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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Obsolete
Number of LABs/CLBs	2125
Number of Logic Elements/Cells	17000
Total RAM Bits	282624
Number of I/O	201
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FTBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfxp2-17e-7ft256c

Introduction

LatticeXP2 devices combine a Look-up Table (LUT) based FPGA fabric with non-volatile Flash cells in an architecture referred to as flexiFLASH.

The flexiFLASH approach provides benefits including instant-on, infinite reconfigurability, on chip storage with FlashBAK embedded block memory and Serial TAG memory and design security. The parts also support Live Update technology with TransFR, 128-bit AES Encryption and Dual-boot technologies.

The LatticeXP2 FPGA fabric was optimized for the new technology from the outset with high performance and low cost in mind. LatticeXP2 devices include LUT-based logic, distributed and embedded memory, Phase Locked Loops (PLLs), pre-engineered source synchronous I/O support and enhanced sysDSP blocks.

Lattice Diamond[®] design software allows large and complex designs to be efficiently implemented using the LatticeXP2 family of FPGA devices. Synthesis library support for LatticeXP2 is available for popular logic synthesis tools. The Diamond software uses the synthesis tool output along with the constraints from its floor planning tools to place and route the design in the LatticeXP2 device. The Diamond tool extracts the timing from the routing and back-annotates it into the design for timing verification.

Lattice provides many pre-designed Intellectual Property (IP) LatticeCORE[™] modules for the LatticeXP2 family. By using these IPs as standardized blocks, designers are free to concentrate on the unique aspects of their design, increasing their productivity.

Modes of Operation

Each slice has up to four potential modes of operation: Logic, Ripple, RAM and ROM.

Logic Mode

In this mode, the LUTs in each slice are configured as LUT4s. A LUT4 has 16 possible input combinations. Four-input logic functions are generated by programming the LUT4. Since there are two LUT4s per slice, a LUT5 can be constructed within one slice. Larger LUTs such as LUT6, LUT7 and LUT8, can be constructed by concatenating two or more slices. Note that a LUT8 requires more than four slices.

Ripple Mode

Ripple mode allows efficient implementation of small arithmetic functions. In ripple mode, the following functions can be implemented by each slice:

- Addition 2-bit
- Subtraction 2-bit
- Add/Subtract 2-bit using dynamic control
- Up counter 2-bit
- Down counter 2-bit
- Up/Down counter with async clear
- Up/Down counter with preload (sync)
- Ripple mode multiplier building block
- Multiplier support
- Comparator functions of A and B inputs
 - A greater-than-or-equal-to B
 - A not-equal-to B
 - A less-than-or-equal-to B

Two carry signals, FCI and FCO, are generated per slice in this mode, allowing fast arithmetic functions to be constructed by concatenating slices.

RAM Mode

In this mode, a 16x4-bit distributed Single Port RAM (SPR) can be constructed using each LUT block in Slice 0 and Slice 2 as a 16x1-bit memory. Slice 1 is used to provide memory address and control signals. A 16x2-bit Pseudo Dual Port RAM (PDPR) memory is created by using one slice as the read-write port and the other companion slice as the read-only port.

The Lattice design tools support the creation of a variety of different size memories. Where appropriate, the software will construct these using distributed memory primitives that represent the capabilities of the PFU. Table 2-3 shows the number of slices required to implement different distributed RAM primitives. For more information on using RAM in LatticeXP2 devices, please see TN1137, [LatticeXP2 Memory Usage Guide](#).

Table 2-3. Number of Slices Required For Implementing Distributed RAM

	SPR 16X4	PDPR 16X4
Number of slices	3	3

Note: SPR = Single Port RAM, PDPR = Pseudo Dual Port RAM

ROM Mode

ROM mode uses the LUT logic; hence, Slices 0 through 3 can be used in the ROM mode. Preloading is accomplished through the programming interface during PFU configuration.

Routing

There are many resources provided in the LatticeXP2 devices to route signals individually or as busses with related control signals. The routing resources consist of switching circuitry, buffers and metal interconnect (routing) segments.

The inter-PFU connections are made with x1 (spans two PFU), x2 (spans three PFU) or x6 (spans seven PFU) connections. The x1 and x2 connections provide fast and efficient connections in horizontal and vertical directions. The x2 and x6 resources are buffered to allow both short and long connections routing between PFUs.

