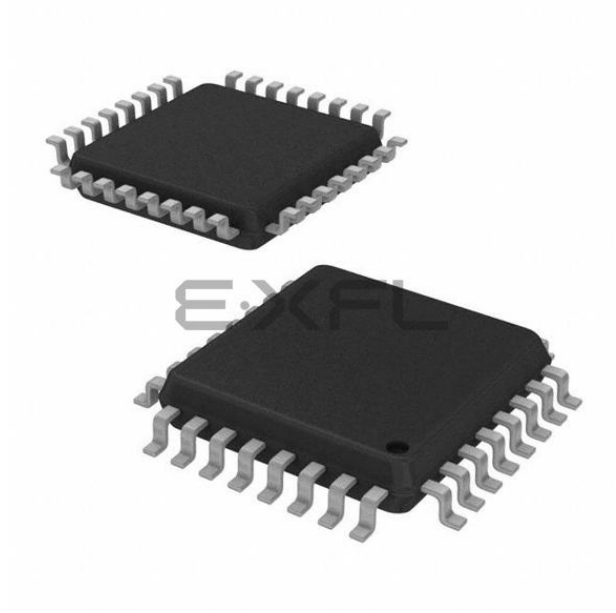

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	ST7
Core Size	8-Bit
Speed	8MHz
Connectivity	SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	24
Program Memory Size	16KB (16K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	3.8V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	32-LQFP
Supplier Device Package	-
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/st72f324k4t6

5 CENTRAL PROCESSING UNIT

5.1 INTRODUCTION

This CPU has a full 8-bit architecture and contains six internal registers allowing efficient 8-bit data manipulation.

5.2 MAIN FEATURES

- Enable executing 63 basic instructions
- Fast 8-bit by 8-bit multiply
- 17 main addressing modes (with indirect addressing mode)
- Two 8-bit index registers
- 16-bit stack pointer
- Low power HALT and WAIT modes
- Priority maskable hardware interrupts
- Non-maskable software/hardware interrupts

5.3 CPU REGISTERS

The 6 CPU registers shown in Figure 8 are not present in the memory mapping and are accessed by specific instructions.

Accumulator (A)

The Accumulator is an 8-bit general purpose register used to hold operands and the results of the arithmetic and logic calculations and to manipulate data.

Index Registers (X and Y)

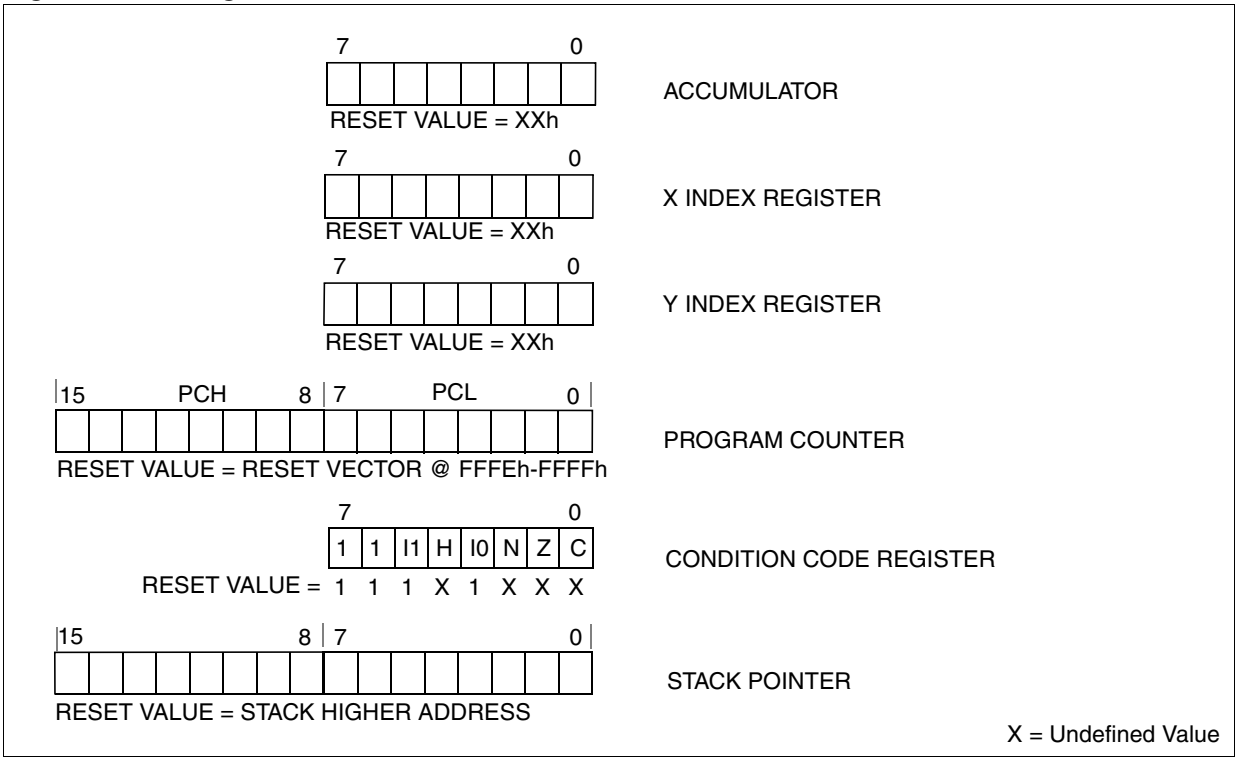
These 8-bit registers are used to create effective addresses or as temporary storage areas for data manipulation. (The Cross-Assembler generates a precede instruction (PRE) to indicate that the following instruction refers to the Y register.)

The Y register is not affected by the interrupt automatic procedures.

Program Counter (PC)

The program counter is a 16-bit register containing the address of the next instruction to be executed by the CPU. It is made of two 8-bit registers PCL (Program Counter Low which is the LSB) and PCH (Program Counter High which is the MSB).

Figure 8. CPU Registers



WATCHDOG TIMER (Cont'd)

10.1.5 Low Power Modes

Mode	Description		
SLOW	No effect on Watchdog.		
WAIT	No effect on Watchdog.		
HALT	**OIE bit in MCCSR register**	**WDGHALT bit in Option Byte**	
	0	0	No Watchdog reset is generated. The MCU enters Halt mode. The Watchdog counter is decremented once and then stops counting and is no longer able to generate a watchdog reset until the MCU receives an external interrupt or a reset. If an external interrupt is received, the Watchdog restarts counting after 256 or 4096 CPU clocks. If a reset is generated, the Watchdog is disabled (reset state) unless Hardware Watchdog is selected by option byte. For application recommendations see Section 10.1.7 below.
	0	1	A reset is generated.
	1	x	No reset is generated. The MCU enters Active Halt mode. The Watchdog counter is not decremented. It stop counting. When the MCU receives an oscillator interrupt or external interrupt, the Watchdog restarts counting immediately. When the MCU receives a reset the Watchdog restarts counting after 256 or 4096 CPU clocks.

10.1.6 Hardware Watchdog Option

If Hardware Watchdog is selected by option byte, the watchdog is always active and the WDGA bit in the WDGCR is not used. Refer to the Option Byte description.

10.1.7 Using Halt Mode with the WDG (WDGHALT option)

The following recommendation applies if Halt mode is used when the watchdog is enabled.

– Before executing the HALT instruction, refresh the WDG counter, to avoid an unexpected WDG reset immediately after waking up the microcontroller.

10.1.8 Interrupts

None.

