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Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

Details

Product Status	Obsolete
Core Processor	ST7
Core Size	8-Bit
Speed	16MHz
Connectivity	SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	22
Program Memory Size	4KB (4K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 6x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Through Hole
Package / Case	32-SDIP (0.400", 10.16mm)
Supplier Device Package	-
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/st72f262g1b6

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

1 INTRODUCTION

The ST72260Gx, ST72262Gx and ST72264Gx devices are members of the ST7 microcontroller family. They can be grouped as follows :

- ST72264Gx devices are designed for mid-range applications with ADC, I²C and SCI interface capabilities.
- ST72262Gx devices target the same range of applications but without 1²C interface or SCI.
- ST72260Gx devices are for applications that do not need ADC, I²C peripherals or SCI.

All devices are based on a common industrystandard 8-bit core, featuring an enhanced instruction set.

The ST72F260G, ST72F262G, and ST72F264G versions feature single-voltage FLASH memory with byte-by-byte In-Circuit Programming (ICP) capabilities.

Under software control, all devices can be placed in WAIT, SLOW, Active-HALT or HALT mode, reducing power consumption when the application is in idle or stand-by state.

The enhanced instruction set and addressing modes of the ST7 offer both power and flexibility to software developers, enabling the design of highly efficient and compact application code. In addition to standard 8-bit data management, all ST7 microcontrollers feature true bit manipulation, 8x8 unsigned multiplication and indirect addressing modes.

For easy reference, all parametric data is located in Section 13 on page 126.

Related Documentation

AN1365: Guidelines for migrating ST72C254 applications to ST72F264



Figure 1. General Block Diagram

F	Pin n'	0			Le	evel		Ро	rt / C	Cont	trol		Main			
32	58	۷	Pin Name	ype	Ŧ	out		Inp	out		Out	put	Function	Alternate Function		
SDIP	SO2	BG		ι Η	Inpi	Outp	float	ndw	int	ana	OD	РР	reset)			
15	13	E2	PC4/OCMP2_B/AIN4	I/O	(CT	x	ei0,	/ei1	х	х	Х	Port C4	Timer B Output Compare 2 or ADC Analog Input 4		
16	14	F3	PC3/ ICAP2_B/AIN3	I/O	(CT	x	ei0	/ei1	х	х	х	Port C3	Timer B Input Capture 2 or ADC Analog Input 3		
17	15	E3	PC2/MCO/AIN2	I/O	(CT	x	ei0	/ei1	х	х	х	Port C2	Main clock output (f _{CPU}) or ADC Analog Input 2		
18	16	F4	PC1/OCMP1_B/AIN1	I/O	(CT	x	ei0	/ei1	х	х	х	Port C1	Timer B Output Compare 1 or ADC Analog Input 1		
19	17	D3	PC0/ICAP1_B/AIN0	I/O	(CT	x	ei0,	/ei1	х	х	х	Port C0	Timer B Input Capture 1 or ADC Analog Input 0		
20	18	E4	PA7/TDO	I/O	C_T	HS	Χ	е	i0		Х	Х	Port A7	SCI output		
21	19	F5	PA6/SDAI	I/O	C_T	HS	Х		ei0		Т		Port A6	I ² C DATA		
22	20	F6	PA5 /RDI	I/O	C_T	HS	Х	е	i0		Х	Х	Port A5	SCI input		
23	21	E6	PA4/SCLI	I/O	C_{T}	HS	Х		ei0		Т		Port A4	I ² C CLOCK		
24		E5	NC													
25		D6	NC								Ν	ot C	onnected			
		D5	NC													
26	22	C6	PA3	I/O	C_T	HS	Χ	е	i0		Х	Х	Port A3			
27	23	D4	PA2	I/O	C_{T}	HS	Х	е	i0		Х	Х	Port A2			
		C5	NC		•	•		•			NI					
		B6	NC								IN		Connected			
28	24	A6	PA1/ICCDATA	I/O	C_T	HS	Х	е	i0		Х	Х	Port A1	In Circuit Communication Data		
29	25	A5	PA0/ICCCLK	I/O	CT	HS	Х	е	i0		х	Х	Port A0	In Circuit Communication Clock		
30	26	B5	ICCSEL	Ι	C_T		Х						ICC mode	e pin, must be tied low		
31	27	A4	V _{SS}	S									Ground			
32	28	B4	V _{DD}	S									Main pow	ver supply		

Notes:

1. In the interrupt input column, "eiX" defines the associated external interrupt vector. If the weak pull-up column (wpu) is merged with the interrupt column (int), then the I/O configuration is a pull-up interrupt input, otherwise the configuration is a floating interrupt input. Port C is mapped to ei0 or ei1 by option byte.

2. In the open drain output column, "T" defines a true open drain I/O (P-Buffer and protection diode to V_{DD} are not implemented). See Section 9 "I/O PORTS" on page 38 for more details.

3. OSC1 and OSC2 pins connect a crystal or ceramic resonator, or an external source to the on-chip oscillator see Section 2 "PIN DESCRIPTION" on page 6 and Section 6.2 "MULTI-OSCILLATOR (MO)" on page 21 for more details.