The LatticeXP2 family has an enhanced routing architecture to produce a compact design. The Diamond design tool takes the output of the synthesis tool and places and routes the design. Generally, the place and route tool is completely automatic, although an interactive routing editor is available to optimize the design.

sysCLOCK Phase Locked Loops (PLL)

The sysCLOCK PLLs provide the ability to synthesize clock frequencies. The LatticeXP2 family supports between two and four full featured General Purpose PLLs (GPLL). The architecture of the GPLL is shown in Figure 2-4.

CLKI, the PLL reference frequency, is provided either from the pin or from routing; it feeds into the Input Clock Divider block. CLKFB, the feedback signal, is generated from CLKOP (the primary clock output) or from a user clock pin/logic. CLKFB feeds into the Feedback Divider and is used to multiply the reference frequency.

Both the input path and feedback signals enter the Voltage Controlled Oscillator (VCO) block. The phase and frequency of the VCO are determined from the input path and feedback signals. A LOCK signal is generated by the VCO to indicate that the VCO is locked with the input clock signal.

The output of the VCO feeds into the CLKOP Divider, a post-scalar divider. The duty cycle of the CLKOP Divider output can be fine tuned using the Duty Trim block, which creates the CLKOP signal. By allowing the VCO to operate at higher frequencies than CLKOP, the frequency range of the GPLL is expanded. The output of the CLKOP Divider is passed through the CLKOK Divider, a secondary clock divider, to generate lower frequencies for the CLKOK output. For applications that require even lower frequencies, the CLKOP signal is passed through a divide-by-three divider to produce the CLKOK2 output. The CLKOK2 output is provided for applications that use source synchronous logic. The Phase/Duty Cycle/Duty Trim block is used to adjust the phase and duty cycle of the CLKOP Divider output to generate the CLKOS signal. The phase/duty cycle setting can be pre-programmed or dynamically adjusted.

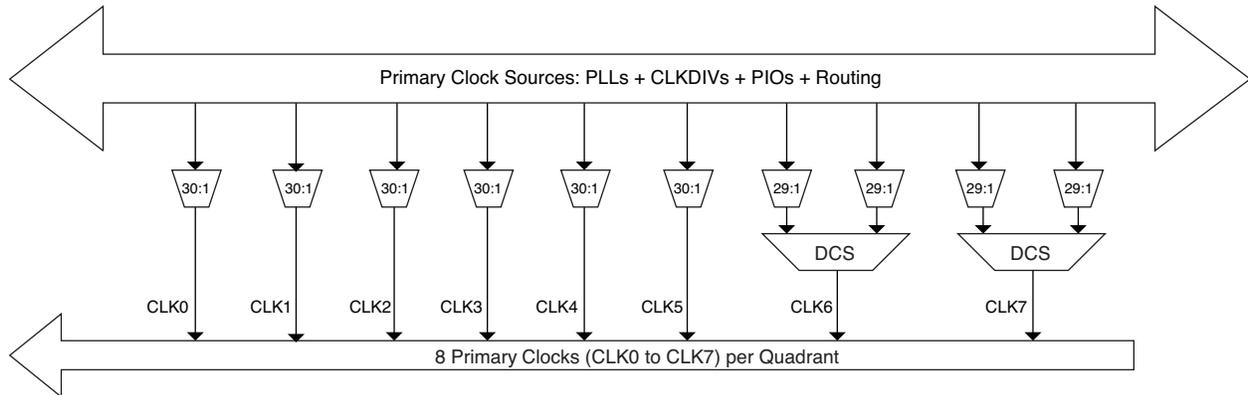
The clock outputs from the GPLL; CLKOP, CLKOK, CLKOK2 and CLKOS, are fed to the clock distribution network.

For further information on the GPLL please see TN1126, [LatticeXP2 sysCLOCK PLL Design and Usage Guide](#).

Primary Clock Routing

The clock routing structure in LatticeXP2 devices consists of a network of eight primary clock lines (CLK0 through CLK7) per quadrant. The primary clocks of each quadrant are generated from muxes located in the center of the device. All the clock sources are connected to these muxes. Figure 2-9 shows the clock routing for one quadrant. Each quadrant mux is identical. If desired, any clock can be routed globally.

