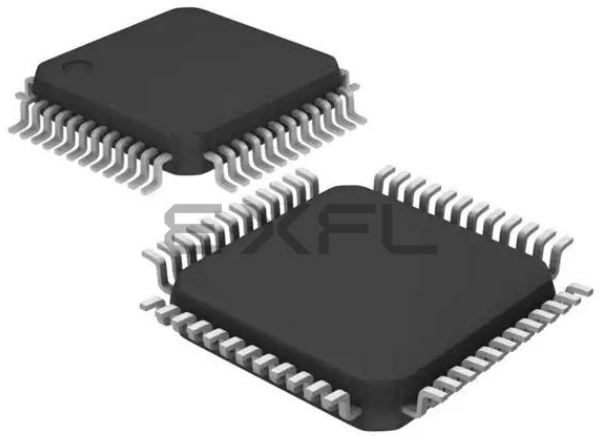

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Details	
Product Status	Active
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	CANbus, I²C, LINbus, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	39
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 16x12b
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	48-LQFP
Supplier Device Package	48-LQFP (7x7)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mc9s08dv32aclf

Section Number	Title	Page

Chapter 16
Timer Pulse-Width Modulator (S08TPMV3)

Chapter 17
Development Support

Table 4-1. Reset and Interrupt Vectors

Address (High/Low)	Vector	Vector Name
0xFFCE:0xFFCF	IIC	Viic
0xFFD0:0xFFD1	ADC Conversion	Vadc
0xFFD2:0xFFD3	Port A, Port B, Port D	Vport
0xFFD4:0xFFD5	SCI2 Transmit	Vsci2tx
0xFFD6:0xFFD7	SCI2 Receive	Vsci2rx
0xFFD8:0xFFD9	SCI2 Error	Vsci2err
0xFFDA:0xFFDB	SCI1 Transmit	Vsci1tx
0xFFDC:0xFFDD	SCI1 Receive	Vsci1rx
0xFFDE:0xFFDF	SCI1 Error	Vsci1err
0xFFE0:0xFFE1	SPI	Vspi
0xFFE2:0xFFE3	TPM2 Overflow	Vtpm2ovf
0xFFE4:0xFFE5	TPM2 Channel 1	Vtpm2ch1
0xFFE6:0xFFE7	TPM2 Channel 0	Vtpm2ch0
0xFFE8:0xFFE9	TPM1 Overflow	Vtpm1ovf
0xFFEA:0xFFEB	TPM1 Channel 5	Vtpm1ch5
0xFFEC:0xFFED	TPM1 Channel 4	Vtpm1ch4
0xFFEE:0xFFEF	TPM1 Channel 3	Vtpm1ch3
0xFFF0:0xFFF1	TPM1 Channel 2	Vtpm1ch2
0xFFF2:0xFFF3	TPM1 Channel 1	Vtpm1ch1
0xFFF4:0xFFF5	TPM1 Channel 0	Vtpm1ch0
0xFFF6:0xFFF7	MCG Loss of lock	Vlol
0xFFF8:0xFFF9	Low-Voltage Detect	Vlvd
0xFFFA:0xFFFB	IRQ	Virq
0xFFFC:0xFFFD	SWI	Vswi
0xFFFE:0xFFFF	Reset	Vreset

Table 4-2. Direct-Page Register Summary (Sheet 1 of 3)

Address	Register Name	Bit 7	6	5	4	3	2	1	Bit 0
0x0000	PTAD	PTAD7	PTAD6	PTAD5	PTAD4	PTAD3	PTAD2	PTAD1	PTAD0
0x0001	PTADD	PTADD7	PTADD6	PTADD5	PTADD4	PTADD3	PTADD2	PTADD1	PTADD0
0x0002	PTBD	PTBD7	PTBD6	PTBD5	PTBD4	PTBD3	PTBD2	PTBD1	PTBD0
0x0003	PTBDD	PTBDD7	PTBDD6	PTBDD5	PTBDD4	PTBDD3	PTBDD2	PTBDD1	PTBDD0
0x0004	PTCD	PTCD7	PTCD6	PTCD5	PTCD4	PTCD3	PTCD2	PTCD1	PTCD0
0x0005	PTCDD	PTCDD7	PTCDD6	PTCDD5	PTCDD4	PTCDD3	PTCDD2	PTCDD1	PTCDD0
0x0006	PTDD	PTDD7	PTDD6	PTDD5	PTDD4	PTDD3	PTDD2	PTDD1	PTDD0
0x0007	PTDDD	PTDDD7	PTDDD6	PTDDD5	PTDDD4	PTDDD3	PTDDD2	PTDDD1	PTDDD0
0x0008	PTED	PTED7	PTED6	PTED5	PTED4	PTED3	PTED2	PTED1	PTED0
0x0009	PTEDD	PTEDD7	PTEDD6	PTEDD5	PTEDD4	PTEDD3	PTEDD2	PTEDD1	PTEDD0
0x000A	PTFD	PTFD7	PTFD6	PTFD5	PTFD4	PTFD3	PTFD2	PTFD1	PTFD0
0x000B	PTFDD	PTFDD7	PTFDD6	PTFDD5	PTFDD4	PTFDD3	PTFDD2	PTFDD1	PTFDD0
0x000C	PTGD	0	0	PTGD5	PTGD4	PTGD3	PTGD2	PTGD1	PTGD0
0x000D	PTGDD	0	0	PTGDD5	PTGDD4	PTGDD3	PTGDD2	PTGDD1	PTGDD0
0x000E	ACMP1SC	ACME	ACBGS	ACF	ACIE	ACO	ACOPE	ACMOD1	ACMOD0
0x000F	ACMP2SC	ACME	ACBGS	ACF	ACIE	ACO	ACOPE	ACMOD1	ACMOD0
0x0010	ADCSC1	COCO	AIEN	ADCO	ADCH				
0x0011	ADCSC2	ADACT	ADTRG	ACFE	ACFGT	0	0	—	—
0x0012	ADCRH	0	0	0	0	ADR11	ADR10	ADR9	ADR8
0x0013	ADCRL	ADR7	ADR6	ADR5	ADR4	ADR3	ADR2	ADR1	ADR0
0x0014	ADCCVH	0	0	0	0	ADCV11	ADCV10	ADCV9	ADCV8
0x0015	ADCCVL	ADCV7	ADCV6	ADCV5	ADCV4	ADCV3	ADCV2	ADCV1	ADCV0
0x0016	ADCCFG	ADLPC	ADIV		ADLSMP	MODE		ADICLK	
0x0017	APCTL1	ADPC7	ADPC6	ADPC5	ADPC4	ADPC3	ADPC2	ADPC1	ADPC0
0x0018	APCTL2	ADPC15	ADPC14	ADPC13	ADPC12	ADPC11	ADPC10	ADPC9	ADPC8
0x0019	APCTL3	ADPC23	ADPC22	ADPC21	ADPC20	ADPC19	ADPC18	ADPC17	ADPC16
0x001A– 0x001B	Reserved	— —	— —	— —	— —	— —	— —	— —	— —
0x001C	IRQSC	0	IRQPDD	IRQEDG	IRQPE	IRQF	IRQACK	IRQIE	IRQMOD
0x001D– 0x001F	Reserved	— —	— —	— —	— —	— —	— —	— —	— —
0x0020	TPM1SC	TOF	TOIE	CPWMS	CLKSB	CLKSA	PS2	PS1	PS0
0x0021	TPM1CNTH	Bit 15	14	13	12	11	10	9	Bit 8
0x0022	TPM1CNTL	Bit 7	6	5	4	3	2	1	Bit 0
0x0023	TPM1MODH	Bit 15	14	13	12	11	10	9	Bit 8
0x0024	TPM1MODL	Bit 7	6	5	4	3	2	1	Bit 0
0x0025	TPM1C0SC	CH0F	CH0IE	MS0B	MS0A	ELS0B	ELS0A	0	0
0x0026	TPM1C0VH	Bit 15	14	13	12	11	10	9	Bit 8
0x0027	TPM1C0VL	Bit 7	6	5	4	3	2	1	Bit 0

