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Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

Details

Details	
Product Status	Obsolete
Core Processor	Z8
Core Size	8-Bit
Speed	10MHz
Connectivity	-
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	14
Program Memory Size	1KB (1K x 8)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	64 x 8
Voltage - Supply (Vcc/Vdd)	4.5V ~ 5.5V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	18-SOIC (0.295", 7.50mm Width)
Supplier Device Package	-
Purchase URL	https://www.e-xfl.com/product-detail/zilog/z8pe003sz010eg

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong



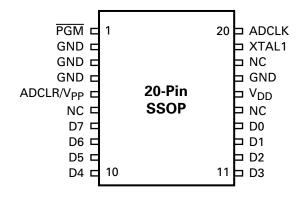


Figure 6. 20-Pin SSOP Pin Identification/EPROM Programming Mode

Pin # Symbol		Function	Direction
1	PGM	Program Mode	Input
2–4	GND	Ground	
5	ADCLR/V _{PP}	Clear Clock/Program Voltage	Input
6	NC	No Connection	
7–10	D7–D4	Data 7,6,5,4	Input/Output
11–14	D3-D0	Data 3,2,1,0	Input/Output
15	NC	No Connection	
16	V _{DD}	Power Supply	
17	GND	Ground	
18	NC	No Connection	
19	XTAL1	1-MHz Clock	Input
20	ADCLK	Address Clock	Input

Table 4. EPROM Programming Mode

STANDARD TEST CONDITIONS

The characteristics listed below apply for standard test conditions as noted. All voltages are referenced to Ground. Positive current flows into the referenced pin (Figure 7).

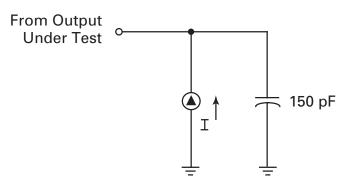


Figure 7. Test Load Diagram

CAPACITANCE

 T_{A} = 25°C, V_{CC} = GND = 0V, f = 1.0 MHz, unmeasured pins returned to GND.

Parameter	Min	Max
Input capacitance	0	12 pF
Output capacitance	0	12 pF
I/O capacitance	0	12 pF

T _A = 0°C to +70°C Standard Temperatures								
Sym	Parameter	V _{CC} ¹	Min	Max	Typical ² @ 25°C		Conditions	Notes
I _{CC}	Supply Current	3.0V		2.5	2.0	mA	@ 10 MHz	5,6
		5.5V		6.0	3.5	mA	@ 10 MHz	5,6
I _{CC1}	Standby Current	3.0V		2.0	1.0	mA	HALT mode V _{IN} = 0V, V _{CC} @ 10 MHz	5,6
		5.5V		4.0	2.5	mA	HALT mode V _{IN} = 0V, V _{CC} @ 10 MHz	5,6
I _{CC2}	Standby Current			500	150	nA	STOP mode V _{IN} = 0V, V _{CC}	7

Table 5. DC Electrical Characteristics (Continued)

Notes:

1. The V_{CC} voltage specification of 3.0V guarantees 3.0V; the V_{CC} voltage specification of 5.5V guarantees 5.0V \pm 0.5V. 2. Typical values are measured at V_{CC} = 3.3V and V_{CC} = 5.0V; V_{SS} = 0V = GND. 3. For the analog comparator input when the analog comparator is enabled.

4. No protection diode is provided from the pin to V_{CC}. External protection is recommended.

5. All outputs are unloaded and all inputs are at the V_{CC} or V_{SS} level.

6. CL1 = CL2 = 22 pF.

7. Same as note 5, except inputs are at V_{CC} .

		T _A = −40°C to +105°C Extended Temperatures			Typical ²			
Sym	Parameter	V _{CC} ¹	Min	Max	@ 25°C	Units	Conditions	Notes
I _{CC1}	Standby Current	4.5V		2.0	1.0	mA	HALT mode V _{IN} = 0V, V _{CC} @ 10 MHz	5,6
		5.5V		2.0	1.0	mA	HALT mode V _{IN} = 0V, V _{CC} @ 10 MHz	5,6
I _{CC2}	Standby Current	4.5V		700	250	nA	STOP mode V _{IN} = 0V,V _{CC}	7
_		5.5V		700	250	nA	STOP mode V _{IN} = 0V,V _{CC}	7

Table 6 DC Electrical Characteristics (Continued)

Notes:

1. The V_{CC} voltage specification of 4.5V and 5.5V guarantees 5.0V \pm 0.5V. 2. Typical values are measured at V_{CC} = 5.0V; V_{SS} = 0V = GND. 3. For analog comparator input when analog comparator is enabled.

