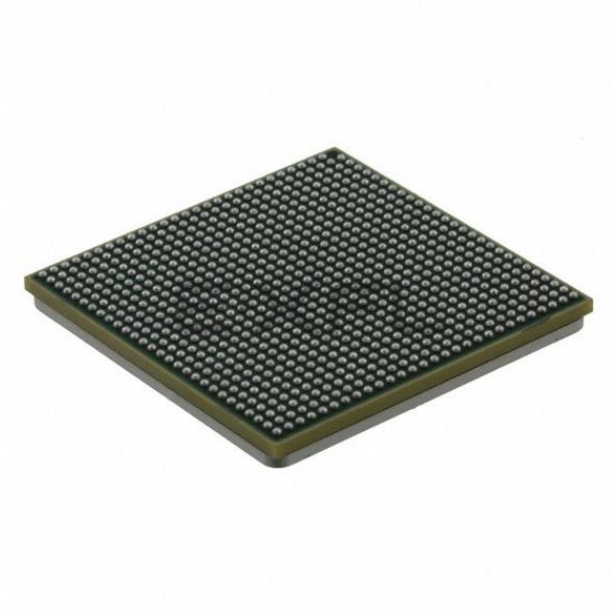

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	600MHz
Co-Processors/DSP	-
RAM Controllers	DDR2, DDR3
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	SATA 3Gbps (1)
USB	USB 2.0 (2)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8535bvtakga

1.1 Pin Map

See Table 1 for the MPC8535E pinout, which is a subset of the MPC8536E.

