NXP USA Inc. - MCF52110CVM80 Datasheet

What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	Coldfire V2
Core Size	32-Bit Single-Core
Speed	80MHz
Connectivity	I²C, SPI, UART/USART
Peripherals	DMA, LVD, POR, PWM, WDT
Number of I/O	56
Program Memory Size	128KB (128K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	16K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 3.6V
Data Converters	A/D 8x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	81-LBGA
Supplier Device Package	81-MAPBGA (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mcf52110cvm80

1.1 Block Diagram

Figure 1 shows a top-level block diagram of the device. Package options for this family are described later in this document.

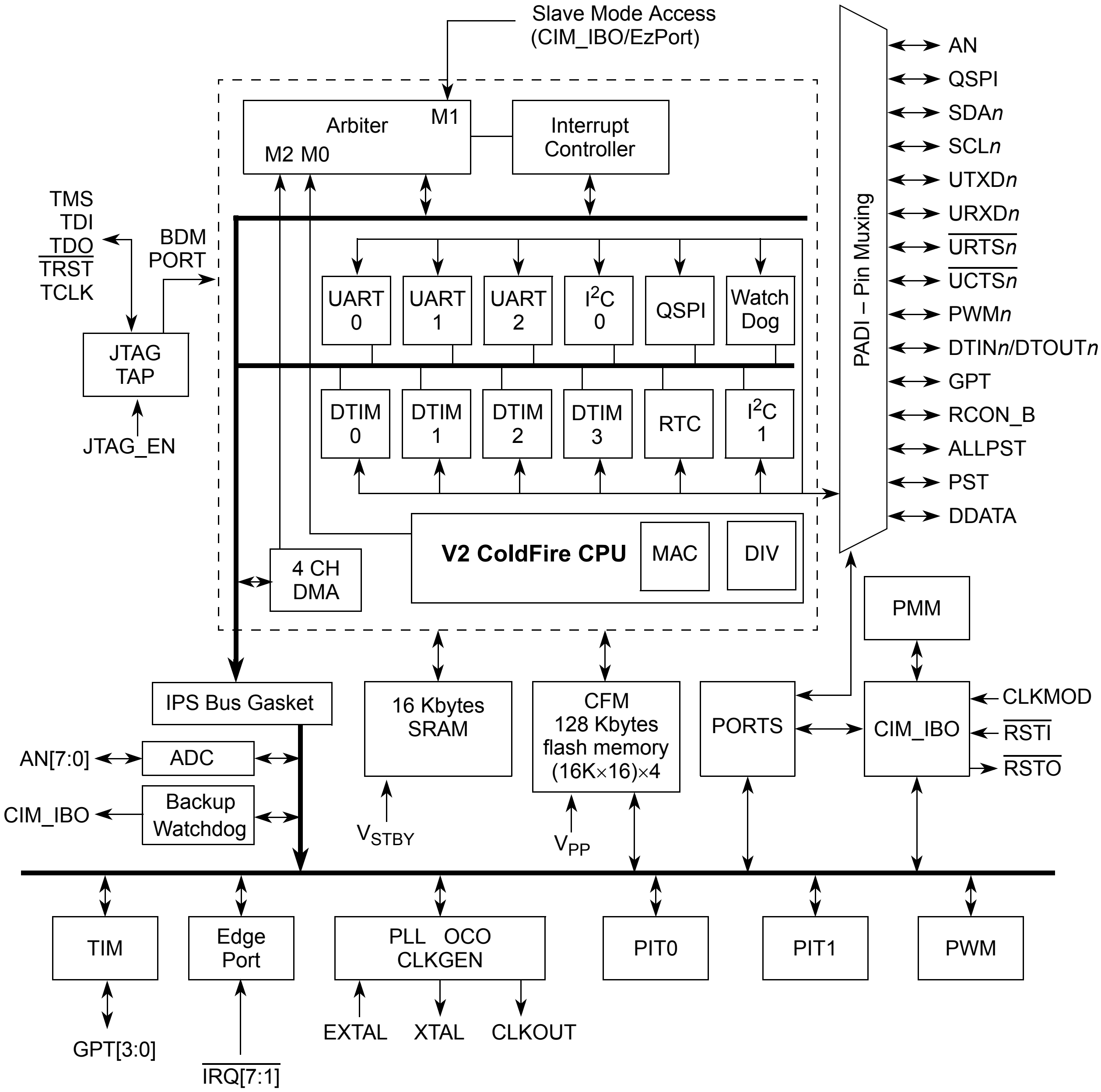


Figure 1. MCF52110 Block Diagram

1.2 Features

1.2.1 Feature Overview

The MCF52110 family includes the following features:

  - Minimum 1.125 µs conversion time
  - Simultaneous sampling of two channels for motor control applications
  - Single-scan or continuous operation
  - Optional interrupts on conversion complete, zero crossing (sign change), or under/over low/high limit
  - Unused analog channels can be used as digital I/O
- Four 32-bit timers with DMA support
  - 12.5 ns resolution at 80 MHz
  - Programmable sources for clock input, including an external clock option
  - Programmable prescaler
  - Input capture capability with programmable trigger edge on input pin
  - Output compare with programmable mode for the output pin
  - Free run and restart modes
  - Maskable interrupts on input capture or output compare
  - DMA trigger capability on input capture or output compare
- Four-channel general purpose timer
  - 16-bit architecture
  - Programmable prescaler
  - Output pulse-widths variable from microseconds to seconds
  - Single 16-bit input pulse accumulator
  - Toggle-on-overflow feature for pulse-width modulator (PWM) generation
  - One dual-mode pulse accumulation channel
- Pulse-width modulation timer
  - Support for PCM mode (resulting in superior signal quality compared to conventional PWM)
  - Operates as eight channels with 8-bit resolution or four channels with 16-bit resolution
  - Programmable period and duty cycle
  - Programmable enable/disable for each channel
  - Software selectable polarity for each channel
  - Period and duty cycle are double buffered. Change takes effect when the end of the current period is reached (PWM counter reaches zero) or when the channel is disabled.
  - Programmable center or left aligned outputs on individual channels
  - Four clock sources (A, B, SA, and SB) provide for a wide range of frequencies
  - Emergency shutdown
- Two periodic interrupt timers (PITs)
  - 16-bit counter
  - Selectable as free running or count down
- Real-Time Clock (RTC)
  - Maintains system time-of-day clock
  - Provides stopwatch and alarm interrupt functions
- Software watchdog timer
  - 32-bit counter
  - Low-power mode support
- Backup watchdog timer (BWT)
  - Independent timer that can be used to help software recover from runaway code
  - 16-bit counter
  - Low-power mode support

  - Unused peripheral pins may be used as extra GPIO
- JTAG support for system level board testing

1.2.2 V2 Core Overview

The version 2 ColdFire processor core is comprised of two separate pipelines decoupled by an instruction buffer. The two-stage instruction fetch pipeline (IFP) is responsible for instruction-address generation and instruction fetch. The instruction buffer is a first-in-first-out (FIFO) buffer that holds prefetched instructions awaiting execution in the operand execution pipeline (OEP). The OEP includes two pipeline stages. The first stage decodes instructions and selects operands (DSOC); the second stage (AGEX) performs instruction execution and calculates operand effective addresses, if needed.