4. No protection diode is provided from the pin to V_{CC} . External protection is recommended.

5. All outputs are unloaded and all inputs are at V_{CC} or V_{SS} level.

6. CL1 = CL2 = 22 pF.

7. Same as note 5, except inputs are at V_{CC} .

AC ELECTRICAL CHARACTERISTICS

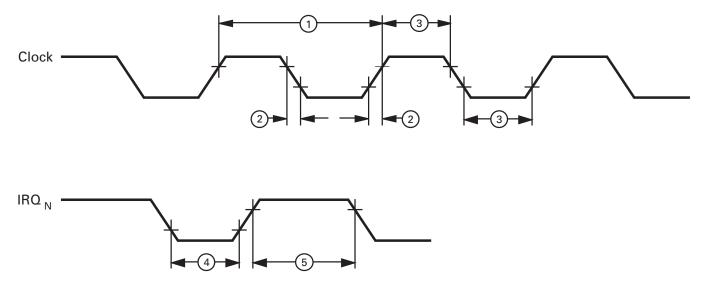


Figure 8. AC Electrical Timing Diagram

Table 7. Additional Timing

No			T _A = 0°C to +70°C T _A = -40°C to +105°C @ 10 MHz					
	Symbol	Parameter	V _{CC} ¹	Min	Max	Units	Notes	
1	T _P C	Input Clock Period	3.0V	100	DC	ns	2	
		-	5.5V	100	DC	ns	2	
2	T _R C,T _F C	Clock Input Rise and Fall Times	3.0V		15	ns	2	
		-	5.5V		15	ns	2	
3	T _W C	Input Clock Width	3.0V	50		ns	2	
		-	5.5V	50		ns	2	
4	T _W IL	Int. Request Input Low Time	3.0V	70		ns	2	
			5.5V	70		ns	2	
5	T _W IH	Int. Request Input High Time	3.0V	5TpC			2	
		-	5.5V	5TpC			2	
6	T _{WSM}	STOP mode Recovery Width	3.0V	25		ns		
		Spec.	5.5V	25		ns		
7	T _{OST}	Oscillator Start-Up Time	3.0V		5TpC			
		-	5.5V		5TpC			
8	T _{POR}	Power-On Reset Time	3.0V	128 T _P C + T _{OST}				
		-	5.5V					

Notes:

1. The V_{DD} voltage specification of 3.0V guarantees 3.0V. The V_{DD} voltage specification of 5.5V guarantees 5.0V \pm 0.5V. 2. Timing Reference uses 0.7 V_{CC} for a logical 1 and 0.2 V_{CC} for a logical 0.

Note: The WDT can only be disabled via software if the first instruction out of the $\overline{\text{RESET}}$ performs this function. Logic within the device detects that it is in the process of executing the first instruction after the processor leaves $\overline{\text{RE-SET}}$. During the execution of this instruction, the upper five bits of the TCTLHI register can be written. After this first instruction, hardware does not allow the upper five bits of this register to be written.

The TCTLHI bits for control of the WDT are described below:

WDT Time Select (D6, D5, D4). Bits 6, 5, and 4 determine the time-out period. Table 13 indicates the range of time-out values that can be obtained. The default values of D6, D5, and D4 are 001, which sets the WDT to its minimum time-out period when coming out of RESET.

WDT During HALT (D7). This bit determines whether or not the WDT is active during HALT mode. A 1 indicates active during HALT mode. A 0 prevents the WDT from resetting the part while halted. Coming out of RESET, the WDT is enabled during HALT mode. **STOP MODE (D3).** Coming out of $\overline{\text{RESET}}$, the device STOP mode is disabled. If an application requires use of STOP mode, bit D3 must be cleared immediately at leaving RESET. If bit D3 is set, the STOP instruction executes as a NOP. If bit D3 is cleared, the STOP instruction enters STOP mode.

Bits 2, 1 and 0. These bits are reserved and must be 0.

D6

0

0

0

0

1

1

1

1

1

1

0

1

;	D5	D4	Crystal Clocks* to Timeout	Time-Out Using a 10-MHz Crystal
	0	0	Disabled	Disabled
	0	1	65,536 TpC	6.55 ms
	1	0	131,072 TpC	13.11 ms
	1	1	262,144 TpC	26.21 ms
	0	0	524,288 TpC	52.43 ms
	0	1	1,048,576 TpC	104.86 ms

Table 13. WDT Time-Out

Note: *TpC is an XTAL clock cycle. The default at reset is 001.

209.72 ms

838.86 ms

2,097,152 TpC

8,388,608 TpC

POWER-DOWN MODES

In addition to the standard RUN mode, the Z8Plus MCU supports two Power-Down modes to minimize device cur-

rent consumption. The two modes supported are HALT and STOP.

HALT MODE OPERATION

The HALT mode suspends instruction execution and turns off the internal CPU clock. The on-chip oscillator circuit remains active so the internal clock continues to run and is applied to the timers and interrupt logic.

To enter HALT mode, the device only requires a HALT instruction. It is *not* necessary to execute a NOP instruction immediately before the HALT instruction.

7F HALT ; enter HALT mode

HALT mode can be exited by servicing an external or internal interrupt. The first instruction executed is the interrupt service routine. At completion of the interrupt service routine, the user program continues from the instruction after the HALT instruction.