4: For details refer to Section 13.8 on page 144

Address	Block	Register Label	Register Name	Reset Status	Remarks
0031h 0032h 0033h 0034h 0035h 0036h 0037h 0038h 0039h 003Ah 003Ah 003Bh 003Ch 003Dh 003Fh	TIMER A	TACR2 TACR1 TASCSR TAIC1HR TAIC1LR TAOC1LR TAOC1LR TACLR TACLR TAACHR TAACLR TAACLR TAIC2HR TAIC2LR TAOC2LR TAOC2LR	Timer A Control Register 2 Timer A Control Register 1 Timer A Control/Status Register Timer A Input Capture 1 High Register Timer A Input Capture 1 Low Register Timer A Output Compare 1 High Register Timer A Output Compare 1 Low Register Timer A Counter High Register Timer A Counter High Register Timer A Counter Low Register Timer A Alternate Counter Low Register Timer A Alternate Counter Low Register Timer A Input Capture 2 High Register Timer A Input Capture 2 Low Register Timer A Output Compare 2 Low Register	00h 00h xxh xxh 80h 00h FFh FCh FCh FCh xxh xxh 80h 00h	R/W R/W Read Only Read Only R/W Read Only Read Only Read Only Read Only Read Only Read Only Read Only Read Only Read Only Read Only R/W R/W
0040h		MISCR2	Miscellanous register 2	00h	R/W
0041h 0042h 0043h 0044h 0045h 0046h 0047h 0048h 0049h 004Ah 004Bh 004Ch 004Ch 004Ch 004Ch 004Ch 004Ch 004Ch 0050h 0051h 0052h 0053h 0055h 0056h	TIMER B	TBCR2 TBCR1 TBSCSR TBIC1HR TBIC1LR TBOC1HR TBOC1LR TBC1LR TBCLR TBCLR TBCLR TBACHR TBACLR TBIC2HR TBIC2HR TBIC2LR TBIC2LR TBOC2LR SCISR SCIDR SCIBRR SCICR1 SCICR2 SCIERPR SCIETPR	Timer B Control Register 2 Timer B Control Register 1 Timer B Control/Status Register Timer B Input Capture 1 High Register Timer B Input Capture 1 Low Register Timer B Output Compare 1 High Register Timer B Output Compare 1 Low Register Timer B Counter High Register Timer B Counter Low Register Timer B Alternate Counter High Register Timer B Alternate Counter Low Register Timer B Alternate Counter Low Register Timer B Input Capture 2 High Register Timer B Output Compare 2 High Register Timer B Output Compare 2 Low Register Timer B Output Compare 2 Low Register SCI Status Register SCI Data Register SCI Control Register1 SCI Control Register2 SCI Extended Receive Prescaler Register	00h 00h xxh xxh xxh 80h 00h FFh FCh FCh FCh xxh xxh 80h 00h C0h xxh 00h x000 0000h 00h 00h	R/W R/W R/W Read Only Read Only Read Only Read Only Read Only Read Only Read Only Read Only R/W R/W R/W R/W R/W R/W R/W R/W R/W R/W
0057h to 006Eh		L	Reserved (24 Bytes)	<u> </u>	
006Fh 0070h 0071h	ADC	ADCDRL ADCDRH ADCCSR	Data Register Low ³⁾ Data Register High ³⁾ Control/Status Register	00h 00h 00h	Read Only Read Only R/W
0072h	FLASH	FCSR	Flash Control Register	00h	R/W
0073h to 007Fh			Reserved (13 Bytes)		

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6 SUPPLY, RESET AND CLOCK MANAGEMENT

The device includes a range of utility features for securing the application in critical situations (for example in case of a power brown-out), and reducing the number of external components. An overview is shown in Figure 10.

For more details, refer to dedicated parametric section.

Main Features

- Optional PLL for multiplying the frequency by 2 (not to be used with internal RC oscillator)
- Reset Sequence Manager (RSM)
- Multi-Oscillator Clock Management (MO)
 - 4 Crystal/Ceramic resonator oscillators
 - 1 Internal RC oscillator
- System Integrity Management (SI)
 - Main supply Low Voltage Detector (LVD)
 - Auxiliary Voltage Detector (AVD) with interrupt capability for monitoring the main supply

6.1 PHASE LOCKED LOOP

If the clock frequency input to the PLL is in the 2 to 4 MHz range, the PLL can be used to multiply the frequency by two to obtain an f_{OSC2} of 4 to 8 MHz.

The PLL is enabled by option byte. If the PLL is disabled, then $f_{OSC2} = f_{OSC}/2$.

Caution: The PLL is not recommended for applications where timing accuracy is required. See "PLL Characteristics" on page 139.