Figure 2-9. Per Quadrant Primary Clock Selection

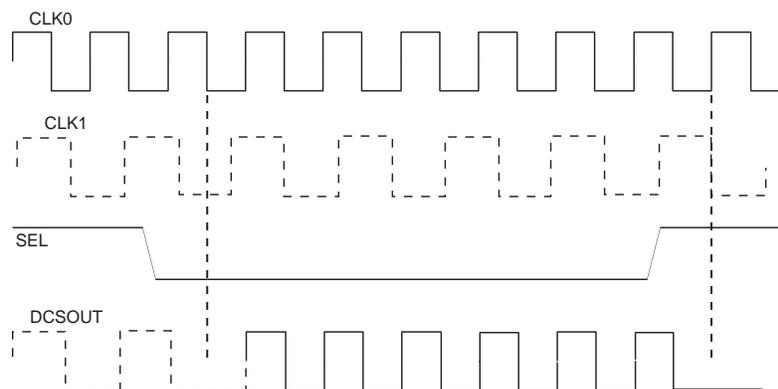


Dynamic Clock Select (DCS)

The DCS is a smart multiplexer function available in the primary clock routing. It switches between two independent input clock sources without any glitches or runt pulses. This is achieved irrespective of when the select signal is toggled. There are two DCS blocks per quadrant; in total, eight DCS blocks per device. The inputs to the DCS block come from the center muxes. The output of the DCS is connected to primary clocks CLK6 and CLK7 (see Figure 2-9).

Figure 2-10 shows the timing waveforms of the default DCS operating mode. The DCS block can be programmed to other modes. For more information on the DCS, please see TN1126, [LatticeXP2 sysCLOCK PLL Design and Usage Guide](#).

