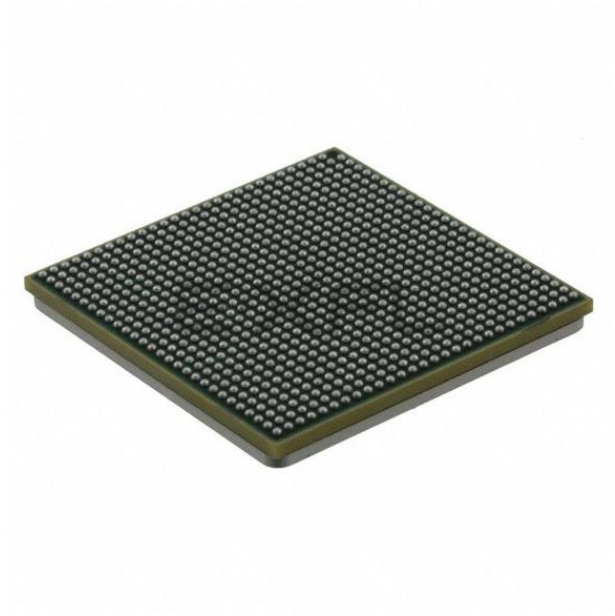

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC e500
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	800MHz
Co-Processors/DSP	-
RAM Controllers	DDR2, DDR3
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100/1000Mbps (2)
SATA	SATA 3Gbps (1)
USB	USB 2.0 (2)
Voltage - I/O	1.8V, 2.5V, 3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	-
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc8535cvtanga

This figure shows the major functional units within the chip.

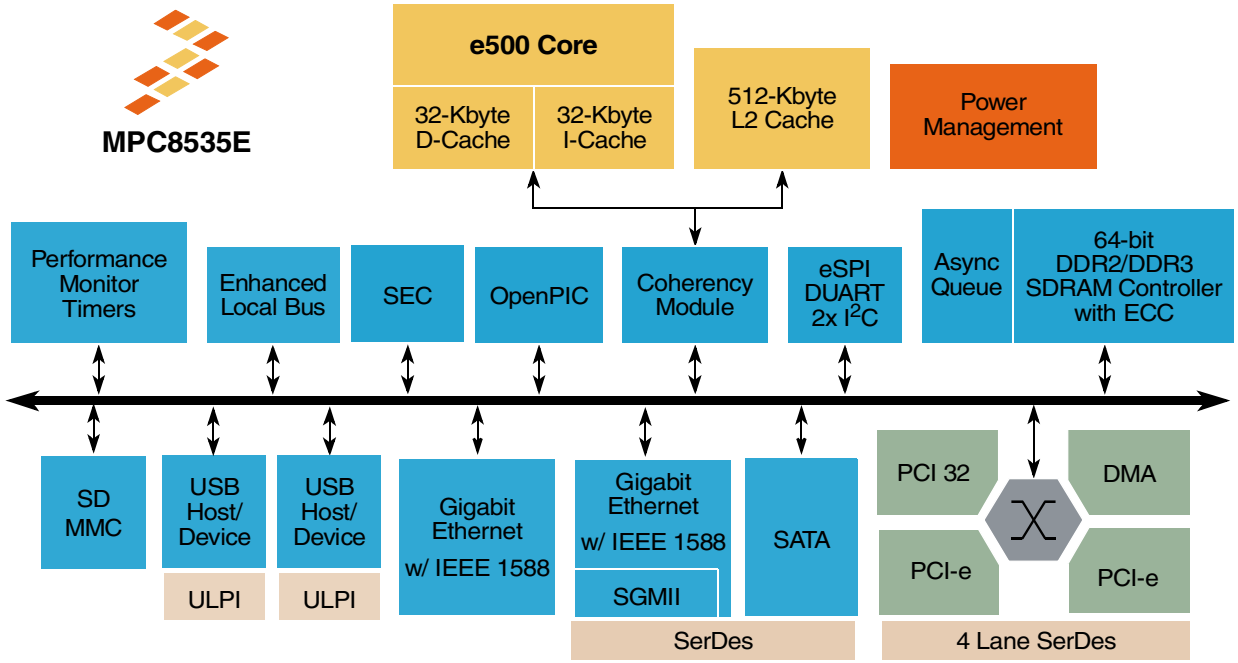


Figure 1. Chip Block Diagram

1 Pin Assignments and Reset States

NOTE

The naming convention of TSEC1 and TSEC3 is used to allow the splitting voltage rails for the eTSEC blocks and to ease the port of existing PowerQUICC III software

NOTE

The UART_SOUT[0:1] and $\overline{\text{TEST\_SEL}}$ pins must be set to a proper state during POR configuration. See Table 1 for more details.