Figure 9. PLL Block Diagram



Figure 10. Clock, Reset and Supply Block Diagram



7 INTERRUPTS

7.1 INTRODUCTION

The ST7 enhanced interrupt management provides the following features:

- Hardware interrupts
- Software interrupt (TRAP)
- Nested or concurrent interrupt management with flexible interrupt priority and level management:
 - Up to 4 software programmable nesting levels
 - Up to 16 interrupt vectors fixed by hardware
- 2 non-maskable events: RESET and TRAP
 This interrupt management is based on:
- Bit 5 and bit 3 of the CPU CC register (I1:0),
- Interrupt software priority registers (ISPRx),
- Fixed interrupt vector addresses located at the high addresses of the memory map (FFE0h to FFFFh) sorted by hardware priority order.

This enhanced interrupt controller guarantees full upward compatibility with the standard (not nested) ST7 interrupt controller.

7.2 MASKING AND PROCESSING FLOW

The interrupt masking is managed by the I1 and I0 bits of the CC register and the ISPRx registers which give the interrupt software priority level of each interrupt vector (see Table 4). The processing flow is shown in Figure 16

Figure 16. Interrupt Processing Flowchart

When an interrupt request has to be serviced:

- Normal processing is suspended at the end of the current instruction execution.
- The PC, X, A and CC registers are saved onto the stack.
- I1 and I0 bits of CC register are set according to the corresponding values in the ISPRx registers of the serviced interrupt vector.
- The PC is then loaded with the interrupt vector of the interrupt to service and the first instruction of the interrupt service routine is fetched (refer to "Interrupt Mapping" table for vector addresses).

The interrupt service routine should end with the IRET instruction which causes the contents of the saved registers to be recovered from the stack.

Note: As a consequence of the IRET instruction, the I1 and I0 bits will be restored from the stack and the program in the previous level will resume.

Table 4. Interrupt Software Priority Levels

Interrupt software priority	Level	l1	10
Level 0 (main)	Low	1	0
Level 1		0	1
Level 2	•	0	0
Level 3 (= interrupt disable)	High	1	1



INTERRUPTS (Cont'd)

7.5 INTERRUPT REGISTER DESCRIPTION

CPU CC REGISTER INTERRUPT BITS

Read/Write

5/

Reset Value: 111x 1010 (xAh)

7							0
1	1	11	н	10	Ν	z	с

Bit 5, 3 = 11, 10 Software Interrupt Priority

These two bits indicate the current interrupt software priority.

Interrupt Software Priority	Level	11	10
Level 0 (main)	Low	1	0
Level 1		0	1
Level 2	. ★	0	0
Level 3 (= interrupt disable*)	High	1	1

These two bits are set/cleared by hardware when entering in interrupt. The loaded value is given by the corresponding bits in the interrupt software priority registers (ISPRx).

They can be also set/cleared by software with the RIM, SIM, HALT, WFI, IRET and PUSH/POP instructions (see "Interrupt Dedicated Instruction Set" table).

*Note: TRAP and RESET events are non maskable sources and can interrupt a level 3 program.

INTERRUPT SOFTWARE PRIORITY REGISTERS (ISPRX)

Read/Write (bits 7:4 of **ISPR3** are read only) Reset Value: 1111 1111 (FFh)

_	7				-	-	-	0
ISPR0	l1_3	10_3	l1_2	10_2	11_1	10_1	l1_0	10_0
ISPR1	l1_7	10_7	l1_6	I0_6	l1_5	10_5	11_4	I0_4
ISPR2	11_11	10_11	11_10	10_10	l1_9	10_9	l1_8	I0_8
ISPR3	1	1	1	1	11_13	10_13	11_12	10_12

These four registers contain the interrupt software priority of each interrupt vector.

 Each interrupt vector (except RESET and TRAP) has corresponding bits in these registers where its own software priority is stored. This correspondance is shown in the following table.

Vector Address	ISPRx Bits
FFFBh-FFFAh	ei0
FFF9h-FFF8h	ei1
FFE1h-FFE0h	Not used

- Each I1_x and I0_x bit value in the ISPRx registers has the same meaning as the I1 and I0 bits in the CC register.
- Level 0 can not be written (l1_x=1, l0_x=0). In this case, the previously stored value is kept. (example: previous=CFh, write=64h, result=44h)

The RESET and TRAP vectors have no software priorities. When one is serviced, the I1 and I0 bits of the CC register are both set.

Caution: If the $I1_x$ and $I0_x$ bits are modified while the interrupt x is executed the following behaviour has to be considered: If the interrupt x is still pending (new interrupt or flag not cleared) and the new software priority is higher than the previous one, the interrupt x is re-entered. Otherwise, the software priority stays unchanged up to the next interrupt request (after the IRET of the interrupt x).

8 POWER SAVING MODES

8.1 INTRODUCTION

To give a large measure of flexibility to the application in terms of power consumption, three main power saving modes are implemented in the ST7 (see Figure 20).

After a RESET the normal operating mode is selected by default (RUN mode). This mode drives the device (CPU and embedded peripherals) by means of a master clock which is based on the main oscillator frequency divided by 2 (f_{CPU}).

From Run mode, the different power saving modes may be selected by setting the relevant register bits or by calling the specific ST7 software instruction whose action depends on the oscillator status.