4.5.7 Block Protection

The block protection feature prevents the protected region of Flash from program or erase changes. Block protection is controlled through the Flash protection register (FPROT). The FPS bits determine the protected region of Flash. See Section 4.5.10.4, “Flash Protection Register (FPROT and NVPROT).”

After exit from reset, FPROT is loaded with the contents of the NVPROT location, which is in the nonvolatile register block of the Flash memory. Any FPROT write that attempts to decrease the size of the protected region will be ignored. Because NVPROT is within the last sector of Flash, if any amount of memory is protected, NVPROT is itself protected and cannot be unprotected (intentionally or unintentionally) by the application software. FPROT can be written through background debug commands, which provides a way to erase and reprogram protected Flash memory.

One use for block protection is to block protect an area of Flash memory for a bootloader program. this bootloader program can call a routine outside of Flash that can be used to sector erase the rest of the Flash memory and reprogram it. The bootloader is protected even if MCU power is lost during an erase and reprogram operation.

4.5.8 Vector Redirection

While any Flash is block protected, the reset and interrupt vectors will be protected. Vector redirection allows users to modify interrupt vector information without unprotecting bootloader and reset vector space. Vector redirection is enabled by programming the FNORED bit in the NVOPT register located at address 0xFFBF to 0. For redirection to occur, at least some portion of the Flash memory must be block protected by programming the NVPROT register located at address 0xFFBD. All interrupt vectors (memory locations 0xFFC0–0xFFFD) are redirected, though the reset vector (0xFFFE:0xFFFF) is not.

For example, if 1536 bytes of Flash are protected, the protected address region is from 0xFA00 through 0xFFFF. The interrupt vectors (0xFFC0–0xFFFD) are redirected to the locations 0xF9C0–0xF9FD. If vector redirection is enabled and an interrupt occurs, the values in the locations 0xF9E0:0xF9E1 are used for the vector instead of the values in the locations 0xFFE0:0xFFE1. This allows the user to reprogram the unprotected portion of the Flash with new program code including new interrupt vector values while leaving the protected area, which includes the default vector locations, unchanged.

4.5.9 Security

The MC9S08DV60 Series includes circuitry to prevent unauthorized access to the contents of Flash and RAM memory. When security is engaged, Flash and RAM are considered secure resources. Direct-page registers, high-page registers, and the background debug controller are considered unsecured resources. Programs executing within secure memory have normal access to any MCU memory locations and resources. Attempts to access a secure memory location with a program executing from an unsecured memory space or through the background debug interface are blocked (writes are ignored and reads return all 0s).

Security is engaged or disengaged based on the state of two register bits (SEC[1:0]) in the FOPT register. During reset, the contents of the nonvolatile location NVOPT are copied from Flash into the working FOPT register in high-page register space. A user engages security by programming the NVOPT location, which can be performed at the same time the Flash memory is programmed. The 1:0 state disengages

5.8.4 System Options Register 1 (SOPT1)

This high page register is a write-once register so only the first write after reset is honored. It can be read at any time. Any subsequent attempt to write to SOPT1 (intentionally or unintentionally) is ignored to avoid accidental changes to these sensitive settings. This register should be written during the user's reset initialization program to set the desired controls even if the desired settings are the same as the reset settings.

	7	6	5	4	3	2	1	0
R	COPT		STOPE	SCI2PS	IICPS	0	0	0
W								
Reset:	1	1	0	0	0	0	0	0

= Unimplemented or Reserved

Figure 5-5. System Options Register 1 (SOPT1)

Table 5-5. SOPT1 Register Field Descriptions

Field	Description
7:6 COPT[1:0]	**COP Watchdog Timeout** — These write-once bits select the timeout period of the COP. COPT along with COPCLKS in SOPT2 defines the COP timeout period. See Table 5-6.
5 STOPE	**Stop Mode Enable** — This write-once bit is used to enable stop mode. If stop mode is disabled and a user program attempts to execute a STOP instruction, an illegal opcode reset is forced. 0 Stop mode disabled. 1 Stop mode enabled.
4 SCI2PS	**SCI2 Pin Select**— This write-once bit selects the location of the RxD2 and TxD2 pins of the SCI2 module. 0 TxD2 on PTF0, RxD2 on PTF1. 1 TxD2 on PTE6, RxD2 on PTE7.
3 IICPS	**IIC Pin Select**— This write-once bit selects the location of the SCL and SDA pins of the IIC module. 0 SCL on PTF2, SDA on PTF3. 1 SCL on PTE4, SDA on PTE5.

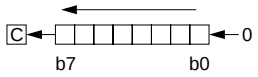
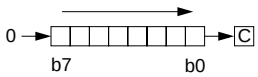
Table 5-6. COP Configuration Options

Control Bits		Clock Source	COP Window[1] Opens (COPW = 1)	COP Overflow Count
COPCLKS	COPT[1:0]			
N/A	0:0	N/A	N/A	COP is disabled
0	0:1	1 kHz	N/A	2^5 cycles (32 ms[2])
0	1:0	1 kHz	N/A	2^8 cycles (256 ms[1])
0	1:1	1 kHz	N/A	2^{10} cycles (1.024 s[1])
1	0:1	Bus	6144 cycles	2^{13} cycles
1	1:0	Bus	49,152 cycles	2^{16} cycles
1	1:1	Bus	196,608 cycles	2^{18} cycles

1 Windowed COP operation requires the user to clear the COP timer in the last 25% of the selected timeout period. This column displays the minimum number of clock counts required before the COP timer can be reset when in windowed COP mode (COPW = 1).