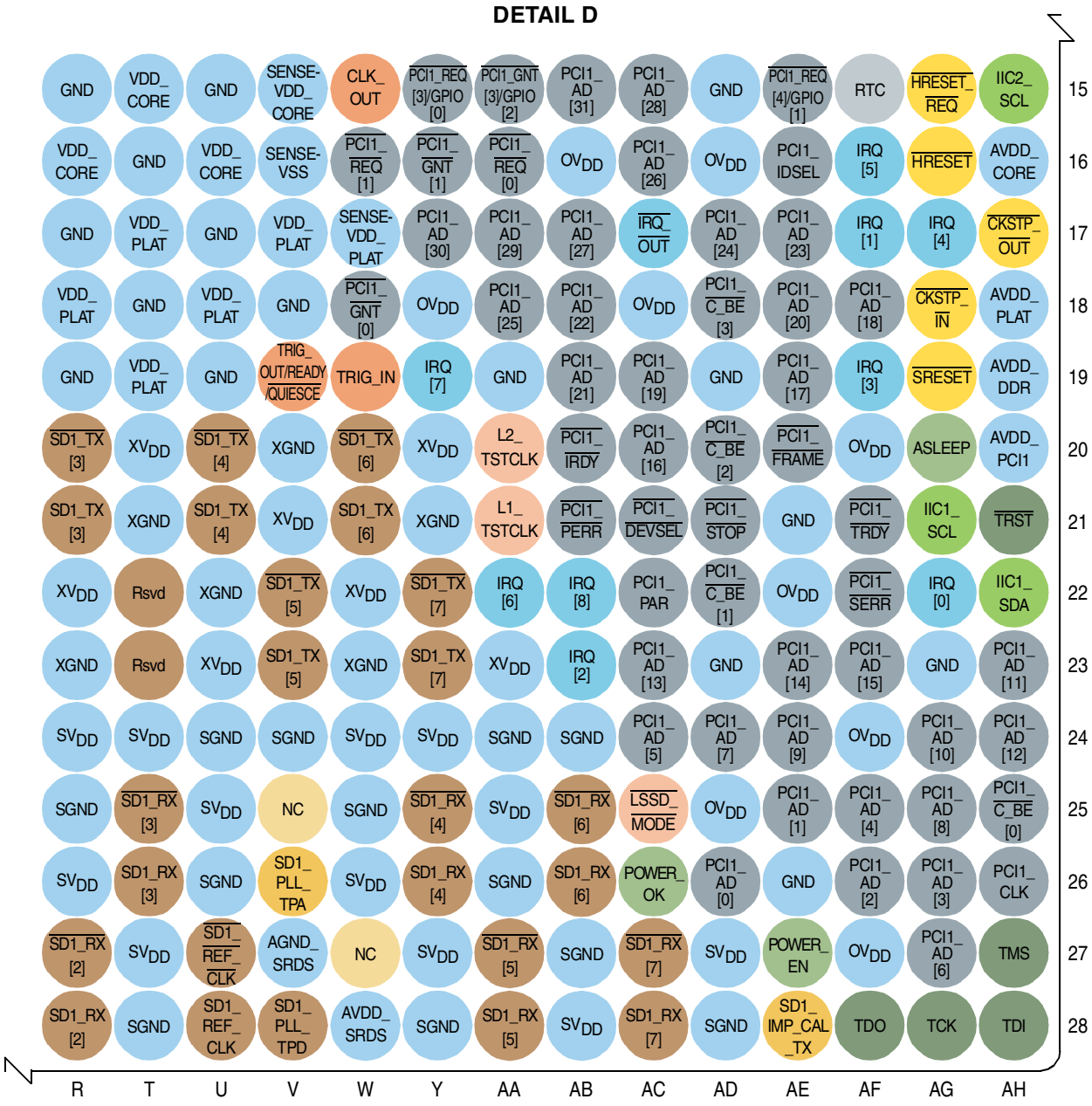


Figure 6. Chip Pin Map Detail D

Table 1. Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes
TSEC3_TX_CLK	Transmit clock In	U10	I	TV_{DD}	—
TSEC3_GTX_CLK	Transmit clock Out	U5	O	TV_{DD}	—
TSEC3_CRS	Carrier sense	T10	I/O	TV_{DD}	17
TSEC3_COL	Collision detect	T9	I	TV_{DD}	—
TSEC3_RXD[7:0]	Receive data	U12,U13,U6,V6,V1,U3, U2,V3	I	TV_{DD}	—
TSEC3_RX_DV	Receive data valid	V2	I	TV_{DD}	—
TSEC3_RX_ER	Receive data error	T4	I	TV_{DD}	—
TSEC3_RX_CLK	Receive clock	U1	I	TV_{DD}	—
IEEE 1588					
TSEC_1588_CLK	Clock In	W9	I	LV_{DD}	29
TSEC_1588_TRIG_IN[0:1]	Trigger In	W8,W7	I	LV_{DD}	29
TSEC_1588_TRIG_OUT[0:1]	Trigger Out	U11,W10	O	LV_{DD}	5,9,29
TSEC_1588_CLK_OUT	Clock Out	V10	O	LV_{DD}	5,9,29
TSEC_1588_PULSE_OUT1	Pulse Out1	V11	O	LV_{DD}	5,9,29
TSEC_1588_PULSE_OUT2	Pulse Out2	T11	O	LV_{DD}	5,9,29
eSDHC					
SDHC_CMD	Command line	AH10	I/O	OV_{DD}	29
$\overline{\text{SDHC\_CD}}$/GPIO[4]	Card detection	AH11	I	OV_{DD}	—
SDHC_DAT[0:3]	Data line	AG12,AH12,AH13, AG11	I/O	OV_{DD}	29
SDHC_DAT[4:7] / SPI_CS[0:3]	8-bit MMC Data line / SPI chip select	AE8,AC10,AF9,AA10	I/O	OV_{DD}	29
SDHC_CLK	SD/MMC/SDIO clock	AG13	I/O	OV_{DD}	29
$\overline{\text{SDHC\_WP}}$/GPIO[5]	Card write protection	AG10	I	OV_{DD}	1, 32
eSPI					
SPI_MOSI	Master Out Slave In	AF8	I/O	OV_{DD}	29
SPI_MISO	Master In Slave Out	AD9	I	OV_{DD}	29
SPI_CLK	eSPI clock	AD8	I/O	OV_{DD}	29
SPI_CS[0:3] / SDHC_DAT[4:7]	eSPI chip select / SDHC 8-bit MMC data	AE8,AC10,AF9,AA10	I/O	OV_{DD}	29
DUART					
$\overline{\text{UART\_CTS}}$[0:1]	Clear to send	AE11,Y12	I	OV_{DD}	29
$\overline{\text{UART\_RTS}}$[0:1]	Ready to send	AB12,AD12	O	OV_{DD}	29
UART_SIN[0:1]	Receive data	AC12,AF12	I	OV_{DD}	29

Table 1. Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes
TCK	Test clock	AG28	I	OV_{DD}	—
TDI	Test data in	AH28	I	OV_{DD}	12
TDO	Test data out	AF28	O	OV_{DD}	11
TMS	Test mode select	AH27	I	OV_{DD}	12
$\overline{\text{TRST}}$	Test reset	AH21	I	OV_{DD}	12
DFT					
L1_TSTCLK	L1 test clock	AA21	I	OV_{DD}	19
L2_TSTCLK	L2 test clock	AA20	I	OV_{DD}	19
$\overline{\text{LSSD\_MODE}}$	LSSD Mode	AC25	I	OV_{DD}	19
$\overline{\text{TEST\_SEL}}$	Test select	AA13	I	OV_{DD}	19
Power Management					
ASLEEP	Asleep	AG20	O	OV_{DD}	9,16,22
POWER_OK	Power OK	AC26	I	OV_{DD}	—
POWER_EN	Power enable	AE27	O	OV_{DD}	—
Power and Ground Signals					
OVDD	General I/O supply	Y18,AG2,AD4,AB16, AF6,AC18,AB13,AD10, AE14,AD16,AD25, AF27,AE22,AF11, AF20,AF24	—	OV_{DD}	—
LVDD	GMAC 1 I/O supply	AA7, AA4	Power for TSEC1 interfaces	LV_{DD}	—
TVDD	GMAC 3 I/O supply	V4,U7	Power for TSEC3 interfaces	TV_{DD}	—
GVDD	SSTL2 DDR supply	B1,B11,C7,C9,C14, C17,D4,D6,R3,D15,E2, E8,C24,E18,F5,E14, C21,G3,G7,G9,G11, H5,H12,E22,F15,J10, K3,K12,K14,H14,D20, E11,M1,N5	Power for DDR DRAM I/O	GV_{DD}	—
BVDD	Local bus I/O supply	L23,J18,J23,J19,F20, F23,H26,J21	Power for Local Bus	BV_{DD}	—
SVDD	SerDes 1 core logic supply	M27,N25,P28,R24, R26,T24,T27,U25, W24,W26,Y24,Y27, AA25,AB28,AD27	—	SV_{DD}	—

Table 1. Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes
XVDD	SerDes 1 transceiver supply	M21,N23,P20,R22,T20, U23,V21,W22,Y20, AA23	—	XV_{DD}	—
S2VDD	SerDes 2 core logic supply	R6,N7,M9	—	$S2V_{DD}$	—
X2VDD	SerDes 2 transceiver supply	R11,N12,L11	—	$X2V_{DD}$	—
VDD_CORE	Core, L2 logic supply	P13,U16,L16,M15,N14, R14,P15,N16,M13, U14,T13,L14,T15,R16, K13	—	V_{DD_CORE}	—
VDD_PLAT	Platform logic supply	T19,T17,V17,U18,R18, N18,M19,P19,P17,M17	—	V_{DD_PLAT}	—
AVDD_CORE	CPU PLL supply	AH16	—	AV_{DD_CORE}	20,28
AVDD_PLAT	Platform PLL supply	AH18	—	AV_{DD_PLAT}	20
AVDD_DDR	DDR PLL supply	AH19	—	AV_{DD_DDR}	20
AVDD_LBIU	Local Bus PLL supply	C28	—	AV_{DD_LBIU}	20
AVDD_PCI1	PCI PLL supply	AH20	—	AV_{DD_PCI1}	20
AVDD_SRDS	SerDes 1 PLL supply	W28	—	AV_{DD_SRDS}	20
AVDD_SRDS2	SerDes 2 PLL supply	T1	—	AV_{DD_SRDS2}	20
SENSEVDD_CORE	—	V15	—	V_{DD_CORE}	13
SENSEVDD_PLAT	—	W17	—	V_{DD_PLAT}	13
GND	Ground	D5,AE7,F4,D26,D23, C12,C15,E20,D8,B10, AF3,E3,J14,K21,F8,A3, F16,E12,E15,D17,L1, F21,H1,G13,G15,G18, C6,A14,A7,G25,H4, C20,J12,J15,J17,F27, M5,J27,K11,L26,K7, K8,T14,V14,M16,M18, P14,N15,N17,N19,N2, P5,P16,P18,M14,R15, R17,R19,T16,T18,L17, U15,U17,U19,V18,C27, Y13,AE26,AA19,AE21, B28,AC11,AD19,AD23, L15,AD15,AG23,AE9, A27,V7,Y7,AC5,U4,Y4, AE12,AB9,AA14,N13, R13,L13	—	—	—
XGND	SerDes 1Transceiver pad GND (xpadvss)	M20,M24,N22,P21, R23,T21,U22,V20, W23, Y21	—	—	—

Table 1. Pinout Listing (continued)

Signal	Signal Name	Package Pin Number	Pin Type	Power Supply	Notes

Notes:

1. All multiplexed signals may be listed only once and may not re-occur.
2. Recommend a weak pull-up resistor (2–10 KΩ) be placed on this pin to OV_{DD}.
3. This pin must always be pulled-high.
4. This pin is an open drain signal.
5. This pin is a reset configuration pin. It has a weak internal pull-up P-FET which is enabled only when the processor is in the reset state. This pull-up is designed such that it can be overpowered by an external 4.7-kΩ pull-down resistor. However, if the signal is intended to be high after reset, and if there is any device on the net which might pull down the value of the net at reset, then a pullup or active driver is needed.
6. Treat these pins as no connects (NC) unless using debug address functionality.
7. The value of LA[28:31] during reset sets the CCB clock to SYSCLK PLL ratio. These pins require 4.7-kΩ pull-up or pull-down resistors. See Section 22.2, "CCB/SYSCLK PLL Ratio."
8. The value of LALE, LGPL2 and LBCTL at reset set the e500 core clock to CCB Clock PLL ratio. These pins require 4.7-kΩ pull-up or pull-down resistors. See the Section 22.3, "e500 Core PLL Ratio."
9. Functionally, this pin is an output, but structurally it is an I/O because it either samples configuration input during reset or because it has other manufacturing test functions. This pin will therefore be described as an I/O for boundary scan.
10. For proper state of these signals during reset, UART_SOUT[1] must be pulled down to GND through a resistor. UART_SOUT[0] can be pulled up or left without a resistor. However, if there is any device on the net which might pull down the value of the net at reset, then a pullup is needed on UART_SOUT[0].
11. This output is actively driven during reset rather than being three-stated during reset.
12. These JTAG pins have weak internal pull-up P-FETs that are always enabled.
13. These pins are connected to the V_{DD_CORE}/V_{DD_PLAT}/GND planes internally and may be used by the core power supply to improve tracking and regulation.
15. These pins have other manufacturing or debug test functions. It's recommended to add both pull-up resistor pads to OVDD and pull-down resistor pads to GND on board to support future debug testing when needed.
16. If this pin is connected to a device that pulls down during reset, an external pull-up is required to drive this pin to a safe state during reset.
17. This pin is only an output in FIFO mode when used as Rx Flow Control.
18. Do not connect.
19. These must be pulled up (100 Ω- 1 kΩ) to OVDD.
20. Independent supplies derived from board VDD.
21. Recommend a pull-up resistor (1 KΩ) be placed on this pin to OV_{DD}.
22. The following pins must NOT be pulled down during power-on reset: MDVAL, UART_SOUT[0], EC_MDC, TSEC1_TXD[3], TSEC3_TXD[7], HRESET_REQ, TRIG_OUT/READY/$\overline{\text{QUIESCE}}$, MSRCID[2:4], ASLEEP.
23. This pin requires an external 4.7-kΩ pull-down resistor to prevent PHY from seeing a valid Transmit Enable before it is actively driven.
24. General-Purpose POR configuration of user system.

This figure shows the DDR SDRAM output timing for the MCK to MDQS skew measurement (t_{DDKHMH}).

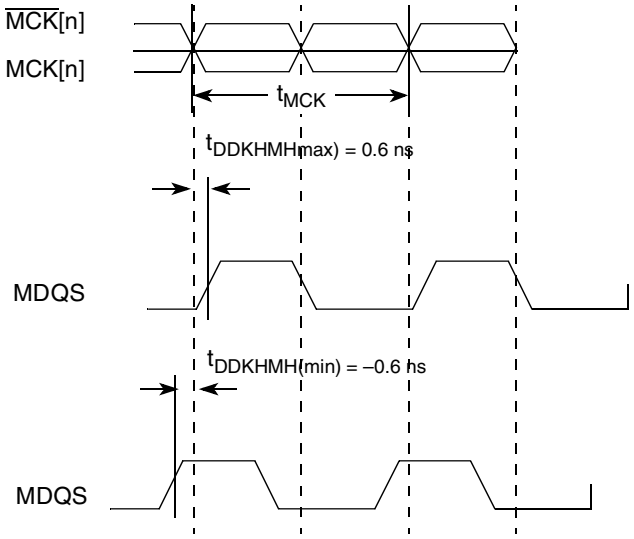


Figure 9. Timing Diagram for tDDKHMH

This figure shows the DDR SDRAM output timing diagram.

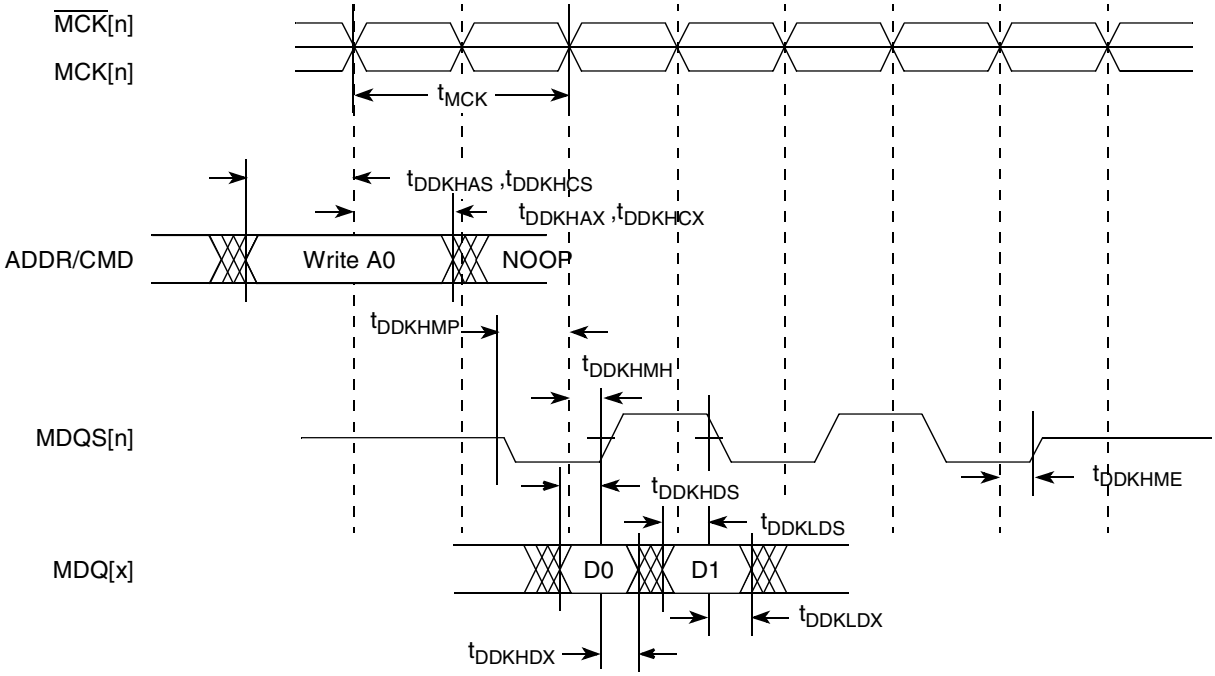


Figure 10. DDR SDRAM Output Timing Diagram

This figure shows the GMII receive AC timing diagram.

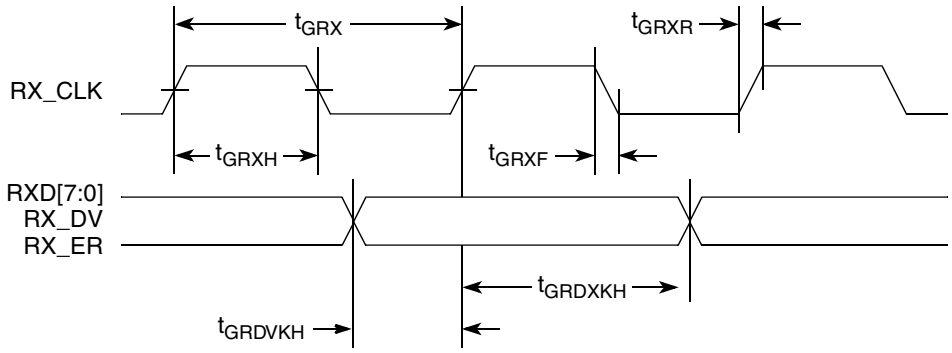


Figure 18. GMII Receive AC Timing Diagram

2.9.2.3 MII AC Timing Specifications

This section describes the MII transmit and receive AC timing specifications.

2.9.2.3.1 MII Transmit AC Timing Specifications

This table provides the MII transmit AC timing specifications.