10.1.9 Register Description

CONTROL REGISTER (WDGCR)

Read/Write

Reset Value: 0111 1111 (7Fh)

7							0
WDGA	T6	T5	T4	T3	T2	T1	T0

Bit 7 = **WDGA** *Activation bit.*
This bit is set by software and only cleared by hardware after a reset. When WDGA = 1, the watchdog can generate a reset.
0: Watchdog disabled
1: Watchdog enabled

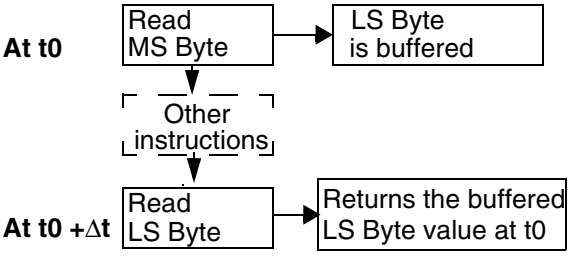
Note: This bit is not used if the hardware watchdog option is enabled by option byte.

Bit 6:0 = **T[6:0]** *7-bit counter (MSB to LSB).*
These bits contain the value of the watchdog counter. It is decremented every 16384 f_{OSC2} cycles (approx.). A reset is produced when it rolls over from 40h to 3Fh (T6 becomes cleared).

16-BIT TIMER (Cont'd)

16-bit read sequence: (from either the Counter Register or the Alternate Counter Register).

Beginning of the sequence



Sequence completed

The user must read the MS Byte first, then the LS Byte value is buffered automatically.

This buffered value remains unchanged until the 16-bit read sequence is completed, even if the user reads the MS Byte several times.

After a complete reading sequence, if only the CLR register or ACLR register are read, they return the LS Byte of the count value at the time of the read.

Whatever the timer mode used (input capture, output compare, one pulse mode or PWM mode) an overflow occurs when the counter rolls over from FFFFh to 0000h then:

– The TOF bit of the SR register is set.

– A timer interrupt is generated if:

  – TOIE bit of the CR1 register is set and

  – I bit of the CC register is cleared.

If one of these conditions is false, the interrupt remains pending to be issued as soon as they are both true.

Clearing the overflow interrupt request is done in two steps:

1. Reading the SR register while the TOF bit is set.
2. An access (read or write) to the CLR register.

Notes: The TOF bit is not cleared by accesses to ACLR register. The advantage of accessing the ACLR register rather than the CLR register is that it allows simultaneous use of the overflow function and reading the free running counter at random times (for example, to measure elapsed time) without the risk of clearing the TOF bit erroneously.

The timer is not affected by WAIT mode.

In HALT mode, the counter stops counting until the mode is exited. Counting then resumes from the previous count (MCU awakened by an interrupt) or from the reset count (MCU awakened by a Reset).

10.3.3.2 External Clock

The external clock (where available) is selected if CC0=1 and CC1=1 in the CR2 register.

The status of the EXEDG bit in the CR2 register determines the type of level transition on the external clock pin EXTCLK that will trigger the free running counter.

The counter is synchronized with the falling edge of the internal CPU clock.

A minimum of four falling edges of the CPU clock must occur between two consecutive active edges of the external clock; thus the external clock frequency must be less than a quarter of the CPU clock frequency.

16-BIT TIMER (Cont'd)

Figure 39. Input Capture Block Diagram

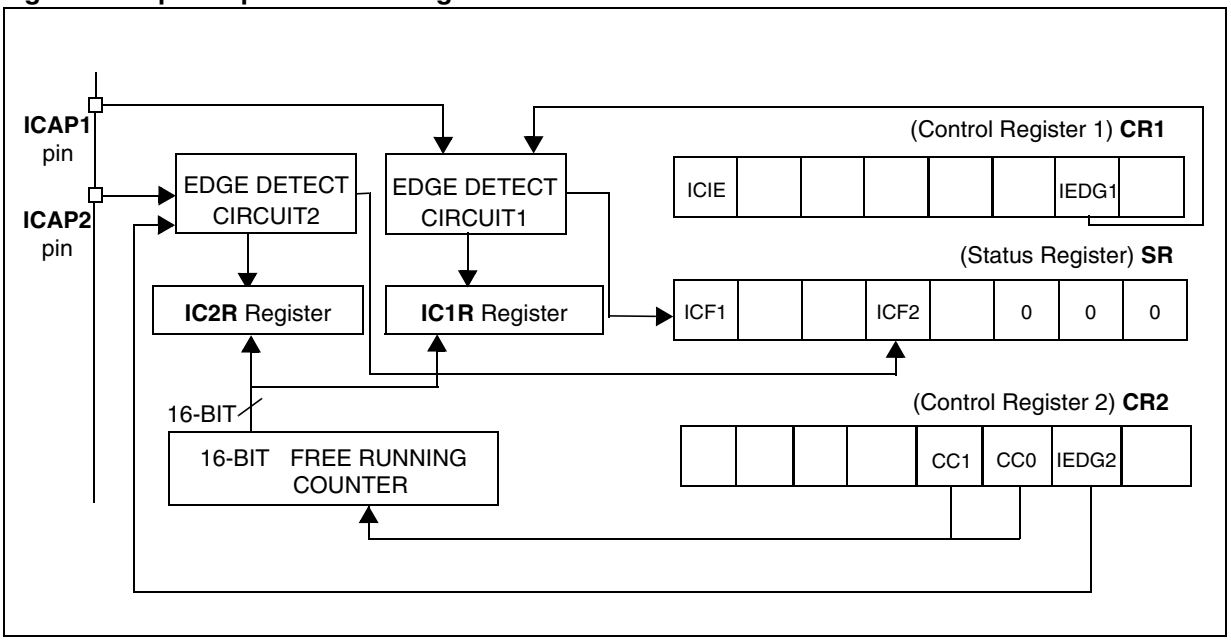
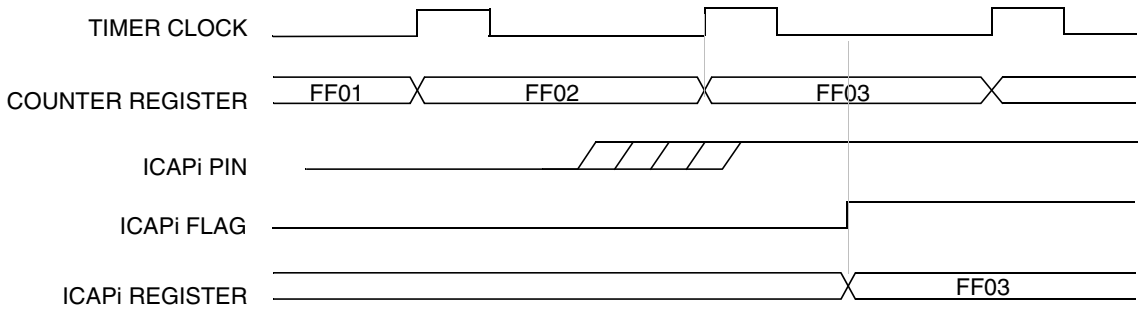


Figure 40. Input Capture Timing Diagram



Note: The rising edge is the active edge.

16-BIT TIMER (Cont'd)

CONTROL REGISTER 2 (CR2)

Read/Write

Reset Value: 0000 0000 (00h)

7							0
OC1E	OC2E	OPM	PWM	CC1	CC0	IEDG2	EXEDG

Bit 7 = **OC1E** *Output Compare 1 Pin Enable.*
This bit is used only to output the signal from the timer on the OCMP1 pin (OLV1 in Output Compare mode, both OLV1 and OLV2 in PWM and one-pulse mode). Whatever the value of the OC1E bit, the Output Compare 1 function of the timer remains active.
0: OCMP1 pin alternate function disabled (I/O pin free for general-purpose I/O).
1: OCMP1 pin alternate function enabled.