Figure 20. Power Saving Mode Transitions

8.2 SLOW MODE

This mode has two targets:

- To reduce power consumption by decreasing the internal clock in the device,
- To adapt the internal clock frequency (f_{CPU}) to the available supply voltage.

SLOW mode is controlled by three bits in the MISR1 register: the SMS bit which enables or disables Slow mode and two CPx bits which select the internal slow frequency (f_{CPU}).

In this mode, the oscillator frequency can be divided by 4, 8, 16 or 32 instead of 2 in normal operating mode. The CPU and peripherals are clocked at this lower frequency.

Note: SLOW-WAIT mode is activated when enterring the WAIT mode while the device is already in SLOW mode.

Figure 21. SLOW Mode Clock Transitions



POWER SAVING MODES (Cont'd)

8.3 WAIT MODE

WAIT mode places the MCU in a low power consumption mode by stopping the CPU.

This power saving mode is selected by calling the "WFI" ST7 software instruction.

All peripherals remain active. During WAIT mode, the I [1:0] bits in the CC register are forced to '10b', to enable all interrupts. All other registers and memory remain unchanged. The MCU remains in WAIT mode until an interrupt or Reset occurs, whereupon the Program Counter branches to the starting address of the interrupt or Reset service routine.

The MCU will remain in WAIT mode until a Reset or an Interrupt occurs, causing it to wake up.

Refer to Figure 22.



Note:

1. Before servicing an interrupt, the CC register is pushed on the stack. The I[1:0] bits in the CC register are set during the interrupt routine and cleared when the CC register is popped.

MISCELLANEOUS REGISTERS (Cont'd)

MISCELLANEOUS REGISTER 2 (MISCR2)

Read/Write

Reset Value: 0000 0000 (00h)

7							0
0	0	0	0	MOD	SOD	SSM	SSI

Caution: This register has been provided for compatibility with the ST72254 family only. The same bits are available in the SPICSR register. New applications must use the SPICSR register. Do not use both registers, this will cause the SPI to malfunction.

Bits 7:4 = **Reserved** *always read as 0*

Bits 3 = **MOD** *SPI Master Output Disable* This bit is set and cleared by software. When set, it disables the SPI Master (MOSI) output signal. 0: SPI Master Output enabled. 1: SPI Master Output disabled.

Bit 2 = **SOD** *SPI Slave Output Disable* This bit is set and cleared by software. When set it disable the SPI Slave (MISO) output signal. 0: SPI Slave Output enabled. 1: SPI Slave Output disabled.

Bit 1 = **SSM** \overline{SS} mode selection

This bit is set and cleared by software.

- 0: Normal mode the level of the SPI SS signal is input from the external SS pin.___
- 1: I/O mode, the level of the SPI SS signal is read from the SSI bit.

Bit $0 = SSI \overline{SS}$ internal mode This bit replaces the \overline{SS} pin of the SPI when the SSM bit is set to 1. (see SPI description). It is set and cleared by software.

Address (Hex.)	Register Label	7	6	5	4	3	2	1	0
0020h	MISCR1 Reset Value	IS11 0	IS10 0	MCO 0	IS01 0	IS00 0	CP1 0	CP0 0	SMS 0
0040h	MISCR2 Reset Value	0	0	0	0	MOD 0	SOD 0	SSM 0	SSI 0

Table 11. Miscellaneous Register Map and Reset Values

16-BIT TIMER (Cont'd)

11.3.3.5 One Pulse Mode

One Pulse mode enables the generation of a pulse when an external event occurs. This mode is selected via the OPM bit in the CR2 register.

The one pulse mode uses the Input Capture1 function and the Output Compare1 function.

Procedure:

To use one pulse mode:

- 1. Load the OC1R register with the value corresponding to the length of the pulse (see the formula in the opposite column).
- 2. Select the following in the CR1 register:
 - Using the OLVL1 bit, select the level to be applied to the OCMP1 pin after the pulse.
 - Using the OLVL2 bit, select the level to be applied to the OCMP1 pin during the pulse.
 - Select the edge of the active transition on the ICAP1 pin with the IEDG1 bit (the ICAP1 pin must be configured as floating input).
- 3. Select the following in the CR2 register:
 - Set the OC1E bit, the OCMP1 pin is then dedicated to the Output Compare 1 function.
 - Set the OPM bit.
 - Select the timer clock CC[1:0] (see Table 14 Clock Control Bits).



When a valid event occurs on the ICAP1 pin, the counter value is loaded in the ICR1 register. The counter is then initialized to FFFCh, the OLVL2 bit is output on the OCMP1 pin and the ICF1 bit is set.

Because the ICF1 bit is set when an active edge occurs, an interrupt can be generated if the ICIE bit is set.