Table 30. MII Transmit AC Timing Specifications

At recommended operating conditions with L/TV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
TX_CLK clock period 10 Mbps	t_{MTX}	—	400	—	ns
TX_CLK clock period 100 Mbps	t_{MTX}	—	40	—	ns
TX_CLK duty cycle	t_{MTXH}/t_{MTX}	35	—	65	%
TX_CLK to MII data TXD[3:0], TX_ER, TX_EN delay	t_{MTKHDX}	1	5	15	ns
TX_CLK data clock rise (20%-80%)	t_{MTXR}	1.0	—	4.0	ns
TX_CLK data clock fall (80%-20%)	t_{MTXF}	1.0	—	4.0	ns

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MTKHDX} symbolizes MII transmit timing (MT) for the time t_{MTX} clock reference (K) going high (H) until data outputs (D) are invalid (X). Note that, in general, the clock reference symbol representation is based on two to three letters representing the clock of a particular functional. For example, the subscript of t_{MTX} represents the MII(M) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

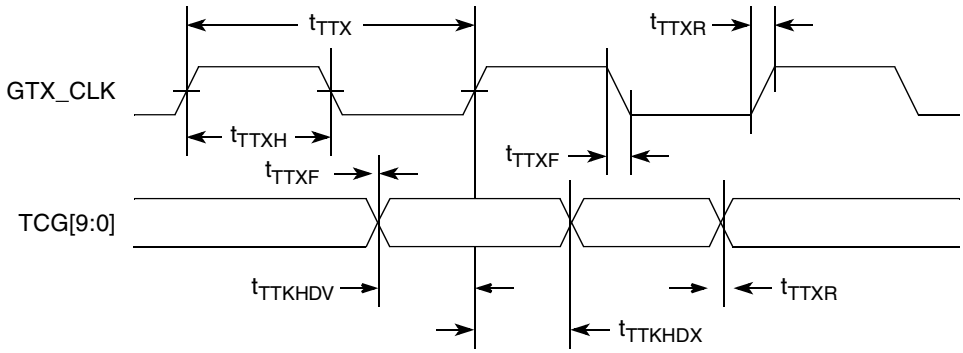


Figure 22. TBI Transmit AC Timing Diagram

2.9.2.4.2 TBI Receive AC Timing Specifications

This table provides the TBI receive AC timing specifications.

Table 33. TBI Receive AC Timing Specifications

At recommended operating conditions with L/TV_{DD} of 3.3 V ± 5%.

Parameter/Condition[2]	Symbol [1]	Min	Typ	Max	Unit
Clock period for TBI Receive Clock 0, 1	t_{TRX}	—	16.0	—	ns
Skew for TBI Receive Clock 0, 1	t_{SKTRX}	7.5	—	8.5	ns
Duty cycle for TBI Receive Clock 0, 1	t_{TRXH}/t_{TRX}	40	—	60	%
RCG[9:0] setup time to rising edge of TBI Receive Clock 0, 1	t_{TRDVKH}	2.5	—	—	ns
RCG[9:0] hold time to rising edge of TBI Receive Clock 0, 1	t_{TRDXKH}	1.5	—	—	ns
Clock rise time (20%-80%) for TBI Receive Clock 0, 1	t_{TRXR}	0.7	—	2.4	ns
Clock fall time (80%-20%) for TBI Receive Clock 0, 1	t_{TRXF}	0.7	—	2.4	ns

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{TRDVKH} symbolizes TBI receive timing (TR) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{TRX} clock reference (K) going to the high (H) state or setup time. Also, t_{TRDXKH} symbolizes TBI receive timing (TR) with respect to the time data input signals (D) went invalid (X) relative to the t_{TRX} clock reference (K) going to the high (H) state. Note that, in general, the clock reference symbol representation is based on three letters representing the clock of a particular functional. For example, the subscript of t_{TRX} represents the TBI (T) receive (RX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall). For symbols representing skews, the subscript is skew (SK) followed by the clock that is being skewed (TRX).
2. The signals “TBI Receive Clock 0” and “TBI Receive Clock 1” refer to TSECn_RX_CLK and TSECn_TX_CLK pins respectively. These two clock signals are also referred as PMA_RX_CLK[0:1].

Table 36. RMII Transmit AC Timing Specifications (continued)

At recommended operating conditions with L/TV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
Rise time TSECn_TX_CLK (20%–80%)	t_{RMTR}	1.0	—	2.0	ns
Fall time TSECn_TX_CLK (80%–20%)	t_{RMTF}	1.0	—	2.0	ns
TSECn_TX_CLK to RMII data TXD[1:0], TX_EN delay	t_{RMTDX}	2.0	—	10.0	ns

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state)(reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{MTKHDX} symbolizes MII transmit timing (MT) for the time t_{MTX} clock reference (K) going high (H) until data outputs (D) are invalid (X). Note that, in general, the clock reference symbol representation is based on two to three letters representing the clock of a particular functional. For example, the subscript of t_{MTX} represents the MII(M) transmit (TX) clock. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).

This figure shows the RMII transmit AC timing diagram.

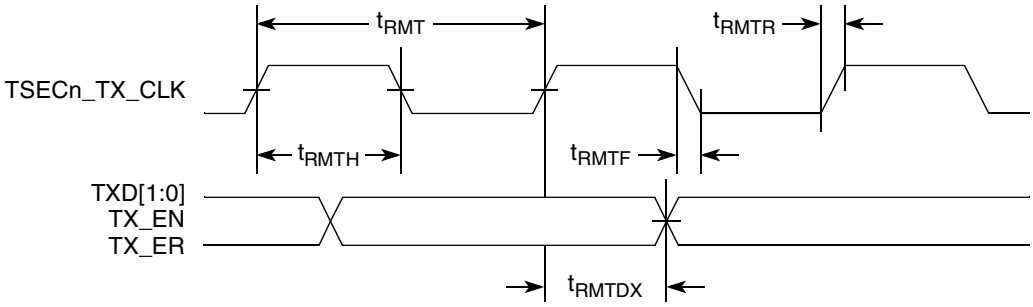


Figure 26. RMII Transmit AC Timing Diagram

2.9.2.7.2 RMII Receive AC Timing Specifications

Table 37. RMII Receive AC Timing Specifications

At recommended operating conditions with L/TV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol [1]	Min	Typ	Max	Unit
TSECn_RX_CLK clock period	t_{RMR}	15.0	20.0	25.0	ns
TSECn_RX_CLK duty cycle	t_{RMRH}	35	50	65	%
TSECn_RX_CLK peak-to-peak jitter	t_{RMRJ}	—	—	250	ps
Rise time TSECn_RX_CLK (20%–80%)	t_{RMRR}	1.0	—	2.0	ns

The IEEE 1588 AC timing specifications are in the following table.

Table 43. eTSEC IEEE 1588 AC Timing Specifications

At recommended operating conditions with L/TV_{DD} of 3.3 V ± 5%.

Parameter/Condition	Symbol	Min	Typ	Max	Unit	Note
TSEC_1588_CLK clock period	$t_{T1588CLK}$	3.8	—	T_{TX_CLK}*7	ns	1
TSEC_1588_CLK duty cycle	$t_{T1588CLKH}$ /$t_{T1588CLK}$	40	50	60	%	—
TSEC_1588_CLK peak-to-peak jitter	$t_{T1588CLKINJ}$	—	—	250	ps	—
Rise time eTSEC_1588_CLK (20%–80%)	$t_{T1588CLKINR}$	1.0	—	2.0	ns	—
Fall time eTSEC_1588_CLK (80%–20%)	$t_{T1588CLKINF}$	1.0	—	2.0	ns	—
TSEC_1588_CLK_OUT clock period	$t_{T1588CLKOUT}$	2*$t_{T1588CLK}$	—	—	ns	—
TSEC_1588_CLK_OUT duty cycle	$t_{T1588CLKOTH}$ /$t_{T1588CLKOUT}$	30	50	70	%	—
TSEC_1588_PULSE_OUT	$t_{T1588OV}$	0.5	—	3.0	ns	—
TSEC_1588_TRIG_IN pulse width	$t_{T1588TRIGH}$	2*$t_{T1588CLK\_MAX}$	—	—	ns	2

Note:

1. When TMR_CTRL[CKSEL]=00, the external TSEC_1588_CLK input is selected as the 1588 timer reference clock source, with the timing defined in the Table above. The maximum value of $t_{T1588CLK}$ is defined in terms of T_{TX_CLK}, which is the maximum clock cycle period of the equivalent interface speed that the eTSEC1 port is running.
   When eTSEC1 is configured to operate in the parallel mode, the T_{TX_CLK} is the maximum clock period of the TSEC1_TX_CLK. When eTSEC1 operates in SGMII mode, the maximum value of $t_{T1588CLK}$ is defined in terms of the recovered clock from SGMII SerDes. For example, for 10/100/1000 Mbps modes, the maximum value of $t_{T1588CLK}$ will be 2800, 280, and 56 ns respectively.
   See the *MPC8536E PowerQUICC III Integrated Communications Processor Reference Manual* for a description of TMR_CTRL registers.
2. It need to be at least two times of clock period of clock selected by TMR_CTRL[CKSEL]. See the *MPC8536E PowerQUICC III Integrated Processor Reference Manual* for a description of TMR_CTRL registers.