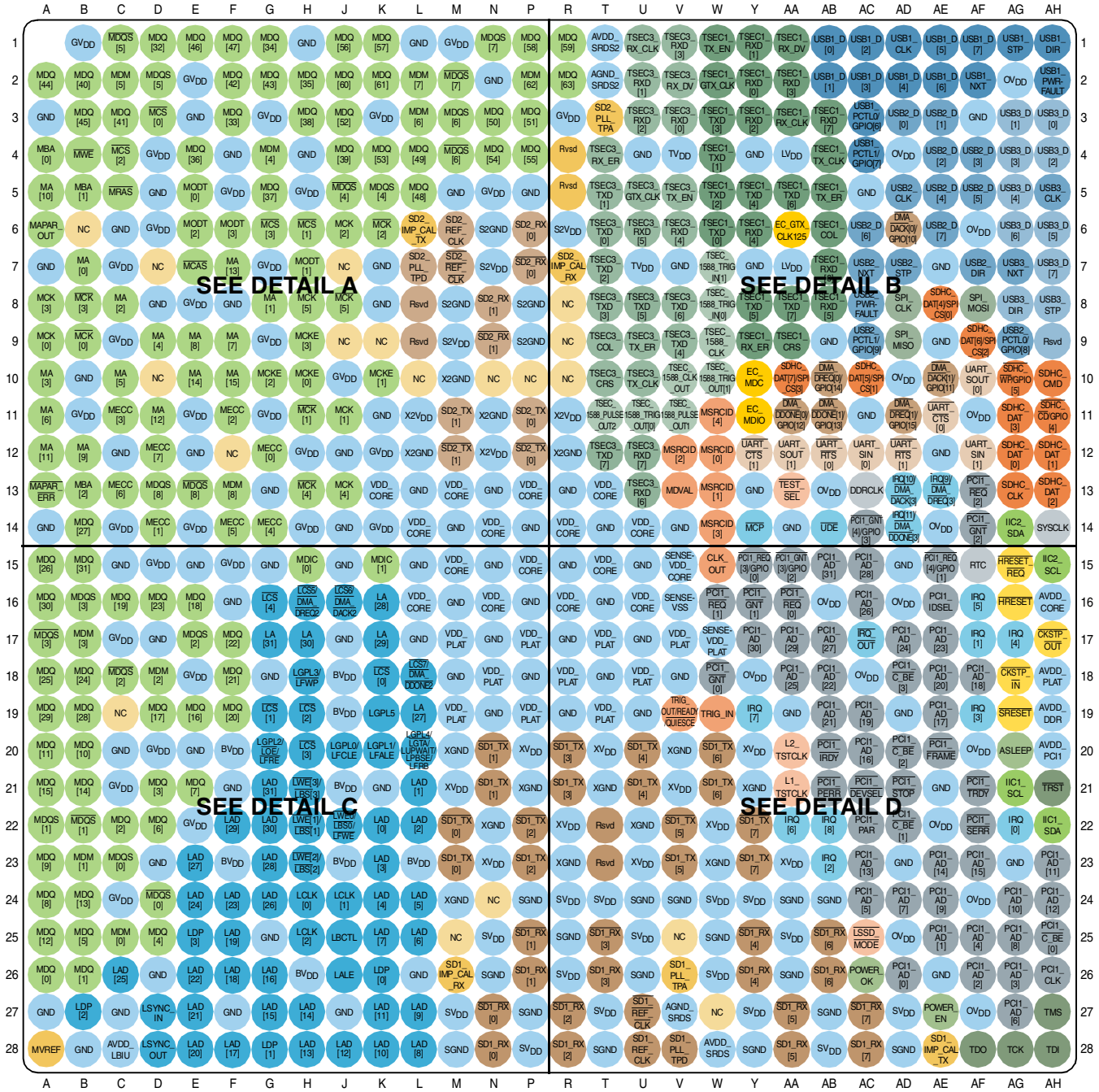


Figure 2. Chip Pin Map Bottom View

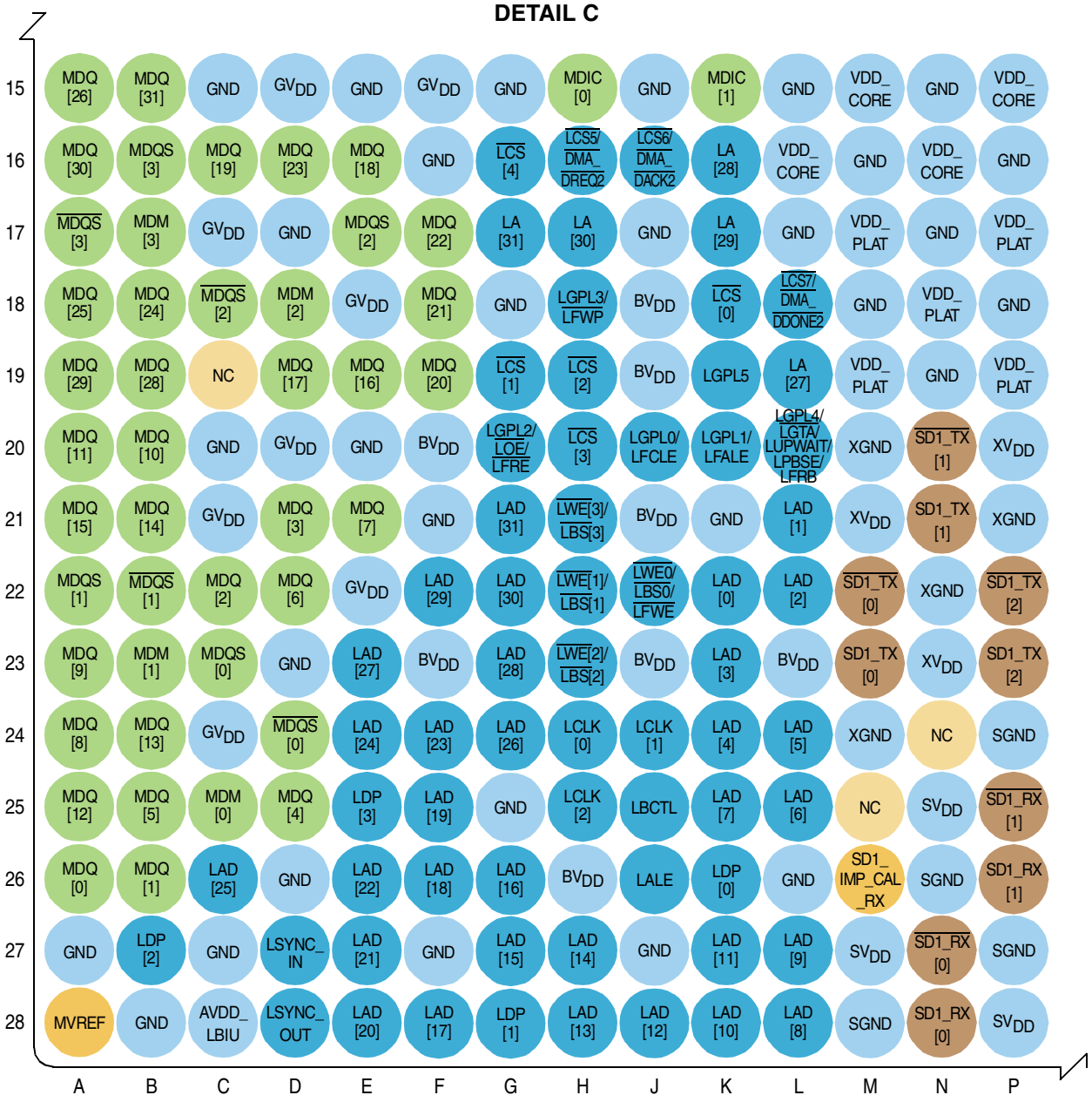


Figure 5. Chip Pin Map Detail C

Table 1. Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes
Programmable Interrupt Controller					
$\overline{\text{MCP}}$	Machine check processor	Y14	I	OV_{DD}	—
$\overline{\text{UDE}}$	Unconditional debug event	AB14	I	OV_{DD}	—
IRQ[0:8]	External interrupts	AG22,AF17,AB23, AF19,AG17,AF16, AA22,Y19,AB22	I	OV_{DD}	—
IRQ[9]/$\overline{\text{DMA\_DREQ[3]}}$	External interrupt/DMA request	AE13	I	OV_{DD}	1
IRQ[10]/$\overline{\text{DMA\_DACK[3]}}$	External interrupt/DMA Ack	AD13	I/O	OV_{DD}	1
IRQ[11]/$\overline{\text{DMA\_DDONE[3]}}$	External interrupt/DMA done	AD14	I/O	OV_{DD}	1
$\overline{\text{IRQ\_OUT}}$	Interrupt output	AC17	O	OV_{DD}	2,4
Ethernet Management Interface					
EC_MDC	Management data clock	Y10	O	OV_{DD}	5,9,22
EC_MDIO	Management data In/Out	Y11	I/O	OV_{DD}	—
Gigabit Reference Clock					
EC_GTX_CLK125	Reference clock	AA6	I	LV_{DD}	31
Three-Speed Ethernet Controller (Gigabit Ethernet 1)					
TSEC1_TXD[7:0]	Transmit data	AA8,AA5,Y8,Y5,W3, W5,W4,W6	O	LV_{DD}	5,9,22
TSEC1_TX_EN	Transmit Enable	W1	O	LV_{DD}	23
TSEC1_TX_ER	Transmit Error	AB5	O	LV_{DD}	5,9
TSEC1_TX_CLK	Transmit clock In	AB4	I	LV_{DD}	—
TSEC1_GTX_CLK	Transmit clock Out	W2	O	LV_{DD}	—
TSEC1_CRS	Carrier sense	AA9	I/O	LV_{DD}	17
TSEC1_COL	Collision detect	AB6	I	LV_{DD}	—
TSEC1_RXD[7:0]	Receive data	AB3,AB7,AB8,Y6,AA2, Y3,Y1,Y2	I	LV_{DD}	—
TSEC1_RX_DV	Receive data valid	AA1	I	LV_{DD}	—
TSEC1_RX_ER	Receive data error	Y9	I	LV_{DD}	—
TSEC1_RX_CLK	Receive clock	AA3	I	LV_{DD}	—
Three-Speed Ethernet Controller (Gigabit Ethernet 3)					
TSEC3_TXD[7:0]	Transmit data	T12,V8,U8,V9,T8,T7, T5,T6	O	TV_{DD}	5,9,22
TSEC3_TX_EN	Transmit Enable	V5	O	TV_{DD}	23
TSEC3_TX_ER	Transmit Error	U9	O	TV_{DD}	5,9

Table 1. Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes
General-Purpose Input/Output					
GPIO[0:1]/PCI1_REQ[3:4]	GPIO/PCI request	Y15,AE15	I/O	OV_{DD}	—
GPIO[2:3]/PCI1_GNT[3:4]	GPIO/PCI grant	AA15,AC14	I/O	OV_{DD}	—
GPIO[4]/SDHC_CD	GPIO/SDHC card detection	AH11	I/O	OV_{DD}	—
GPIO[5]/SDHC_WP	GPIO/SDHC write protection	AG10	I/O	OV_{DD}	32
GPIO[6]/USB1_PCTL0	GPIO/USB1 PCTL0	AC3	I/O	OV_{DD}	—
GPIO[7]/USB1_PCTL1	GPIO/USB1 PCTL1	AC4	I/O	OV_{DD}	—
GPIO[8]/USB2_PCTL0	GPIO/USB2 PCTL0	AG9	I/O	OV_{DD}	—
GPIO[9]/USB2_PCTL1	GPIO/USB2 PCTL1	AC9	I/O	OV_{DD}	—
GPIO[10:11] /DMA_DACK[0:1]	GPIO/DMA Ack	AD6,AE10	I/O	OV_{DD}	—
GPIO[12:13] /DMA_DDONE[0:1]	GPIO/DMA done	AA11,AB11	I/O	OV_{DD}	—
GPIO[14:15] /DMA_DREQ[0:1]	GPIO/DMA request	AB10,AD11	I/O	OV_{DD}	—
System Control					
HRESET	Hard reset	AG16	I	OV_{DD}	—
HRESET_REQ	Hard reset - request	AG15	O	OV_{DD}	22
SRESET	Soft reset	AG19	I	OV_{DD}	—
CKSTP_IN	CheckStop in	AG18	I	OV_{DD}	—
CKSTP_OUT	CheckStop Output	AH17	O	OV_{DD}	2,4
Debug					
TRIG_IN	Trigger in	W19	I	OV_{DD}	—
TRIG_OUT/READY /QUIESCE	Trigger out/Ready/Quiesce	V19	O	OV_{DD}	22
MSRCID[0:1]	Memory debug source port ID	W12,W13	O	OV_{DD}	6,9
MSRCID[2:4]	Memory debug source port ID	V12, W14,W11	O	OV_{DD}	6,9,22
MDVAL	Memory debug data valid	V13	O	OV_{DD}	6,22
CLK_OUT	Clock Out	W15	O	OV_{DD}	11
Clock					
RTC	Real time clock	AF15	I	OV_{DD}	—
SYSCLK	System clock / PCI clock	AH14	I	OV_{DD}	—
DDRCLK	DDR clock	AC13	I	OV_{DD}	30
JTAG					

Table 1. Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes
SGND	SerDes 1 Transceiver core logic GND (xcorevss)	M28,N26,P24,P27, R25,T28,U24,U26,V24, W25,Y28,AA24,AA26, AB24,AB27,AD28	—	—	—
X2GND	SerDes 2 Transceiver pad GND (xpadvss)	R12,M10,N11,L12	—	—	—
S2GND	SerDes 2 Transceiver core logic GND (xcorevss)	P8,P9,N6,M8	—	—	—
AGND_SRDS	SerDes 1 PLL GND	V27	—	—	—
AGND_SRDS2	SerDes 2 PLL GND	T2	—	—	—
SENSEVSS	GND Sensing	V16	—	—	13
Analog Signals					
MVREF	SSTL2 reference voltage	A28	Reference voltage for DDR	GVDD/2	—
SD1_IMP_CAL_RX	Rx impedance calibration	M26	—	200Ω (±1%) to GND	—
SD1_IMP_CAL_TX	Tx impedance calibration	AE28	—	100Ω (±1%) to GND	—
SD1_PLL_TPA	PLL test point analog	V26	—	AVDD_SRDS analog	18
SD2_IMP_CAL_RX	Rx impedance calibration	R7	—	200Ω (±1%) to GND	—
SD2_IMP_CAL_TX	Tx impedance calibration	L6	—	100Ω (±1%) to GND	—
SD2_PLL_TPA	PLL test point analog	T3	—	AVDD_SRDS2 analog	18
Reserved	—	R4	—	—	—
Reserved	—	R5	—	—	—
No Connect Pins					
NC	—	C19,D7,D10,L10,R10, B6,F12,J7,P10,M25, W27,N24,N10,R8,J9, K9,V25,R9	—	—	—

2.1.1 Absolute Maximum Ratings

This table provides the absolute maximum ratings.