Clearing the Input Capture interrupt request (i.e. clearing the ICF*i* bit) is done in two steps:

1. Reading the SR register while the ICF*i* bit is set.

2. An access (read or write) to the ICiLR register.

The OC1R register value required for a specific timing application can be calculated using the following formula:

Where:

t = Pulse period (in seconds)

f_{CPU} = CPU clock frequency (in hertz)

PRESC = Timer prescaler factor (2, 4 or 8 depending on the CC[1:0] bits, see Table 14 Clock Control Bits)

If the timer clock is an external clock the formula is:

Where:

t = Pulse period (in seconds)

f_{EXT} = External timer clock frequency (in hertz)

When the value of the counter is equal to the value of the contents of the OC1R register, the OLVL1 bit is output on the OCMP1 pin, (See Figure 44).

Notes:

- 1. The OCF1 bit cannot be set by hardware in one pulse mode but the OCF2 bit can generate an Output Compare interrupt.
- 2. When the Pulse Width Modulation (PWM) and One Pulse Mode (OPM) bits are both set, the PWM mode is the only active one.
- 3. If OLVL1=OLVL2 a continuous signal will be seen on the OCMP1 pin.
- 4. The ICAP1 pin can not be used to perform input capture. The ICAP2 pin can be used to perform input capture (ICF2 can be set and IC2R can be loaded) but the user must take care that the counter is reset each time a valid edge occurs on the ICAP1 pin and ICF1 can also generates interrupt if ICIE is set.
- 5. When one pulse mode is used OC1R is dedicated to this mode. Nevertheless OC2R and OCF2 can be used to indicate a period of time has been elapsed but cannot generate an output waveform because the level OLVL2 is dedicated to the one pulse mode.

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16-BIT TIMER (Cont'd) CONTROL/STATUS REGISTER (CSR)

Read Only

Reset Value: 0000 0000 (00h)

The three least significant bits are not used.

7							0
ICF1	OCF1	TOF	ICF2	OCF2	TIMD	0	0

Bit 7 = ICF1 Input Capture Flag 1.

0: No input capture (reset value).

1: An input capture has occurred on the ICAP1 pin or the counter has reached the OC2R value in PWM mode. To clear this bit, first read the SR register, then read or write the low byte of the IC1R (IC1LR) register.

Bit 6 = OCF1 Output Compare Flag 1.

0: No match (reset value).

1: The content of the free running counter has matched the content of the OC1R register. To clear this bit, first read the SR register, then read or write the low byte of the OC1R (OC1LR) register.

Bit 5 = **TOF** *Timer Overflow Flag.*

- 0: No timer overflow (reset value).
- 1: The free running counter rolled over from FFFFh to 0000h. To clear this bit, first read the SR register, then read or write the low byte of the CR (CLR) register.

Note: Reading or writing the ACLR register does not clear TOF.

Bit 4 = ICF2 Input Capture Flag 2.

- 0: No input capture (reset value).
- 1: An input capture has occurred on the ICAP2 pin. To clear this bit, first read the SR register, then read or write the low byte of the IC2R (IC2LR) register.

Bit 3 = **OCF2** *Output Compare Flag 2.*

- 0: No match (reset value).
- 1: The content of the free running counter has matched the content of the OC2R register. To clear this bit, first read the SR register, then read or write the low byte of the OC2R (OC2LR) register.

Bit 2 = **TIMD** *Timer disable.*

This bit is set and cleared by software. When set, it freezes the timer prescaler and counter and disabled the output functions (OCMP1 and OCMP2 pins) to reduce power consumption. Access to the timer registers is still available, allowing the timer configuration to be changed while it is disabled. 0: Timer enabled

1: Timer prescaler, counter and outputs disabled

Bits 1:0 = Reserved, must be kept cleared.

SERIAL COMMUNICATIONS INTERFACE (Cont'd)

11.5.4.7 Parity Control

Parity control (generation of parity bit in transmission and parity checking in reception) can be enabled by setting the PCE bit in the SCICR1 register. Depending on the frame length defined by the M bit, the possible SCI frame formats are as listed in Table 18.

Table 18. Frame Form	nats
----------------------	------

M bit	PCE bit	SCI frame
0	0	SB 8 bit data STB
0	1	SB 7-bit data PB STB
1	0	SB 9-bit data STB
1	1	SB 8-bit data PB STB

Legend: SB = Start Bit, STB = Stop Bit,

PB = Parity Bit

Note: In case of wake up by an address mark, the MSB bit of the data is taken into account and not the parity bit

Even parity: the parity bit is calculated to obtain an even number of "1s" inside the frame made of the 7 or 8 LSB bits (depending on whether M is equal to 0 or 1) and the parity bit.

Ex: data=00110101; 4 bits set => parity bit will be 0 if even parity is selected (PS bit = 0).

Odd parity: the parity bit is calculated to obtain an odd number of "1s" inside the frame made of the 7 or 8 LSB bits (depending on whether M is equal to 0 or 1) and the parity bit.

Ex: data=00110101; 4 bits set => parity bit will be 1 if odd parity is selected (PS bit = 1).

Transmission mode: If the PCE bit is set then the MSB bit of the data written in the data register is not transmitted but is changed by the parity bit.

Reception mode: If the PCE bit is set then the interface checks if the received data byte has an even number of "1s" if even parity is selected (PS=0) or an odd number of "1s" if odd parity is selected (PS=1). If the parity check fails, the PE flag is set in the SCISR register and an interrupt is generated if PIE is set in the SCICR1 register.