2.10 Ethernet Management Interface Electrical Characteristics

The electrical characteristics specified here apply to MII management interface signals EC_MDIO (management data input/output) and EC_MDC (management data clock). The electrical characteristics for GMII, SGMII, RGMII, RMII, TBI and RTBI are specified in Section 2.9, “Ethernet: Enhanced Three-Speed Ethernet (eTSEC), MII Management”

2.18 GPIO

This section describes the DC and AC electrical specifications for the GPIO interface of the chip.

2.18.1 GPIO DC Electrical Characteristics

This table provides the DC electrical characteristics for the GPIO interface.

Table 65. GPIO DC Electrical Characteristics

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	– 0.3	0.8	V
Input current (V_{IN} [1] = 0 V or V_{IN} = V_{DD})	I_{IN}	—	±5	μA
High-level output voltage (OV_{DD} = min, I_{OH} = –2 mA)	V_{OH}	2.4	—	V
Low-level output voltage (OV_{DD} = min, I_{OL} = 2 mA)	V_{OL}	—	0.4	V

Note:

1. The symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

2.18.2 GPIO AC Electrical Specifications

This table provides the GPIO input and output AC timing specifications.

Table 66. GPIO Input and Output AC Timing Specifications[1]

Characteristic	Symbol [2]	Min	Unit	Notes
GPIO inputs—minimum pulse width	t_{PIWID}	7.5	ns	3
GPIO outputs—minimum pulse width	t_{GTOWID}	12	ns	—

Notes:

1. Input specifications are measured from the 50% level of the signal to the 50% level of the rising edge of CLKIN. Timings are measured at the pin.
2. GPIO inputs and outputs are asynchronous to any visible clock. GPIO outputs should be synchronized before use by any external synchronous logic. GPIO inputs are required to be valid for at least t_{PIWID} ns to ensure proper operation.
3. The minimum pulse width is a function of the MPX/Platform clock. The minimum pulse width must be greater than or equal to 4 times the MPX/Platform clock period.

This figure provides the AC test load for the GPIO.

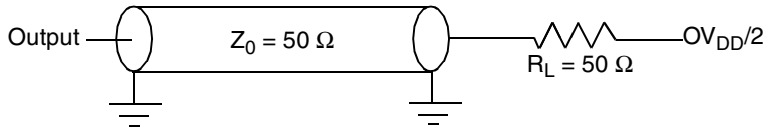


Figure 53. GPIO AC Test Load

2.19 PCI

This section describes the DC and AC electrical specifications for the PCI bus of the chip.

2.19.1 PCI DC Electrical Characteristics

This table provides the DC electrical characteristics for the PCI interface.

Table 67. PCI DC Electrical Characteristics [1]

Parameter	Symbol	Min	Max	Unit
High-level input voltage	V_{IH}	2	OV_{DD} + 0.3	V
Low-level input voltage	V_{IL}	–0.3	0.8	V
Input current (V_{IN} [2] = 0 V or V_{IN} = V_{DD})	I_{IN}	—	±5	μA
High-level output voltage (OV_{DD} = min, I_{OH} = –2 mA)	V_{OH}	2.4	—	V
Low-level output voltage (OV_{DD} = min, I_{OL} = 2 mA)	V_{OL}	—	0.4	V

Notes:

1. Ranges listed do not meet the full range of the DC specifications of the *PCI 2.2 Local Bus Specifications*.
2. The symbol V_{IN}, in this case, represents the OV_{IN} symbol referenced in Table 1 and Table 2.

2.19.2 PCI AC Electrical Specifications

This section describes the general AC timing parameters of the PCI bus. Note that the SYSCLK signal is used as the PCI input clock. This table provides the PCI AC timing specifications at 66 MHz.

Table 68. PCI AC Timing Specifications at 66 MHz

Parameter	Symbol [1]	Min	Max	Unit	Notes
SYSCLK to output valid	t_{PCKHOV}	—	6.0	ns	2, 3
Output hold from SYSCLK	t_{PCKHOX}	2.0	—	ns	2
SYSCLK to output high impedance	t_{PCKHOZ}	—	14	ns	2, 4
Input setup to SYSCLK	t_{PCIVKH}	3.0	—	ns	2, 5
Input hold from SYSCLK	t_{PCIXKH}	0	—	ns	2, 5
$\overline{\text{REQ64}}$ to $\overline{\text{HRESET}}$ [9] setup time	t_{PCRVRH}	$10 \times t_{SYS}$	—	clocks	6, 7
$\overline{\text{HRESET}}$ to $\overline{\text{REQ64}}$ hold time	t_{PCRHRX}	0	50	ns	7

This figure shows the SerDes reference clock connection reference circuits for LVPECL type clock driver. Since LVPECL driver's DC levels (both common mode voltages and output swing) are incompatible with chip's SerDes reference clock input's DC requirement, AC-coupling has to be used. This figure assumes that the LVPECL clock driver's output impedance is 50Ω. R1 is used to DC-bias the LVPECL outputs prior to AC-coupling. Its value could be ranged from 140Ω to 240Ω depending on clock driver vendor's requirement. R2 is used together with the SerDes reference clock receiver's 50-Ω termination resistor to attenuate the LVPECL output's differential peak level such that it meets the chip's SerDes reference clock's differential input amplitude requirement (between 200mV and 800mV differential peak). For example, if the LVPECL output's differential peak is 900mV and the desired SerDes reference clock input amplitude is selected as 600mV, the attenuation factor is 0.67, which requires R2 = 25Ω. Please consult clock driver chip manufacturer to verify whether this connection scheme is compatible with a particular clock driver chip.

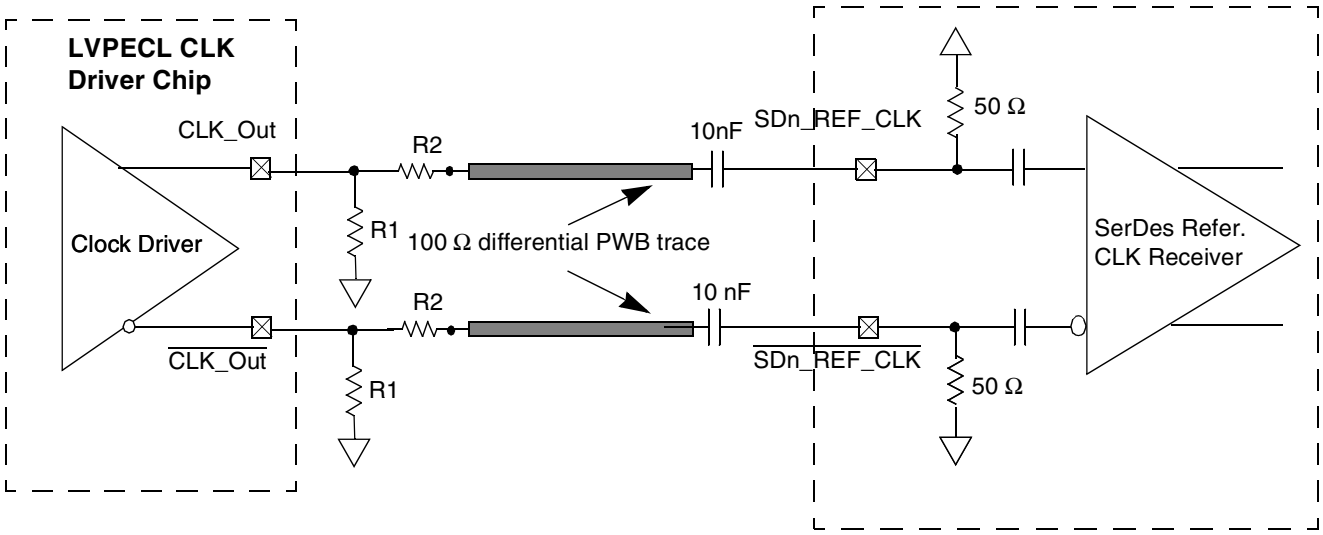


Figure 64. AC-Coupled Differential Connection with LVPECL Clock Driver (Reference Only)

This figure shows the SerDes reference clock connection reference circuits for a single-ended clock driver. It assumes the DC levels of the clock driver are compatible with chip's SerDes reference clock input's DC requirement.