The HALT mode can also be exited via a **RESET** activation or a Watch-Dog Timer (WDT) time-out. In these cases, program execution restarts at 0020H, the reset restart address. The STOP mode provides the lowest possible device standby current. This instruction turns off the on-chip oscillator and internal system clock.

To enter the STOP mode, the Z8Plus only requires a STOP instruction. It is *not* necessary to execute a NOP instruction immediately before the STOP instruction.

6F STOP ;enter STOP mode

The STOP mode is exited by any one of the following resets: POR or a Stop-Mode Recovery source. At reset generation, the processor always restarts the application program at address 0020H, and the STOP mode flag is set. Reading the STOP mode flag does not clear it. The user must clear the STOP mode flag with software.

Note: Failure to clear the STOP mode flag can result in undefined behavior.

The Z8Plus provides a dedicated Stop-Mode Recovery (SMR) circuit. In this case, a low-level applied to input pin

(SMR) circuit. In this case, a low-level applied to input pin PB0 (I/O Port B, bit 0) triggers an SMR. To use this mode, pin PB0 must be configured as an input and the special function selected before the STOP mode is entered. The Low level on PB0 must be held for a minimum pulse width T_{WSM} . Program execution starts at address 20h, after the POR delay.

Notes: 1. The PB0 input, when used for Stop-Mode Recovery, does not initialize the control registers.

The STOP mode current (I_{CC2}) is minimized when:

- V_{CC} is at the low end of the device's operating range
- Output current sourcing is minimized
- All inputs (digital and analog) are at the Low or High rail voltages
- 2. For detailed information about flag settings, see the <u>Z8Plus User's Manual</u>.

OSCILLATOR OPERATION

The Z8Plus MCU uses a Pierce oscillator with an internal feedback resistor (Figure 14). The advantages of this circuit are low-cost, large output signal, low-power level in the crystal, stability with respect to V_{CC} and temperature, and low impedances (not disturbed by stray effects).

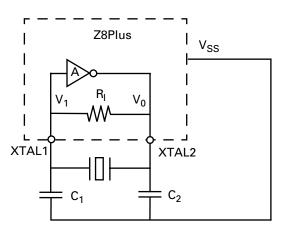


Figure 14. Pierce Oscillator with Internal Feedback Circuit

One drawback to the Pierce oscillator is the requirement for high gain in the amplifier to compensate for feedback path losses. The oscillator amplifies its own noise at start-up until it settles at the frequency that satisfies the gain/phase requirements. A \times B = 1; where A = VO/VI is the gain of the amplifier, and B = VI/VO is the gain of the feedback element. The total phase shift around the loop is forced to 0 (360 degrees). V_{IN} must be in phase with itself; therefore, the amplifier/inverter provides a 180-degree phase shift, and the feedback element is forced to provide the other 180-degree phase shift.

R1 is a resistive component placed from output to input of the amplifier. The purpose of this feedback is to bias the amplifier in its linear region and provide the start-up transition.

Capacitor C2, combined with the amplifier output resistance, provides a small phase shift. It also provides some attenuation of overtones.

Capacitor C_1 , combined with the crystal resistance, provides an additional phase shift.

Start-up time may be affected if C_1 and C_2 are increased dramatically in size. As C_1 and C_2 increase, the start-up time increases until the oscillator reaches a point where it ceases to operate.

For fast and reliable oscillator start-up over the manufacturing process range, the load capacitors should be sized as low as possible without resulting in overtone operation.

Layout

Traces connecting crystal, caps, and the Z8Plus oscillator pins should be as short and wide as possible to reduce parasitic inductance and resistance. The components (caps, the crystal, and resistors) should be placed as close as possible to the oscillator pins of the Z8Plus.

The traces from the oscillator pins of the integrated circuit (IC) and the ground side of the lead caps should be guarded from all other traces (clock, V_{CC} , address/data lines, and system ground) to reduce cross talk and noise injection. Guarding is usually accomplished by keeping other traces and system ground trace planes away from the oscillator circuit, and by placing a Z8Plus device V_{SS} ground ring around the traces/components. The ground side of the oscillator lead caps should be connected to a single trace to the Z8Plus device V_{SS} (GND) pin. It should not be shared with any other system-ground trace or components except at the Z8Plus device V_{SS} pin. The objective is to prevent differential system ground noise injection into the oscillator (Figure 15).

Indications of an Unreliable Design

There are two major indicators that are used in working designs to determine their reliability over full lot and temperature variations. They are:

Start-Up Time. If start-up time is excessive, or varies widely from unit to unit, there is probably a gain problem. To fix the problem, the C_1 and C_2 capacitors require reduction. The amplifier gain is either not adequate at frequency, or the crystal R's are too large.

