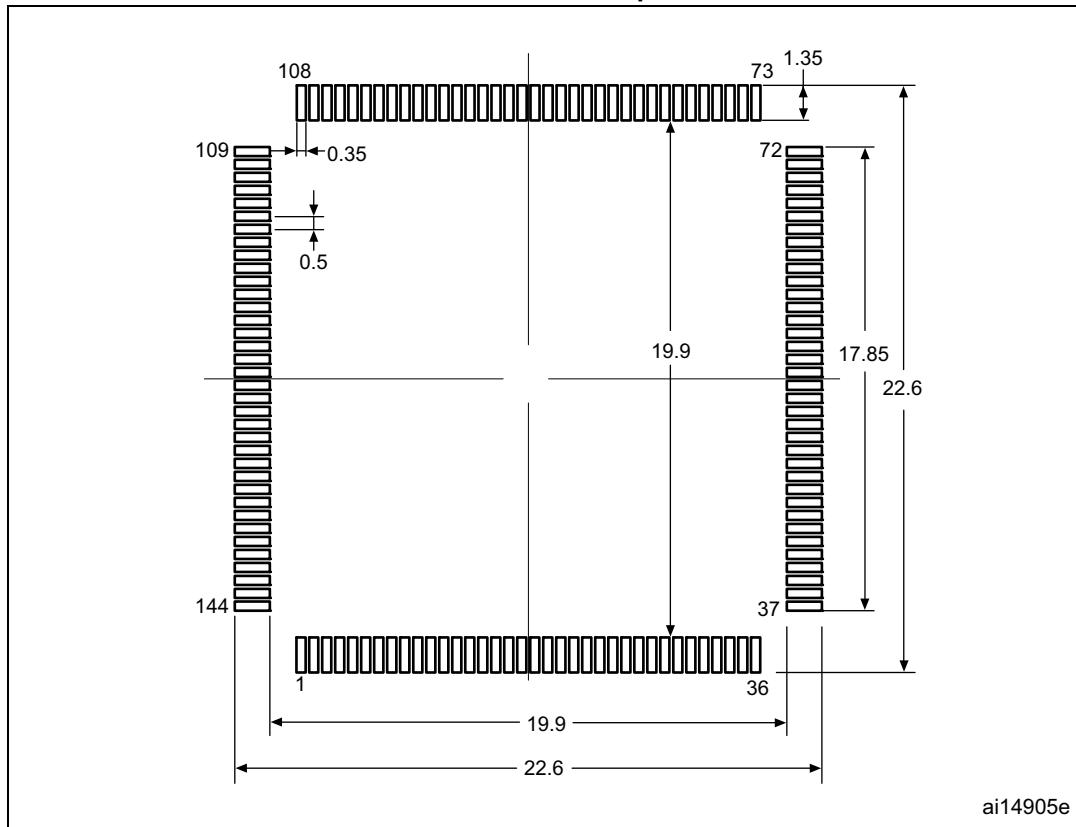
1. Parts marked as "ES", "E" or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering Samples to run qualification activity.

**Table 94. LQFP144 - 144-pin, 20 x 20 mm low-profile quad flat package mechanical data**

<b>Symbol</b>	<b>millimeters</b>			<b>inches<sup>(1)</sup></b>		
	<b>Min</b>	<b>Typ</b>	<b>Max</b>	<b>Min</b>	<b>Typ</b>	<b>Max</b>
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	21.800	22.000	22.200	0.8583	0.8661	0.874
D1	19.800	20.000	20.200	0.7795	0.7874	0.7953
D3	-	17.500	-	-	0.689	-
E	21.800	22.000	22.200	0.8583	0.8661	0.8740
E1	19.800	20.000	20.200	0.7795	0.7874	0.7953
E3	-	17.500	-	-	0.6890	-
e	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.080	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 85. LQFP144 - 144-pin, 20 x 20 mm low-profile quad flat package recommended footprint



1. Dimensions are in millimeters.

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## 6.7 Thermal characteristics

The maximum chip-junction temperature,  $T_J$  max, in degrees Celsius, may be calculated using the following equation:

$$T_J \text{ max} = T_A \text{ max} + (P_D \text{ max} \times \Theta_{JA})$$

Where:

- $T_A$  max is the maximum ambient temperature in °C,
- $\Theta_{JA}$  is the package junction-to-ambient thermal resistance, in °C/W,
- $P_D$  max is the sum of  $P_{INT}$  max and  $P_{I/O}$  max ( $P_D$  max =  $P_{INT}$  max +  $P_{I/O}$  max),
- $P_{INT}$  max is the product of  $I_{DD}$  and  $V_{DD}$ , expressed in Watts. This is the maximum chip internal power.