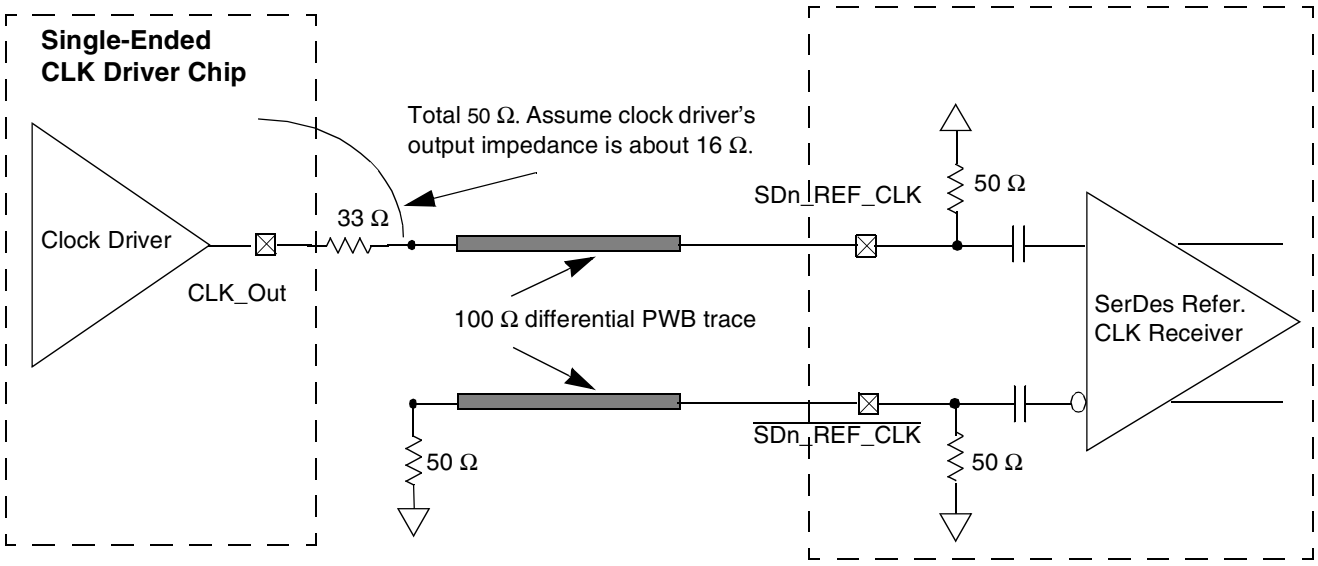


Figure 65. Single-Ended Connection (Reference Only)

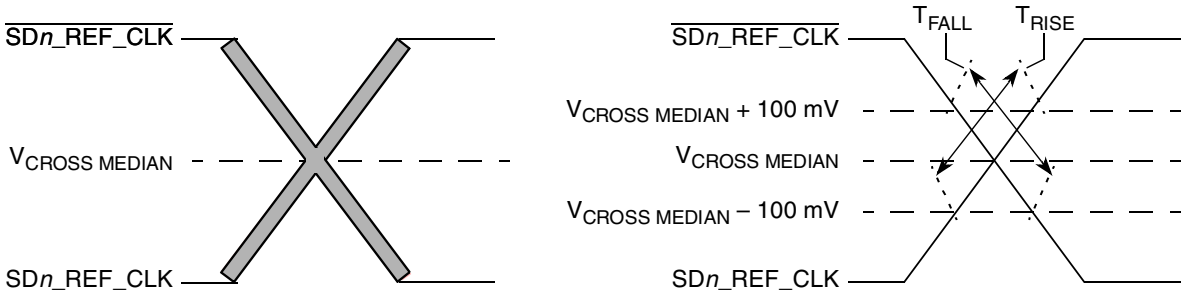


Figure 67. Single-Ended Measurement Points for Rise and Fall Time Matching

The other detailed AC requirements of the SerDes Reference Clocks is defined by each interface protocol based on application usage. See the following sections for detailed information:

- Section 2.9.3.2, "AC Requirements for SGMII SD2_REF_CLK and SD2_REF_CLK"
- Section 2.21.2, "AC Requirements for PCI Express SerDes Clocks"

2.20.2.4.1 Spread Spectrum Clock

SD1_REF_CLK/$\overline{\text{SD1\_REF\_CLK}}$ were designed to work with a spread spectrum clock (+0 to -0.5% spreading at 30–33 kHz rate is allowed), assuming both ends have same reference clock. For better results, a source without significant unintended modulation should be used.

SD2_REF_CLK/$\overline{\text{SD2\_REF\_CLK}}$ are not intended to be used with, and should not be clocked by, a spread spectrum clock source.

2.20.3 SerDes Transmitter and Receiver Reference Circuits

This figure shows the reference circuits for SerDes data lane's transmitter and receiver.

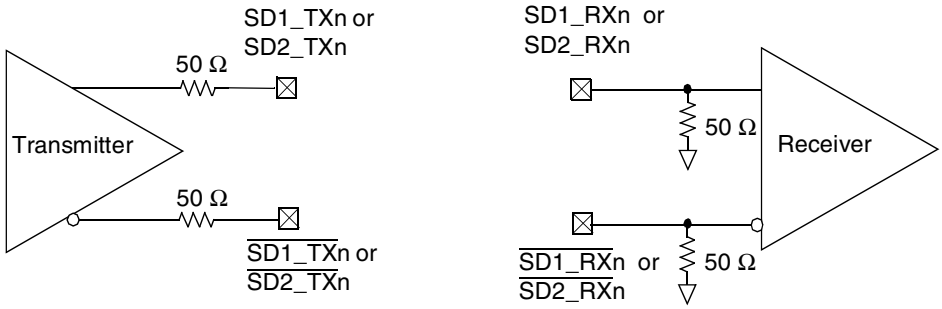


Figure 68. SerDes Transmitter and Receiver Reference Circuits

The DC and AC specification of SerDes data lanes are defined in each interface protocol section below (PCI Express, SATA or SGMII) in this document based on the application usage:

- Section 2.9.3, "SGMII Interface Electrical Characteristics"
- Section 2.21, "PCI Express"
- Section 2.16, "Serial ATA (SATA)"

Please note that external AC Coupling capacitor is required for the above three serial transmission protocols with the capacitor value defined in specification of each protocol section.

2.21.4.3 Differential Receiver (RX) Input Specifications

This table defines the specifications for the differential input at all receivers (RXs). The parameters are specified at the component pins.