Output Level. The signal at the amplifier output should swing from ground to V_{CC} to indicate adequate gain in the amplifier. As the oscillator starts up, the signal amplitude grows until clipping occurs. At that point, the loop gain is effectively reduced to unity, and constant oscillation is achieved. A signal of less than 2.5 volts peak-to-peak is an indication that low gain can be a problem. Either C₁ or C₂ should be made smaller, or a low-resistance crystal should be used.

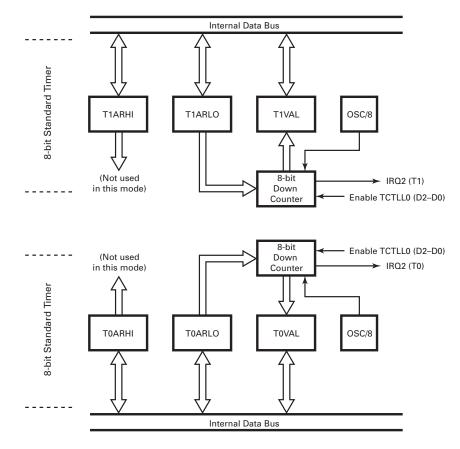


Figure 20. 8-Bit Standard Timers

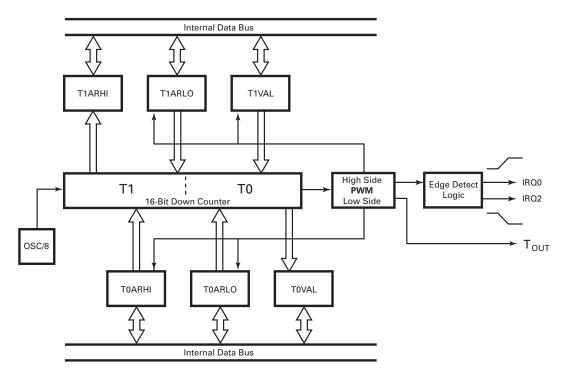
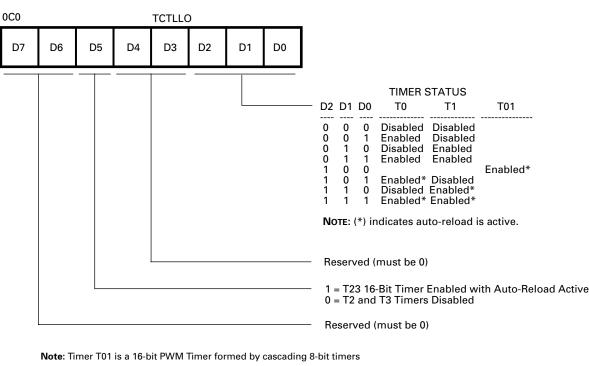


Figure 21. 16-Bit Standard PWM Timer



T1 (MSB) and T0 (LSB). T23 is a standard 16-bit timer formed by cascading 8-bit timers T3 (MSB) and T2 (LSB).



A pair of READ/WRITE registers is utilized for each 8-bit timer. One register is defined to contain the auto-initialization value for the timer. The second register contains the current value for the timer. When a timer is enabled, the timer decrements the value in its count register and continues decrementing until it reaches 0. An interrupt is generated, and the contents of the auto-initialization register are optionally copied into the count value register. If auto-initialization is not enabled, the timer stops counting when the value reaches 0. Control logic clears the appropriate control register bit to disable the timer. This operation is referred to as a *single-shot*. If auto-initialization is enabled, the timer counts from the initialization value. Software must not attempt to use timer registers for any other function.

User software is allowed to write to any WRITE register at any time; however, care should be taken if timer registers are updated while the timer is enabled. If software changes the count value while the timer is in operation, the timer continues counting from the updated value. **Note:** Unpredictable behavior can occur if the value updates at the same time that the timer reaches 0.

Similarly, if user software changes the initialization value register while the timer is active, the next time that the timer reaches 0, the timer initializes to the changed value.

Note: Unpredictable behavior can occur if the initialization value register is changed while the timer is in the process of being initialized.

The initialization value is determined by the exact timing of the WRITE operation. In all cases, the Z8Plus assigns a higher priority to the software WRITE than to a decrementer write-back. However, when hardware clears a control register bit for a timer that is configured for single-shot operation, the clearing of the control bit overrides a software WRITE. A READ of either register can be conducted at any time, with no effect on the functionality of the timer. If a timer pair is defined to operate as a single 16-bit entity, the entire 16-bit value must reach 0 before an interrupt is generated. In this case, a single interrupt is generated, and the interrupt corresponds to the even 8-bit timer.

Example: Timers T2 and T3 are cascaded to form a single 16bit timer. The interrupt for the combined timer is defined to be generated by timer T2 rather than T3. When a timer pair is specified to act as a single 16bit timer, the even timer registers in the pair (timer T0 or T2) is defined to hold the timer's least significant byte. In contrast, the odd timer in the pair holds the timer's most significant byte.

In parallel with the posting of the interrupt request, the interrupting timer's count value is initialized by copying the contents of the auto-initialization value register to the count value register.