2 Values shown in milliseconds based on t_{LPO} = 1 ms. See t_{LPO} in the appendix Section A.12.1, "Control Timing," for the tolerance of this value.

Table 7-2. Instruction Set Summary (Sheet 5 of 9)

Source Form	Operation	Address Mode	Object Code	Cycles	Cyc-by-Cyc Details	Affect on CCR V 1 1 H	Affect on CCR I N Z C
INC *oprx8*,X	Increment M ← (M) + $01	DIR	3C dd	5	rfwpp	↕ 1 1 –	– ↕ ↕ –
INCA	A ← (A) + $01	INH	4C	1	p		
INCX	X ← (X) + $01	INH	5C	1	p		
INC *oprx8*,X	M ← (M) + $01	IX1	6C ff	5	rfwpp		
INC ,X	M ← (M) + $01	IX	7C	4	rfwp		
INC *oprx8*,SP	M ← (M) + $01	SP1	9E 6C ff	6	prfwpp		
JMP *opr8a*	Jump PC ← Jump Address	DIR	BC dd	3	ppp	– 1 1 –	– – – –
JMP *opr16a*		EXT	CC hh ll	4	pppp		
JMP *oprx16*,X		IX2	DC ee ff	4	pppp		
JMP *oprx8*,X		IX1	EC ff	3	ppp		
JMP ,X		IX	FC	3	ppp		
JSR *opr8a*	Jump to Subroutine PC ← (PC) + *n* (*n* = 1, 2, or 3) Push (PCL); SP ← (SP) – $0001 Push (PCH); SP ← (SP) – $0001 PC ← Unconditional Address	DIR	BD dd	5	ssppp	– 1 1 –	– – – –
JSR *opr16a*		EXT	CD hh ll	6	pssppp		
JSR *oprx16*,X		IX2	DD ee ff	6	pssppp		
JSR *oprx8*,X		IX1	ED ff	5	ssppp		
JSR ,X		IX	FD	5	ssppp		
LDA *#opr8i*	Load Accumulator from Memory A ← (M)	IMM	A6 ii	2	pp	0 1 1 –	– ↕ ↕ –
LDA *opr8a*		DIR	B6 dd	3	rpp		
LDA *opr16a*		EXT	C6 hh ll	4	prpp		
LDA *oprx16*,X		IX2	D6 ee ff	4	prpp		
LDA *oprx8*,X		IX1	E6 ff	3	rpp		
LDA ,X		IX	F6	3	rfp		
LDA *oprx16*,SP		SP2	9E D6 ee ff	5	pprpp		
LDA *oprx8*,SP		SP1	9E E6 ff	4	prpp		
LDHX *#opr16i*	Load Index Register (H:X) H:X ← (M:M + $0001)	IMM	45 jj kk	3	ppp	0 1 1 –	– ↕ ↕ –
LDHX *opr8a*		DIR	55 dd	4	rrpp		
LDHX *opr16a*		EXT	32 hh ll	5	prrpp		
LDHX ,X		IX	9E AE	5	prrfp		
LDHX *oprx16*,X		IX2	9E BE ee ff	6	pprrpp		
LDHX *oprx8*,X		IX1	9E CE ff	5	prrpp		
LDHX *oprx8*,SP		SP1	9E FE ff	5	prrpp		
LDX *#opr8i*	Load X (Index Register Low) from Memory X ← (M)	IMM	AE ii	2	pp	0 1 1 –	– ↕ ↕ –
LDX *opr8a*		DIR	BE dd	3	rpp		
LDX *opr16a*		EXT	CE hh ll	4	prpp		
LDX *oprx16*,X		IX2	DE ee ff	4	prpp		
LDX *oprx8*,X		IX1	EE ff	3	rpp		
LDX ,X		IX	FE	3	rfp		
LDX *oprx16*,SP		SP2	9E DE ee ff	5	pprpp		
LDX *oprx8*,SP		SP1	9E EE ff	4	prpp		
LSL *oprx8*,X	Logical Shift Left  (Same as ASL)	DIR	38 dd	5	rfwpp	↕ 1 1 –	– ↕ ↕ ↕
LSL *oprx8*,SP		INH	48	1	p		
LSLA		INH	58	1	p		
LSLX		IX1	68 ff	5	rfwpp		
LSL ,X		IX	78	4	rfwp		
LSL *oprx8*,SP		SP1	9E 68 ff	6	prfwpp		
LSR *oprx8*,X	Logical Shift Right  	DIR	34 dd	5	rfwpp	↕ 1 1 –	– 0 ↕ ↕
LSR *oprx8*,SP		INH	44	1	p		
LSR*X*		INH	54	1	p		
LSR *oprx8*,X		IX1	64 ff	5	rfwpp		
LSR ,X		IX	74	4	rfwp		
LSR *oprx8*,SP		SP1	9E 64 ff	6	prfwpp		

10.1.5 Temperature Sensor

To use the on-chip temperature sensor, the user must perform the following:

- Configure ADC for long sample with a maximum of 1 MHz clock
- Convert the bandgap voltage reference channel (AD27)
  - By converting the digital value of the bandgap voltage reference channel using the value of V_{BG} the user can determine V_{DD}. For value of bandgap voltage, see Section A.6, “DC Characteristics”.
- Convert the temperature sensor channel (AD26)
  - By using the calculated value of V_{DD}, convert the digital value of AD26 into a voltage, V_{TEMP}

Equation 10-1 provides an approximate transfer function of the temperature sensor.

$$Temp = 25 - ((V_{TEMP} - V_{TEMP25}) \div m) \qquad \textit{Eqn. 10-1}$$

where:

- V_{TEMP} is the voltage of the temperature sensor channel at the ambient temperature.
- V_{TEMP25} is the voltage of the temperature sensor channel at 25°C.
- m is the hot or cold voltage versus temperature slope in V/°C.

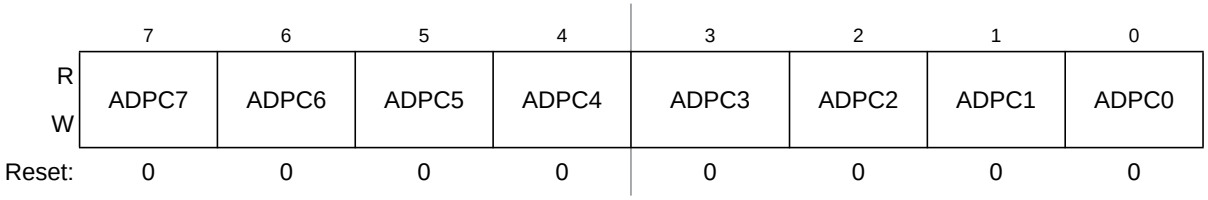
For temperature calculations, use the V_{TEMP25} and m values from the ADC Electricals table.