Bit 6 = **OC2E** *Output Compare 2 Pin Enable.*
This bit is used only to output the signal from the timer on the OCMP2 pin (OLV2 in Output Compare mode). Whatever the value of the OC2E bit, the Output Compare 2 function of the timer remains active.
0: OCMP2 pin alternate function disabled (I/O pin free for general-purpose I/O).
1: OCMP2 pin alternate function enabled.

Note: In Flash devices, this bit is not available for Timer A. It must be kept at its reset value.

Bit 5 = **OPM** *One Pulse Mode.*
0: One Pulse Mode is not active.
1: One Pulse Mode is active, the ICAP1 pin can be used to trigger one pulse on the OCMP1 pin; the active transition is given by the IEDG1 bit. The length of the generated pulse depends on the contents of the OC1R register.

Bit 4 = **PWM** *Pulse Width Modulation.*
0: PWM mode is not active.
1: PWM mode is active, the OCMP1 pin outputs a programmable cyclic signal; the length of the pulse depends on the value of OC1R register; the period depends on the value of OC2R register.

Bit 3, 2 = **CC[1:0]** *Clock Control.*

The timer clock mode depends on these bits:

Table 16. Clock Control Bits

Timer Clock	CC1	CC0
f_{CPU} / 4	0	0
f_{CPU} / 2	0	1
f_{CPU} / 8	1	0
External Clock (where available)	1	1

Note: If the external clock pin is not available, programming the external clock configuration stops the counter.

Bit 1 = **IEDG2** *Input Edge 2.*
This bit determines which type of level transition on the ICAP2 pin will trigger the capture.
0: A falling edge triggers the capture.
1: A rising edge triggers the capture.

Bit 0 = **EXEDG** *External Clock Edge.*
This bit determines which type of level transition on the external clock pin EXTCLK will trigger the counter register.
0: A falling edge triggers the counter register.
1: A rising edge triggers the counter register.

SERIAL PERIPHERAL INTERFACE (Cont'd)

10.4.3.3 Master Mode Operation

In master mode, the serial clock is output on the SCK pin. The clock frequency, polarity and phase are configured by software (refer to the description of the SPICSR register).

Note: The idle state of SCK must correspond to the polarity selected in the SPICSR register (by pulling up SCK if CPOL=1 or pulling down SCK if CPOL=0).

To operate the SPI in master mode, perform the following steps in order (if the SPICSR register is not written first, the SPICR register setting (MSTR bit) may be not taken into account):

1. Write to the SPICR register:
   - Select the clock frequency by configuring the SPR[2:0] bits.
   - Select the clock polarity and clock phase by configuring the CPOL and CPHA bits. Figure 50 shows the four possible configurations.
     Note: The slave must have the same CPOL and CPHA settings as the master.
2. Write to the SPICSR register:
   - Either set the SSM bit and set the SSI bit or clear the SSM bit and tie the SS pin high for the complete byte transmit sequence.
3. Write to the SPICR register:
   - Set the MSTR and SPE bits
     Note: MSTR and SPE bits remain set only if SS is high).

The transmit sequence begins when software writes a byte in the SPIDR register.

10.4.3.4 Master Mode Transmit Sequence

When software writes to the SPIDR register, the data byte is loaded into the 8-bit shift register and then shifted out serially to the MOSI pin most significant bit first.

When data transfer is complete:

- The SPIF bit is set by hardware
- An interrupt request is generated if the SPIE bit is set and the interrupt mask in the CCR register is cleared.

Clearing the SPIF bit is performed by the following software sequence:

1. An access to the SPICSR register while the SPIF bit is set
2. A read to the SPIDR register.

Note: While the SPIF bit is set, all writes to the SPIDR register are inhibited until the SPICSR register is read.

10.4.3.5 Slave Mode Operation

In slave mode, the serial clock is received on the SCK pin from the master device.

To operate the SPI in slave mode:

1. Write to the SPICSR register to perform the following actions:
   - Select the clock polarity and clock phase by configuring the CPOL and CPHA bits (see Figure 50).
     Note: The slave must have the same CPOL and CPHA settings as the master.
   - Manage the $\overline{SS}$ pin as described in Section 10.4.3.2 and Figure 48. If CPHA=1 $\overline{SS}$ must be held low continuously. If CPHA=0 $\overline{SS}$ must be held low during byte transmission and pulled up between each byte to let the slave write in the shift register.
2. Write to the SPICR register to clear the MSTR bit and set the SPE bit to enable the SPI I/O functions.

10.4.3.6 Slave Mode Transmit Sequence

When software writes to the SPIDR register, the data byte is loaded into the 8-bit shift register and then shifted out serially to the MISO pin most significant bit first.

The transmit sequence begins when the slave device receives the clock signal and the most significant bit of the data on its MOSI pin.

When data transfer is complete:

- The SPIF bit is set by hardware
- An interrupt request is generated if SPIE bit is set and interrupt mask in the CCR register is cleared.

Clearing the SPIF bit is performed by the following software sequence:

1. An access to the SPICSR register while the SPIF bit is set.
2. A write or a read to the SPIDR register.

Notes: While the SPIF bit is set, all writes to the SPIDR register are inhibited until the SPICSR register is read.

The SPIF bit can be cleared during a second transmission; however, it must be cleared before the second SPIF bit in order to prevent an Overrun condition (see Section 10.4.5.2).

SERIAL PERIPHERAL INTERFACE (Cont'd)

10.4.5 Error Flags

10.4.5.1 Master Mode Fault (MODF)

Master mode fault occurs when the master device has its $\overline{SS}$ pin pulled low.

When a Master mode fault occurs:

– The MODF bit is set and an SPI interrupt request is generated if the SPIE bit is set.

– The SPE bit is reset. This blocks all output from the device and disables the SPI peripheral.

– The MSTR bit is reset, thus forcing the device into slave mode.

Clearing the MODF bit is done through a software sequence:

1. A read access to the SPICSR register while the MODF bit is set.

2. A write to the SPICR register.

Notes: To avoid any conflicts in an application with multiple slaves, the $\overline{SS}$ pin must be pulled high during the MODF bit clearing sequence. The SPE and MSTR bits may be restored to their original state during or after this clearing sequence.

Hardware does not allow the user to set the SPE and MSTR bits while the MODF bit is set except in the MODF bit clearing sequence.

10.4.5.2 Overrun Condition (OVR)

An overrun condition occurs, when the master device has sent a data byte and the slave device has not cleared the SPIF bit issued from the previously transmitted byte.

When an Overrun occurs:

– The OVR bit is set and an interrupt request is generated if the SPIE bit is set.

In this case, the receiver buffer contains the byte sent after the SPIF bit was last cleared. A read to the SPIDR register returns this byte. All other bytes are lost.

The OVR bit is cleared by reading the SPICSR register.

10.4.5.3 Write Collision Error (WCOL)

A write collision occurs when the software tries to write to the SPIDR register while a data transfer is taking place with an external device. When this happens, the transfer continues uninterrupted; and the software write will be unsuccessful.

Write collisions can occur both in master and slave mode. See also Section 10.4.3.2 Slave Select Management.