The V2 core implements the ColdFire instruction set architecture revision A+ with support for a separate user stack pointer register and four new instructions to assist in bit processing. Additionally, the core includes the multiply-accumulate (MAC) unit for improved signal processing capabilities. The MAC implements a three-stage arithmetic pipeline, optimized for 16x16 bit operations, with support for one 32-bit accumulator. Supported operands include 16- and 32-bit signed and unsigned integers, signed fractional operands, and a complete set of instructions to process these data types. The MAC provides support for execution of DSP operations within the context of a single processor at a minimal hardware cost.

1.2.3 Integrated Debug Module

The ColdFire processor core debug interface is provided to support system debugging with low-cost debug and emulator development tools. Through a standard debug interface, access to debug information and real-time tracing capability is provided on 100-lead packages. This allows the processor and system to be debugged at full speed without the need for costly in-circuit emulators.

The on-chip breakpoint resources include a total of nine programmable 32-bit registers: an address and an address mask register, a data and a data mask register, four PC registers, and one PC mask register. These registers can be accessed through the dedicated debug serial communication channel or from the processor's supervisor mode programming model. The breakpoint registers can be configured to generate triggers by combining the address, data, and PC conditions in a variety of single- or dual-level definitions. The trigger event can be programmed to generate a processor halt or initiate a debug interrupt exception. This device implements revision B+ of the ColdFire Debug Architecture.

The processor's interrupt servicing options during emulator mode allow real-time critical interrupt service routines to be serviced while processing a debug interrupt event. This ensures the system continues to operate even during debugging.

To support program trace, the V2 debug module provides processor status (PST[3:0]) and debug data (DDATA[3:0]) ports. These buses and the PSTCLK output provide execution status, captured operand data, and branch target addresses defining processor activity at the CPU's clock rate. The device includes a new debug signal, ALLPST. This signal is the logical AND of the processor status (PST[3:0]) signals and is useful for detecting when the processor is in a halted state (PST[3:0] = 1111).

The full debug/trace interface is available only on the 100-pin packages. However, every product features the dedicated debug serial communication channel (DSI, DSO, DSCLK) and the ALLPST signal.

1.2.4 JTAG

The processor supports circuit board test strategies based on the Test Technology Committee of IEEE and the Joint Test Action Group (JTAG). The test logic includes a test access port (TAP) consisting of a 16-state controller, an instruction register, and three test registers (a 1-bit bypass register, a 256-bit boundary-scan register, and a 32-bit ID register). The boundary scan register links the device's pins into one shift register. Test logic, implemented using static logic design, is independent of the device system logic.

The device implementation can:

- Perform boundary-scan operations to test circuit board electrical continuity
- Sample system pins during operation and transparently shift out the result in the boundary scan register

- Bypass the device for a given circuit board test by effectively reducing the boundary-scan register to a single bit
- Disable the output drive to pins during circuit-board testing
- Drive output pins to stable levels

1.2.5 On-Chip Memories

1.2.5.1 SRAM

The dual-ported SRAM module provides a general-purpose 16-Kbyte memory block that the ColdFire core can access in a single cycle. The location of the memory block can be set to any 16-Kbyte boundary within the 4-Gbyte address space. This memory is ideal for storing critical code or data structures and for use as the system stack. Because the SRAM module is physically connected to the processor's high-speed local bus, it can quickly service core-initiated accesses or memory-referencing commands from the debug module.

The SRAM module is also accessible by the DMA. The dual-ported nature of the SRAM makes it ideal for implementing applications with double-buffer schemes, where the processor and a DMA device operate in alternate regions of the SRAM to maximize system performance.

1.2.5.2 Flash Memory

The ColdFire flash module (CFM) is a non-volatile memory (NVM) module that connects to the processor’s high-speed local bus. The CFM is constructed with up to four banks of 16-Kbyte×16-bit flash memory arrays to generate up to 128 Kbytes of 32-bit flash memory. These electrically erasable and programmable arrays serve as non-volatile program and data memory. The flash memory is ideal for program and data storage for single-chip applications, allowing for field reprogramming without requiring an external high voltage source. The CFM interfaces to the ColdFire core through an optimized read-only memory controller that supports interleaved accesses from the 2-cycle flash memory arrays. A backdoor mapping of the flash memory is used for all program, erase, and verify operations, as well as providing a read datapath for the DMA. Flash memory may also be programmed via the EzPort, which is a serial flash memory programming interface that allows the flash memory to be read, erased and programmed by an external controller in a format compatible with most SPI bus flash memory chips.

1.2.6 Power Management

The device incorporates several low-power modes of operation entered under program control and exited by several external trigger events. An integrated power-on reset (POR) circuit monitors the input supply and forces an MCU reset as the supply voltage rises. The low voltage detector (LVD) monitors the supply voltage and is configurable to force a reset or interrupt condition if it falls below the LVD trip point. The RAM standby switch provides power to RAM when the supply voltage to the chip falls below the standby battery voltage. The peripheral clocks may be controlled on an individual basis for power reduction.

1.2.7 UARTs

The device has three full-duplex UARTs that function independently. The three UARTs can be clocked by the system bus clock, eliminating the need for an external clock source. On smaller packages, the third UART is multiplexed with other digital I/O functions. The UARTs are capable of generating DMA requests as well as interrupts.

1.2.8 I^2C Bus

The processor includes two I^2C modules. The I^2C bus is an industry-standard, two-wire, bidirectional serial bus that provides a simple, efficient method of data exchange and minimizes the interconnection between devices. This bus is suitable for applications requiring occasional communications over a short distance between many devices.