In application code, the user reads the temperature sensor channel, calculates V_{TEMP}, and compares to V_{TEMP25}. If V_{TEMP} is greater than V_{TEMP25} the cold slope value is applied in Equation 10-1. If V_{TEMP} is less than V_{TEMP25} the hot slope value is applied in Equation 10-1. To improve accuracy the user should calibrate the bandgap voltage reference and temperature sensor.

Calibrating at 25°C will improve accuracy to ± 4.5°C.

Calibration at three points, -40°C, 25°C, and 125°C will improve accuracy to ± 2.5°C. Once calibration has been completed, the user will need to calculate the slope for both hot and cold. In application code, the user would then calculate the temperature using Equation 10-1 as detailed above and then determine if the temperature is above or below 25°C. Once determined if the temperature is above or below 25°C, the user can recalculate the temperature using the hot or cold slope value obtained during calibration.

used to control the pins associated with channels 0–7 of the ADC module.



	7	6	5	4	3	2	1	0
R W	ADPC7	ADPC6	ADPC5	ADPC4	ADPC3	ADPC2	ADPC1	ADPC0
Reset:	0	0	0	0	0	0	0	0

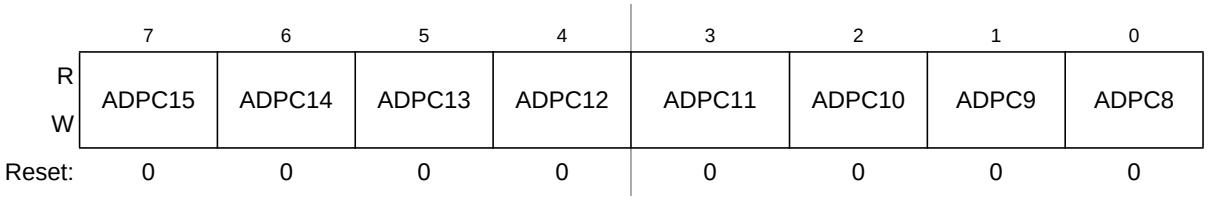
Figure 10-10. Pin Control 1 Register (APCTL1)

Table 10-10. APCTL1 Register Field Descriptions

Field	Description
7 ADPC7	ADC Pin Control 7. ADPC7 controls the pin associated with channel AD7. 0 AD7 pin I/O control enabled 1 AD7 pin I/O control disabled
6 ADPC6	ADC Pin Control 6. ADPC6 controls the pin associated with channel AD6. 0 AD6 pin I/O control enabled 1 AD6 pin I/O control disabled
5 ADPC5	ADC Pin Control 5. ADPC5 controls the pin associated with channel AD5. 0 AD5 pin I/O control enabled 1 AD5 pin I/O control disabled
4 ADPC4	ADC Pin Control 4. ADPC4 controls the pin associated with channel AD4. 0 AD4 pin I/O control enabled 1 AD4 pin I/O control disabled
3 ADPC3	ADC Pin Control 3. ADPC3 controls the pin associated with channel AD3. 0 AD3 pin I/O control enabled 1 AD3 pin I/O control disabled
2 ADPC2	ADC Pin Control 2. ADPC2 controls the pin associated with channel AD2. 0 AD2 pin I/O control enabled 1 AD2 pin I/O control disabled
1 ADPC1	ADC Pin Control 1. ADPC1 controls the pin associated with channel AD1. 0 AD1 pin I/O control enabled 1 AD1 pin I/O control disabled
0 ADPC0	ADC Pin Control 0. ADPC0 controls the pin associated with channel AD0. 0 AD0 pin I/O control enabled 1 AD0 pin I/O control disabled

10.3.9 Pin Control 2 Register (APCTL2)

APCTL2 controls channels 8–15 of the ADC module.



	7	6	5	4	3	2	1	0
R W	ADPC15	ADPC14	ADPC13	ADPC12	ADPC11	ADPC10	ADPC9	ADPC8
Reset:	0	0	0	0	0	0	0	0

Figure 10-11. Pin Control 2 Register (APCTL2)

ADCSC1 = 0x41 (%01000001)

```
Bit 7    COCO     0        Read-only flag which is set when a conversion completes
Bit 6    AIEN     1        Conversion complete interrupt enabled
Bit 5    ADCO     0        One conversion only (continuous conversions disabled)
Bit 4:0  ADCH     00001    Input channel 1 selected as ADC input channel
```

ADCRH/L = 0xxx

Holds results of conversion. Read high byte (ADCRH) before low byte (ADCRL) so that conversion data cannot be overwritten with data from the next conversion.

ADCCVH/L = 0xxx

Holds compare value when compare function enabled

APCTL1=0x02

AD1 pin I/O control disabled. All other AD pins remain general purpose I/O pins

APCTL2=0x00

All other AD pins remain general purpose I/O pins

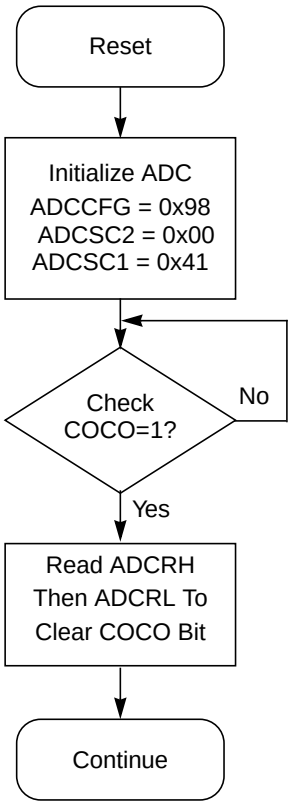


Figure 10-13. Initialization Flowchart for Example

Table 12-5. Baud Rate Prescaler

BRP5	BRP4	BRP3	BRP2	BRP1	BRP0	Prescaler value (P)
0	0	0	0	0	0	1
0	0	0	0	0	1	2
0	0	0	0	1	0	3
0	0	0	0	1	1	4
:	:	:	:	:	:	:
1	1	1	1	1	1	64

12.3.4 MSCAN Bus Timing Register 1 (CANBTR1)

The CANBTR1 register configures various CAN bus timing parameters of the MSCAN module.