Note: Any time that a timer pair is defined to act as a single 16bit timer, the auto-reload function is performed automatically.

All 16-bit timers continue counting while their interrupt requests are active and operate independently of each other.

If interrupts are disabled for a long period of time, it is possible for the timer to decrement to 0 again before its initial interrupt is responded to. This condition is termed a degenerate case, and hardware is not required to detect it.

When the timer control register is written, all timers that are enabled by the WRITE begin counting from the value in the count register. In this case, an auto-initialization is not performed. All timers can receive an internal clock source input only. Each enabled timer is updated every 8th XTAL clock cycle.

If T0 and T1 are defined to work independently, then each works as an 8-bit timer with a single auto-initialization register (T0ARLO for T0, and T1ARLO for T1). Each timer asserts its predefined interrupt when it times out, optionally performing the auto-initialization function. If T0 and T1 are cascaded to form a single 16-bit timer, then the single 16bit timer is capable of performing as a Pulse-Width Modulator (PWM). This timer is referred to as T01 to distinguish it as having special functionality that is not available when T0 and T1 act independently.

When **T01** is enabled, it can use a pair of 16-bit auto-initialization registers. In this mode, one 16-bit auto-initialization value is composed of the concatenation of T1ARLO and T0ARLO. The second auto-initialization value is composed of the concatenation of T1ARHI and T0ARHI. When T01 times out, it alternately initializes its count value using the Low auto-init pair, followed by the High auto-init pair. This functionality corresponds to a PWM. That is, the T1 interrupt defines the end of the High section of the waveform, and the T0 interrupt marks the end of the Low portion of the PWM waveform.

The PWM begins counting with whatever data is held in the count registers. After this value expires, the first reload depends on the state of the PB1 pin if T_{OUT} mode is selected. Otherwise, the Low value is applied first.

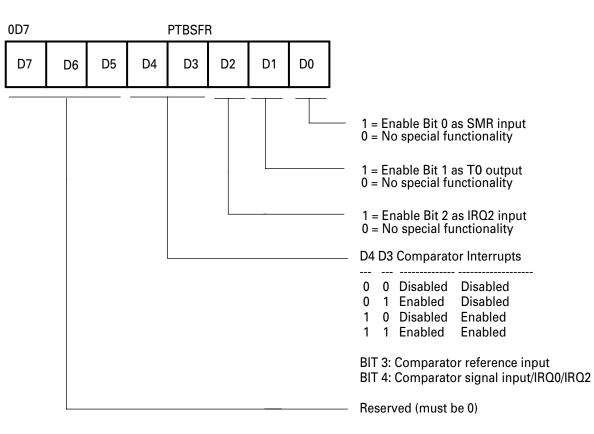
After the auto-initialization is completed, decrementing occurs for the number of counts defined by the PWM_LO registers. When decrementing again reaches 0, the T0 interrupt is asserted; and auto-init using the PWM_HI registers occurs. Decrementing occurs for the number of counts defined by the PWM_HI registers until reaching 0. From there, the T1 interrupt IRQ2 is asserted, and the cycle begins again.

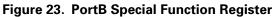
The internal timers can be used to trigger external events by toggling the PB1 output when generating an interrupt. This functionality can only be achieved in conjunction with the port unit defining the appropriate pin as an output signal with the timer output special function enabled. In this mode, the port output is toggled when the timer count reaches 0, and continues toggling each time that the timer times out.

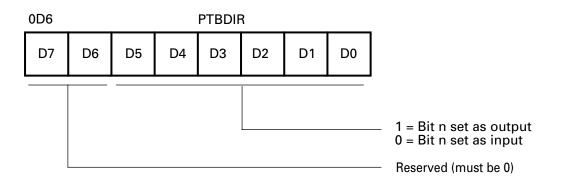
T_{OUT} Mode

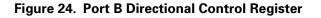
The PortB special function register PTBSFR (0D7H; Figure 23) is used in conjunction with the Port B directional control register PTBDIR (0D6; Figure 24) to configure PB1 for T_{OUT} operation for T0. In order for T_{OUT} to function, PB1 must be defined as an output line by setting PTBDIR bit 1 to 1. Configured in this way, PB1 is capable of being a clock output for T0, toggling the PB1 output pin on each T0 timeout.

At end-of-count, the interrupt request line (IRQ0), clocks a toggle flip-flop. The output of this flip-flop drives the T_{OUT} line, PB1. In all cases, when T0 reaches its end-of-count, T_{OUT} toggles to its opposite state (Figure 25). If, for example, T0 is in Continuous Counting Mode, T_{OUT} exhibits a 50-percent duty cycle output. If the timer pair is selected (T01) as a PWM, the duty cycle depends on the High and Low reload values. At the end of each High time, PB1 toggles Low. At the end of each Low time, PB1 toggles HI.