Table 2. Absolute Maximum Ratings[1]

Characteristic		Symbol	Max Value	Unit	Notes
Core supply voltage		V_{DD_CORE}	–0.3 to 1.21	V	—
Platform supply voltage		V_{DD_PLAT}	–0.3 to 1.1	V	—
PLL core supply voltage		AV_{DD_CORE}	–0.3 to 1.21	V	—
PLL other supply voltage		AV_{DD}	–0.3 to 1.1	V	—
Core power supply for SerDes transceivers		SV_{DD}, $S2V_{DD}$	–0.3 to 1.1	V	—
Pad power supply for SerDes transceivers and PCI Express		XV_{DD}, $X2V_{DD}$	–0.3 to 1.1	V	—
DDR SDRAM Controller I/O supply voltage	DDR2 SDRAM Interface	GV_{DD}	–0.3 to 1.98	V	—
	DDR3 SDRAM Interface		–0.3 to 1.65		
Three-speed Ethernet I/O		LV_{DD} (eTSEC1)	–0.3 to 3.63 –0.3 to 2.75	V	2
		TV_{DD} (eTSEC3)	–0.3 to 3.63 –0.3 to 2.75	V	2
PCI, DUART, system control and power management, I^2C, USB, eSDHC, eSPI and JTAG I/O voltage, MII management voltage		OV_{DD}	–0.3 to 3.63	V	—
Local bus I/O voltage		BV_{DD}	–0.3 to 3.63 –0.3 to 2.75 –0.3 to 1.98	V	—
Input voltage	DDR2/DDR3 DRAM signals	MV_{IN}	–0.3 to (GV_{DD} + 0.3)	V	3
	DDR2/DDR3 DRAM reference	MV_{REF}	–0.3 to (GV_{DD} + 0.3)	V	—
	Three-speed Ethernet signals	LV_{IN} TV_{IN}	–0.3 to (LV_{DD} + 0.3) –0.3 to (TV_{DD} + 0.3)	V	3
	Local bus signals	BV_{IN}	–0.3 to (BV_{DD} + 0.3)	—	—
	PCI, DUART, SYSCLK, system control and power management, I^2C, and JTAG signals	OV_{IN}	–0.3 to (OV_{DD} + 0.3)	V	3
Storage temperature range		T_{STG}	–55 to 150	^0C	—

Notes:

1. Functional and tested operating conditions are given in Table 3. Absolute maximum ratings are stress ratings only, and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect chip reliability or cause permanent damage to the chip.
2. The 3.63-V maximum is only supported when the port is configured in GMII, MII, RMII or TBI modes; otherwise the 2.75V maximum applies. See Section 2.9.2, "FIFO, GMII, MII, TBI, RGMII, RMII, and RTBI AC Timing Specifications," for details on the recommended operating conditions per protocol.
3. (M,L,O)V_{IN} and MV_{REF} may overshoot/undershoot to a voltage and for a maximum duration as shown in Figure 7.

2.1.2 Recommended Operating Conditions

This table provides the recommended operating conditions for this chip. Note that the values in this table are the recommended and tested operating conditions. Proper chip operation outside these conditions is not guaranteed.

Table 5. Power Dissipation (continued)[5]

Power Mode	Core Frequency	CCB Frequency	DDR Frequency	V_{DD} Platform	V_{DD} Core	Junction Temperature	Core Power		Platform Power[9]		Notes
	(MHz)	(MHz)	(MHz)	(V)	(V)	(°C)	mean[7]	Max	mean[7]	Max	
Maximum (A)	1250	500	500	1.0	1.0	105 / 90	—	5.3/4.4	—	5.0/4.0	1, 3, 8
Thermal (W)							—	4.4/3.6	—	5.0/4.0	1, 4, 8
Typical (W)						65	2.2		1.7		1
Doze (W)							1.6	2.4	1.5	2.1	1
Nap (W)							0.8	1.6	1.5	2.1	1
Sleep (W)							0.8	1.6	1.1	1.7	1
Deep Sleep (W)						35	0	0	0.6	1.2	1, 6

Notes:

1. These values specify the power consumption at nominal voltage and apply to all valid processor bus frequencies and configurations. The values do not include power dissipation for I/O supplies.
2. Typical power is an average value measured at the nominal recommended core voltage (V_{DD}) and 65°C junction temperature (see Table 3) while running the Dhrystone benchmark.
3. Maximum power is the maximum power measured with the worst process and recommended core and platform voltage (V_{DD}) at maximum operating junction temperature (see Table 3) while running a smoke test which includes an entirely L1-cache-resident, contrived sequence of instructions which keep the execution unit maximally busy.
4. Thermal power is the maximum power measured with worst case process and recommended core and platform voltage (V_{DD}) at maximum operating junction temperature (see Table 3) while running the Dhrystone benchmark.
6. Maximum power is the maximum number measured with USB1, eTSEC1, and DDR blocks enabled. The Mean power is the mean power measured with only external interrupts enabled and DDR in self refresh.
7. Mean power is provided for information purposes only and is the mean power consumed by a statistically significant range of devices.
8. Maximum operating junction temperature (see Table 3) for Commercial Tier is 90 ^{0}C, for Industrial Tier is 105 ^{0}C.
9. Platform power is the power supplied to all the V_{DD_PLAT} pins.

See Section 2.23.6.1, "SYSCLK to Platform Frequency Options," for the full range of CCB frequencies that the chip supports.

This figure shows the GMII transmit AC timing diagram.

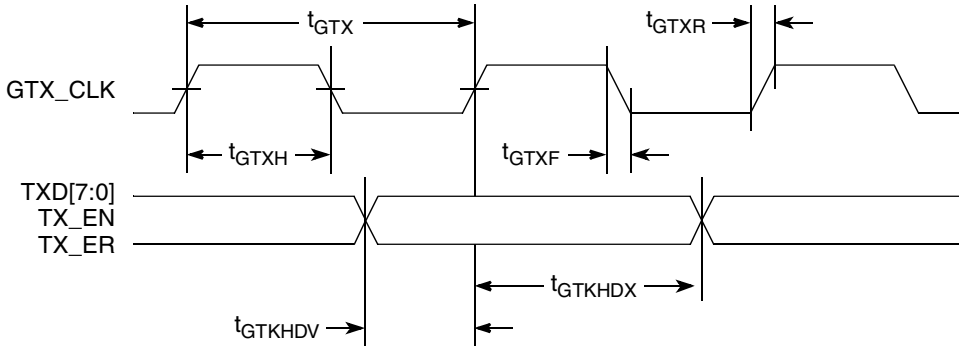


Figure 16. GMII Transmit AC Timing Diagram

2.9.2.2.2 GMII Receive AC Timing Specifications

This table provides the GMII receive AC timing specifications.

Table 29. GMII Receive AC Timing Specifications

At recommended operating conditions with L/TV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
RX_CLK clock period	t_{GRX}	—	8.0	—	ns
RX_CLK duty cycle	t_{GRXH}/t_{GRX}	35	—	65	%
RXD[7:0], RX_DV, RX_ER setup time to RX_CLK	t_{GRDVKH}	2.0	—	—	ns
RXD[7:0], RX_DV, RX_ER hold time to RX_CLK	t_{GRDXKH}	0	—	—	ns
RX_CLK clock rise (20%-80%)	t_{GRXR}	—	—	1.0	ns
RX_CLK clock fall time (80%-20%)	t_{GRXF}	—	—	1.0	ns

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{GRDVKH} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) reaching the valid state (V) relative to the t_{RX} clock reference (K) going to the high state (H) or setup time. Also, t_{GRDXKL} symbolizes GMII receive timing (GR) with respect to the time data input signals (D) went invalid (X) relative to the t_{GRX} clock reference (K) going to the low (L) state or hold time. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{GRX} represents the GMII (G) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

This figure provides the AC test load for eTSEC.