	7	6	5	4	3	2	1	0
R W	SAMP	TSEG22	TSEG21	TSEG20	TSEG13	TSEG12	TSEG11	TSEG10
Reset:	0	0	0	0	0	0	0	0

Figure 12-7. MSCAN Bus Timing Register 1 (CANBTR1)

Read: Anytime
Write: Anytime in initialization mode (INITRQ = 1 and INITAK = 1)

Table 12-6. CANBTR1 Register Field Descriptions

Field	Description
7 SAMP	**Sampling** — This bit determines the number of CAN bus samples taken per bit time. 0 One sample per bit. 1 Three samples per bit[1]. If SAMP = 0, the resulting bit value is equal to the value of the single bit positioned at the sample point. If SAMP = 1, the resulting bit value is determined by using majority rule on the three total samples. For higher bit rates, it is recommended that only one sample is taken per bit time (SAMP = 0).
6:4 TSEG2[2:0]	**Time Segment 2** — Time segments within the bit time fix the number of clock cycles per bit time and the location of the sample point (see Figure 12-43). Time segment 2 (TSEG2) values are programmable as shown in Table 12-7.
3:0 TSEG1[3:0]	**Time Segment 1** — Time segments within the bit time fix the number of clock cycles per bit time and the location of the sample point (see Figure 12-43). Time segment 1 (TSEG1) values are programmable as shown in Table 12-8.

1 In this case, PHASE_SEG1 must be at least 2 time quanta (Tq).

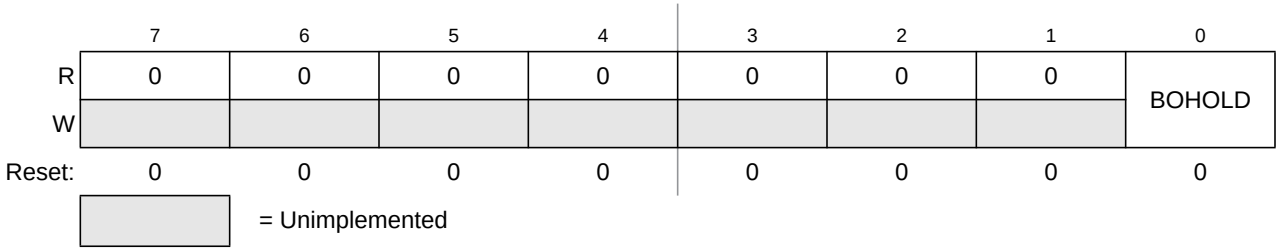


Figure 12-16. MSCAN Miscellaneous Register (CANMISC)

Read: Anytime
Write: Anytime; write of '1' clears flag; write of '0' ignored

Table 12-19. CANMISC Register Field Descriptions

Field	Description
0 BOHOLD	**Bus-off State Hold Until User Request** — If BORM is set in Section 12.3.2, "MSCAN Control Register 1 (CANCTL1), this bit indicates whether the module has entered the bus-off state. Clearing this bit requests the recovery from bus-off. Refer to Section 12.6.2, "Bus-Off Recovery," for details. 0 Module is not bus-off or recovery has been requested by user in bus-off state 1 Module is bus-off and holds this state until user request

12.3.13 MSCAN Receive Error Counter (CANRXERR)

This register reflects the status of the MSCAN receive error counter.

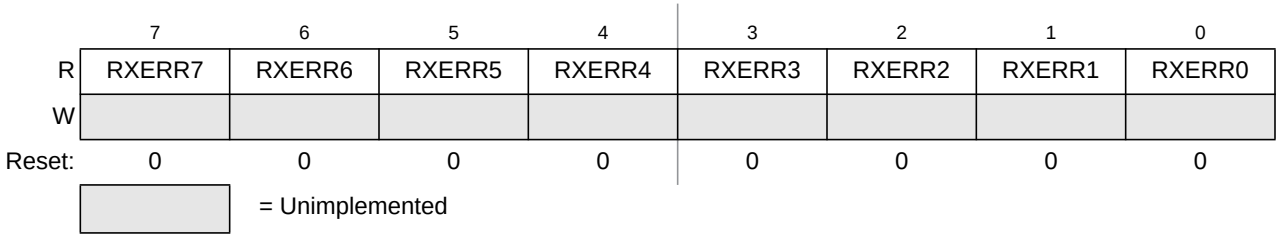


Figure 12-17. MSCAN Receive Error Counter (CANRXERR)

Read: Only when in sleep mode (SLPRQ = 1 and SLPAK = 1) or initialization mode (INITRQ = 1 and INITAK = 1)

Write: Unimplemented

NOTE

Reading this register when in any other mode other than sleep or initialization mode may return an incorrect value. For MCUs with dual CPUs, this may result in a CPU fault condition.

Writing to this register when in special modes can alter the MSCAN functionality.

Table 12-33. Data Length Codes

Data Length Code				Data Byte Count
DLC3	DLC2	DLC1	DLC0	
0	0	0	0	0
0	0	0	1	1
0	0	1	0	2
0	0	1	1	3
0	1	0	0	4
0	1	0	1	5
0	1	1	0	6
0	1	1	1	7
1	0	0	0	8

12.4.5 Transmit Buffer Priority Register (TBPR)

This register defines the local priority of the associated message transmit buffer. The local priority is used for the internal prioritization process of the MSCAN and is defined to be highest for the smallest binary number. The MSCAN implements the following internal prioritization mechanisms:

- All transmission buffers with a cleared TXEx flag participate in the prioritization immediately before the SOF (start of frame) is sent.
- The transmission buffer with the lowest local priority field wins the prioritization.

In cases of more than one buffer having the same lowest priority, the message buffer with the lower index number wins.

	7	6	5	4	3	2	1	0
R W	PRIO7	PRIO6	PRIO5	PRIO4	PRIO3	PRIO2	PRIO1	PRIO0
Reset:	0	0	0	0	0	0	0	0

Figure 12-35. Transmit Buffer Priority Register (TBPR)

Read: Anytime when TXEx flag is set (see Section 12.3.6, "MSCAN Transmitter Flag Register (CANTFLG)") and the corresponding transmit buffer is selected in CANTBSEL (see Section 12.3.10, "MSCAN Transmit Buffer Selection Register (CANTBSEL)").

Write: Anytime when TXEx flag is set (see Section 12.3.6, "MSCAN Transmitter Flag Register (CANTFLG)") and the corresponding transmit buffer is selected in CANTBSEL (see Section 12.3.10, "MSCAN Transmit Buffer Selection Register (CANTBSEL)").

12.4.6 Time Stamp Register (TSRH–TSRL)

If the TIME bit is enabled, the MSCAN will write a time stamp to the respective registers in the active transmit or receive buffer as soon as a message has been acknowledged on the CAN bus (see

- Four identifier acceptance filters, each to be applied to
  - a) the 14 most significant bits of the extended identifier plus the SRR and IDE bits of CAN 2.0B messages or
  - b) the 11 bits of the standard identifier, the RTR and IDE bits of CAN 2.0A/B messages. Figure 12-40 shows how the first 32-bit filter bank (CANIDAR0–CANIDA3, CANIDMR0–3CANIDMR) produces filter 0 and 1 hits. Similarly, the second filter bank (CANIDAR4–CANIDAR7, CANIDMR4–CANIDMR7) produces filter 2 and 3 hits.
- Eight identifier acceptance filters, each to be applied to the first 8 bits of the identifier. This mode implements eight independent filters for the first 8 bits of a CAN 2.0A/B compliant standard identifier or a CAN 2.0B compliant extended identifier. Figure 12-41 shows how the first 32-bit filter bank (CANIDAR0–CANIDAR3, CANIDMR0–CANIDMR3) produces filter 0 to 3 hits. Similarly, the second filter bank (CANIDAR4–CANIDAR7, CANIDMR4–CANIDMR7) produces filter 4 to 7 hits.
- Closed filter. No CAN message is copied into the foreground buffer RxFG, and the RXF flag is never set.