1.2.9 QSPI

The queued serial peripheral interface (QSPI) provides a synchronous serial peripheral interface with queued transfer capability. It allows up to 16 transfers to be queued at once, minimizing the need for CPU intervention between transfers.

1.2.10 Fast ADC

The fast ADC consists of an eight-channel input select multiplexer and two independent sample and hold (S/H) circuits feeding separate 12-bit ADCs. The two separate converters store their results in accessible buffers for further processing.

The ADC can be configured to perform a single scan and halt, a scan when triggered, or a programmed scan sequence repeatedly until manually stopped.

The ADC can be configured for sequential or simultaneous conversion. When configured for sequential conversions, up to eight channels can be sampled and stored in any order specified by the channel list register. Both ADCs may be required during a scan, depending on the inputs to be sampled.

During a simultaneous conversion, both S/H circuits are used to capture two different channels at the same time. This configuration requires that a single channel may not be sampled by both S/H circuits simultaneously.

Optional interrupts can be generated at the end of the scan sequence if a channel is out of range (measures below the low threshold limit or above the high threshold limit set in the limit registers) or at several different zero crossing conditions.

1.2.11 DMA Timers (DTIM0–DTIM3)

There are four independent, DMA transfer capable 32-bit timers (DTIM0, DTIM1, DTIM2, and DTIM3) on the device. Each module incorporates a 32-bit timer with a separate register set for configuration and control. The timers can be configured to operate from the system clock or from an external clock source using one of the DTIN*n* signals. If the system clock is selected, it can be divided by 16 or 1. The input clock is further divided by a user-programmable 8-bit prescaler that clocks the actual timer counter register (TCR*n*). Each of these timers can be configured for input capture or reference (output) compare mode. Timer events may optionally cause interrupt requests or DMA transfers.

1.2.12 General Purpose Timer (GPT)

The general purpose timer (GPT) is a four-channel timer module consisting of a 16-bit programmable counter driven by a seven-stage programmable prescaler. Each of the four channels can be configured for input capture or output compare. Additionally, channel three, can be configured as a pulse accumulator.

A timer overflow function allows software to extend the timing capability of the system beyond the 16-bit range of the counter. The input capture and output compare functions allow simultaneous input waveform measurements and output waveform generation. The input capture function can capture the time of a selected transition edge. The output compare function can generate output waveforms and timer software delays. The 16-bit pulse accumulator can operate as a simple event counter or a gated time accumulator.

1.2.13 Periodic Interrupt Timers (PIT0 and PIT1)

The two periodic interrupt timers (PIT0 and PIT1) are 16-bit timers that provide interrupts at regular intervals with minimal processor intervention. Each timer can count down from the value written in its PIT modulus register or it can be a free-running down-counter.

Table 3. Pin Functions by Primary and Alternate Purpose (continued)

Pin Group	Primary Function	Secondary Function	Tertiary Function	Quaternary Function	Drive Strength / Control[1]	Slew Rate / Control[1]	Pull-up / Pull-down[2]	Pin on 100 LQFP	Pin on 81 MAPBGA	Pin on 64 LQFP/QFN
QSPI	QSPI_DIN/ EZPD	SDA1	URXD1	GPIO	PDSR[2]	PSRR[2]	—	16	F3	12
	QSPI_DOUT/ EZPQ	SCL1	UTXD1	GPIO	PDSR[1]	PSRR[1]	—	17	G1	13
	QSPI_CLK/ EZPCK	SCL0	$\overline{\text{URTS1}}$	GPIO	PDSR[3]	PSRR[3]	pull-up[7]	18	G2	14
	QSPI_CS3	SYNCA	SYNCB	GPIO	PDSR[7]	PSRR[7]		12	F1	—
	QSPI_CS2	SYNCB	—	GPIO	PDSR[6]	PSRR[6]		13	F2	—
	QSPI_CS1	—	—	GPIO	PDSR[5]	PSRR[5]	—	19	H2	—
	QSPI_CS0	SDA0	$\overline{\text{UCTS1}}$	GPIO	PDSR[4]	PSRR[4]	pull-up[7]	20	H1	15
Reset[8]	$\overline{\text{RSTI}}$	—	—	—	N/A	N/A	pull-up[8]	96	A3	59
	$\overline{\text{RSTO}}$	—	—	—	high	FAST	—	97	B3	60
Test	TEST	—	—	—	N/A	N/A	pull-down	5	C2	3
Timers, 16-bit	GPT3	—	PWM7	GPIO	PDSR[23]	PSRR[23]	pull-up[9]	62	D8	43
	GPT2	—	PWM5	GPIO	PDSR[22]	PSRR[22]	pull-up[9]	61	D9	42
	GPT1	—	PWM3	GPIO	PDSR[21]	PSRR[21]	pull-up[9]	59	E9	41
	GPT0	—	PWM1	GPIO	PDSR[20]	PSRR[20]	pull-up[9]	58	F7	40
Timers, 32-bit	DTIN3	DTOUT3	PWM6	GPIO	PDSR[19]	PSRR[19]	—	32	H3	19
	DTIN2	DTOUT2	PWM4	GPIO	PDSR[18]	PSRR[18]	—	31	J3	18
	DTIN1	DTOUT1	PWM2	GPIO	PDSR[17]	PSRR[17]	—	37	G4	23
	DTIN0	DTOUT0	PWM0	GPIO	PDSR[16]	PSRR[16]	—	36	H4	22
UART 0	$\overline{\text{UCTS0}}$	—	—	GPIO	PDSR[11]	PSRR[11]	—	6	C1	4
	$\overline{\text{URTS0}}$	—	—	GPIO	PDSR[10]	PSRR[10]	—	9	D3	7
	URXD0	RTC_EXTAL	—	GPIO	PDSR[9]	PSRR[9]	—	7	D1	5
	UTXD0	RTC_XTAL	—	GPIO	PDSR[8]	PSRR[8]	—	8	D2	6

Table 3. Pin Functions by Primary and Alternate Purpose (continued)