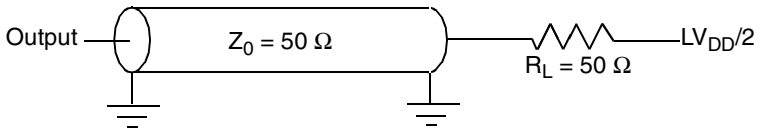


Figure 17. eTSEC AC Test Load

A timing diagram for TBI receive appears in the following figure.

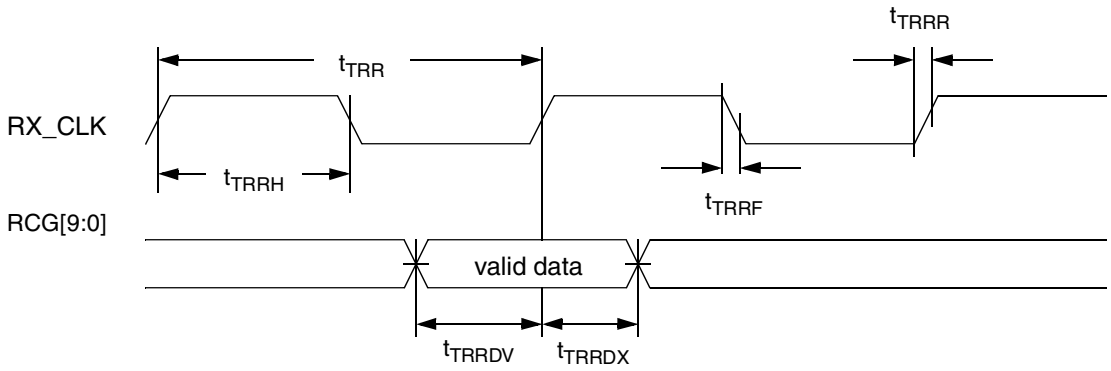


Figure 24. TBI Single-Clock Mode Receive AC Timing Diagram

2.9.2.6 RGMII and RTBI AC Timing Specifications

This table presents the RGMII and RTBI AC timing specifications.

Table 35. RGMII and RTBI AC Timing Specifications

At recommended operating conditions with L/TV_{DD} of 2.5 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)	t_{SKRGT_TX}	–500	0	500	ps
Data to clock input skew (at receiver) [2]	t_{SKRGT_RX}	1.0	—	2.8	ns
Clock period duration [3]	t_{RGT}	7.2	8.0	8.8	ns
Duty cycle for 1000BASE-T[4]	t_{RGTH}/t_{RGT}	45	—	55	%
Duty cycle for 10BASE-T and 100BASE-TX [3, 4]	t_{RGTH}/t_{RGT}	40	50	60	%
Rise time (20%–80%)	t_{RGTR}	—	—	0.75	ns
Fall time (20%–80%)	t_{RGTF}	—	—	0.75	ns

Notes:

1. Note that, in general, the clock reference symbol representation for this section is based on the symbols RGT to represent RGMII and RTBI timing. For example, the subscript of t_{RGT} represents the TBI (T) receive (RX) clock. Note also that the notation for rise (R) and fall (F) times follows the clock symbol that is being represented. For symbols representing skews, the subscript is skew (SK) followed by the clock that is being skewed (RGT).
2. This implies that PC board design will require clocks to be routed such that an additional trace delay of greater than 1.5 ns will be added to the associated clock signal.
3. For 10 and 100 Mbps, t_{RGT} scales to 400 ns ± 40 ns and 40 ns ± 4 ns, respectively.
4. Duty cycle may be stretched/shrunk during speed changes or while transition to a received packet's clock domains as long as the minimum duty cycle is not violated and stretching occurs for no more than three t_{RGT} of the lowest speed transitioned between.

2.18 GPIO

This section describes the DC and AC electrical specifications for the GPIO interface of the chip.

2.18.1 GPIO DC Electrical Characteristics

This table provides the DC electrical characteristics for the GPIO interface.

Table 65. GPIO DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	– 0.3	0.8	V
Input current (V_{IN} [1] = 0 V or V_{IN} = V_{DD})	I_{IN}	—	±5	μA
High-level output voltage (OV_{DD} = min, I_{OH} = –2 mA)	V_{OH}	2.4	—	V
Low-level output voltage (OV_{DD} = min, I_{OL} = 2 mA)	V_{OL}	—	0.4	V

Note:

1. The symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

2.18.2 GPIO AC Electrical Specifications

This table provides the GPIO input and output AC timing specifications.

Table 66. GPIO Input and Output AC Timing Specifications[1]

Characteristic	Symbol [2]	Min	Unit	Notes
GPIO inputs—minimum pulse width	t_{PIWID}	7.5	ns	3
GPIO outputs—minimum pulse width	t_{GTOWID}	12	ns	—

Notes:

1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. GPIO inputs and outputs are asynchronous to any visible clock. GPIO outputs should be synchronized before use by any external synchronous logic. GPIO inputs are required to be valid for at least t_{PIWID} ns to ensure proper operation.
3. The minimum pulse width is a function of the MPX/Platform clock. The minimum pulse width must be greater than or equal to 4 times the MPX/Platform clock period.

This figure provides the AC test load for the GPIO.

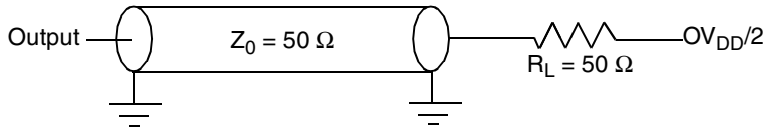


Figure 53. GPIO AC Test Load

peak. For example, the output differential peak-peak voltage can also be calculated as $V_{TX\text{-}DIFFp\text{-}p} = 2*|V_{OD}|$.

6. Common Mode Voltage, V_{cm}

The Common Mode Voltage is equal to one half of the sum of the voltages between each conductor of a balanced interchange circuit and ground. In this example, for SerDes output, $V_{cm\_out} = V_{SDn\_TX} + V_{\overline{SDn\_TX}} = (A + B) / 2$, which is the arithmetic mean of the two complimentary output voltages within a differential pair. In a system, the common mode voltage may often differ from one component's output to the other's input. Sometimes, it may be even different between the receiver input and driver output circuits within the same component. It is also referred as the DC offset in some occasion.