Figure 2-10. DCS Waveforms

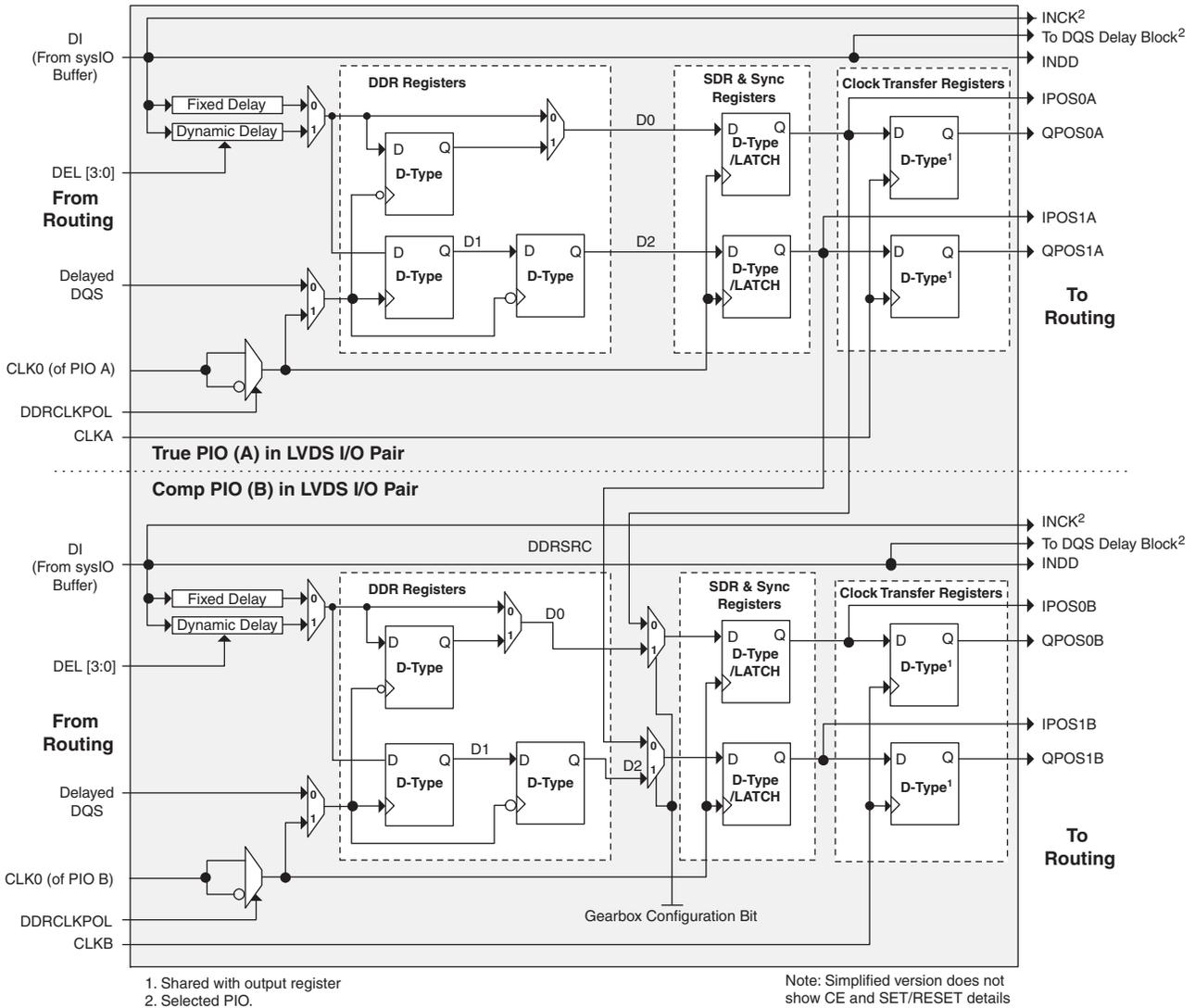


Secondary Clock/Control Routing

Secondary clocks in the LatticeXP2 devices are region-based resources. The benefit of region-based resources is the relatively low injection delay and skew within the region, as compared to primary clocks. EBR rows, DSP rows and a special vertical routing channel bound the secondary clock regions. This special vertical routing channel aligns with either the left edge of the center DSP block in the DSP row or the center of the DSP row. Figure 2-11 shows this special vertical routing channel and the eight secondary clock regions for the LatticeXP2-40.

The signal DDRCLKPOL controls the polarity of the clock used in the synchronization registers. It ensures adequate timing when data is transferred from the DQS to system clock domain. For further discussion on this topic, see the DDR Memory section of this data sheet.

Figure 2-26. Input Register Block



Output Register Block

The output register block provides the ability to register signals from the core of the device before they are passed to the sysIO buffers. The blocks on the PIOs on the left, right and bottom contain registers for SDR operation that are combined with an additional latch for DDR operation. Figure 2-27 shows the diagram of the Output Register Block for PIOs.

In SDR mode, ONEG0 feeds one of the flip-flops that then feeds the output. The flip-flop can be configured as a D-type or latch. In DDR mode, ONEG0 and OPOS0 are fed into registers on the positive edge of the clock. At the next clock cycle the registered OPOS0 is latched. A multiplexer running off the same clock cycle selects the correct register to feed the output (D0).

By combining output blocks of the complementary PIOs and sharing some registers from input blocks, a gearbox function can be implemented, to take four data streams ONEG0A, ONEG1A, ONEG1B and ONEG1B. Figure 2-27

DLL Calibrated DQS Delay Block

Source synchronous interfaces generally require the input clock to be adjusted in order to correctly capture data at the input register. For most interfaces a PLL is used for this adjustment. However, in DDR memories the clock, referred to as DQS, is not free-running, and this approach cannot be used. The DQS Delay block provides the required clock alignment for DDR memory interfaces.

The DQS signal (selected PIOs only, as shown in Figure 2-30) feeds from the PAD through a DQS delay element to a dedicated DQS routing resource. The DQS signal also feeds polarity control logic which controls the polarity of the clock to the sync registers in the input register blocks. Figure 2-30 and Figure 2-31 show how the DQS transition signals are routed to the PIOs.

The temperature, voltage and process variations of the DQS delay block are compensated by a set of 6-bit bus calibration signals from two dedicated DLLs (DDR_DLL) on opposite sides of the device. Each DLL compensates DQS delays in its half of the device as shown in Figure 2-30. The DLL loop is compensated for temperature, voltage and process variations by the system clock and feedback loop.