Pin Group	Primary Function	Secondary Function	Tertiary Function	Quaternary Function	Drive Strength / Control[1]	Slew Rate / Control[1]	Pull-up / Pull-down[2]	Pin on 100 LQFP	Pin on 81 MAPBGA	Pin on 64 LQFP/QFN
UART 1	$\overline{\text{UCTS1}}$	SYNCA	URXD2	GPIO	PDSR[15]	PSRR[15]	—	98	C3	61
	$\overline{\text{URTS1}}$	SYNCB	UTXD2	GPIO	PDSR[14]	PSRR[14]	—	4	B1	2
	URXD1	SDA1	—	GPIO	PDSR[13]	PSRR[13]	—	100	B2	63
	UTXD1	SCL1	—	GPIO	PDSR[12]	PSRR[12]	—	99	A2	62
UART 2	$\overline{\text{UCTS2}}$	SCL1	—	GPIO	PDSR[27]	PSRR[27]	—	27	—	—
	$\overline{\text{URTS2}}$	SDA1	—	GPIO	PDSR[26]	PSRR[26]	—	30	—	—
	URXD2	—	—	GPIO	PDSR[25]	PSRR[25]	—	28	—	—
	UTXD2	—	—	GPIO	PDSR[24]	PSRR[24]	—	29	—	—
VSTBY	VSTBY	—	—	—	N/A	N/A	—	55	F8	37
VDD	VDD	—	—	—	N/A	N/A	—	1,2,14,22, 23,34,41, 57,68,81,93	D5,E3–E7, F5	1,10,20,39,52
VSS	VSS	—	—	—	N/A	N/A	—	3,15,24,25,35,42,56, 67,75,82,92	A1,A9,D4,D6,F4,F6,J1	11,21,38, 53,64

[1] The PDSR and PSSR registers are described in the General Purpose I/O chapter. All programmable signals default to 2 mA drive and FAST slew rate in normal (single-chip) mode.

[2] All signals have a pull-up in GPIO mode.

[3] These signals are multiplexed on other pins.

[4] For primary and GPIO functions only.

[5] Only when JTAG mode is enabled.

[6] CLKMOD0 and CLKMOD1 have internal pull-down resistors; however, the use of external resistors is very strongly recommended.

[7] For secondary and GPIO functions only.

[8] $\overline{\text{RSTI}}$ has an internal pull-up resistor; however, the use of an external resistor is very strongly recommended.

[9] For GPIO function. Primary Function has pull-up control within the GPT module.

1.3 Reset Signals

Table 4 describes signals used to reset the chip or as a reset indication.

Table 4. Reset Signals

Signal Name	Abbreviation	Function	I/O
Reset In	$\overline{\text{RSTI}}$	Primary reset input to the device. Asserting $\overline{\text{RSTI}}$ for at least 8 CPU clock cycles immediately resets the CPU and peripherals.	I
Reset Out	$\overline{\text{RSTO}}$	Driven low for 1024 CPU clocks after the reset source has deasserted.	O

1.4 PLL and Clock Signals

Table 5 describes signals used to support the on-chip clock generation circuitry.

Table 5. PLL and Clock Signals

Signal Name	Abbreviation	Function	I/O
External Clock In	EXTAL	Crystal oscillator or external clock input except when the on-chip relaxation oscillator is used.	I
Crystal	XTAL	Crystal oscillator output except when CLKMOD0=0, then sampled as part of the clock mode selection mechanism.	O
Clock Out	CLKOUT	This output signal reflects the internal system clock.	O

1.5 Mode Selection

Table 6 describes signals used in mode selection; Table 7 describes the particular clocking modes.

Table 6. Mode Selection Signals

Signal Name	Abbreviation	Function	I/O
Clock Mode Selection	CLKMOD[1:0]	Selects the clock boot mode.	I
Reset Configuration	RCON	The Serial Flash Programming mode is entered by asserting the $\overline{\text{RCON}}$ pin (with the TEST pin negated) as the chip comes out of reset. During this mode, the EzPort has access to the flash memory which can be programmed from an external device.	
Test	TEST	Reserved for factory testing only and in normal modes of operation should be connected to VSS to prevent unintentional activation of test functions.	I

Table 7. Clocking Modes

CLKMOD[1:0]	XTAL	Configure the clock mode.
00	0	PLL disabled, clock driven by external oscillator
00	1	PLL disabled, clock driven by on-chip oscillator
01	N/A	PLL disabled, clock driven by crystal
10	0	PLL in normal mode, clock driven by external oscillator[1]
10	1	Reserved[2]
11	N/A	PLL in normal mode, clock driven by crystal

Table 16. Debug Support Signals (continued)

Signal Name	Abbreviation	Function	I/O
Development Serial Input	DSI	Development Serial Input - Internally synchronized input that provides data input for the serial communication port to the debug module, after the DSCLK has been seen as high (logic 1).	I
Development Serial Output	DSO	Development Serial Output - Provides serial output communication for debug module responses. DSO is registered internally. The output is delayed from the validation of DSCLK high.	O
Debug Data	DDATA[3:0]	Display captured processor data and breakpoint status. The CLKOUT signal can be used by the development system to know when to sample DDATA[3:0].	O
Processor Status Clock	PSTCLK	Processor Status Clock - Delayed version of the processor clock. Its rising edge appears in the center of valid PST and DDATA output. PSTCLK indicates when the development system should sample PST and DDATA values. If real-time trace is not used, setting CSR[PCD] keeps PSTCLK, and PST and DDATA outputs from toggling without disabling triggers. Non-quiescent operation can be reenabled by clearing CSR[PCD], although the external development systems must resynchronize with the PST and DDATA outputs. PSTCLK starts clocking only when the first non-zero PST value (0xC, 0xD, or 0xF) occurs during system reset exception processing.	O
Processor Status Outputs	PST[3:0]	Indicate core status. Debug mode timing is synchronous with the processor clock; status is unrelated to the current bus transfer. The CLKOUT signal can be used by the development system to know when to sample PST[3:0].	O
All Processor Status Outputs	ALLPST	Logical AND of PST[3:0]. The CLKOUT signal can be used by the development system to know when to sample ALLPST.	O

1.15 EzPort Signal Descriptions

Table 17 contains a list of EzPort external signals.

Table 17. EzPort Signal Descriptions

Signal Name	Abbreviation	Function	I/O
EzPort Clock	EZPCK	Shift clock for EzPort transfers.	I
EzPort Chip Select	EZPCS	Chip select for signalling the start and end of serial transfers.	I
EzPort Serial Data In	EZPD	EZPD is sampled on the rising edge of EZPCK.	I
EzPort Serial Data Out	EZPQ	EZPQ transitions on the falling edge of EZPCK.	O

[1] θ_{JA} and Ψ_{jt} parameters are simulated in conformance with EIA/JESD Standard 51-2 for natural convection. Freescale recommends the use of θ_{JA} and power dissipation specifications in the system design to prevent device junction temperatures from exceeding the rated specification. System designers should be aware that device junction temperatures can be significantly influenced by board layout and surrounding devices. Conformance to the device junction temperature specification can be verified by physical measurement in the customer's system using the Ψ_{jt} parameter, the device power dissipation, and the method described in EIA/JESD Standard 51-2.