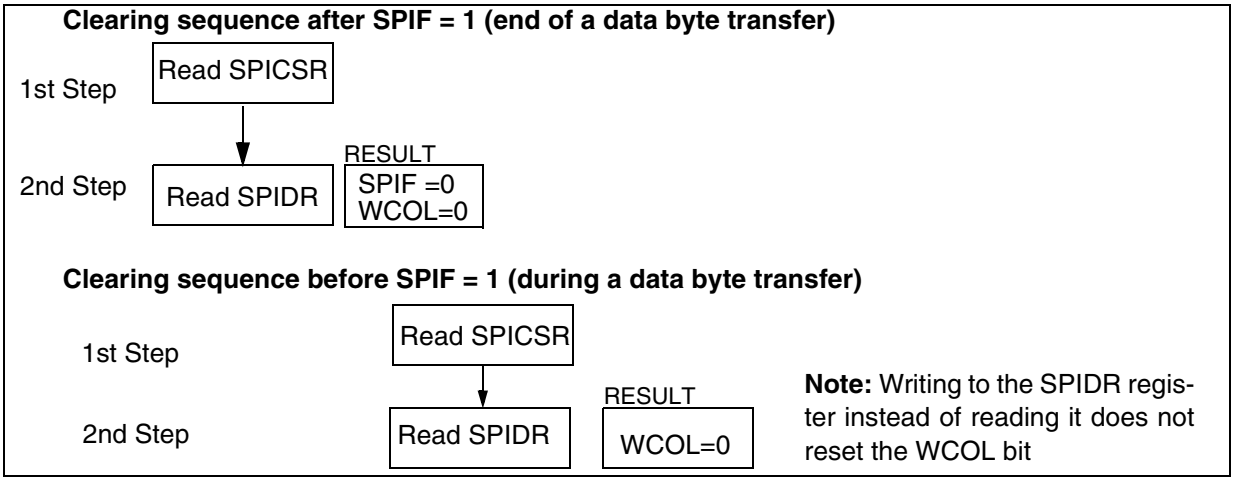
Note: a "read collision" will never occur since the received data byte is placed in a buffer in which access is always synchronous with the MCU operation.

The WCOL bit in the SPICSR register is set if a write collision occurs.

No SPI interrupt is generated when the WCOL bit is set (the WCOL bit is a status flag only).

Clearing the WCOL bit is done through a software sequence (see Figure 51).

Figure 51. Clearing the WCOL bit (Write Collision Flag) Software Sequence



SERIAL COMMUNICATIONS INTERFACE (Cont'd)

10.5.4.3 Receiver

The SCI can receive data words of either 8 or 9 bits. When the M bit is set, word length is 9 bits and the MSB is stored in the R8 bit in the SCICR1 register.

Character reception

During a SCI reception, data shifts in least significant bit first through the RDI pin. In this mode, the SCIDR register consists or a buffer (RDR) between the internal bus and the received shift register (see Figure 1.).

Procedure

– Select the M bit to define the word length.

– Select the desired baud rate using the SCIBRR and the SCIERPR registers.

– Set the RE bit, this enables the receiver which begins searching for a start bit.

When a character is received:

– The RDRF bit is set. It indicates that the content of the shift register is transferred to the RDR.

– An interrupt is generated if the RIE bit is set and the I bit is cleared in the CCR register.

– The error flags can be set if a frame error, noise or an overrun error has been detected during reception.

Clearing the RDRF bit is performed by the following software sequence done by:

1. An access to the SCISR register

2. A read to the SCIDR register.

The RDRF bit must be cleared before the end of the reception of the next character to avoid an overrun error.

Break Character

When a break character is received, the SCI handles it as a framing error.

Idle Character

When a idle frame is detected, there is the same procedure as a data received character plus an interrupt if the ILIE bit is set and the I bit is cleared in the CCR register.

Overrun Error

An overrun error occurs when a character is received when RDRF has not been reset. Data can not be transferred from the shift register to the RDR register as long as the RDRF bit is not cleared.

When an overrun error occurs:

– The OR bit is set.

– The RDR content is not lost.

– The shift register is overwritten.

– An interrupt is generated if the RIE bit is set and the I bit is cleared in the CCR register.

The OR bit is reset by an access to the SCISR register followed by a SCIDR register read operation.

Noise Error

Oversampling techniques are used for data recovery by discriminating between valid incoming data and noise. Normal data bits are considered valid if three consecutive samples (8th, 9th, 10th) have the same bit value, otherwise the NF flag is set. In the case of start bit detection, the NF flag is set on the basis of an algorithm combining both valid edge detection and three samples (8th, 9th, 10th). Therefore, to prevent the NF flag getting set during start bit reception, there should be a valid edge detection as well as three valid samples.

When noise is detected in a frame:

– The NF flag is set at the rising edge of the RDRF bit.

– Data is transferred from the Shift register to the SCIDR register.

– No interrupt is generated. However this bit rises at the same time as the RDRF bit which itself generates an interrupt.

The NF flag is reset by a SCISR register read operation followed by a SCIDR register read operation.

During reception, if a false start bit is detected (e.g. 8th, 9th, 10th samples are 011,101,110), the frame is discarded and the receiving sequence is not started for this frame. There is no RDRF bit set for this frame and the NF flag is set internally (not accessible to the user). This NF flag is accessible along with the RDRF bit when a next valid frame is received.

Note: If the application Start Bit is not long enough to match the above requirements, then the NF Flag may get set due to the short Start Bit. In this case, the NF flag may be ignored by the application software when the first valid byte is received.

See also Section 0.1.4.10 .

INSTRUCTION SET OVERVIEW (Cont'd)

Mnemo	Description	Function/Example	Dst	Src	I1	H	I0	N	Z	C
JRULE	Jump if (C + Z = 1)	Unsigned <=								
LD	Load	dst <= src	reg, M	M, reg				N	Z	
MUL	Multiply	X,A = X * A	A, X, Y	X, Y, A		0				0
NEG	Negate (2's compl)	neg $10	reg, M					N	Z	C
NOP	No Operation									
OR	OR operation	A = A + M	A	M				N	Z	
POP	Pop from the Stack	pop reg	reg	M						
		pop CC	CC	M	I1	H	I0	N	Z	C
PUSH	Push onto the Stack	push Y	M	reg, CC						
RCF	Reset carry flag	C = 0								0
RET	Subroutine Return									
RIM	Enable Interrupts	I1:0 = 10 (level 0)			1		0			
RLC	Rotate left true C	C <= A <= C	reg, M					N	Z	C
RRC	Rotate right true C	C => A => C	reg, M					N	Z	C
RSP	Reset Stack Pointer	S = Max allowed								
SBC	Substract with Carry	A = A - M - C	A	M				N	Z	C
SCF	Set carry flag	C = 1								1
SIM	Disable Interrupts	I1:0 = 11 (level 3)			1		1			
SLA	Shift left Arithmetic	C <= A <= 0	reg, M					N	Z	C
SLL	Shift left Logic	C <= A <= 0	reg, M					N	Z	C
SRL	Shift right Logic	0 => A => C	reg, M					0	Z	C
SRA	Shift right Arithmetic	A7 => A => C	reg, M					N	Z	C
SUB	Substraction	A = A - M	A	M				N	Z	C
SWAP	SWAP nibbles	A7-A4 <=> A3-A0	reg, M					N	Z	
TNZ	Test for Neg & Zero	tnz lbl1						N	Z	
TRAP	S/W trap	S/W interrupt			1		1			
WFI	Wait for Interrupt				1		0			
XOR	Exclusive OR	A = A XOR M	A	M				N	Z	

SUPPLY CURRENT CHARACTERISTICS (Cont'd)