11.5.4.8 SCI Clock Tolerance

During reception, each bit is sampled 16 times. The majority of the 8th, 9th and 10th samples is considered as the bit value. For a valid bit detection, all the three samples should have the same value otherwise the noise flag (NF) is set. For example: if the 8th, 9th and 10th samples are 0, 1 and 1 respectively, then the bit value will be "1", but the Noise Flag bit is be set because the three samples values are not the same.

Consequently, the bit length must be long enough so that the 8th, 9th and 10th samples have the desired bit value. This means the clock frequency should not vary more than 6/16 (37.5%) within one bit. The sampling clock is resynchronized at each start bit, so that when receiving 10 bits (one start bit, 1 data byte, 1 stop bit), the clock deviation must not exceed 3.75%.

Note: The internal sampling clock of the microcontroller samples the pin value on every falling edge. Therefore, the internal sampling clock and the time the application expects the sampling to take place may be out of sync. For example: If the baud rate is 15.625 kbaud (bit length is 64µs), then the 8th, 9th and 10th samples will be at 28µs, 32µs & 36µs respectively (the first sample starting ideally at 0µs). But if the falling edge of the internal clock occurs just before the pin value changes, the samples would then be out of sync by ~4us. This means the entire bit length must be at least 40µs (36µs for the 10th sample + 4µs for synchronization with the internal sampling clock).

SERIAL COMMUNICATIONS INTERFACE (Cont'd)

11.5.4.9 Clock Deviation Causes

The causes which contribute to the total deviation are:

- D_{TRA}: Deviation due to transmitter error (Local oscillator error of the transmitter or the transmitter is transmitting at a different baud rate).
- D_{QUANT}: Error due to the baud rate quantisation of the receiver.
- D_{REC}: Deviation of the local oscillator of the receiver: This deviation can occur during the reception of one complete SCI message assuming that the deviation has been compensated at the beginning of the message.
- D_{TCL}: Deviation due to the transmission line (generally due to the transceivers)

All the deviations of the system should be added and compared to the SCI clock tolerance:

 $\mathsf{D}_{\mathsf{TRA}} + \mathsf{D}_{\mathsf{QUANT}} + \mathsf{D}_{\mathsf{REC}} + \mathsf{D}_{\mathsf{TCL}} < 3.75\%$

11.5.4.10 Noise Error Causes

See also description of Noise error in Section 11.5.4.3.

Start bit

The noise flag (NF) is set during start bit reception if one of the following conditions occurs:

- 1. A valid falling edge is not detected. A falling edge is considered to be valid if the 3 consecutive samples before the falling edge occurs are detected as '1' and, after the falling edge occurs, during the sampling of the 16 samples, if one of the samples numbered 3, 5 or 7 is detected as a "1".
- 2. During sampling of the 16 samples, if one of the samples numbered 8, 9 or 10 is detected as a "1".

Therefore, a valid Start Bit must satisfy both the above conditions to prevent the Noise Flag getting set.

Data Bits

The noise flag (NF) is set during normal data bit reception if the following condition occurs:

 During the sampling of 16 samples, if all three samples numbered 8, 9 and 10 are not the same. The majority of the 8th, 9th and 10th samples is considered as the bit value.

Therefore, a valid Data Bit must have samples 8, 9 and 10 at the same value to prevent the Noise Flag getting set.



Figure 56. Bit Sampling in Reception Mode

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SERIAL COMMUNICATIONS INTERFACE (Cont'd) CONTROL REGISTER 1 (SCICR1)

Read/Write

Reset Value: x000 0000 (x0h)

7						0	
R8	Т8	SCID	М	WAKE	PCE	PS	PIE

Bit 7 = R8 Receive data bit 8.

This bit is used to store the 9th bit of the received word when M=1.

Bit 6 = T8 Transmit data bit 8.

This bit is used to store the 9th bit of the transmitted word when M=1.

Bit 5 = **SCID** *Disabled for low power consumption* When this bit is set the SCI prescalers and outputs are stopped and the end of the current byte transfer in order to reduce power consumption. This bit is set and cleared by software.

0: SCI enabled

1: SCI prescaler and outputs disabled

Bit $4 = \mathbf{M}$ Word length. This bit determines the word length. It is set or cleared by software.

0: 1 Start bit, 8 Data bits, 1 Stop bit

1: 1 Start bit, 9 Data bits, 1 Stop bit

Note: The M bit must not be modified during a data transfer (both transmission and reception).

Bit 3 = WAKE Wake-Up method.

This bit determines the SCI Wake-Up method, it is set or cleared by software. 0: Idle Line 1: Address Mark

Bit 2 = **PCE** *Parity control enable.*

This bit selects the hardware parity control (generation and detection). When the parity control is enabled, the computed parity is inserted at the MSB position (9th bit if M=1; 8th bit if M=0) and parity is checked on the received data. This bit is set and cleared by software. Once it is set, PCE is active after the current byte (in reception and in transmission).