[2] Per JEDEC JESD51-2 with the single-layer board (JESD51-3) horizontal.

[3] Per JEDEC JESD51-6 with the board JESD51-7) horizontal.

[4] Thermal resistance between the die and the printed circuit board in conformance with JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.

[5] Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1).

[6] Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2. When Greek letters are not available, the thermal characterization parameter is written in conformance with Psi-JT.

The average chip-junction temperature (T_J) in °C can be obtained from:

$$T_J = T_A + (P_D \times \Theta_{JMA}) \quad (1)$$

Where:

T_A = ambient temperature, °C
Θ_{JA} = package thermal resistance, junction-to-ambient, °C/W
P_D = $P_{INT} + P_{I/O}$
P_{INT} = chip internal power, $I_{DD} \times V_{DD}$, watts
$P_{I/O}$ = power dissipation on input and output pins — user determined, watts

For most applications $P_{I/O} < P_{INT}$ and can be ignored. An approximate relationship between P_D and T_J (if $P_{I/O}$ is neglected) is:

$$P_D = K \div (T_J + 273°C) \quad (2)$$

Solving equations 1 and 2 for K gives:

$$K = P_D \times (T_A + 273\ °C) + \Theta_{JMA} \times P_D^{\ 2} \quad (3)$$

where K is a constant pertaining to the particular part. K can be determined from equation (3) by measuring P_D (at equilibrium) for a known T_A. Using this value of K, the values of P_D and T_J can be obtained by solving equations (1) and (2) iteratively for any value of T_A.

2.4 Flash Memory Characteristics

The flash memory characteristics are shown in Table 23 and Table 24.

Table 23. SGFM Flash Program and Erase Characteristics

(V_{DD} = 3.0 to 3.6 V)

Parameter	Symbol	Min	Typ	Max	Unit
System clock (read only)	$f_{sys(R)}$	0	—	50–80[1]	MHz
System clock (program/erase)[2]	$f_{sys(P/E)}$	0.15	—	102.4	MHz

[1] Depending on packaging; see the orderable part number summary.

[2] Refer to the flash memory section for more information

Table 24. SGFM Flash Module Life Characteristics

(V_{DD} = 3.0 to 3.6 V)

Parameter	Symbol	Value	Unit
Maximum number of guaranteed program/erase cycles[1] before failure	P/E	10,000[2]	Cycles
Data retention at average operating temperature of 85°C	Retention	10	Years

[1] A program/erase cycle is defined as switching the bits from 1 → 0 → 1.

[2] Reprogramming of a flash memory array block prior to erase is not required.

2.5 EzPort Electrical Specifications

Table 25. EzPort Electrical Specifications

Name	Characteristic	Min	Max	Unit
EP1	EPCK frequency of operation (all commands except READ)	—	f_{sys} / 2	MHz
EP1a	EPCK frequency of operation (READ command)	—	f_{sys} / 8	MHz
EP2	EPCS_b negation to next EPCS_b assertion	$2 \times T_{cyc}$	—	ns
EP3	EPCS_B input valid to EPCK high (setup)	5	—	ns
EP4	EPCK high to EPCS_B input invalid (hold)	5	—	ns
EP5	EPD input valid to EPCK high (setup)	2	—	ns
EP6	EPCK high to EPD input invalid (hold)	5	—	ns
EP7	EPCK low to EPQ output valid (out setup)	—	12	ns
EP8	EPCK low to EPQ output invalid (out hold)	0	—	ns
EP9	EPCS_B negation to EPQ tri-state	—	12	ns

Table 27. DC Electrical Specifications (continued)[1]

Characteristic	Symbol	Min	Max	Unit
Output high voltage (high drive) I_{OH} = -5 mA	V_{OH}	V_{DD} – 0.5	—	V
Output low voltage (high drive) I_{OL} = 5 mA	V_{OL}	—	0.5	V
Output high voltage (low drive) I_{OH} = -2 mA	V_{OH}	V_{DD} - 0.5	—	V
Output low voltage (low drive) I_{OL} = 2 mA	V_{OL}	—	0.5	V
Weak internal pull Up device current, tested at V_{IL} Max.[3]	I_{APU}	–10	–130	µA
Input Capacitance [4] • All input-only pins • All input/output (three-state) pins	C_{in}	 — —	 7 7	pF

[1] Refer to Table 28 for additional PLL specifications.

[2] Only for pins: IRQ1, IRQ2. IRQ3, IRQ4, IRQ5, IRQ6. IRQ7, RSTIN_B, RCON_B, PCS0, SCK, I2C_SDA, I2C_SCL, TCLK, TRST_B, TEST

[3] Refer to Table 3 for pins having internal pull-up devices.

[4] This parameter is characterized before qualification rather than 100% tested.

2.8 Clock Source Electrical Specifications

Table 28. Oscillator and PLL Electrical Specifications

(V_{DD} and V_{DDPLL} = 2.7 to 3.6 V, V_{SS} = V_{SSPLL} = 0 V)

Characteristic	Symbol	Min	Max	Unit
Clock Source Frequency Range of EXTAL Frequency Range • Crystal • External[1]	 $f_{crystal}$ f_{ext}	 1 0	 25.0[2] 66.67 or 80	MHz
PLL reference frequency range	f_{ref_pll}	2	10.0	MHz
System frequency [3] • External clock mode • On-chip PLL frequency	f_{sys}	 0 f_{ref} / 32	 66.67 or 80[4] 66.67 or 80[4]	MHz
Loss of reference frequency [5, 7]	f_{LOR}	100	1000	kHz
Self clocked mode frequency [6]	f_{SCM}	1	5	MHz
Crystal start-up time [7, 8]	t_{cst}	—	0.1	ms
EXTAL input high voltage • External reference	V_{IHEXT}	 2.0	 3.0[2]	V
EXTAL input low voltage • External reference	V_{ILEXT}	 V_{SS}	 0.8	V
PLL lock time[4,9]	t_{lpll}	—	500	µs
Duty cycle of reference [4]	t_{dc}	40	60	% f_{ref}

Table 28. Oscillator and PLL Electrical Specifications (continued)

(V_{DD} and V_{DDPLL} = 2.7 to 3.6 V, V_{SS} = V_{SSPLL} = 0 V)

Characteristic	Symbol	Min	Max	Unit
Frequency un-LOCK range	f_{UL}	–1.5	1.5	% f_{ref}
Frequency LOCK range	f_{LCK}	–0.75	0.75	% f_{ref}
CLKOUT period jitter [4, 5, 10 ,11], measured at f_{SYS} Max • Peak-to-peak (clock edge to clock edge) • Long term (averaged over 2 ms interval)	C_{jitter}	 — —	 10 .01	% f_{sys}
On-chip oscillator frequency	f_{oco}	7.84	8.16	MHz

[1] In external clock mode, it is possible to run the chip directly from an external clock source without enabling the PLL.