Table 72. Differential Receiver (RX) Input Specifications

Symbol	Parameter	Min	Nom	Max	Units	Comments
UI	Unit Interval	399.88	400	400.12	ps	Each UI is 400 ps ± 300 ppm. UI does not account for Spread Spectrum Clock dictated variations. See Note 1.
$V_{RX-DIFFp-p}$	Differential Peak-to-Peak Output Voltage	0.175	—	1.200	V	$V_{RX-DIFFp-p} = 2*\|V_{RX-D+} - V_{RX-D-}\|$ See Note 2.
T_{RX-EYE}	Minimum Receiver Eye Width	0.4	—	—	UI	The maximum interconnect media and Transmitter jitter that can be tolerated by the Receiver can be derived as $T_{RX-MAX-JITTER} = 1 - T_{RX-EYE} = 0.6$ UI. See Notes 2 and 3.
$T_{RX-EYE-MEDIAN-to-MAX-JITTER}$	Maximum time between the jitter median and maximum deviation from the median.	—	—	0.3	UI	Jitter is defined as the measurement variation of the crossing points ($V_{RX-DIFFp-p}$ = 0 V) in relation to a recovered TX UI. A recovered TX UI is calculated over 3500 consecutive unit intervals of sample data. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the TX UI. See Notes 2, 3 and 7.
$V_{RX-CM-ACp}$	AC Peak Common Mode Input Voltage	—	—	150	mV	$V_{RX-CM-ACp} = \|V_{RXD+} - V_{RXD-}\|/2 + V_{RX-CM-DC}$ $V_{RX-CM-DC} = DC_{(avg)}$ of $\|V_{RX-D+} + V_{RX-D-}\|/2$ See Note 2
$RL_{RX-DIFF}$	Differential Return Loss	15	—	—	dB	Measured over 50 MHz to 1.25 GHz with the D+ and D- lines biased at +300 mV and –300 mV, respectively. See Note 4
RL_{RX-CM}	Common Mode Return Loss	6	—	—	dB	Measured over 50 MHz to 1.25 GHz with the D+ and D- lines biased at 0 V. See Note 4
$Z_{RX-DIFF-DC}$	DC Differential Input Impedance	80	100	120	Ω	RX DC Differential mode impedance. See Note 5
Z_{RX-DC}	DC Input Impedance	40	50	60	Ω	Required RX D+ as well as D- DC Impedance (50 ± 20% tolerance). See Notes 2 and 5.
$Z_{RX-HIGH-IMP-DC}$	Powered Down DC Input Impedance	200 k	—	—	Ω	Required RX D+ as well as D– DC Impedance when the Receiver terminations do not have power. See Note 6.
$V_{RX-IDLE-DET-DIFFp-p}$	Electrical Idle Detect Threshold	65	—	175	mV	$V_{RX-IDLE-DET-DIFFp-p} = 2*\|V_{RX-D+} - V_{RX-D-}\|$ Measured at the package pins of the Receiver
$T_{RX-IDLE-DET-DIFF-ENTERTIME}$	Unexpected Electrical Idle Enter Detect Threshold Integration Time	—	—	10	ms	An unexpected Electrical Idle ($V_{RX-DIFFp-p} < V_{RX-IDLE-DET-DIFFp-p}$) must be recognized no longer than $T_{RX-IDLE-DET-DIFF-ENTERING}$ to signal an unexpected idle condition.

Table 72. Differential Receiver (RX) Input Specifications (continued)

Symbol	Parameter	Min	Nom	Max	Units	Comments
$L_{TX-SKEW}$	Total Skew	—	—	20	ns	Skew across all lanes on a Link. This includes variation in the length of SKP ordered set (for example, COM and one to five Symbols) at the RX as well as any delay differences arising from the interconnect itself.

Notes:
1. No test load is necessarily associated with this value.
2. Specified at the measurement point and measured over any 250 consecutive UIs. The test load in Figure 71 should be used as the RX device when taking measurements (also refer to the Receiver compliance eye diagram shown in Figure 70). If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as a reference for the eye diagram.
3. A T_{RX-EYE} = 0.40 UI provides for a total sum of 0.60 UI deterministic and random jitter budget for the Transmitter and interconnect collected any 250 consecutive UIs. The $T_{RX-EYE-MEDIAN-to-MAX-JITTER}$ specification ensures a jitter distribution in which the median and the maximum deviation from the median is less than half of the total. UI jitter budget collected over any 250 consecutive TX UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. If the clocks to the RX and TX are not derived from the same reference clock, the TX UI recovered from 3500 consecutive UI must be used as the reference for the eye diagram.
4. The Receiver input impedance shall result in a differential return loss greater than or equal to 15 dB with the D+ line biased to 300 mV and the D- line biased to -300 mV and a common mode return loss greater than or equal to 6 dB (no bias required) over a frequency range of 50 MHz to 1.25 GHz. This input impedance requirement applies to all valid input levels. The reference impedance for return loss measurements for is 50 ohms to ground for both the D+ and D- line (that is, as measured by a Vector Network Analyzer with 50 ohm probes - see Figure 71). Note: that the series capacitors CTX is optional for the return loss measurement.
5. Impedance during all LTSSM states. When transitioning from a Fundamental Reset to Detect (the initial state of the LTSSM) there is a 5 ms transition time before Receiver termination values must be met on all un-configured Lanes of a Port.
6. The RX DC Common Mode Impedance that exists when no power is present or Fundamental Reset is asserted. This helps ensure that the Receiver Detect circuit will not falsely assume a Receiver is powered on when it is not. This term must be measured at 300 mV above the RX ground.
7. It is recommended that the recovered TX UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function. Least squares and median deviation fits have worked well with experimental and simulated data.

2.22 Receiver Compliance Eye Diagrams

The RX eye diagram in Figure 70 is specified using the passive compliance/test measurement load (see Figure 71) in place of any real PCI Express RX component.

Note: In general, the minimum Receiver eye diagram measured with the compliance/test measurement load (see Figure 71) will be larger than the minimum Receiver eye diagram measured over a range of systems at the input Receiver of any real PCI Express component. The degraded eye diagram at the input Receiver is due to traces internal to the package as well as silicon parasitic characteristics which cause the real PCI Express component to vary in impedance from the compliance/test measurement load. The input Receiver eye diagram is implementation specific and is not specified. RX component designer should provide additional margin to adequately compensate for the degraded minimum Receiver eye diagram (shown in Figure 70) expected at the input Receiver based on some adequate combination of system simulations and the Return Loss measured looking into the RX package and silicon. The RX eye diagram must be aligned in time using the jitter median to locate the center of the eye diagram.

The eye diagram must be valid for any 250 consecutive UIs.

To measure Z_0 for the single-ended drivers, an external resistor is connected from the chip pad to OV_{DD} or GND. Then, the value of each resistor is varied until the pad voltage is $OV_{DD}/2$ (see Figure 77). The output impedance is the average of two components, the resistances of the pull-up and pull-down devices. When data is held high, SW1 is closed (SW2 is open) and R_P is trimmed until the voltage at the pad equals $OV_{DD}/2$. R_P then becomes the resistance of the pull-up devices. R_P and R_N are designed to be close to each other in value. Then, $Z_0 = (R_P + R_N)/2$.

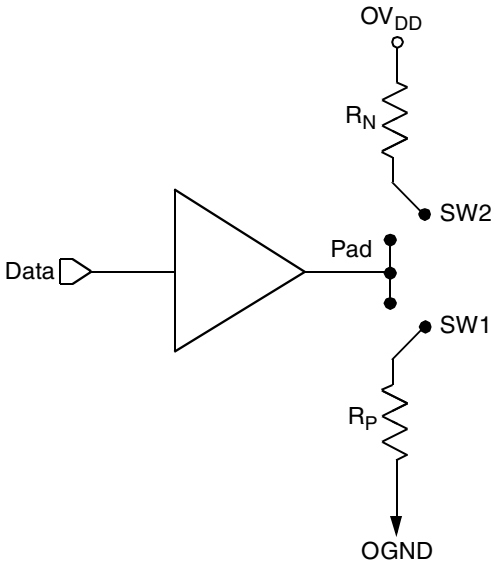


Figure 77. Driver Impedance Measurement

This table summarizes the signal impedance targets. The driver impedances are targeted at minimum V_{DD}, nominal OV_{DD}, 105°C.