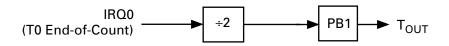


Figure 25. Timer T0 Output Through T_{OUT}

For port bits configured as an output by means of the directional control register, the value held in the corresponding bit of the Output Value Register is driven directly onto

READ/WRITE OPERATIONS

The control for each port is done on a bit-by-bit basis. All bits are capable of operating as inputs or outputs, depending on the setting of the port's directional control register. If configured as an input, each bit is provided a Schmitt-trigger. The output of the Schmitt-trigger is latched twice to perform a synchronization function, and the output of the synchronizer is fed to the port input register, which can be read by software.

A WRITE to a port input register carries the effect of updating the contents of the input register, but subsequent READs do not necessarily return the same value that was written. If the bit in question is defined as an input, the input register for that bit position contains the current synchronized input value. WRITEs to that bit position are overwritten on the next clock cycle with the newly sampled input data. However, if the particular bit is programmed as an output, the input register for that bit retains the software-updated value. The port bits that are programmed as outputs do not sample the value being driven out.

Any bit in either port can be defined as an output by setting the appropriate bit in the directional control register. In this instance, the value held in the appropriate bit of the port output register is driven directly onto the output pin.

Note: The preceding result does not necessarily reflect the actual output value. If an external error is holding an output pin either High or Low against the output driver, the software READ returns the *requested* value, not the actual state caused by the contention. When a bit is defined as an output, the Schmitt-trigger on the input is disabled to save power.

the output pin. The opposite register bit for a given pin (the output register bit for an input pin and the input register bit for an output pin) holds their previous value. These bits are not changed and do not exhibit any effect on the hardware.

Updates to the output register take effect based on the timing of the internal instruction pipeline; however, this timing is referenced to the rising edge of the clock. The output register can be read at any time, and returns the current output value that is held. No restrictions are placed on the timing of READs and/or WRITEs to any of the port registers with respect to the others.

Note: Care should be taken when updating the directional control and special function registers.

When updating a directional control register, the special function register (SFR) should first be disabled. If this precaution is not taken, unpredicted events could occur as a result of the change in the port I/O status. This precaution is especially important when defining changes in Port B, as the unpredicted event referred to above could be one or more interrupts. Clearing of the SFR register should be the first step in configuring the port, while setting the SFR register should be the final step in the port configuration process. To ensure unpredictable results, the SFR register should not be written until the pins are being driven appropriately, and all initialization is completed.

PORT A

Port A is a general-purpose port. Figure 27 features a block diagram of Port A. Each of its lines can be independently programmed as input or output via the Port A directional control register (PTADIR at 0D2H) as seen in Figure 26. A bit set to a 1 in PTADIR configures the corresponding bit in Port A as an output, while a bit cleared to 0 configures the corresponding bit in Port A as an input.

The input buffers are Schmitt-triggered. Bits programmed as outputs can be individually programmed as either push-

pull or open-drain by setting the corresponding bit in the special function register (PTASFR, Figure 26).

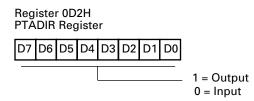


Figure 27. Port A Directional Control Register

PORT B

Port B Description

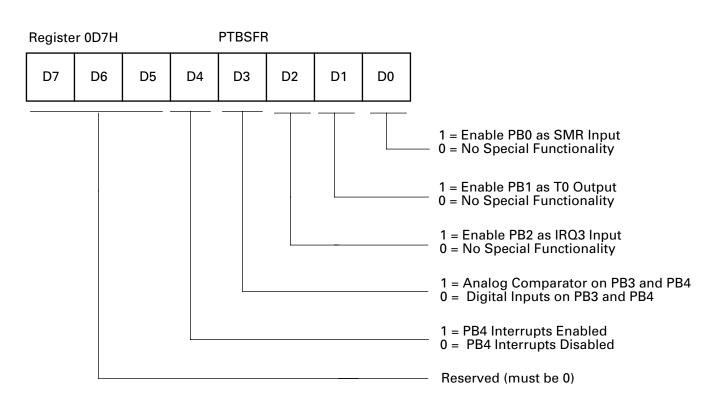
Port B is a 6-bit (bidirectional), CMOS-compatible I/O port. These six I/O lines can be configured under software control to be an input or output. Each bit is configured independently from the other bits. That is, one bit may be set to INPUT while another bit is set to OUTPUT.

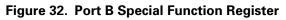
In addition to standard input/output capability, five pins of Port B provide special functionality as indicated in Table 15.

Special functionality is invoked via the Port B special function register. Port B, bit 5, is an open-drain-only pin when in output mode. There is no high-side driver on the output stage, nor is there any high-side protection device, because PB5 acts as the V_{PP} pin for EPROM programming mode. The user should always place an external protection diode on this pin. See Figure 32.