[2] This value has been updated.

[3] All internal registers retain data at 0 Hz.

[4] Depending on packaging; see the orderable part number summary.

[5] Loss of Reference Frequency is the reference frequency detected internally, which transitions the PLL into self clocked mode.

[6] Self clocked mode frequency is the frequency at which the PLL operates when the reference frequency falls below f_{LOR} with default MFD/RFD settings.

[7] This parameter is characterized before qualification rather than 100% tested.

[8] Proper PC board layout procedures must be followed to achieve specifications.

[9] This specification applies to the period required for the PLL to relock after changing the MFD frequency control bits in the synthesizer control register (SYNCR).

[10] Jitter is the average deviation from the programmed frequency measured over the specified interval at maximum f_{sys}. Measurements are made with the device powered by filtered supplies and clocked by a stable external clock signal. Noise injected into the PLL circuitry via V_{DDPLL} and V_{SSPLL} and variation in crystal oscillator frequency increase the C_{jitter} percentage for a given interval.

[11] Based on slow system clock of 40 MHz measured at f_{sys} max.

2.9 General Purpose I/O Timing

GPIO can be configured for certain pins of the QSPI, DDR Control, timer, UART, and Interrupt interfaces. When in GPIO mode, the timing specification for these pins is given in Table 29 and Figure 5.

The GPIO timing is met under the following load test conditions:

- 50 pF / 50 Ω for high drive
- 25 pF / 25 Ω for low drive

Table 29. GPIO Timing

NUM	Characteristic	Symbol	Min	Max	Unit
G1	CLKOUT High to GPIO Output Valid	t_{CHPOV}	—	10	ns
G2	CLKOUT High to GPIO Output Invalid	t_{CHPOI}	1.5	—	ns
G3	GPIO Input Valid to CLKOUT High	t_{PVCH}	9	—	ns
G4	CLKOUT High to GPIO Input Invalid	t_{CHPI}	1.5	—	ns

Table 36. JTAG and Boundary Scan Timing

Num	Characteristics[1]	Symbol	Min	Max	Unit
J1	TCLK frequency of operation	f_{JCYC}	DC	1/4	$f_{sys/2}$
J2	TCLK cycle period	t_{JCYC}	$4 \times t_{CYC}$	—	ns
J3	TCLK clock pulse width	t_{JCW}	26	—	ns
J4	TCLK rise and fall times	t_{JCRF}	0	3	ns
J5	Boundary scan input data setup time to TCLK rise	t_{BSDST}	4	—	ns
J6	Boundary scan input data hold time after TCLK rise	t_{BSDHT}	26	—	ns
J7	TCLK low to boundary scan output data valid	t_{BSDV}	0	33	ns
J8	TCLK low to boundary scan output high Z	t_{BSDZ}	0	33	ns
J9	TMS, TDI input data setup time to TCLK rise	t_{TAPBST}	4	—	ns
J10	TMS, TDI Input data hold time after TCLK rise	t_{TAPBHT}	10	—	ns
J11	TCLK low to TDO data valid	t_{TDODV}	0	26	ns
J12	TCLK low to TDO high Z	t_{TDODZ}	0	8	ns
J13	$\overline{\text{TRST}}$ assert time	t_{TRSTAT}	100	—	ns
J14	$\overline{\text{TRST}}$ setup time (negation) to TCLK high	t_{TRSTST}	10	—	ns

[1] JTAG_EN is expected to be a static signal. Hence, it is not associated with any timing.

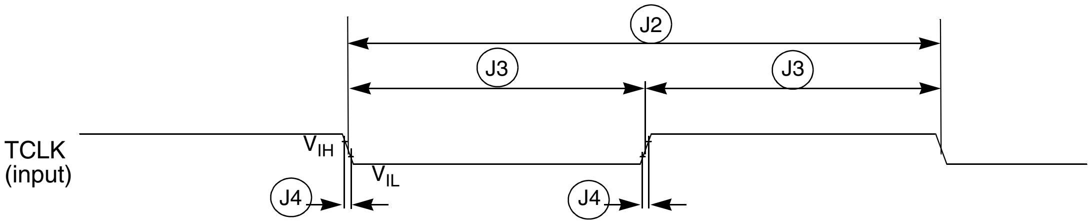


Figure 10. Test Clock Input Timing

2.17 Debug AC Timing Specifications

Table 37 lists specifications for the debug AC timing parameters shown in Figure 15.

Table 37. Debug AC Timing Specification

Num	Characteristic	66/80 MHz		Units
		Min	Max	
D1	PST, DDATA to CLKOUT setup	4	—	ns
D2	CLKOUT to PST, DDATA hold	1.5	—	ns
D3	DSI-to-DSCLK setup	1 × t_{CYC}	—	ns
D4[1]	DSCLK-to-DSO hold	4 × t_{CYC}	—	ns
D5	DSCLK cycle time	5 × t_{CYC}	—	ns
D6	$\overline{\text{BKPT}}$ input data setup time to CLKOUT rise	4	—	ns
D7	$\overline{\text{BKPT}}$ input data hold time to CLKOUT rise	1.5	—	ns
D8	CLKOUT high to $\overline{\text{BKPT}}$ high Z	0.0	10.0	ns

[1] DSCLK and DSI are synchronized internally. D4 is measured from the synchronized DSCLK input relative to the rising edge of CLKOUT.

Figure 14 shows real-time trace timing for the values in Table 37.