12.5.1.1 Power Consumption vs f_{CPU}: Flash Devices

Figure 61. Typical I_{DD} in RUN mode

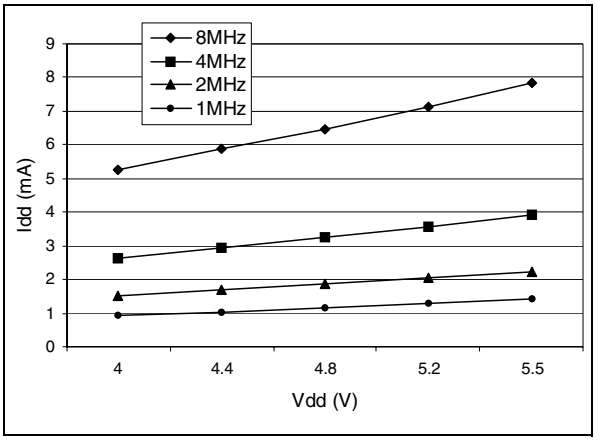


Figure 62. Typical I_{DD} in SLOW mode

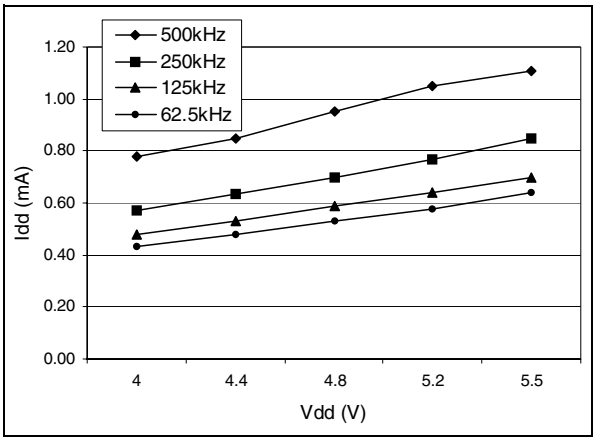


Figure 63. Typical I_{DD} in WAIT mode

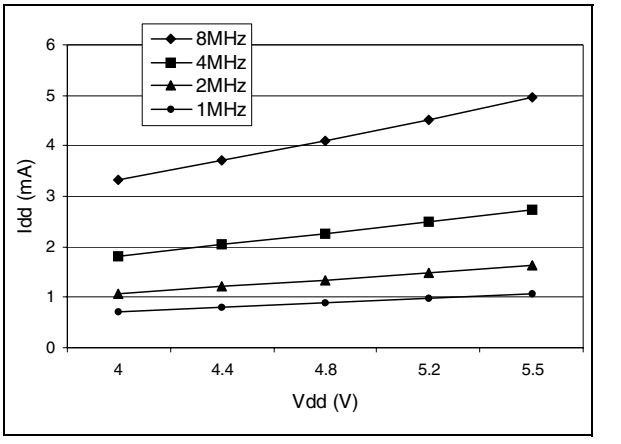
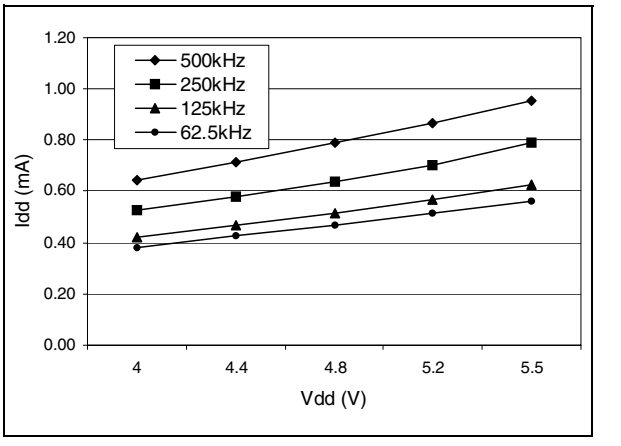


Figure 64. Typ. I_{DD} in SLOW-WAIT mode



12.6 CLOCK AND TIMING CHARACTERISTICS

Subject to general operating conditions for V_{DD}, f_{CPU}, and T_A.

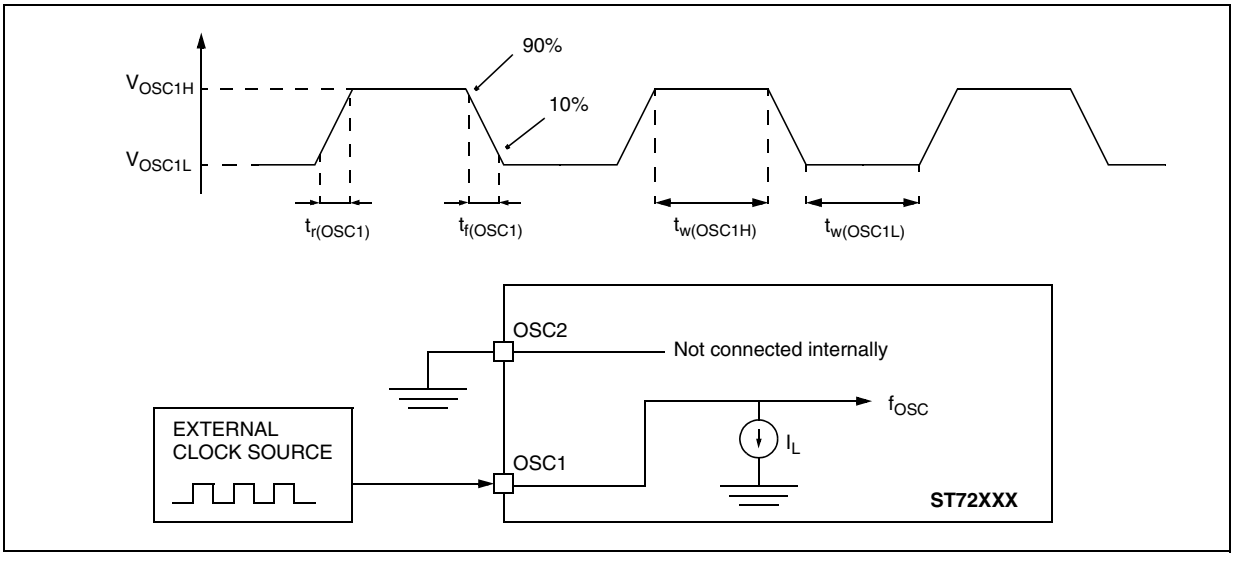
12.6.1 General Timings

Symbol	Parameter	Conditions	Min	Typ [1]	Max	Unit
$t_{c(INST)}$	Instruction cycle time		2	3	12	t_{CPU}
		f_{CPU}=8MHz	250	375	1500	ns
$t_{v(IT)}$	Interrupt reaction time [2] $t_{v(IT)} = \Delta t_{c(INST)} + 10$		10		22	t_{CPU}
		f_{CPU}=8MHz	1.25		2.75	µs

12.6.2 External Clock Source

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{OSC1H}	OSC1 input pin high level voltage	see Figure 65	V_{DD}-1		V_{DD}	V
V_{OSC1L}	OSC1 input pin low level voltage		V_{SS}		V_{SS}+1	
$t_{w(OSC1H)}$ $t_{w(OSC1L)}$	OSC1 high or low time [3]		5			ns
$t_{r(OSC1)}$ $t_{f(OSC1)}$	OSC1 rise or fall time [3]				15	
I_L	OSC1 Input leakage current	$V_{SS} \le V_{IN} \le V_{DD}$			±1	µA

Figure 65. Typical Application with an External Clock Source



Notes:

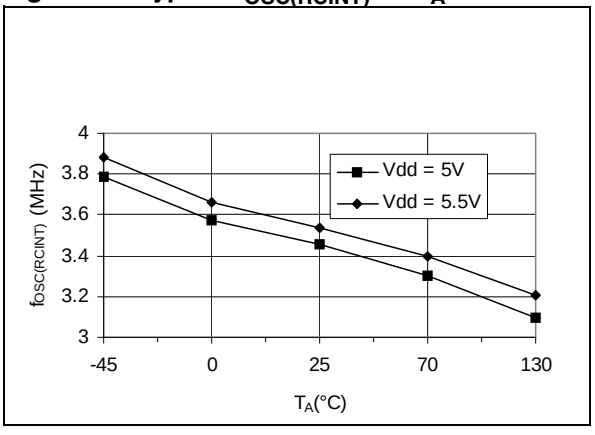
1. Data based on typical application software.
2. Time measured between interrupt event and interrupt vector fetch. $\Delta t_{c(INST)}$ is the number of t_{CPU} cycles needed to finish the current instruction execution.
3. Data based on design simulation and/or technology characteristics, not tested in production.