Figure 2-30. Edge Clock, DLL Calibration and DQS Local Bus Distribution

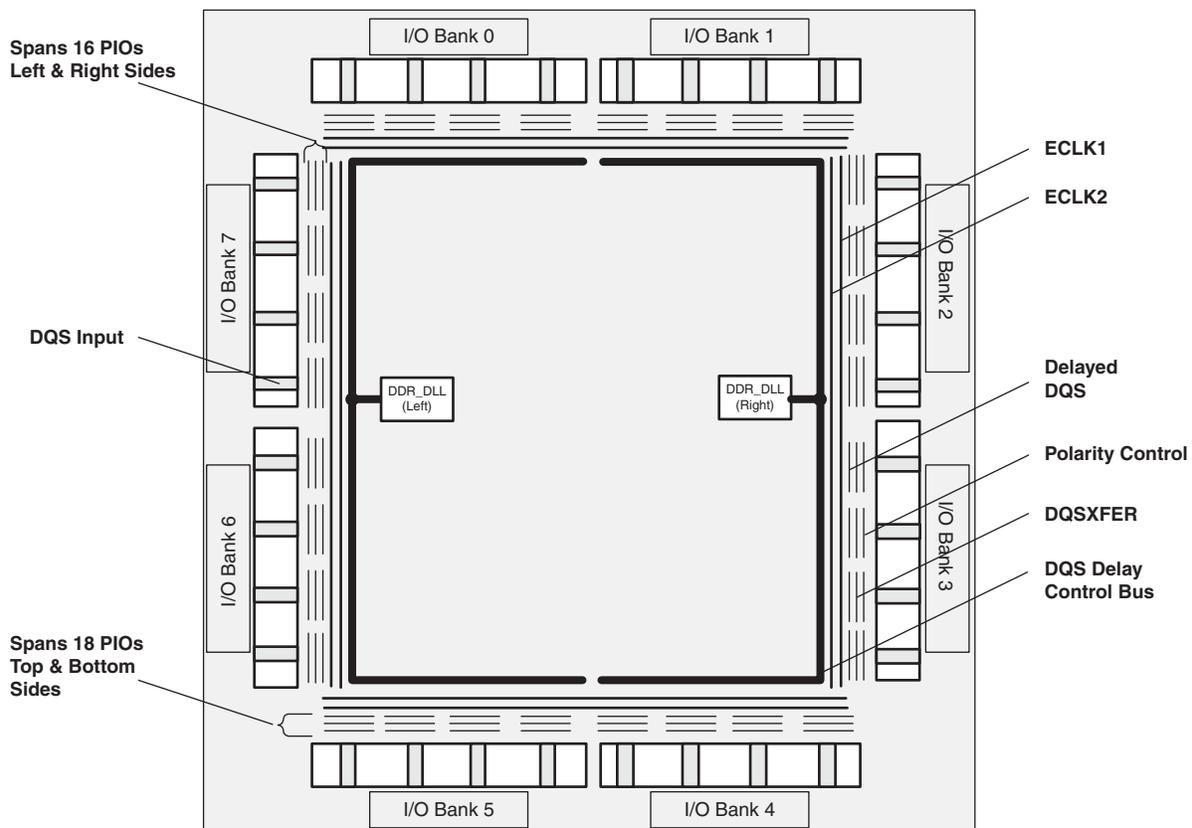


Table 2-12. Supported Input Standards

Input Standard	V _{REF} (Nom.)	V _{CCIO} ¹ (Nom.)
Single Ended Interfaces		
LVTTTL	—	—
LVCMOS33	—	—
LVCMOS25	—	—
LVCMOS18	—	1.8
LVCMOS15	—	1.5
LVCMOS12	—	—
PCI33	—	—
HSTL18 Class I, II	0.9	—
HSTL15 Class I	0.75	—
SSTL33 Class I, II	1.5	—
SSTL25 Class I, II	1.25	—
SSTL18 Class I, II	0.9	—
Differential Interfaces		
Differential SSTL18 Class I, II	—	—
Differential SSTL25 Class I, II	—	—
Differential SSTL33 Class I, II	—	—
Differential HSTL15 Class I	—	—
Differential HSTL18 Class I, II	—	—
LVDS, MLVDS, LVPECL, BLVDS, RSDS	—	—

1. When not specified, V_{CCIO} can be set anywhere in the valid operating range (page 3-1).

Density Shifting

The LatticeXP2 family is designed to ensure that different density devices in the same family and in the same package have the same pinout. Furthermore, the architecture ensures a high success rate when performing design migration from lower density devices to higher density devices. In many cases, it is also possible to shift a lower utilization design targeted for a high-density device to a lower density device. However, the exact details of the final resource utilization will impact the likely success in each case.