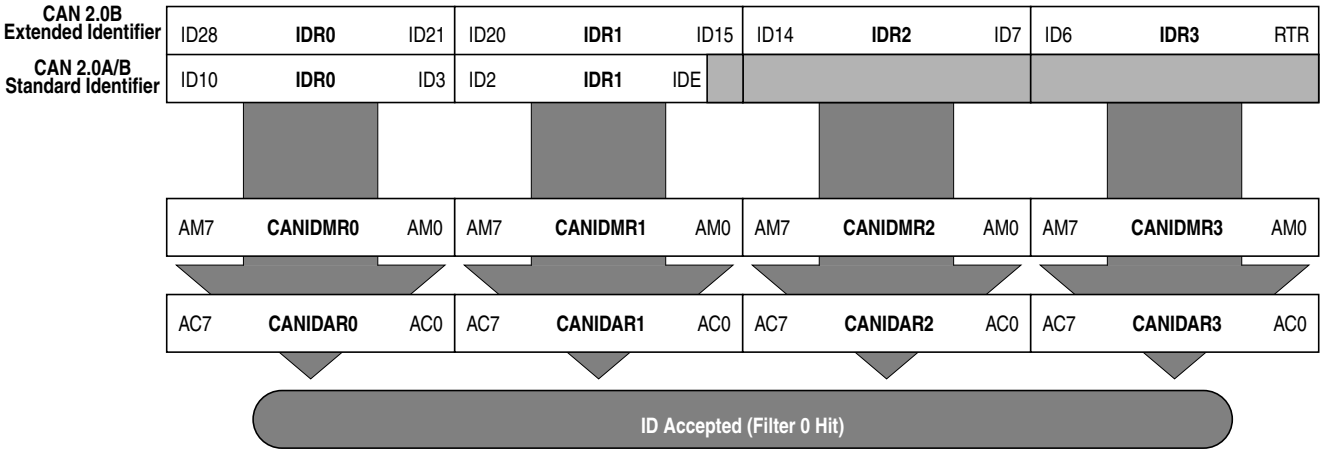


Figure 12-39. 32-bit Maskable Identifier Acceptance Filter

If the bus clock is generated from a PLL, it is recommended to select the oscillator clock rather than the bus clock due to jitter considerations, especially at the faster CAN bus rates. PLL lock may also be too wide to ensure adequate clock tolerance.

For microcontrollers without a clock and reset generator (CRG), CANCLK is driven from the crystal oscillator (oscillator clock).

A programmable prescaler generates the time quanta (Tq) clock from CANCLK. A time quantum is the atomic unit of time handled by the MSCAN.

Eqn. 12-2

$$Tq = \frac{f_{CANCLK}}{(\text{Prescaler value})}$$

A bit time is subdivided into three segments as described in the Bosch CAN specification. (see Figure 12-43):

- SYNC_SEG: This segment has a fixed length of one time quantum. Signal edges are expected to happen within this section.
- Time Segment 1: This segment includes the PROP_SEG and the PHASE_SEG1 of the CAN standard. It can be programmed by setting the parameter TSEG1 to consist of 4 to 16 time quanta.
- Time Segment 2: This segment represents the PHASE_SEG2 of the CAN standard. It can be programmed by setting the TSEG2 parameter to be 2 to 8 time quanta long.

Eqn. 12-3

$$\text{Bit Rate} = \frac{f_{Tq}}{(\text{number of Time Quanta})}$$

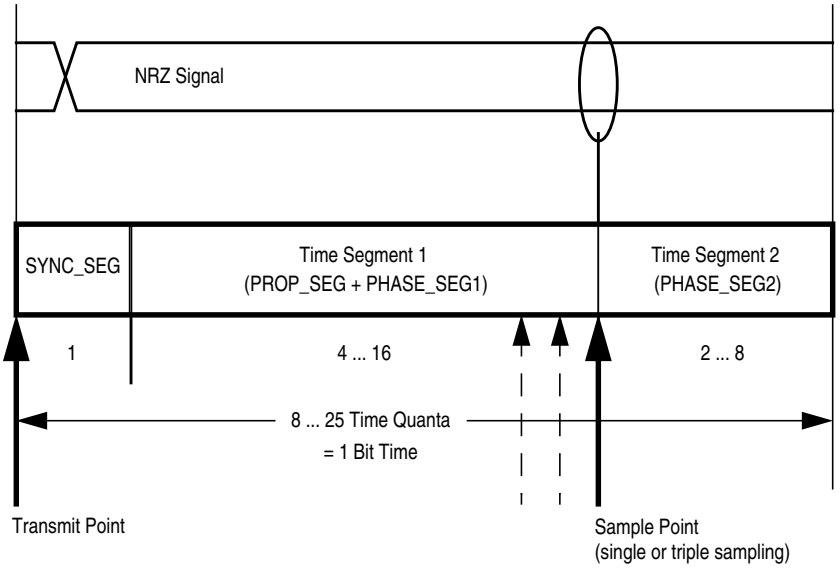


Figure 12-43. Segments within the Bit Time

	7	6	5	4	3	2	1	0
R W	Bit 15	14	13	12	11	10	9	Bit 8
Reset	0	0	0	0	0	0	0	0

Figure 16-13. TPM Channel Value Register High (TPMxCnVH)

	7	6	5	4	3	2	1	0
R W	Bit 7	6	5	4	3	2	1	Bit 0
Reset	0	0	0	0	0	0	0	0

Figure 16-14. TPM Channel Value Register Low (TPMxCnVL)

In input capture mode, reading either byte (TPMxCnVH or TPMxCnVL) latches the contents of both bytes into a buffer where they remain latched until the other half is read. This latching mechanism also resets (becomes unlatched) when the TPMxCnSC register is written (whether BDM mode is active or not). Any write to the channel registers will be ignored during the input capture mode.

When BDM is active, the coherency mechanism is frozen (unless reset by writing to TPMxCnSC register) such that the buffer latches remain in the state they were in when the BDM became active, even if one or both halves of the channel register are read while BDM is active. This assures that if the user was in the middle of reading a 16-bit register when BDM became active, it will read the appropriate value from the other half of the 16-bit value after returning to normal execution. The value read from the TPMxCnVH and TPMxCnVL registers in BDM mode is the value of these registers and not the value of their read buffer.