CLOCK CHARACTERISTICS (Cont'd)

12.6.4 RC Oscillators

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{OSC\ (RCINT)}$	Internal RC oscillator frequency See Figure 67	T_A=25°C, V_{DD}=5V	2	3.5	5.6	MHz

Figure 67. Typical $f_{OSC(RCINT)}$ vs T_A



Note: To reduce disturbance to the RC oscillator, it is recommended to place decoupling capacitors between V_{DD} and V_{SS} as shown in Figure 86

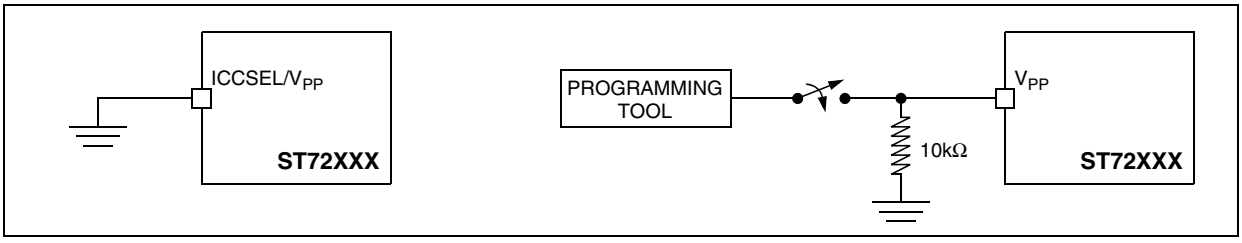
CONTROL PIN CHARACTERISTICS (Cont'd)

12.10.2 ICCSEL/V_{PP} Pin

Subject to general operating conditions for V_{DD}, f_{CPU}, and T_A unless otherwise specified.

Symbol	Parameter	Conditions	Min	Max	Unit
V_{IL}	Input low level voltage [1]		V_{SS}	0.2	V
V_{IH}	Input high level voltage [1]		V_{DD}-0.1	12.6	
I_L	Input leakage current	$V_{IN}=V_{SS}$		±1	µA

Figure 79. Two typical Applications with ICCSEL/V_{PP} Pin [2]



Notes:

1. Data based on design simulation and/or technology characteristics, not tested in production.
2. When ICC mode is not required by the application ICCSEL/V_{PP} pin must be tied to V_{SS}.

ADC CHARACTERISTICS (Cont'd)

Figure 83. R_{AIN} max. vs f_{ADC} with C_{AIN}=0pF[1)]

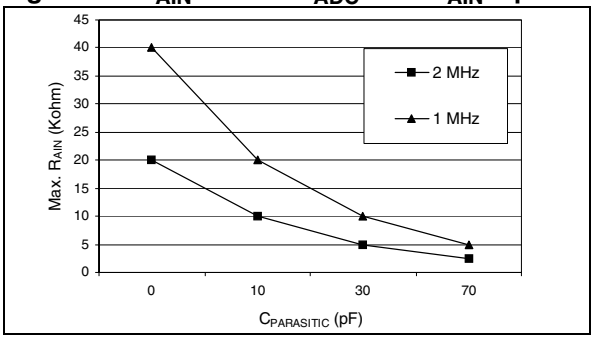


Figure 84. Recommended C_{AIN} & R_{AIN} values.[2)]

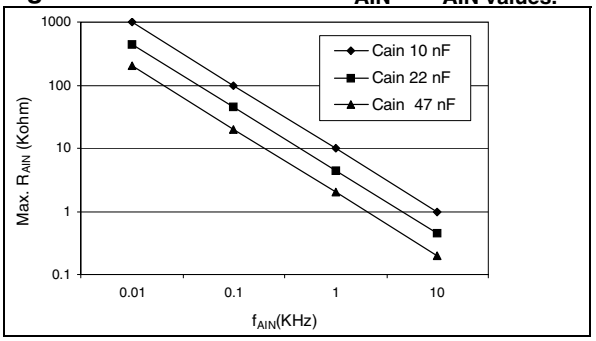
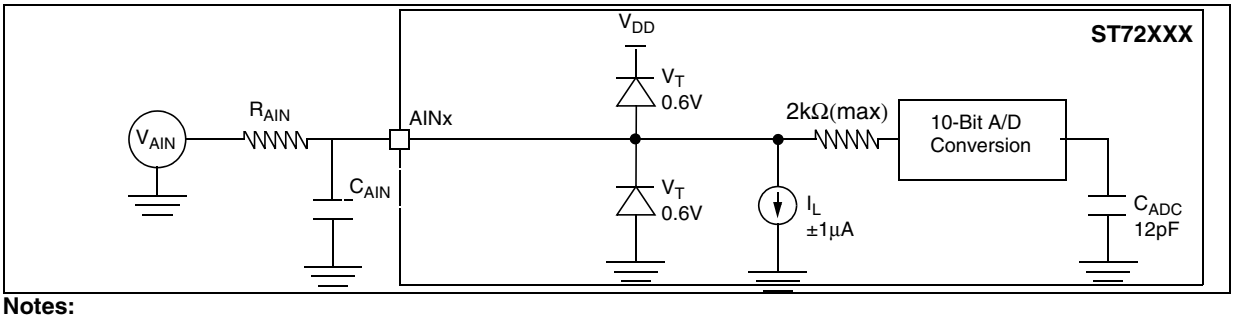


Figure 85. Typical A/D Converter Application



Notes:

1. $C_{PARASITIC}$ represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (3pF). A high $C_{PARASITIC}$ value will downgrade conversion accuracy. To remedy this, f_{ADC} should be reduced.

2. This graph shows that depending on the input signal variation (f_{AIN}), C_{AIN} can be increased for stabilization time and decreased to allow the use of a larger serial resistor (R_{AIN}).

10-BIT ADC CHARACTERISTICS (Cont'd)

12.13.3 ADC Accuracy

Conditions: V_{DD}=5V [1)]

Symbol	Parameter	Conditions	Flash Devices		Unit
			Typ	Max[2)]	
$\|E_T\|$	Total unadjusted error [1)]		4	6	LSB
$\|E_O\|$	Offset error [1)]		3	5	
$\|E_G\|$	Gain Error [1)]		0.5	4.5	
$\|E_D\|$	Differential linearity error [1)]	CPU in run mode @ f_{ADC} 2 MHz.	1.5	4.5	
$\|E_L\|$	Integral linearity error [1)]	CPU in run mode @ f_{ADC} 2 MHz.	1.5	4.5	