Table 81. Impedance Characteristics

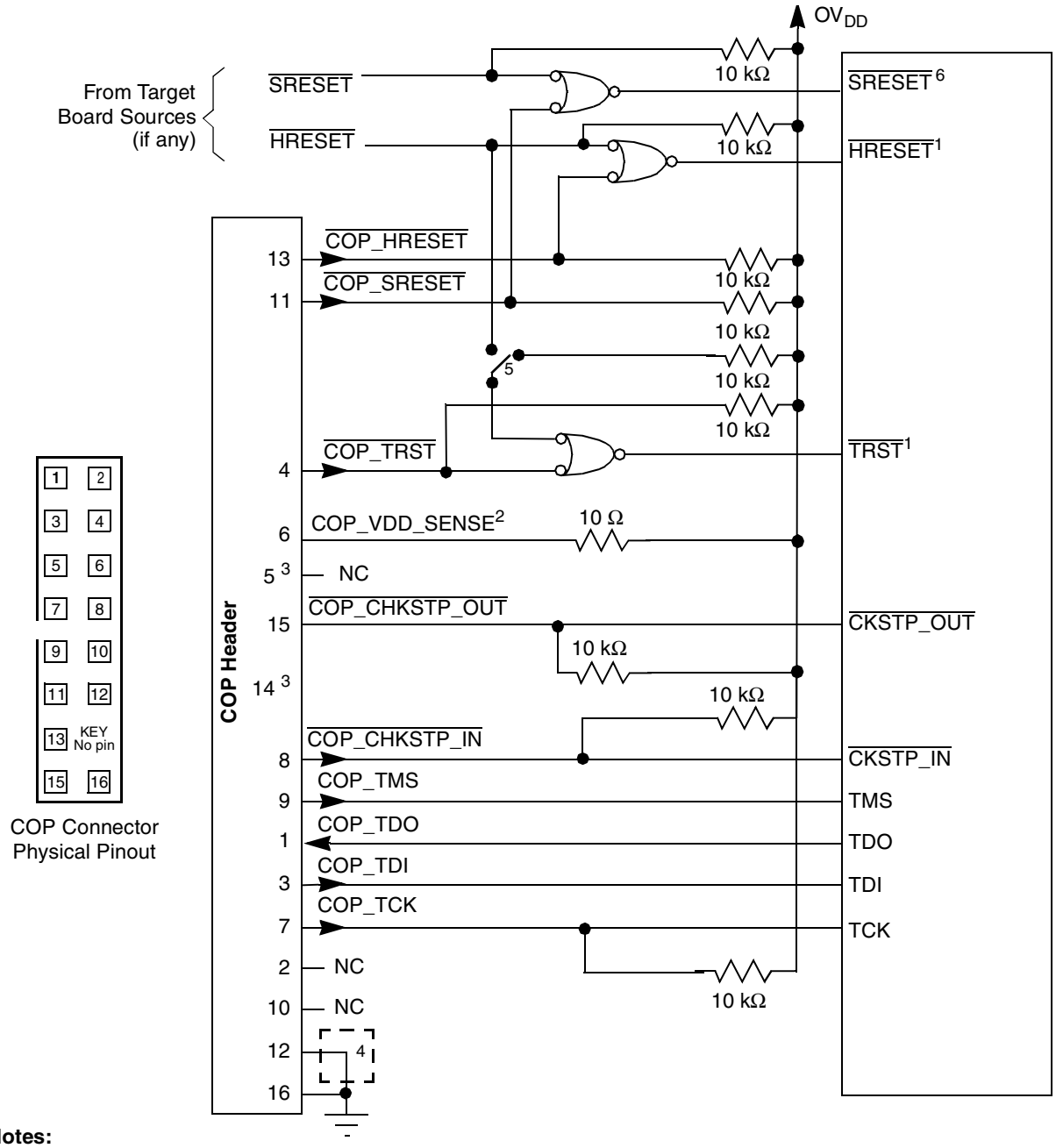
Impedance	Local Bus, Ethernet, DUART, Control, Configuration, Power Management	PCI	DDR DRAM	Symbol	Unit
R_N	45 Target	45 Target (cfg_pci_impd=1) 25 Target (cfg_pci_impd=0)	18 Target (full strength mode) 36 Target (full strength mode)	Z_0	Ω
R_P	45 Target	45 Target (cfg_pci_impd=1) 25 Target (cfg_pci_impd=0)	18 Target (full strength mode) 36 Target (full strength mode)	Z_0	Ω

Note: Nominal supply voltages. See Table 1.

3.9 Configuration Pin Muxing

The chip provides the user with power-on configuration options which can be set through the use of external pull-up or pull-down resistors of 4.7 kΩ on certain output pins (see customer visible configuration pins). These pins are generally used as output only pins in normal operation.

While $\overline{\text{HRESET}}$ is asserted however, these pins are treated as inputs. The value presented on these pins while $\overline{\text{HRESET}}$ is asserted, is latched when $\overline{\text{HRESET}}$ deasserts, at which time the input receiver is disabled and the I/O circuit takes on its normal function. Most of these sampled configuration pins are equipped with an on-chip gated resistor of approximately 20 kΩ. This value should permit the 4.7-kΩ resistor to pull the configuration pin to a valid logic low level. The pull-up resistor is enabled only during $\overline{\text{HRESET}}$ (and for platform /system clocks after $\overline{\text{HRESET}}$ deassertion to ensure capture of the reset value). When the input receiver is disabled the pull-up is also, thus allowing functional operation of the pin as an output with minimal signal quality or delay disruption. The default value for all configuration bits treated this way has been encoded such that a high voltage level puts the chip into the default state and external resistors are needed only when non-default settings are required by the user.



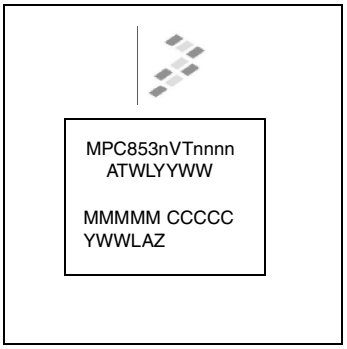
Notes:

1. The COP port and target board should be able to independently assert $\overline{\text{HRESET}}$ and $\overline{\text{TRST}}$ to the processor in order to fully control the processor as shown here.
2. Populate this with a 10 Ω resistor for short-circuit/current-limiting protection.
3. The KEY location (pin 14) is not physically present on the COP header.
4. Although pin 12 is defined as a No-Connect, some debug tools may use pin 12 as an additional GND pin for improved signal integrity.
5. This switch is included as a precaution for BSDL testing. The switch should be closed to position A during BSDL testing to avoid accidentally asserting the $\overline{\text{TRST}}$ line. If BSDL testing is not being performed, this switch should be closed to position B.
6. Asserting $\overline{\text{SRESET}}$ causes a machine check interrupt to the e500 core.

Figure 78. JTAG Interface Connection

4.2 Part Marking

Parts are marked as in the example shown in the following figure.



FC-PBGA

Notes:

MMMMM is the 5-digit mask number.
ATWLYYWW is the traceability code.
CCCCC is the country of assembly. This space is left blank if parts are assembled in the United States.

Figure 80. Part Marking for FC-PBGA

4.3 Part Numbering

These tables list all part numbers that are offered for the chip.

Table 83. MPC8535 Part Numbers Commercial Tier

Core/Platform/ DDR (MHz)	Standard Temp Without Security	Standard Temp With Security	Notes
600/400/400	MPC8535AVTAKG(A)	MPC8535EAVTAKG(A)	—
800/400/400	MPC8535AVTANG(A)	MPC8535EAVTANG(A)	—
1000/400/400	MPC8535AVTAQG(A)	MPC8535EAVTAQG(A)	—
1250/500/500	MPC8535AVTATH(A)	MPC8535EAVTATH(A)	—
1250/500/667	MPC8535AVTATLA	MPC8535EAVTATLA	—

Table 84. MPC8535 Part Numbers Industrial Tier

Core/Platform/ DDR (MHz)	Standard Temp Without Security	Standard Temp With Security	Extended Temp Without Security	Extended Temp With Security	Notes
600/400/400	MPC8535BVTAKG(A)	MPC8535EBVTAKG(A)	MPC8535CVTAKG(A)	MPC8535ECVTAKG(A)	1
800/400/400	MPC8535BVTANG(A)	MPC8535EBVTANG(A)	MPC8535CVTANG(A)	MPC8535ECVTANG(A)	
1000/400/400	MPC8535BVTAQG(A)	MPC8535EBVTAQG(A)	MPC8535CVTAQG(A)	MPC8535ECVTAQG(A)	
1250/500/500	MPC8535BVTATH(A)	MPC8535EBVTATH(A)	MPC8535CVTATH(A)	MPC8535ECVTATH(A)	
1250/500/667	MPC8535BVTATLA	MPC8535EBVTATLA	MPC8535CVTATLA	MPC8535ECVTATLA	

Note:

1. The last letter A indicates a Rev 1.2 silicon. It would be Rev 1.0 or Rev 1.1 silicon without a letter A