$P_{I/O}$  max represents the maximum power dissipation on output pins where:

$$P_{I/O} \text{ max} = \sum (V_{OL} \times I_{OL}) + \sum ((V_{DD} - V_{OH}) \times I_{OH}),$$

taking into account the actual  $V_{OL}$  /  $I_{OL}$  and  $V_{OH}$  /  $I_{OH}$  of the I/Os at low and high level in the application.

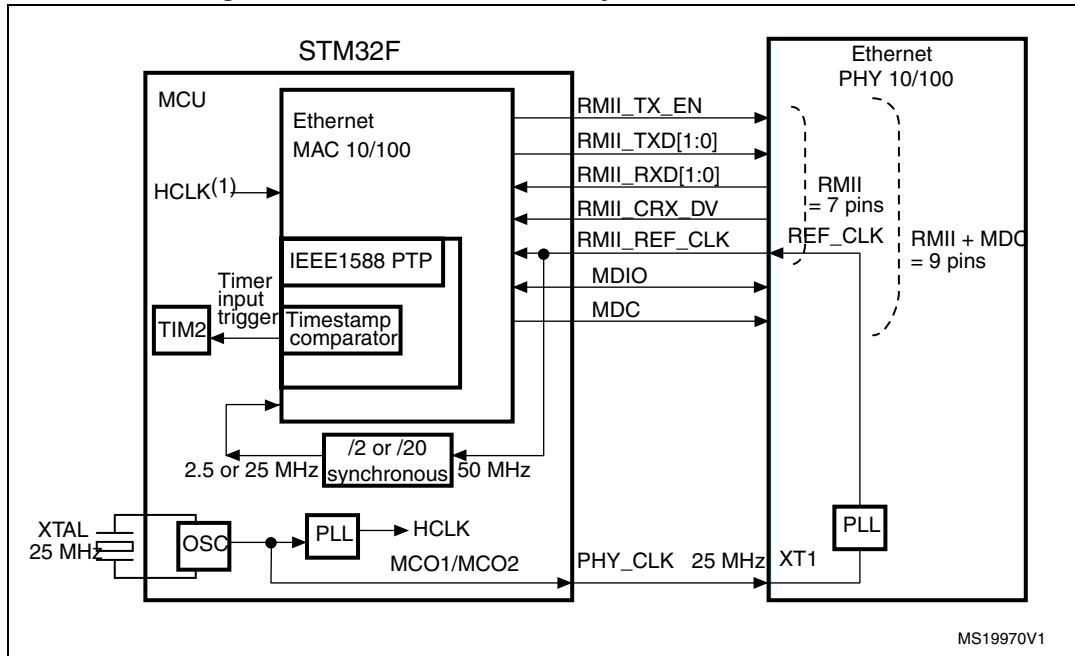
Table 98. Package thermal characteristics

Symbol	Parameter	Value	Unit
$\Theta_{JA}$	<b>Thermal resistance junction-ambient</b> LQFP64 - 10 × 10 mm / 0.5 mm pitch	46	°C/W
	<b>Thermal resistance junction-ambient</b> LQFP100 - 14 × 14 mm / 0.5 mm pitch	43	
	<b>Thermal resistance junction-ambient</b> LQFP144 - 20 × 20 mm / 0.5 mm pitch	40	
	<b>Thermal resistance junction-ambient</b> LQFP176 - 24 × 24 mm / 0.5 mm pitch	38	
	<b>Thermal resistance junction-ambient</b> UFBGA176 - 10× 10 mm / 0.65 mm pitch	39	
	<b>Thermal resistance junction-ambient</b> WLCSP90 - 0.400 mm pitch	38.1	

### Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from [www.jedec.org](http://www.jedec.org).

Figure 99. RMII with a 25 MHz crystal and PHY with PLL



1.  $f_{HCLK}$  must be greater than 25 MHz.
2. The 25 MHz (PHY\_CLK) must be derived directly from the HSE oscillator, before the PLL block.