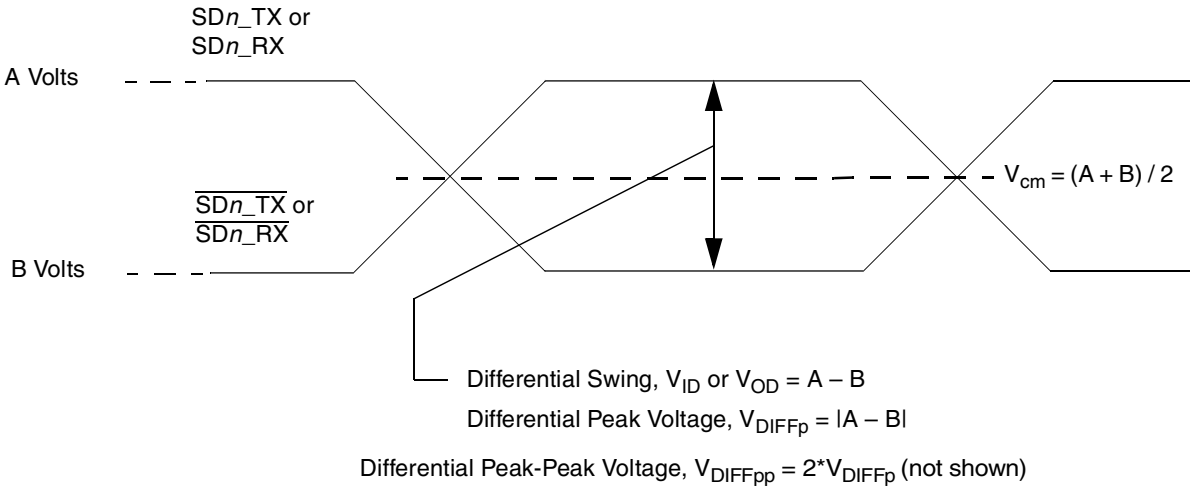


Figure 57. Differential Voltage Definitions for Transmitter or Receiver

To illustrate these definitions using real values, consider the case of a CML (Current Mode Logic) transmitter that has a common mode voltage of 2.25 V and each of its outputs, TD and $\overline{TD}$, has a swing that goes between 2.5V and 2.0V. Using these values, the peak-to-peak voltage swing of each signal (TD or $\overline{TD}$) is 500 mV p-p, which is referred as the single-ended swing for each signal. In this example, since the differential signaling environment is fully symmetrical, the transmitter output's differential swing (V_{OD}) has the same amplitude as each signal's single-ended swing. The differential output signal ranges between 500 mV and –500 mV, in other words, V_{OD} is 500 mV in one phase and –500 mV in the other phase. The peak differential voltage (V_{DIFFp}) is 500 mV. The peak-to-peak differential voltage ($V_{DIFFp\text{-}p}$) is 1000 mV p-p.

2.20.2 SerDes Reference Clocks

The SerDes reference clock inputs are applied to an internal PLL whose output creates the clock used by the corresponding SerDes lanes. The SerDes reference clocks for PCI Express are SD1_REF_CLK and, $\overline{SD1\_REF\_CLK}$. The SerDes reference clocks for the SATA and SGMII interfaces are SD2_REF_CLK and, $\overline{SD2\_REF\_CLK}$.

The following sections describe the SerDes reference clock requirements and some application information.

2.20.2.1 SerDes Reference Clock Receiver Characteristics

Figure 58 shows a receiver reference diagram of the SerDes reference clocks.

- The supply voltage requirements for $X2V_{DD}$ are specified in Table 2 and Table 3.
- SerDes Reference Clock Receiver Reference Circuit Structure
  - The SDn_REF_CLK and $\overline{SDn\_REF\_CLK}$ are internally AC-coupled differential inputs as shown in Figure 58. Each differential clock input (SD*n*_REF_CLK or $\overline{SDn\_REF\_CLK}$) has a 50-Ω termination to SGND (xcorevss) followed by on-chip AC-coupling.

— The external reference clock driver must be able to drive this termination.
— The SerDes reference clock input can be either differential or single-ended. See the Differential Mode and Single-ended Mode description below for further detailed requirements.

- The maximum average current requirement that also determines the common mode voltage range
  — When the SerDes reference clock differential inputs are DC coupled externally with the clock driver chip, the maximum average current allowed for each input pin is 8mA. In this case, the exact common mode input voltage is not critical as long as it is within the range allowed by the maximum average current of 8 mA (refer to the following bullet for more detail), since the input is AC-coupled on-chip.
  — This current limitation sets the maximum common mode input voltage to be less than 0.4 V (0.4 V/50 = 8 mA) while the minimum common mode input level is 0.1V above SnGND (xcorevss). For example, a clock with a 50/50 duty cycle can be produced by a clock driver with output driven by its current source from 0mA to 16mA (0–0.8 V), such that each phase of the differential input has a single-ended swing from 0 V to 800mV with the common mode voltage at 400mV.
  — If the device driving the SD*n*_REF_CLK and $\overline{\text{SD}n\text{\_REF\_CLK}}$ inputs cannot drive 50 Ω to SnGND (xcorevss) DC, or it exceeds the maximum input current limitations, then it must be AC-coupled off-chip.
- The input amplitude requirement
  — This requirement is described in detail in the following sections.

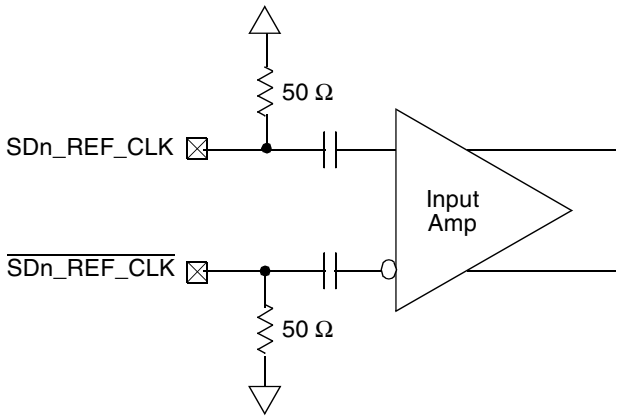


Figure 58. Receiver of SerDes Reference Clocks

This figure shows the SerDes reference clock connection reference circuits for HCSL type clock driver. It assumes that the DC levels of the clock driver chip is compatible with chip's SerDes reference clock input's DC requirement.

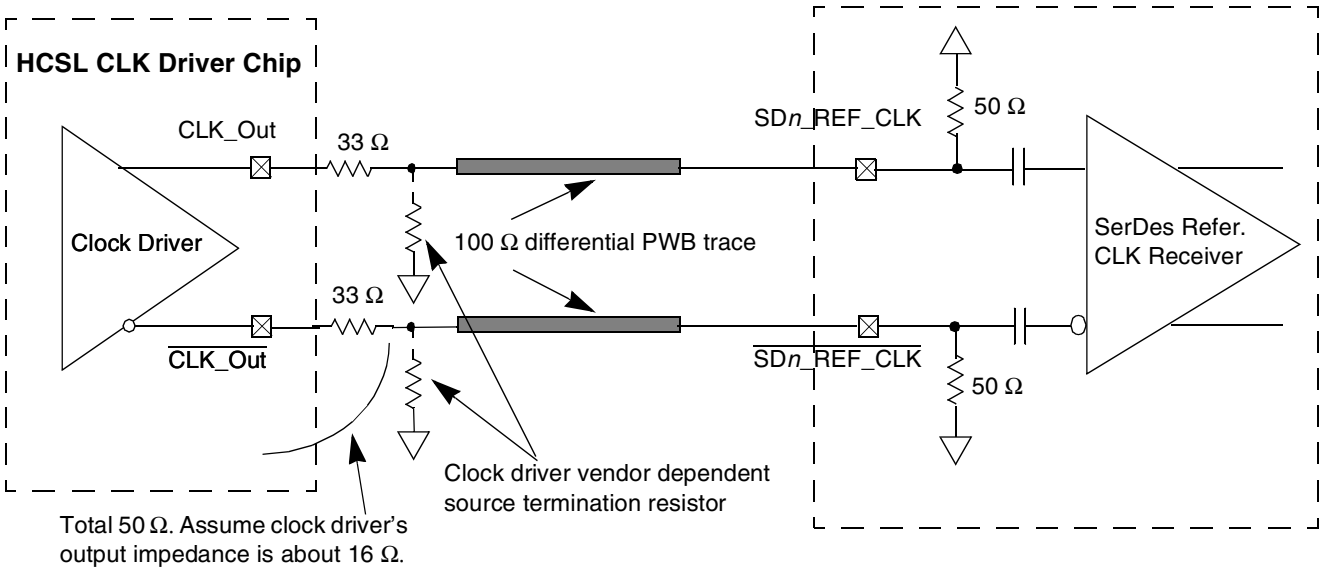


Figure 62. DC-Coupled Differential Connection with HCSL Clock Driver (Reference Only)

This figure shows the SerDes reference clock connection reference circuits for LVDS type clock driver. Since LVDS clock driver's common mode voltage is higher than the chip's SerDes reference clock input's allowed range (100 to 400mV), AC-coupled connection scheme must be used. It assumes the LVDS output driver features 50-Ω termination resistor. It also assumes that the LVDS transmitter establishes its own common mode level without relying on the receiver or other external component.