sysIO Single-Ended DC Electrical Characteristics

Over Recommended Operating Conditions

Input/Output Standard	V _{IL}		V _{IH}		V _{OL}	V _{OH}	I _{OL} ¹ (mA)	I _{OH} ¹ (mA)
	Min. (V)	Max. (V)	Min. (V)	Max. (V)	Max. (V)	Min. (V)		
LVCMOS33	-0.3	0.8	2.0	3.6	0.4	V _{CCIO} - 0.4	20, 16, 12, 8, 4	-20, -16, -12, -8, -4
					0.2	V _{CCIO} - 0.2	0.1	-0.1
LVTTTL33	-0.3	0.8	2.0	3.6	0.4	V _{CCIO} - 0.4	20, 16, 12, 8, 4	-20, -16, -12, -8, -4
					0.2	V _{CCIO} - 0.2	0.1	-0.1
LVCMOS25	-0.3	0.7	1.7	3.6	0.4	V _{CCIO} - 0.4	20, 16, 12, 8, 4	-20, -16, -12, -8, -4
					0.2	V _{CCIO} - 0.2	0.1	-0.1
LVCMOS18	-0.3	0.35 V _{CCIO}	0.65 V _{CCIO}	3.6	0.4	V _{CCIO} - 0.4	16, 12, 8, 4	-16, -12, -8, -4
					0.2	V _{CCIO} - 0.2	0.1	-0.1
LVCMOS15	-0.3	0.35 V _{CCIO}	0.65 V _{CCIO}	3.6	0.4	V _{CCIO} - 0.4	8, 4	-8, -4
					0.2	V _{CCIO} - 0.2	0.1	-0.1
LVCMOS12	-0.3	0.35 V _{CC}	0.65 V _{CC}	3.6	0.4	V _{CCIO} - 0.4	6, 2	-6, -2
					0.2	V _{CCIO} - 0.2	0.1	-0.1
PCI33	-0.3	0.3 V _{CCIO}	0.5 V _{CCIO}	3.6	0.1 V _{CCIO}	0.9 V _{CCIO}	1.5	-0.5
SSTL33_I	-0.3	V _{REF} - 0.2	V _{REF} + 0.2	3.6	0.7	V _{CCIO} - 1.1	8	-8
SSTL33_II	-0.3	V _{REF} - 0.2	V _{REF} + 0.2	3.6	0.5	V _{CCIO} - 0.9	16	-16
SSTL25_I	-0.3	V _{REF} - 0.18	V _{REF} + 0.18	3.6	0.54	V _{CCIO} - 0.62	7.6	-7.6
							12	-12
SSTL25_II	-0.3	V _{REF} - 0.18	V _{REF} + 0.18	3.6	0.35	V _{CCIO} - 0.43	15.2	-15.2
							20	-20
SSTL18_I	-0.3	V _{REF} - 0.125	V _{REF} + 0.125	3.6	0.4	V _{CCIO} - 0.4	6.7	-6.7
SSTL18_II	-0.3	V _{REF} - 0.125	V _{REF} + 0.125	3.6	0.28	V _{CCIO} - 0.28	8	-8
							11	-11
HSTL15_I	-0.3	V _{REF} - 0.1	V _{REF} + 0.1	3.6	0.4	V _{CCIO} - 0.4	4	-4
							8	-8
HSTL18_I	-0.3	V _{REF} - 0.1	V _{REF} + 0.1	3.6	0.4	V _{CCIO} - 0.4	8	-8
							12	-12
HSTL18_II	-0.3	V _{REF} - 0.1	V _{REF} + 0.1	3.6	0.4	V _{CCIO} - 0.4	16	-16

1. The average DC current drawn by I/Os between GND connections, or between the last GND in an I/O bank and the end of an I/O bank, as shown in the logic signal connections table shall not exceed $n * 8\text{mA}$, where n is the number of I/Os between bank GND connections or between the last GND in a bank and the end of a bank.