0: Parity control disabled

1: Parity control enabled

Bit 1 = **PS** Parity selection.

This bit selects the odd or even parity when the parity generation/detection is enabled (PCE bit set). It is set and cleared by software. The parity will be selected after the current byte.

0: Even parity

1: Odd parity

Bit 0 = **PIE** Parity interrupt enable.

This bit enables the interrupt capability of the hardware parity control when a parity error is detected (PE bit set). It is set and cleared by software.

0: Parity error interrupt disabled

1: Parity error interrupt enabled.

I²C BUS INTERFACE (Cont'd)

Acknowledge may be enabled and disabled by software.

The I²C interface address and/or general call address can be selected by software.

The speed of the I^2C interface may be selected between Standard (up to 100KHz) and Fast I^2C (up to 400KHz).

SDA/SCL Line Control

Transmitter mode: the interface holds the clock line low before transmission to wait for the microcontroller to write the byte in the Data Register.

Receiver mode: the interface holds the clock line low after reception to wait for the microcontroller to read the byte in the Data Register. The SCL frequency ($\rm F_{scl}$) is controlled by a programmable clock divider which depends on the $\rm I^2C$ bus mode.

When the I^2C cell is enabled, the SDA and SCL ports must be configured as floating inputs. In this case, the value of the external pull-up resistor used depends on the application.

When the I²C cell is disabled, the SDA and SCL ports revert to being standard I/O port pins.



Figure 58. I²C Interface Block Diagram

13.3 OPERATING CONDITIONS

13.3.1 General Operating Conditions

 $T_A = -40$ to $+85^{\circ}C$ unless otherwise specified.

Symbol	Parameter	Conditions	Min	Max	Unit	
V _{DD}	Supply voltage	f _{OSC} = 8 MHz. max.	2.7	5.5	V	
	Supply voltage	f _{OSC} = 16 MHz. max.	3.3	5.5	v	
f _{OSC}	External clock frequency on OSC1 pin	V _{DD} ≥3.3V	up to 16		MHz	
		V _{DD} ≥2.7V	up to 8			

Figure 64. f_{OSC} Maximum Operating Frequency Versus V_{DD} Supply Voltage



OPERATING CONDITIONS (Cont'd)

13.3.2 Operating Conditions with Low Voltage Detector (LVD)

 $T_A = -40$ to $+85^{\circ}C$ unless otherwise specified

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
	Reset release threshold	High Threshold	4.0 ¹⁾	4.2	4.5	
V _{IT+(LVD)}	$(V_{-}, rico)$	Med. Threshold	3.55 ¹⁾	3.75	4.0	
~ /	(V _{DD} lise)	Low Threshold	2.95 ¹⁾	3.15	3.35	V
	Report generation threshold	High Threshold	3.75	4.0	4.25 ¹⁾	v
V _{IT-(LVD)}		Med. Threshold	3.3	3.55	3.75 ¹⁾	
()	(V _{DD} Iall)	Low Threshold	2.75	3.0	3.15 ¹⁾	
V _{hys(LVD)}	LVD voltage threshold hysteresis	V _{IT+(LVD)} -V _{IT-(LVD)}		200		mV
1/+	$V_{\rm risc}$ time rate $^{1)2)3)$	Flash	20µs/V		20ms/V	
V 'POR	VDD lise line late 2000	ROM	3.75 4.0 4 3.3 3.55 3 2.75 3.0 3 20μs/V 20 20 20μs/V 20 20	∞		
t _{g(VDD)}	Filtered glitch delay on $V_{DD}^{(1)}$	Not detected by the LVD			40	ns

Notes:

1. Data based on characterization results, not tested in production.

2. When Vt_{POR} is faster than 100 μ s/V, the Reset signal is released after a delay of max. 42 μ s after V_{DD} crosses the V_{IT+(LVD)} threshold.

3. Use of LVD with capacitive power supply: with this type of power supply, if power cuts occur in the application, it is recommended to pull V_{DD} down to 0V to ensure optimum restart conditions. Refer to circuit example in Figure 91 on page 151 and note 6.

Figure 65. LVD Startup Behaviour



Note: When the LVD is enabled, the MCU reaches its authorized operating voltage from a reset state. However, in some devices, the reset signal may be undefined until V_{DD} is approximately 2V. As a consequence, the I/Os may toggle when V_{DD} is below this voltage.

Because Flash write access is impossible below this voltage, the Flash memory contents will not be corrupted.

CLOCK AND TIMING CHARACTERISTICS (Cont'd)

13.5.3 Crystal and Ceramic Resonator Oscillators

The ST7 internal clock can be supplied with four different Crystal/Ceramic resonator oscillators. All the information given in this paragraph are based on characterization results with specified typical external components. In the application, the resonator and the load capacitors have to be placed as

close as possible to the oscillator pins in order to minimize output distortion and start-up stabilization time. Refer to the crystal/ceramic resonator manufacturer for more details (frequency, package, accuracy...).