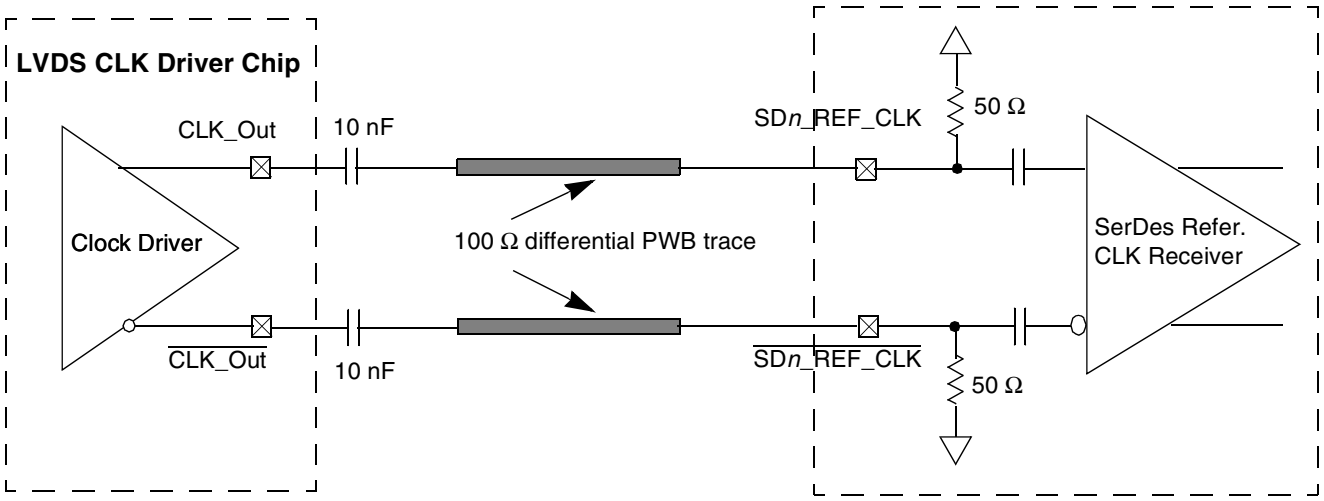


Figure 63. AC-Coupled Differential Connection with LVDS Clock Driver (Reference Only)

This figure shows the SerDes reference clock connection reference circuits for LVPECL type clock driver. Since LVPECL driver's DC levels (both common mode voltages and output swing) are incompatible with chip's SerDes reference clock input's DC requirement, AC-coupling has to be used. This figure assumes that the LVPECL clock driver's output impedance is 50Ω. R1 is used to DC-bias the LVPECL outputs prior to AC-coupling. Its value could be ranged from 140Ω to 240Ω depending on clock driver vendor's requirement. R2 is used together with the SerDes reference clock receiver's 50-Ω termination resistor to attenuate the LVPECL output's differential peak level such that it meets the chip's SerDes reference clock's differential input amplitude requirement (between 200mV and 800mV differential peak). For example, if the LVPECL output's differential peak is 900mV and the desired SerDes reference clock input amplitude is selected as 600mV, the attenuation factor is 0.67, which requires R2 = 25Ω. Please consult clock driver chip manufacturer to verify whether this connection scheme is compatible with a particular clock driver chip.

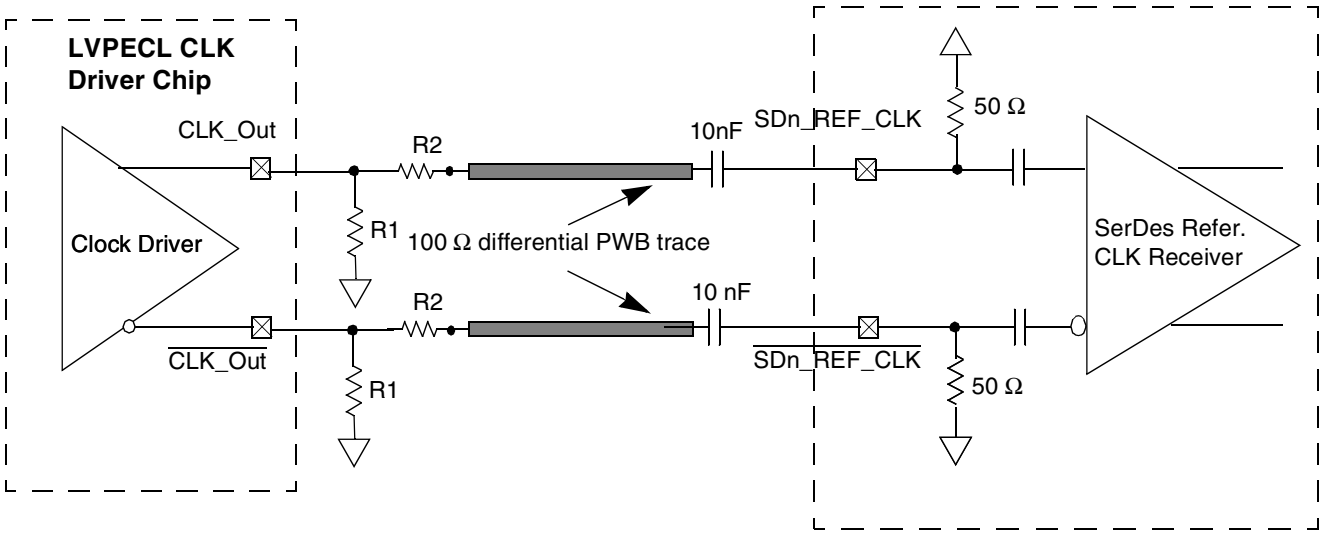


Figure 64. AC-Coupled Differential Connection with LVPECL Clock Driver (Reference Only)

This figure shows the SerDes reference clock connection reference circuits for a single-ended clock driver. It assumes the DC levels of the clock driver are compatible with chip's SerDes reference clock input's DC requirement.

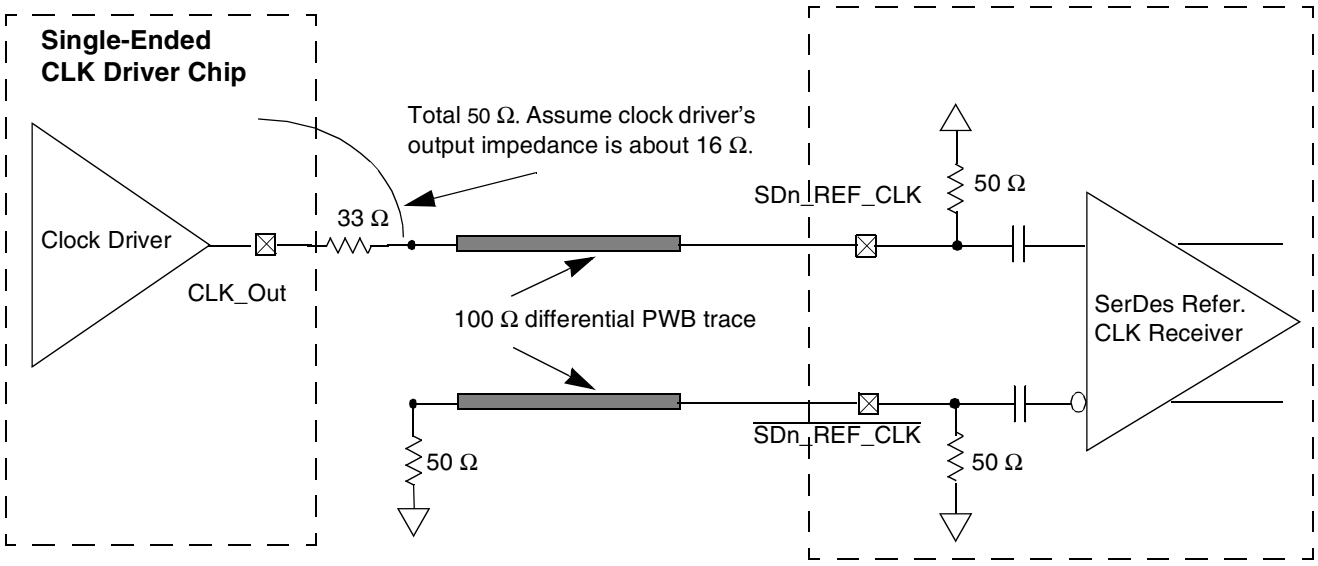


Figure 65. Single-Ended Connection (Reference Only)