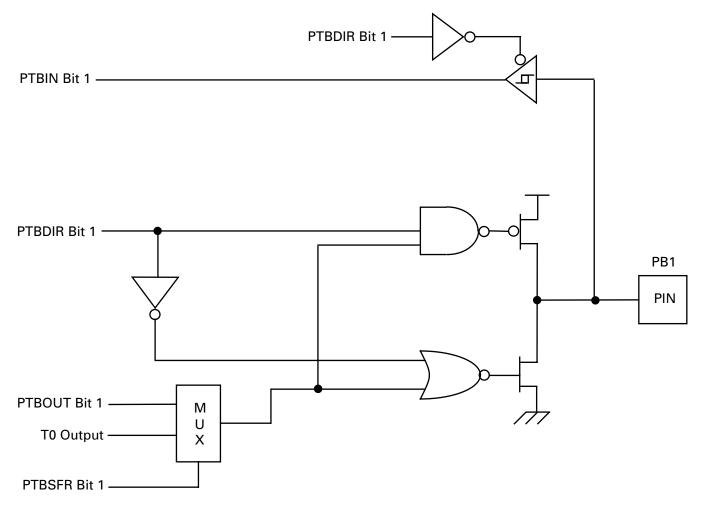
Port Pin	Input Special Function	Output Special Function
PB0	Stop Mode Recovery Input	None
PB1	None	T0 Output
PB2	IRQ3	None
PB3	Comparator Reference Input	None
PB4	Comparator Signal Input/IRQ1/IRQ4	None

Table	15.	Port I	BS	Special	Functions
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PORT B—PIN 1 CONFIGURATION





PORT B—PIN 2 CONFIGURATION

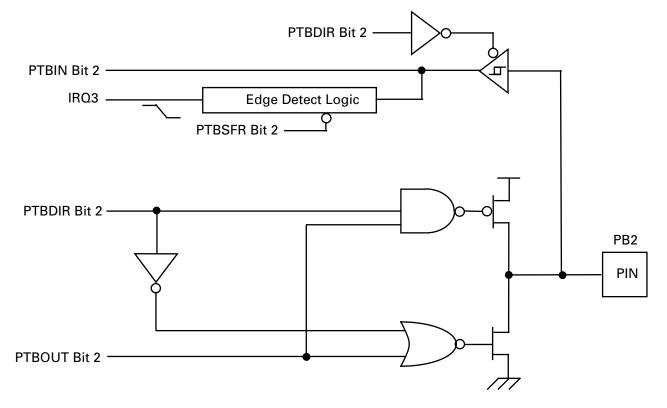


Figure 36. Port B Pin 2 Diagram

PORT B—PINS 3 AND 4 CONFIGURATION

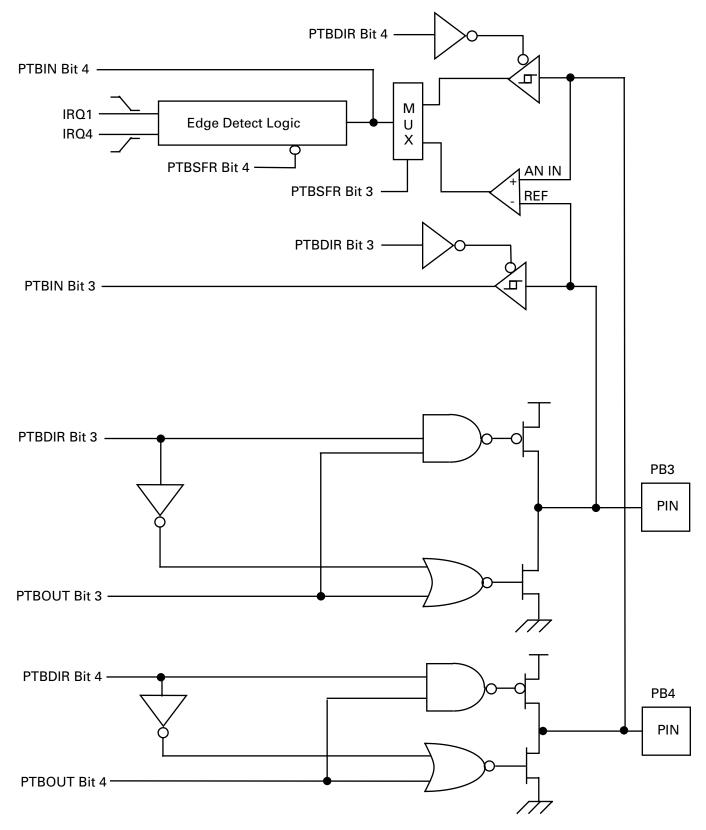
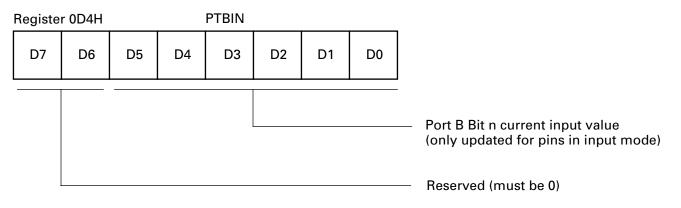
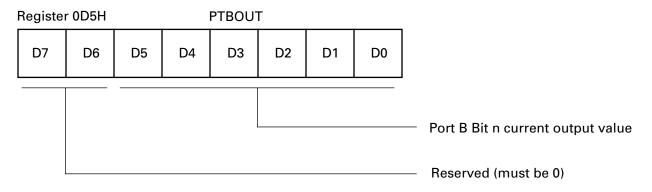