Symbol	Parameter	Cor	Min	Max	Unit	
		VLP : Very Lov	0.032	0.1		
	Oscillator Frequency ¹⁾	LP: Low powe	r oscillator	1	2	
fosc		MP: Medium p	>2	4	MHz	
		MS: Medium s	peed oscillator	>4	8	
		HS: High spee	>8	16		
R _F	Feedback resistor			20	40	kΩ
			VLP oscillator	60	100	
C	Recommended load capacitance ver-	R _S =200Ω	LP oscillator	38	100	
C _{L2}	sus equivalent serial resistance of the	R _S =200Ω	MP oscillator	32	47	pF
	crystal or ceramic resonator (R _S)	R _S =200Ω	MS oscillator	10	47	
		R _S =100Ω	HS oscillator	10	30	

Symbol	Parameter	C	Тур	Max	Unit	
			VLP oscillator	2.5	5	
i ₂	OSC2 driving current	V _{DD} =5V	LP oscillator	80	150	
		V _{IN} =V _{SS}	MP oscillator	160	250	μA
			MS oscillator	310	460	
			HS oscillator	610	900	

Figure 71. Typical Application with a Crystal or Ceramic Resonator



Notes:

1. The oscillator selection can be optimized in terms of supply current using an high quality resonator with small R_S value. Refer to crystal/ceramic resonator manufacturer for more details.

13.8 I/O PORT PIN CHARACTERISTICS

13.8.1 General Characteristics

 $T_A = -40$ to $+85^{\circ}C$ unless otherwise specified

Symbol	Parameter	Cond	litions	Min	Тур	Max	Unit
V _{IL}	Input low level voltage ¹⁾			V _{ss} - 0.3		$0.3 \mathrm{xV}_{\mathrm{DD}}$	V
V _{IH}	Input high level voltage ¹⁾			$0.7 \mathrm{xV}_{\mathrm{DD}}$		V _{DD} + 0.3	v
V _{hys}	Schmitt trigger voltage hysteresis 1)				400		mV
I _{INJ(PIN)} 2)	Injected current on Flash device pins PB0 and PB1					+4	
. ,	Injected Current on other I/O pins					±4	mA
$\Sigma I_{\rm INJ(PIN)}^{2)}$	Total injected current (sum of all I/O and control pins)	V _{DD} =5V				±25	
١L	Input leakage current	V _{SS} ≤V _{IN} ≤V _{DD}				±1	
۱ _S	Static current consumption	Floating inpu	ut mode ³⁾		400		μΛ
D	Weak pull-up equivalent resistor ⁴⁾	V _{IN} =V _{SS}	V _{DD} =5V	50	85	250	kΩ
npU			V _{DD} =3V	170 ¹⁾	190	230 ¹⁾	
C _{IO}	I/O pin capacitance				5		pF
t _{f(IO)out}	Output high to low level fall time 1)	C _L =50pF Between 10% and 90%			25		20
t _{r(IO)out}	Output low to high level rise time ¹⁾				25		115
t _{w(IT)in}	External interrupt pulse time ⁵⁾			1			t _{CPU}

Notes:

1. Data based on characterization results, not tested in production.

2. $I_{INJ(PIN)}$ must never be exceeded. This is implicitly insured if V_{IN} maximum is respected. If V_{IN} maximum cannot be respected, the injection current must be limited externally to the $I_{INJ(PIN)}$ value. A positive injection is induced by $V_{IN} > V_{DD}$ while a negative injection is induced by $V_{IN} < V_{SS}$. For true open-drain pads, there is no positive injection current, and the corresponding V_{IN} maximum must always be respected.

Caution: Negative current injection not allowed on Flash device pins PB0 and PB1.

3. Configuration not recommended, all unused pins must be kept at a fixed voltage: using the output mode of the I/O for example and leaving the I/O unconnected on the board or an external pull-up or pull-down resistor (see Figure 76). Data based on design simulation and/or technology characteristics, not tested in production.

4. The R_{PU} pull-up equivalent resistor is based on a resistive transistor (corresponding I_{PU} current characteristics described in Figure 77).

5. To generate an external interrupt, a minimum pulse width has to be applied on an I/O port pin configured as an external interrupt source.

Figure 76. Two typical Applications with unused I/O Pin configured as input



Note: I/O can be left unconnected if it is configured as output (0 or 1) by the software. This has the advantage of greater EMC robustness and lower cost.

bit) accuracy does not meet the accuracy specified in the data sheet.

Workaround

In order to have the accuracy specified in the datasheet, the first conversion after a ADC switch-on has to be ignored.

16.2.11 Negative injection impact on ADC accuracy

Injecting a negative current on an analog input pins significantly reduces the accuracy of the AD Converter. Whenever necessary, the negative injection should be prevented by the addition of a Schottky diode between the concerned I/Os and ground. Injecting a negative current on digital input pins degrades ADC accuracy especially if performed on a pin close to ADC channel in use.

16.2.12 ADC conversion spurious results

Spurious conversions occur with a rate lower than 50 per million. Such conversions happen when the measured voltage is just between 2 consecutive digital values.

Workaround

A software filter should be implemented to remove erratic conversion results whenever they may cause unwanted consequences.