Table 71. Differential Transmitter (TX) Output Specifications (continued)

Symbol	Parameter	Min	Nom	Max	Units	Comments
$T_{TX-IDLE-SET-TO-IDLE}$	Maximum time to transition to a valid electrical idle after sending an electrical Idle ordered set	—	—	20	UI	After sending an Electrical Idle ordered set, the Transmitter must meet all Electrical Idle Specifications within this time. This is considered a debounce time for the Transmitter to meet Electrical Idle after transitioning from L0.
$T_{TX-IDLE-TO-DIFF-DATA}$	Maximum time to transition to valid TX specifications after leaving an electrical idle condition	—	—	20	UI	Maximum time to meet all TX specifications when transitioning from Electrical Idle to sending differential data. This is considered a debounce time for the TX to meet all TX specifications after leaving Electrical Idle
$RL_{TX-DIFF}$	Differential Return Loss	12	—	—	dB	Measured over 50 MHz to 1.25 GHz. See Note 4
RL_{TX-CM}	Common Mode Return Loss	6	—	—	dB	Measured over 50 MHz to 1.25 GHz. See Note 4
$Z_{TX-DIFF-DC}$	DC Differential TX Impedance	80	100	120	Ω	TX DC Differential mode Low Impedance
Z_{TX-DC}	Transmitter DC Impedance	40	—	—	Ω	Required TX D+ as well as D- DC Impedance during all states
$L_{TX-SKEW}$	Lane-to-Lane Output Skew	—	—	500 + 2 UI	ps	Static skew between any two Transmitter Lanes within a single Link
C_{TX}	AC Coupling Capacitor	75	—	200	nF	All Transmitters shall be AC coupled. The AC coupling is required either within the media or within the transmitting component itself. See Note 8.
$T_{crosslink}$	Crosslink Random Timeout	0	—	1	ms	This random timeout helps resolve conflicts in crosslink configuration by eventually resulting in only one Downstream and one Upstream Port. See Note 7.

Notes:

1. No test load is necessarily associated with this value.
2. Specified at the measurement point into a timing and voltage compliance test load as shown in Figure 52 and measured over any 250 consecutive TX UIs. (Also refer to the transmitter compliance eye diagram shown in Figure 50)
3. A T_{TX-EYE} = 0.70 UI provides for a total sum of deterministic and random jitter budget of $T_{TX-JITTER-MAX}$ = 0.30 UI for the Transmitter collected over any 250 consecutive TX UIs. The $T_{TX-EYE-MEDIAN-to-MAX-JITTER}$ median is less than half of the total TX jitter budget collected over any 250 consecutive TX UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value.
4. The Transmitter input impedance shall result in a differential return loss greater than or equal to 12 dB and a common mode return loss greater than or equal to 6 dB over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements is 50 ohms to ground for both the D+ and D- line (that is, as measured by a Vector Network Analyzer with 50 ohm probes—see Figure 52). Note that the series capacitors C_{TX} is optional for the return loss measurement.
5. Measured between 20-80% at transmitter package pins into a test load as shown in Figure 52 for both V_{TX-D+} and V_{TX-D-}.
6. See Section 4.3.1.8 of the PCI Express Base Specifications Rev 1.0a
7. See Section 4.2.6.3 of the PCI Express Base Specifications Rev 1.0a
8. SerDes transmitter does not have C_{TX} built-in. An external AC Coupling capacitor is required.

2.23.1 Clock Ranges

This table provides the clocking specifications for the processor cores and Table 74 provides the clocking specifications for the memory bus.

Table 73. Processor Core Clocking Specifications

Characteristic	Maximum Processor Core Frequency								Unit	Notes
	600 MHz		800 MHz		1000 MHz		1250 MHz			
	Min	Max	Min	Max	Min	Max	Min	Max		
e500 core processor frequency	600	600	600	800	600	1000	600	1250	MHz	1, 2
CCB frequency	400	400	400	400	333	400	333	500		
DDR Data Rate	400	400	400	400	400	400	400	500		

Notes:

1. **Caution:** The CCB to SYSCLK ratio and e500 core to CCB ratio settings must be chosen such that the resulting SYSCLK frequency, e500 (core) frequency, and CCB frequency do not exceed their respective maximum or minimum operating frequencies. See Section 2.23.2, “CCB/SYSCLK PLL Ratio,” Section 2.23.3, “e500 Core PLL Ratio,“and Section 2.23.4, “DDR/DDRCLK PLL Ratio,” for ratio settings.
2. The processor core frequency speed bins listed also reflect the maximum platform (CCB) and DDR data rate frequency supported by production test. Running CCB and/or DDR data rate higher than the limit shown above, although logically possible via valid clock ratio setting in some condition, is not supported.

The DDR memory controller can run in either synchronous or asynchronous mode. When running in synchronous mode, the memory bus is clocked relative to the platform clock frequency. When running in asynchronous mode, the memory bus is clocked with its own dedicated PLL. This table provides the clocking specifications for the memory bus.

Table 74. Memory Bus Clocking Specifications

Characteristic	Maximum Processor Core Frequency		Unit	Notes
	600, 800, 1000, 1250			
	Min	Max		
DDR Memory bus clock speed	200	250	MHz	1, 2, 3, 4

Notes:

1. **Caution:** The CCB clock to SYSCLK ratio and e500 core to CCB clock ratio settings must be chosen such that the resulting SYSCLK frequency, e500 (core) frequency, and CCB clock frequency do not exceed their respective maximum or minimum operating frequencies. See Section 2.23.2, “CCB/SYSCLK PLL Ratio,” Section 2.23.3, “e500 Core PLL Ratio,” and Section 2.23.4, “DDR/DDRCLK PLL Ratio,” for ratio settings.
2. The Memory bus clock refers to the chip’s memory controllers’ MCK[0:5] and $\overline{\text{MCK}}$[0:5] output clocks, running at half of the DDR data rate.
3. In synchronous mode, the memory bus clock speed is half the platform clock frequency. In other words, the DDR data rate is the same as the platform (CCB) frequency. If the desired DDR data rate is higher than the platform (CCB) frequency, asynchronous mode must be used.
4. In asynchronous mode, the memory bus clock speed is dictated by its own PLL. See Section 2.23.4, “DDR/DDRCLK PLL Ratio.” The memory bus clock speed must be less than or equal to the CCB clock rate which in turn must be less than the DDR data rate.

Table 79. Package Thermal Characteristics (continued)

Characteristic	JEDEC Board	Symbol	Value	Unit	Notes
Junction-to-ambient (@200 ft/min)	Four layer board (2s2p)	$R_{\theta JA}$	14	°C/W	1, 2
Junction-to-board thermal	—	$R_{\theta JB}$	10	°C/W	3
Junction-to-case thermal	—	$R_{\theta JC}$	< 0.1	°C/W	4

Notes

1. Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.
2. Per JEDEC JESD51-2 and JESD51-6 with the board (JESD51-9) horizontal.
3. Thermal resistance between the die and the printed-circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
4. Thermal resistance between the active surface of the die and the case top surface determined by the cold plate method (MIL SPEC-883 Method 1012.1) with the calculated case temperature. Actual thermal resistance is less than 0.1 •C/W

Simulations with heat sinks were done with the package mounted on the 2s2p thermal test board. The thermal interface material was a typical thermal grease such as Dow Corning 340 or Wakefield 120 grease.For system thermal modeling, the chip's thermal model without a lid is shown in Figure 72 The substrate is modeled as a block 29 x 29 x 1.2 mm with an in-plane conductivity of 19.8 W/m•K and a through-plane conductivity of 1.13 W/m•K. The solder balls and air are modeled as a single block 29 x 29 x 0.5 mm with an in-plane conductivity of 0.034 W/m•K and a through plane conductivity of 12.1 W/m•K. The die is modeled as 9.6 x 9.57 mm with a thickness of 0.75 mm. The bump/underfill layer is modeled as a collapsed thermal resistance between the die and substrate assuming a conductivity of 7.5 W/m•K in the thickness dimension of 0.07 mm. The die is centered on the substrate. The thermal model uses approximate dimensions to reduce grid. Please refer to the case outline for actual dimensions.