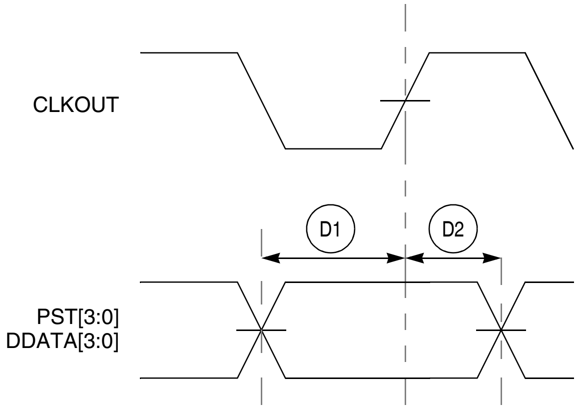


Figure 14. Real-Time Trace AC Timing

Figure 15 shows BDM serial port AC timing for the values in Table 37.

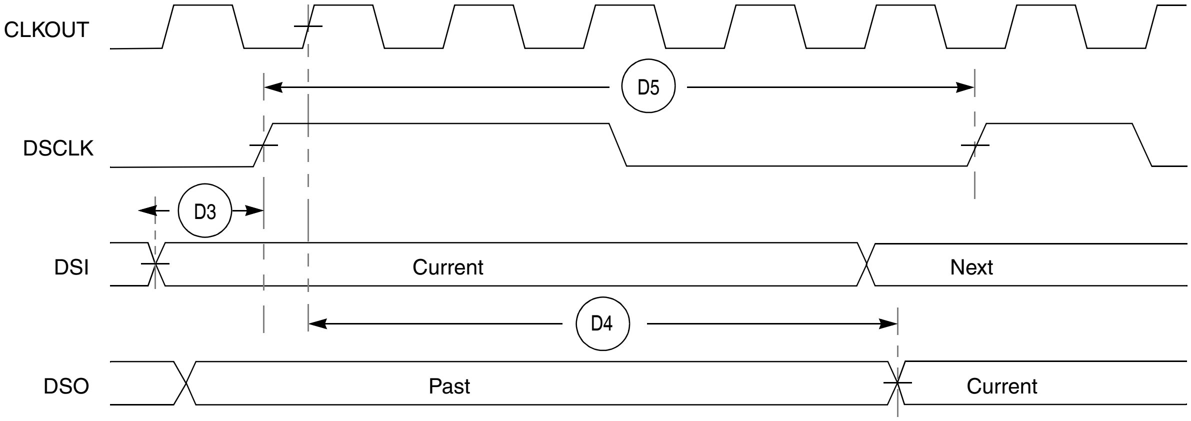


Figure 15. BDM Serial Port AC Timing

3 Mechanical Outline Drawings

This section describes the physical properties of the device and its derivatives.

NOTES:

1. DIMENSIONS ARE IN MILLIMETERS.

2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M-1994.

3. DATUMS A, B AND D TO BE DETERMINED AT DATUM PLANE H.

△4. DIMENSIONS TO BE DETERMINED AT SEATING PLANE C.

△5. THIS DIMENSION DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL NOT CAUSE THE LEAD WIDTH TO EXCEED THE UPPER LIMIT BY MORE THAN 0.08 mm AT MAXIMUM MATERIAL CONDITION. DAMBAR CANNOT BE LOCATED ON THE LOWER RADIUS OR THE FOOT. MINIMUM SPACE BETWEEN PROTRUSION AND ADJACENT LEAD SHALL NOT BE LESS THAN 0.07 mm.

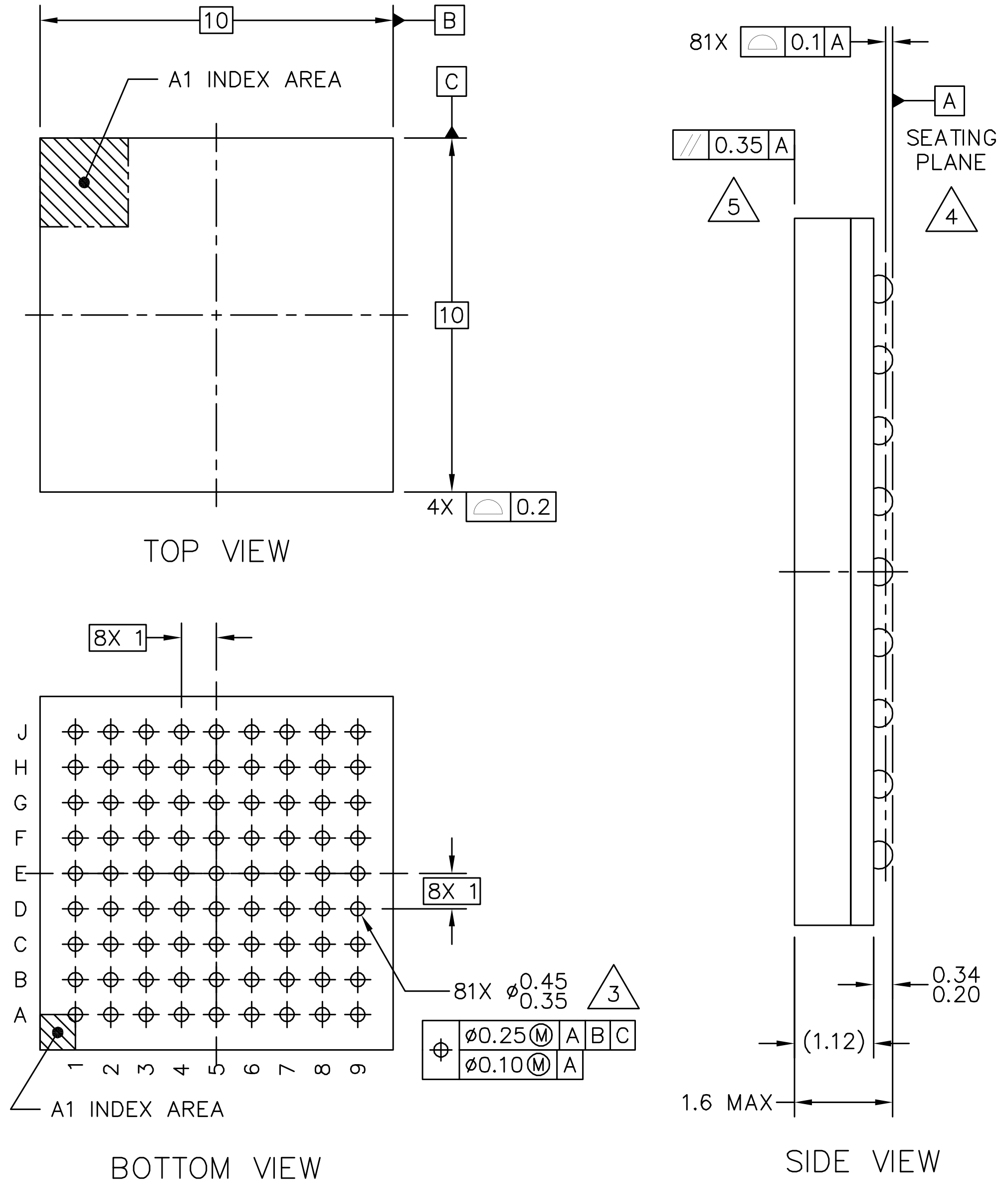
△6. THIS DIMENSION DOES NOT INCLUDE MOLD PROTRUSION. ALLOWABLE PROTRUSION IS 0.25 mm PER SIDE. THIS DIMENSION IS MAXIMUM PLASTIC BODY SIZE DIMENSION INCLUDING MOLD MISMATCH.