In output compare or PWM modes, writing to either byte (TPMxCnVH or TPMxCnVL) latches the value into a buffer. After both bytes are written, they are transferred as a coherent 16-bit value into the timer-channel registers according to the value of CLKSB:CLKSA bits and the selected mode, so:

- If (CLKSB:CLKSA = 0:0), then the registers are updated when the second byte is written.
- If (CLKSB:CLKSA not = 0:0 and in output compare mode) then the registers are updated after the second byte is written and on the next change of the TPM counter (end of the prescaler counting).
- If (CLKSB:CLKSA not = 0:0 and in EPWM or CPWM modes), then the registers are updated after the both bytes were written, and the TPM counter changes from (TPMxMODH:TPMxMODL - 1) to (TPMxMODH:TPMxMODL). If the TPM counter is a free-running counter then the update is made when the TPM counter changes from 0xFFFE to 0xFFFF.

The latching mechanism may be manually reset by writing to the TPMxCnSC register (whether BDM mode is active or not). This latching mechanism allows coherent 16-bit writes in either big-endian or little-endian order which is friendly to various compiler implementations.

When BDM is active, the coherency mechanism is frozen such that the buffer latches remain in the state they were in when the BDM became active even if one or both halves of the channel register are written while BDM is active. Any write to the channel registers bypasses the buffer latches and directly write to the channel register while BDM is active. The values written to the channel register while BDM is active

BDM mode returns the latched value of TPMxCNTH:L from the read buffer instead of the frozen TPM counter value.

— This read coherency mechanism is cleared in TPM v3 in BDM mode if there is a write to TPMxSC, TPMxCNTH or TPMxCNTL. Instead, in these conditions the TPM v2 does not clear this read coherency mechanism.

3. Read of TPMxCnVH:L registers (Section 16.3.5, “TPM Channel Value Registers (TPMxCnVH:TPMxCnVL)”)

— In TPM v3, any read of TPMxCnVH:L registers during BDM mode returns the value of the TPMxCnVH:L register. In TPM v2, if only one byte of the TPMxCnVH:L registers was read before the BDM mode became active, then any read of TPMxCnVH:L registers during BDM mode returns the latched value of TPMxCNTH:L from the read buffer instead of the value in the TPMxCnVH:L registers.

— This read coherency mechanism is cleared in TPM v3 in BDM mode if there is a write to TPMxCnSC. Instead, in this condition the TPM v2 does not clear this read coherency mechanism.

4. Write to TPMxCnVH:L registers

— Input Capture Mode (Section 16.4.2.1, “Input Capture Mode)

In this mode the TPM v3 does not allow the writes to TPMxCnVH:L registers. Instead, the TPM v2 allows these writes.

— Output Compare Mode (Section 16.4.2.2, “Output Compare Mode)

In this mode and if (CLKSB:CLKSA not = 0:0), the TPM v3 updates the TPMxCnVH:L registers with the value of their write buffer at the next change of the TPM counter (end of the prescaler counting) after the second byte is written. Instead, the TPM v2 always updates these registers when their second byte is written.

The following procedure can be used in the TPM v3 to verify if the TPMxCnVH:L registers were updated with the new value that was written to these registers (value in their write buffer).

...

write the new value to TPMxCnVH:L;

read TPMxCnVH and TPMxCnVL registers;

while (the read value of TPMxCnVH:L is different from the new value written to TPMxCnVH:L)

begin

read again TPMxCnVH and TPMxCnVL;

end

...

In this point, the TPMxCnVH:L registers were updated, so the program can continue and, for example, write to TPMxC0SC without cancelling the previous write to TPMxCnVH:L registers.

— Edge-Aligned PWM (Section 16.4.2.3, “Edge-Aligned PWM Mode)

In this mode and if (CLKSB:CLKSA not = 00), the TPM v3 updates the TPMxCnVH:L registers with the value of their write buffer after that the both bytes were written and when the

- Non-intrusive commands can be executed at any time even while the user's program is running. Non-intrusive commands allow a user to read or write MCU memory locations or access status and control registers within the background debug controller.

Typically, a relatively simple interface pod is used to translate commands from a host computer into commands for the custom serial interface to the single-wire background debug system. Depending on the development tool vendor, this interface pod may use a standard RS-232 serial port, a parallel printer port, or some other type of communications such as a universal serial bus (USB) to communicate between the host PC and the pod. The pod typically connects to the target system with ground, the BKGD pin, $\overline{\text{RESET}}$, and sometimes V_{DD}. An open-drain connection to reset allows the host to force a target system reset, which is useful to regain control of a lost target system or to control startup of a target system before the on-chip nonvolatile memory has been programmed. Sometimes V_{DD} can be used to allow the pod to use power from the target system to avoid the need for a separate power supply. However, if the pod is powered separately, it can be connected to a running target system without forcing a target system reset or otherwise disturbing the running application program.

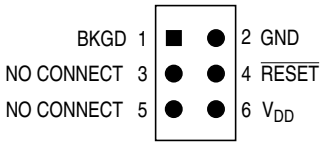


Figure 17-1. BDM Tool Connector

17.2.1 BKGD Pin Description

BKGD is the single-wire background debug interface pin. The primary function of this pin is for bidirectional serial communication of active background mode commands and data. During reset, this pin is used to select between starting in active background mode or starting the user's application program. This pin is also used to request a timed sync response pulse to allow a host development tool to determine the correct clock frequency for background debug serial communications.

BDC serial communications use a custom serial protocol first introduced on the M68HC12 Family of microcontrollers. This protocol assumes the host knows the communication clock rate that is determined by the target BDC clock rate. All communication is initiated and controlled by the host that drives a high-to-low edge to signal the beginning of each bit time. Commands and data are sent most significant bit first (MSB first). For a detailed description of the communications protocol, refer to Section 17.2.2, "Communication Details."

If a host is attempting to communicate with a target MCU that has an unknown BDC clock rate, a SYNC command may be sent to the target MCU to request a timed sync response signal from which the host can determine the correct communication speed.

BKGD is a pseudo-open-drain pin and there is an on-chip pullup so no external pullup resistor is required. Unlike typical open-drain pins, the external RC time constant on this pin, which is influenced by external capacitance, plays almost no role in signal rise time. The custom protocol provides for brief, actively driven speedup pulses to force rapid rise times on this pin without risking harmful drive level conflicts. Refer to Section 17.2.2, "Communication Details," for more detail.