Figure 37. Port B Pins 3 and 4 Diagram

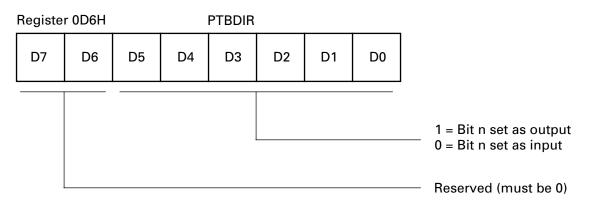
PORT B CONTROL REGISTERS













I/O PORT RESET CONDITIONS

Full Reset

Port A and Port B output value registers are not affected by RESET.

On **RESET**, the Port A and Port B directional control registers are cleared to all zeros, which defines all pins in both ports as inputs.

On **RESET**, the directional control registers redefine all pins as inputs, and the Port A and Port B input value registers

overwrites the previously held data with the current sample of the input pins.

On **RESET**, the Port A and Port B special function registers are cleared to 00h, which deactivates all port special functions.

Note: The SMR and WDT time-out events are *not* full device resets. The port control registers are not affected by either of these events.

ANALOG COMPARATOR

The device includes one on-chip analog comparator. Pin PB4 features a comparator front end. The comparator reference voltage is on pin PB3.

Comparator Description

The on-chip comparator can process an analog signal on PB4 with reference to the voltage on PB3. The analog function is enabled by programming the Port B special function register bits 3 and 4.

When the analog comparator function is enabled, bit 4 of the input register is defined as holding the synchronized output of the comparator, while bit 3 retains a synchronized sample of the reference input.

If the interrupts for PB4 are enabled when the comparator special function is selected, the output of the comparator generates interrupts.

COMPARATOR OPERATION

The comparator output reflects the relationship between the analog input to the reference input. If the voltage on the analog input is higher than the voltage on the reference input, then the comparator output is at a High state. If the voltage on the analog input is lower than the voltage on the reference input, then the analog output is at a Low state.

Comparator Definitions

VICR

The usable voltage range for the positive input and reference input is called the Comparator Input Common Mode Voltage Range (V_{ICR}).

Note: The comparator is not guaranteed to work if the input is outside of the V_{ICR} range.

VOFFSET

The absolute value of the voltage between the positive input and the reference input required to make the comparator output voltage switch is the Comparator Input Offset Voltage (V_{OFFSET}).

Ι_{ΙΟ}

For the CMOS voltage comparator input, the input offset current (I_{10}) is the leakage current of the CMOS input gate.

HALT Mode

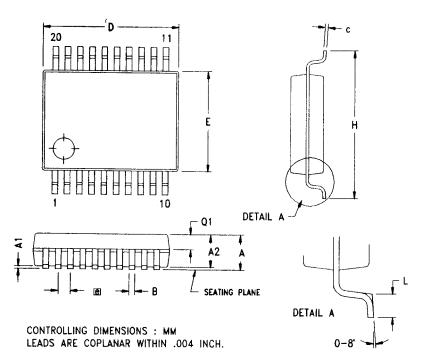
The analog comparator is functional during HALT mode. If the interrupts are enabled, an interrupt generated by the comparator causes a return from HALT mode.

STOP Mode

The analog comparator is disabled during STOP mode. The comparator is powered down to prevent it from drawing any current.

Low Voltage Protection. An on-board Voltage Comparator checks that the V_{CC} is at the required level to ensure correct operation of the device. A reset is globally driven if V_{CC} is below the specified voltage (Low Voltage Protection).

The device functions normally at or above 3.0V under all conditions, and is guaranteed to function normally at supply voltages above the Low Voltage Protection trip point. Below 3.0V, the device functions normally until the Low Volt-



SYMBOL		MILLIMETER		INCH			
	MIN	NOM	MAX	MIN	NOM	MAX	
A	1.73	1.85	1.98	0.068	0.073	0.078	
A1	0.05	0.13	0.21	0.002	0.005	0.008	
A2	1.68	1.73	1.83	0.066	0.068	0.072	
8	0.25	0.30	0.38	0.010	0.012	0.015	
С	0.13	0.15	0.22	0.005	0.006	0.009	
D	7.07	7.20	7.33	0.278	0.283	0.289	
E	5.20	5.30	5.38	0.205	0.209	0.212	
(e)		0.65 TYP			0.0256 TY	0	
Н	7.65	7.80	7.90	0.301	0.307	0.311	
L	0.56	0.75	0.94	0.022	0.030	0.037	
Q1	0.74	0.78	0.82	0.029	0.031	0.032	

Figure 47. 20-Pin SSOP Package Diagram