2.24.2 Recommended Thermal Model

This table shows the chip's thermal model.

Table 80. Thermal Model

Conductivity	Value	Units
Die (9.6x9.6 × 0.85 mm)		
Silicon	Temperature dependent	—
Bump/Underfill (9.6 x 9.6 × 0.07 mm) Collapsed Thermal Resistance		
Kz	7.5	W/m•K
Substrate (29 × 29 × 1.2 mm)		
Kx	19.8	W/m•K
Ky	19.8	
Kz	1.13	
Solder and Air (29 × 29 × 0.5 mm)		
Kx	0.034	W/m•K
Ky	0.034	
Kz	12.1	

This figure shows the PLL power supply filter circuit.

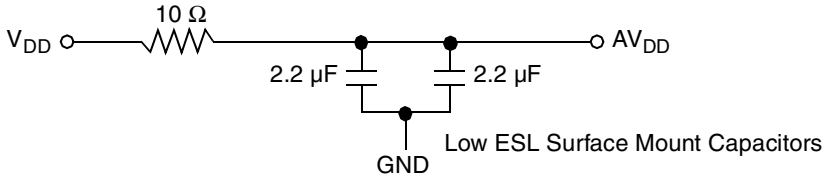
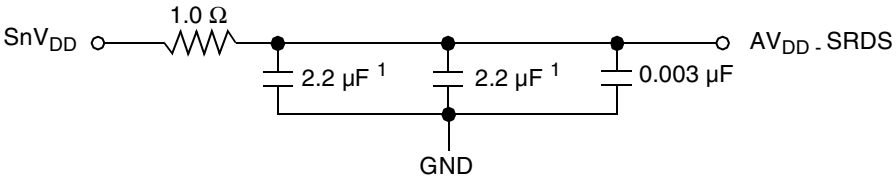


Figure 75. Chip PLL Power Supply Filter Circuit

The AV_{DD}_SRDS*n* signals provides power for the analog portions of the SerDes PLL. To ensure stability of the internal clock, the power supplied to the PLL is filtered using a circuit similar to the one shown in following Figure 76. For maximum effectiveness, the filter circuit is placed as closely as possible to the AV_{DD}_SRDS*n* balls to ensure it filters out as much noise as possible. The ground connection should be near the AV_{DD}_SRDS*n* balls. The 0.003-µF capacitor is closest to the balls, followed by the 1-µF capacitor, and finally the 1 ohm resistor to the board supply plane. The capacitors are connected from AV_{DD}_SRDS*n* to the ground plane. Use ceramic chip capacitors with the highest possible self-resonant frequency. All traces should be kept short, wide and direct.



1. An 0805 sized capacitor is recommended for system initial bring-up

Figure 76. SerDes PLL Power Supply Filter Circuit

Note the following:

- AV_{DD} should be a filtered version of SV_{DD}.
- Signals on the SerDes interface are fed from the XV_{DD} power plane.

3.3 Pin States in Deep Sleep State

In all low power mode by default, all input and output pads remain driven as per normal functional operation. The inputs remain enabled.

The exception is that in Deep Sleep mode, GCR[DEEPSLEEP_Z] can be used to tristate a subset of output pads, and disable the receivers of input pads as defined in Table 1. See the *MPC8536E PowerQUICC III Integrated Processor Reference Manual* for details.

3.4 Decoupling Recommendations

Due to large address and data buses, and high operating frequencies, this chip can generate transient power surges and high frequency noise in its power supply, especially while driving large capacitive loads. This noise must be prevented from reaching other components in the chip system, and the chip itself requires a clean, tightly regulated source of power. Therefore, it is recommended that the system designer place at least one decoupling capacitor at each V_{DD}, TV_{DD}, BV_{DD}, OV_{DD}, GV_{DD}, and LV_{DD} pin of the chip. These decoupling capacitors should receive their power from separate V_{DD},TV_{DD}, BV_{DD}, OV_{DD}, GV_{DD}, and LV_{DD}, and GND power planes in the PCB, utilizing short low impedance traces to minimize inductance. Capacitors must be placed directly under the chip using a standard escape pattern as much as possible. If some caps are to be placed surrounding the part it should be routed with short and large trace to minimize the inductance.

Careful board layout with stubless connections to these pull-down resistors coupled with the large value of the pull-down resistor should minimize the disruption of signal quality or speed for output pins thus configured.

The platform PLL ratio and e500 PLL ratio configuration pins are not equipped with these default pull-up devices.

3.10 JTAG Configuration Signals

Correct operation of the JTAG interface requires configuration of a group of system control pins as demonstrated in Figure 78. Care must be taken to ensure that these pins are maintained at a valid deasserted state under normal operating conditions as most have asynchronous behavior and spurious assertion will give unpredicatable results.

Boundary-scan testing is enabled through the JTAG interface signals. The $\overline{\text{TRST}}$ signal is optional in the IEEE 1149.1 specification, but it is provided on all processors built on Power Architecture technology. The chip requires $\overline{\text{TRST}}$ to be asserted during power-on reset flow to ensure that the JTAG boundary logic does not interfere with normal chip operation. While the TAP controller can be forced to the reset state using only the TCK and TMS signals, generally systems assert $\overline{\text{TRST}}$ during the power-on reset flow. Simply tying $\overline{\text{TRST}}$ to $\overline{\text{HRESET}}$ is not practical because the JTAG interface is also used for accessing the common on-chip processor (COP), which implements the debug interface to the chip.

The COP function of these processors allow a remote computer system (typically, a PC with dedicated hardware and debugging software) to access and control the internal operations of the processor. The COP interface connects primarily through the JTAG port of the processor, with some additional status monitoring signals. The COP port requires the ability to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$ in order to fully control the processor. If the target system has independent reset sources, such as voltage monitors, watchdog timers, power supply failures, or push-button switches, then the COP reset signals must be merged into these signals with logic.

The arrangement shown in Figure 78 allows the COP port to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$, while ensuring that the target can drive $\overline{\text{HRESET}}$ as well.

The COP interface has a standard header, shown in Figure 79, for connection to the target system, and is based on the 0.025" square-post, 0.100" centered header assembly (often called a Berg header). The connector typically has pin 14 removed as a connector key.

The COP header adds many benefits such as breakpoints, watchpoints, register and memory examination/modification, and other standard debugger features. An inexpensive option can be to leave the COP header unpopulated until needed.

There is no standardized way to number the COP header; consequently, many different pin numbers have been observed from emulator vendors. Some are numbered top-to-bottom then left-to-right, while others use left-to-right then top-to-bottom, while still others number the pins counter clockwise from pin 1 (as with an IC). Regardless of the numbering, the signal placement recommended in Figure 79 is common to all known emulators.

3.10.1 Termination of Unused Signals

If the JTAG interface and COP header will not be used, Freescale recommends the following connections:

- $\overline{\text{TRST}}$ should be tied to $\overline{\text{HRESET}}$ through a 0 kΩ isolation resistor so that it is asserted when the system reset signal ($\overline{\text{HRESET}}$) is asserted, ensuring that the JTAG scan chain is initialized during the power-on reset flow. Freescale recommends that the COP header be designed into the system as shown in Figure 78. If this is not possible, the isolation resistor will allow future access to $\overline{\text{TRST}}$ in case a JTAG interface may need to be wired onto the system in future debug situations.
- No pull-up/pull-down is required for TDI, TMS, or TDO.

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