Table A-7. Supply Current Characteristics (continued)

Num	C	Parameter	Symbol	V_{DD} (V)	Typical[1]	Max[2]	Unit
6	C	RTC adder to stop2 or stop3[5], 25°C		5	300	—	nA
				3	300	—	nA
7	C	LVD adder to stop3 (LVDE = LVDSE = 1)		5	110	—	µA
				3	90	—	µA
8	C	Adder to stop3 for oscillator enabled[6] (IRCLKEN = 1 and IREFSTEN = 1 or ERCLKEN = 1 and EREFSTEN = 1)		5	5	—	µA
				3	5	—	µA

1 Typicals are measured at 25°C, unless otherwise noted.

2 Maximum values in this column apply for the full operating temperature range of the device unless otherwise noted.

3 All modules except ADC active, MCG configured for FBE, and does not include any dc loads on port pins

4 Stop currents are tested in production for 25°C on all parts. Tests at other temperatures depend upon the part number suffix and maturity of the product. Freescale may eliminate a test insertion at a particular temperature from the production test flow once sufficient data has been collected and is approved.

5 Most customers are expected to find that auto-wakeup from stop2 or stop3 can be used instead of the higher current wait mode.

6 Values given under the following conditions: low range operation (RANGE = 0), low power mode (HGO = 0).

A.8 Analog Comparator (ACMP) Electricals

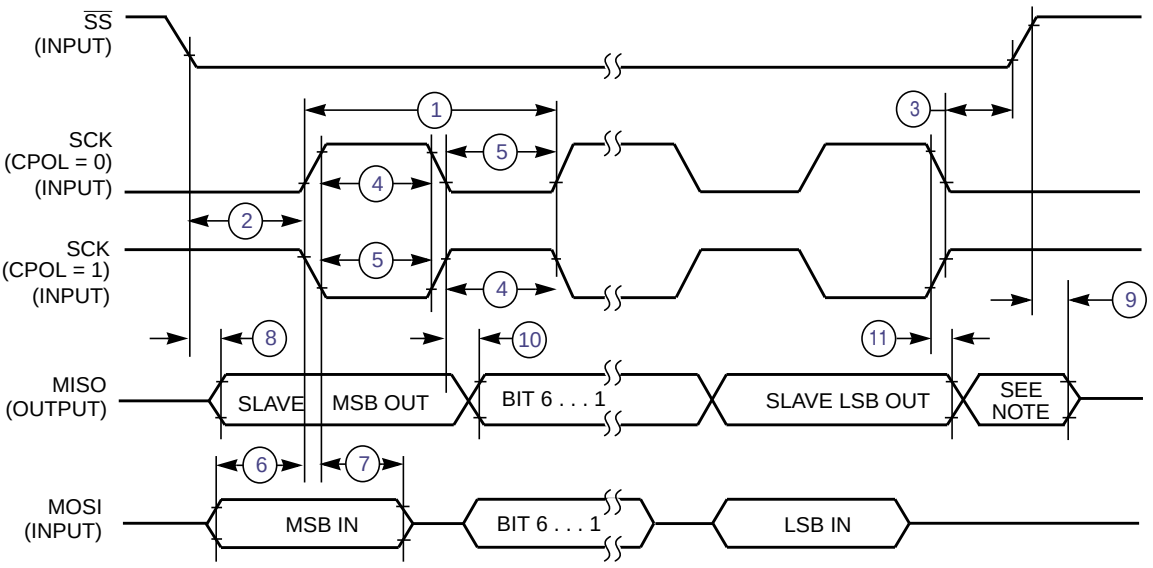
Table A-8. Analog Comparator Electrical Specifications

Num	C	Rating	Symbol	Min	Typical	Max	Unit
1	—	Supply voltage	V_{DD}	2.7	—	5.5	V
2	D	Supply current (active)	I_{DDAC}	—	20	35	µA
3	D	Analog input voltage	V_{AIN}	V_{SS} – 0.3	—	V_{DD}	V
4	D	Analog input offset voltage	V_{AIO}		20	40	mV
5	D	Analog Comparator hysteresis	V_H	3.0	6.0	20.0	mV
6	D	Analog input leakage current	I_{ALKG}	--	--	1.0	µA
7	D	Analog Comparator initialization delay	t_{AINIT}	—	—	1.0	µs

A.9 ADC Characteristics

Table A-9. 12-bit ADC Operating Conditions

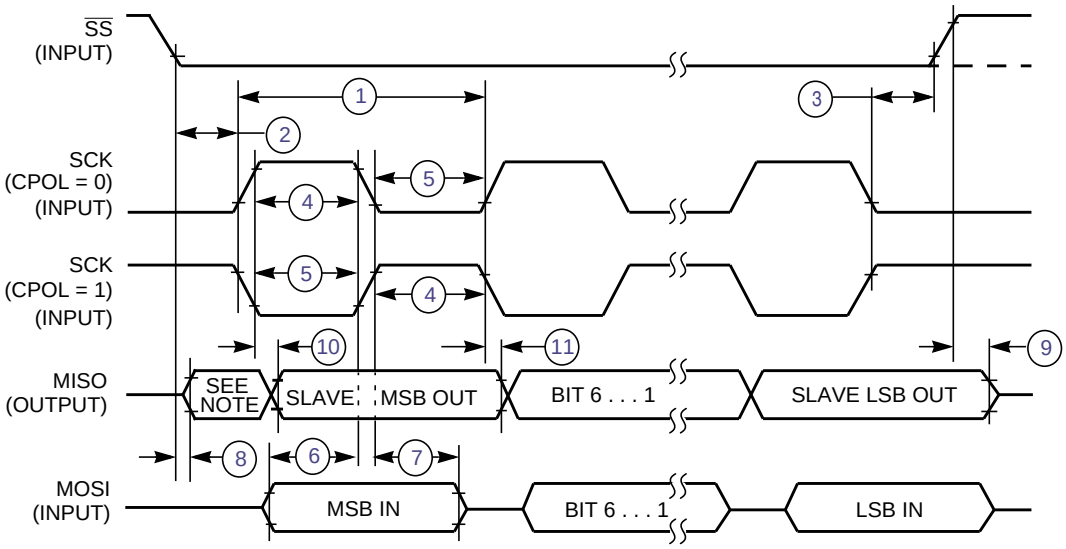
Characteristic	Conditions	Symb	Min	Typ[1]	Max	Unit	Comment
Supply voltage	Absolute	V_{DDAD}	2.7	—	5.5	V	
	Delta to V_{DD} (V_{DD}-V_{DDAD})[2]	ΔV_{DDAD}	-100	0	+100	mV	
Ground voltage	Delta to V_{SS} (V_{SS}-V_{SSAD})[2]	ΔV_{SSAD}	-100	0	+100	mV	



NOTE:

1. Not defined but normally MSB of character just received

Figure A-9. SPI Slave Timing (CPHA = 0)



NOTE:

1. Not defined but normally LSB of character just received

Figure A-10. SPI Slave Timing (CPHA = 1)