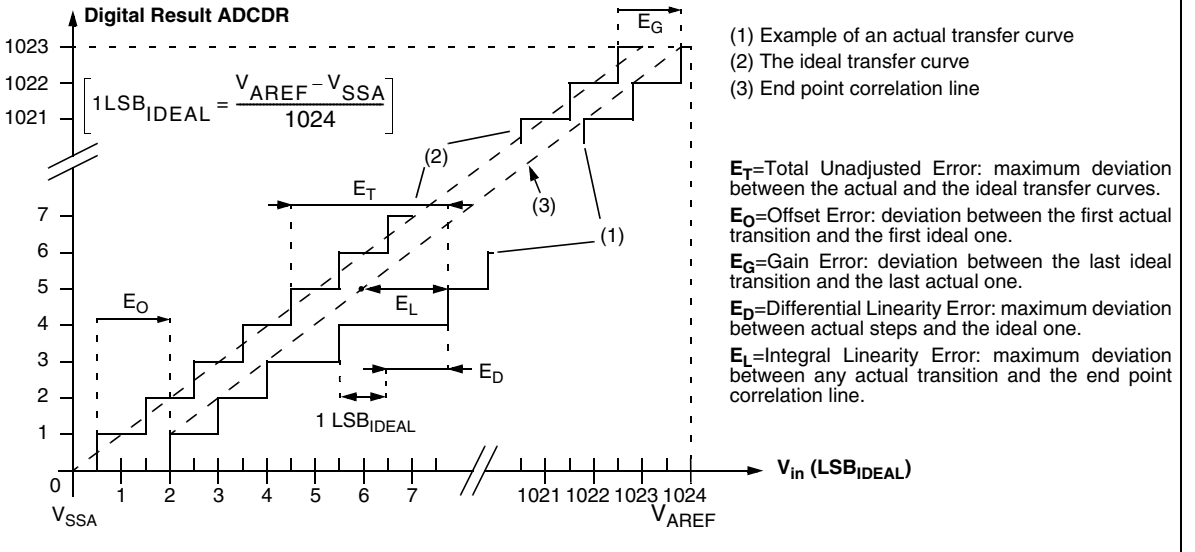
Notes:

1. ADC Accuracy vs. Negative Injection Current: Injecting negative current may reduce the accuracy of the conversion being performed on another analog input.

Any positive injection current within the limits specified for $I_{INJ(PIN)}$ and $\Sigma I_{INJ(PIN)}$ in Section 12.9 does not affect the ADC accuracy.

2. Data based on characterization results, monitored in production to guarantee 99.73% within ± max value from -40°C to 125°C (± 3σ distribution limits).

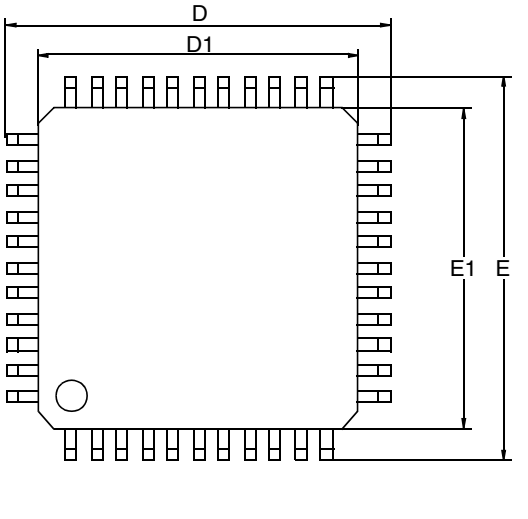
Figure 87. ADC Accuracy Characteristics



13 PACKAGE CHARACTERISTICS

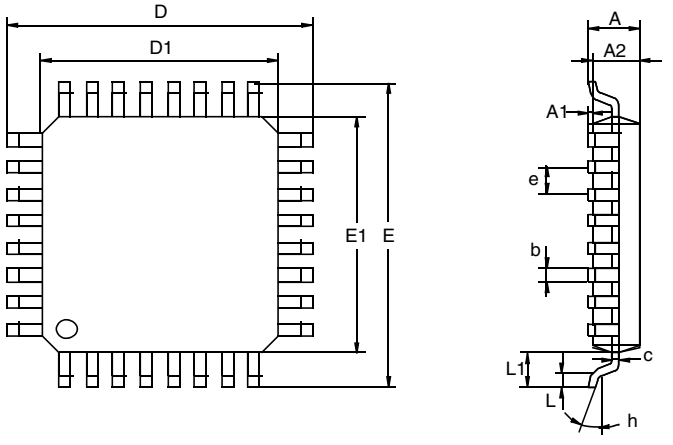
13.1 PACKAGE MECHANICAL DATA

Figure 88. 44-Pin Thin Quad Flat Package



Dim.	mm			inches		
	Min	Typ	Max	Min	Typ	Max
A			1.60			0.063
A1	0.05		0.15	0.002		0.006
A2	1.35	1.40	1.45	0.053	0.055	0.057
b	0.30	0.37	0.45	0.012	0.015	0.018
C	0.09		0.20	0.004	0.000	0.008
D		12.00			0.472	
D1		10.00			0.394	
E		12.00			0.472	
E1		10.00			0.394	
e		0.80			0.031	
θ	0°	3.5°	7°	0°	3.5°	7°
L	0.45	0.60	0.75	0.018	0.024	0.030
L1		1.00			0.039	
	Number of Pins					
N	44					

Figure 89. 32-Pin Thin Quad Flat Package



Dim.	mm			inches		
	Min	Typ	Max	Min	Typ	Max
A			1.60			0.063
A1	0.05		0.15	0.002		0.006
A2	1.35	1.40	1.45	0.053	0.055	0.057
b	0.30	0.37	0.45	0.012	0.015	0.018
C	0.09		0.20	0.004		0.008
D		9.00			0.354	
D1		7.00			0.276	
E		9.00			0.354	
E1		7.00			0.276	
e		0.80			0.031	
θ	0°	3.5°	7°	0°	3.5°	7°
L	0.45	0.60	0.75	0.018	0.024	0.030
L1		1.00			0.039	
	Number of Pins					
N	32					

14.5 ST7 APPLICATION NOTES

Table 30. ST7 Application Notes

IDENTIFICATION	DESCRIPTION
APPLICATION EXAMPLES	
AN1658	SERIAL NUMBERING IMPLEMENTATION
AN1720	MANAGING THE READ-OUT PROTECTION IN FLASH MICROCONTROLLERS
AN1755	A HIGH RESOLUTION/PRECISION THERMOMETER USING ST7 AND NE555
EXAMPLE DRIVERS	
AN 969	SCI COMMUNICATION BETWEEN ST7 AND PC
AN 970	SPI COMMUNICATION BETWEEN ST7 AND EEPROM
AN 972	ST7 SOFTWARE SPI MASTER COMMUNICATION
AN 973	SCI SOFTWARE COMMUNICATION WITH A PC USING ST72251 16-BIT TIMER
AN 974	REAL TIME CLOCK WITH ST7 TIMER OUTPUT COMPARE
AN 976	DRIVING A BUZZER THROUGH ST7 TIMER PWM FUNCTION
AN 979	DRIVING AN ANALOG KEYBOARD WITH THE ST7 ADC
AN 980	ST7 KEYPAD DECODING TECHNIQUES, IMPLEMENTING WAKE-UP ON KEYSTROKE
AN1041	USING ST7 PWM SIGNAL TO GENERATE ANALOG OUTPUT (SINUSOÏD)
AN1044	MULTIPLE INTERRUPT SOURCES MANAGEMENT FOR ST7 MCUS
AN1046	UART EMULATION SOFTWARE
AN1047	MANAGING RECEPTION ERRORS WITH THE ST7 SCI PERIPHERALS
AN1048	ST7 SOFTWARE LCD DRIVER
AN1078	PWM DUTY CYCLE SWITCH IMPLEMENTING TRUE 0% & 100% DUTY CYCLE
AN1445	EMULATED 16 BIT SLAVE SPI
AN1504	STARTING A PWM SIGNAL DIRECTLY AT HIGH LEVEL USING THE ST7 16-BIT TIMER
GENERAL PURPOSE	
AN1476	LOW COST POWER SUPPLY FOR HOME APPLIANCES
AN1709	EMC DESIGN FOR ST MICROCONTROLLERS
AN1752	ST72324 QUICK REFERENCE NOTE
PRODUCT EVALUATION	
AN 910	PERFORMANCE BENCHMARKING
AN 990	ST7 BENEFITS VERSUS INDUSTRY STANDARD
AN1150	BENCHMARK ST72 VS PC16
AN1151	PERFORMANCE COMPARISON BETWEEN ST72254 & PC16F876
AN1278	LIN (LOCAL INTERCONNECT NETWORK) SOLUTIONS
PRODUCT MIGRATION	
AN1131	MIGRATING APPLICATIONS FROM ST72511/311/214/124 TO ST72521/321/324
AN2197	GUIDELINES FOR MIGRATING ST72F324 & ST72F321 APPLICATIONS TO ST72F324B, ST72F321B OR ST72F325
PRODUCT OPTIMIZATION	
AN 982	USING ST7 WITH CERAMIC RESONATOR
AN1014	HOW TO MINIMIZE THE ST7 POWER CONSUMPTION
AN1015	SOFTWARE TECHNIQUES FOR IMPROVING MICROCONTROLLER EMC PERFORMANCE
AN1070	ST7 CHECKSUM SELF-CHECKING CAPABILITY
AN1181	ELECTROSTATIC DISCHARGE SENSITIVE MEASUREMENT
AN1502	EMULATED DATA EEPROM WITH ST7 HDFLASH MEMORY
AN1530	ACCURATE TIMEBASE FOR LOW-COST ST7 APPLICATIONS WITH INTERNAL RC OSCILLATOR
AN1636	UNDERSTANDING AND MINIMIZING ADC CONVERSION ERRORS
PROGRAMMING AND TOOLS	