△7. EXACT SHAPE OF EACH CORNER IS OPTIONAL.

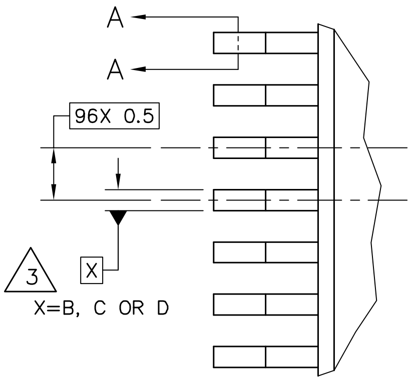
△8. THESE DIMENSIONS APPLY TO THE FLAT SECTION OF THE LEAD BETWEEN 0.1 mm AND 0.25 mm FROM THE LEAD TIP.

© FREESCALE SEMICONDUCTOR, INC. ALL RIGHTS RESERVED.	MECHANICAL OUTLINE	PRINT VERSION NOT TO SCALE
TITLE: 64LD LQFP, 10 X 10 X 1.4 PKG, 0.5 PITCH, CASE OUTLINE	DOCUMENT NO: 98ASS23234W	REV: D
	CASE NUMBER: 840F-02	06 APR 2005
	STANDARD: JEDEC MS-026 BCD	

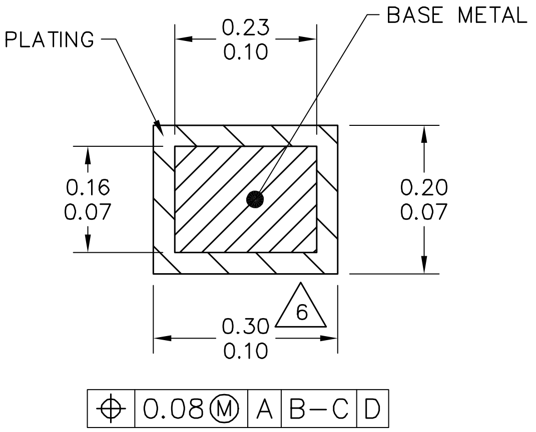
3.3 81 MAPBGA Package



	MECHANICAL OUTLINE	PRINT VERSION NOT TO SCALE
TITLE: PBGA, LOW PROFILE, 81 I/O, 10 X 10 PKG, 1 MM PITCH (MAP)	DOCUMENT NO: 98ASA10670D	REV: 0
	CASE NUMBER: 1662–01	04 FEB 2005
	STANDARD: NON–JEDEC	



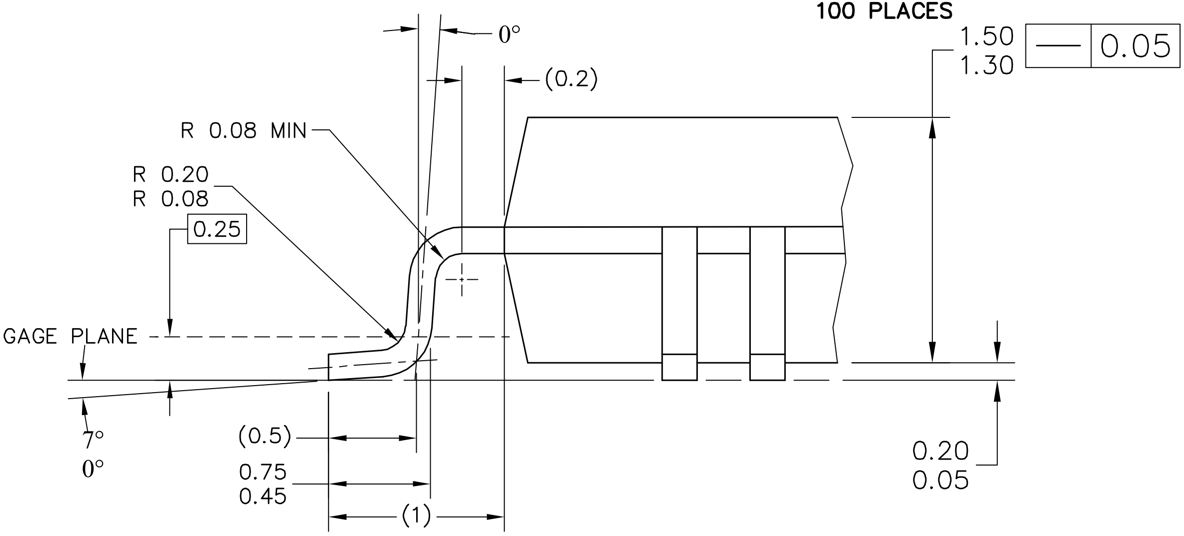
VIEW A



SECTION A–A

ROTATED 90° CW

100 PLACES



VIEW B

© FREESCALE SEMICONDUCTOR, INC. ALL RIGHTS RESERVED.	**MECHANICAL OUTLINE**	PRINT VERSION NOT TO SCALE
TITLE: 100 LEAD LQFP 14 X 14, 0.5 PITCH, 1.4 THICK	DOCUMENT NO: 98ASS23308W	REV: G
	CASE NUMBER: 983–03	07 APR 2005
	STANDARD: NON–JEDEC	

4 Revision History

Table 38. Revision History

Revision	Description
0	Initial public release.
1	• Updated Clock generation features • Changed crystal frequency range maximum to 25 MHz • Updated Table: Clocking Modes and added appropriate footnote • In Table: CLock Source Electrical Specifications, updated the following values: fcrystal, fext, fref_pll, EXTAL input high voltage (External reference)