Table 30. ST7 Application Notes

IDENTIFICATION	DESCRIPTION
AN 978	ST7 VISUAL DEVELOP SOFTWARE KEY DEBUGGING FEATURES
AN 983	KEY FEATURES OF THE COSMIC ST7 C-COMPILER PACKAGE
AN 985	EXECUTING CODE IN ST7 RAM
AN 986	USING THE INDIRECT ADDRESSING MODE WITH ST7
AN 987	ST7 SERIAL TEST CONTROLLER PROGRAMMING
AN 988	STARTING WITH ST7 ASSEMBLY TOOL CHAIN
AN 989	GETTING STARTED WITH THE ST7 HIWARE C TOOLCHAIN
AN1039	ST7 MATH UTILITY ROUTINES
AN1064	WRITING OPTIMIZED HIWARE C LANGUAGE FOR ST7
AN1106	TRANSLATING ASSEMBLY CODE FROM HC05 TO ST7
AN1446	USING THE ST72521 EMULATOR TO DEBUG A ST72324 TARGET APPLICATION
AN1478	PORTING AN ST7 PANTA PROJECT TO CODEWARRIOR IDE
AN1575	ON-BOARD PROGRAMMING METHODS FOR XFLASH AND HDFLASH ST7 MCUS
AN1576	IN-APPLICATION PROGRAMMING (IAP) DRIVERS FOR ST7 HDFLASH OR XFLASH MCUS
AN1635	ST7 CUSTOMER ROM CODE RELEASE INFORMATION
AN1754	DATA LOGGING PROGRAM FOR TESTING ST7 APPLICATIONS VIA ICC
AN1796	FIELD UPDATES FOR FLASH BASED ST7 APPLICATIONS USING A PC COMM PORT
SYSTEM OPTIMIZATION	
AN1711	SOFTWARE TECHNIQUES FOR COMPENSATING ST7 ADC ERRORS

15 KNOWN LIMITATIONS

15.1 ALL DEVICES

15.1.1 External RC option

The External RC clock source option described in previous datasheet revisions is no longer supported and has been removed from this specification.

15.1.2 CSS Function

The Clock Security System function has been removed from the datasheet.

15.1.3 Safe Connection of OSC1/OSC2 Pins

The OSC1 and/or OSC2 pins must not be left unconnected otherwise the ST7 main oscillator may start and, in this configuration, could generate an f_{OSC} clock frequency in excess of the allowed maximum (>16MHz.), putting the ST7 in an unsafe/undefined state. Refer to Section 6.2 on page 24.

15.1.4 Unexpected Reset Fetch

If an interrupt request occurs while a “POP CC” instruction is executed, the interrupt controller does not recognise the source of the interrupt and, by default, passes the RESET vector address to the CPU.

Workaround

To solve this issue, a “POP CC” instruction must always be preceded by a “SIM” instruction.

15.1.5 Clearing active interrupts outside interrupt routine

When an active interrupt request occurs at the same time as the related flag is being cleared, an unwanted reset may occur.

Note: clearing the related interrupt mask will not generate an unwanted reset

Concurrent interrupt context

The symptom does not occur when the interrupts are handled normally, i.e.

when:

– The interrupt flag is cleared within its own interrupt routine

– The interrupt flag is cleared within any interrupt routine

– The interrupt flag is cleared in any part of the code while this interrupt is disabled

If these conditions are not met, the symptom can be avoided by implementing the following sequence:

Perform SIM and RIM operation before and after resetting an active interrupt request.

Example:

SIM

reset interrupt flag

RIM

Nested interrupt context:

The symptom does not occur when the interrupts are handled normally, i.e.

when:

– The interrupt flag is cleared within its own interrupt routine

– The interrupt flag is cleared within any interrupt routine with higher or identical priority level

– The interrupt flag is cleared in any part of the code while this interrupt is disabled

If these conditions are not met, the symptom can be avoided by implementing the following sequence:

PUSH CC

SIM

reset interrupt flag

POP CC

15.1.6 External Interrupt Missed

To avoid any risk of generating a parasitic interrupt, the edge detector is automatically disabled for one clock cycle during an access to either DDR and OR. Any input signal edge during this period will not be detected and will not generate an interrupt.

This case can typically occur if the application refreshes the port configuration registers at intervals during runtime.

Workaround

The workaround is based on software checking the level on the interrupt pin before and after writing to the PxOR or PxDDR registers. If there is a level change (depending on the sensitivity programmed for this pin) the interrupt routine is invoked using the call instruction with three extra PUSH instructions before executing the interrupt routine (this is to make the call compatible with the IRET instruction at the end of the interrupt service routine).

But detection of the level change does ensure that edge occurs during the critical 1 cycle duration and the interrupt has been missed. This may lead to occurrence of same interrupt twice (one hardware and another with software call).

16 IMPORTANT NOTES ON ST72F324B FLASH DEVICES:

With the objective of continuous improvement, ST has developed new ST72F324B devices. These devices are fully compatible with all ROM features and provide an improved price/performance ratio compared to the ST72F324 flash devices.

A summary of the technical improvements is given below.

Refer to separate ST72324B datasheet for the ordering information and full specifications.

16.1 Reset Pin Logic levels

In ST72F324B Flash devices, the V_{IH}/V_{IL} levels for the reset pin are the same as specified for ROM devices

16.2 Wake-Up from Active Halt mode using external interrupts

In ST72F324B Flash devices, any external interrupt that capable of waking-up the MCU from Halt mode can also wake-up the MCU from Active Halt mode. Consequently note 1 below Table 8 on page 36 does not apply to 'B' devices.

16.3 PLL Jitter

In ST72F324B Flash devices, PLL clock accuracy is improved and the jitter is the same as specified for ROM devices

16.4 Active Halt Power Consumption

In ST72F324B Flash devices, the power consumption in Active Halt mode is specified as 230µA max. See Table 12.5.1 on page 120 for test conditions.

16.5 Timer A Registers

In ST72F324B Flash devices, all Timer A registers are present and their functionality is the same as described for ROM devices in the ST72324B datasheet.