sysIO Differential Electrical Characteristics

LVDS

Over Recommended Operating Conditions

Parameter	Description	Test Conditions	Min.	Typ.	Max.	Units
V_{INP} V_{INM}	Input Voltage		0	—	2.4	V
V_{CM}	Input Common Mode Voltage	Half the Sum of the Two Inputs	0.05	—	2.35	V
V_{THD}	Differential Input Threshold	Difference Between the Two Inputs	+/-100	—	—	mV
I_{IN}	Input Current	Power On or Power Off	—	—	+/-10	μ A
V_{OH}	Output High Voltage for V_{OP} or V_{OM}	$R_T = 100$ Ohm	—	1.38	1.60	V
V_{OL}	Output Low Voltage for V_{OP} or V_{OM}	$R_T = 100$ Ohm	0.9V	1.03	—	V
V_{OD}	Output Voltage Differential	$(V_{OP} - V_{OM})$, $R_T = 100$ Ohm	250	350	450	mV
ΔV_{OD}	Change in V_{OD} Between High and Low		—	—	50	mV
V_{OS}	Output Voltage Offset	$(V_{OP} + V_{OM})/2$, $R_T = 100$ Ohm	1.125	1.20	1.375	V
ΔV_{OS}	Change in V_{OS} Between H and L		—	—	50	mV
I_{SA}	Output Short Circuit Current	$V_{OD} = 0V$ Driver Outputs Shorted to Ground	—	—	24	mA
I_{SAB}	Output Short Circuit Current	$V_{OD} = 0V$ Driver Outputs Shorted to Each Other	—	—	12	mA

Differential HSTL and SSTL

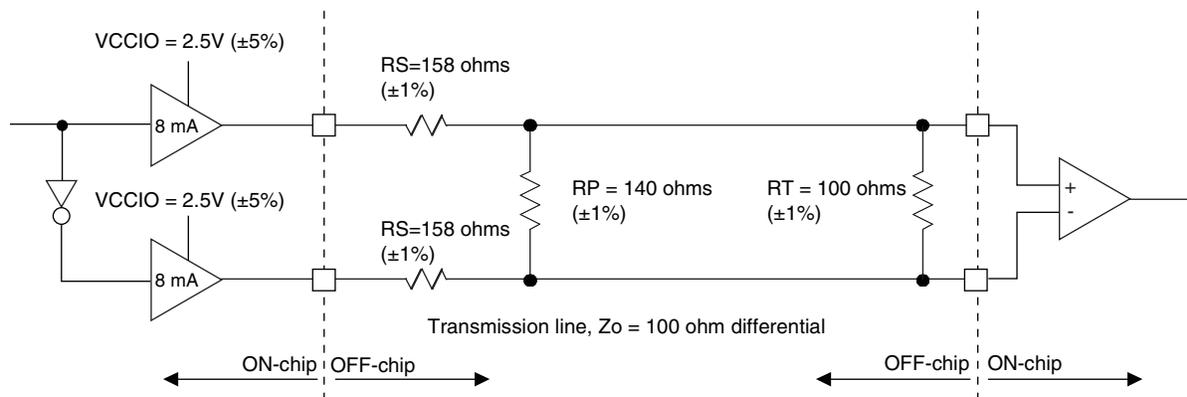
Differential HSTL and SSTL outputs are implemented as a pair of complementary single-ended outputs. All allowable single-ended output classes (class I and class II) are supported in this mode.

For further information on LVPECL, RSDS, MLVDS, BLVDS and other differential interfaces please see details in additional technical notes listed at the end of this data sheet.

LVDS25E

The top and bottom sides of LatticeXP2 devices support LVDS outputs via emulated complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The scheme shown in Figure 3-1 is one possible solution for point-to-point signals.

Figure 3-1. LVDS25E Output Termination Example



Typical Building Block Function Performance¹

Pin-to-Pin Performance (LVCMOS25 12mA Drive)

Function	-7 Timing	Units
Basic Functions		
16-bit Decoder	4.4	ns
32-bit Decoder	5.2	ns
64-bit Decoder	5.6	ns
4:1 MUX	3.7	ns
8:1 MUX	3.9	ns
16:1 MUX	4.3	ns
32:1 MUX	4.5	ns

Register-to-Register Performance

Function	-7 Timing	Units
Basic Functions		
16-bit Decoder	521	MHz
32-bit Decoder	537	MHz
64-bit Decoder	484	MHz
4:1 MUX	744	MHz
8:1 MUX	678	MHz
16:1 MUX	616	MHz
32:1 MUX	529	MHz
8-bit Adder	570	MHz
16-bit Adder	507	MHz
64-bit Adder	293	MHz
16-bit Counter	541	MHz
32-bit Counter	440	MHz
64-bit Counter	321	MHz
64-bit Accumulator	261	MHz
Embedded Memory Functions		
512x36 Single Port RAM, EBR Output Registers	315	MHz
1024x18 True-Dual Port RAM (Write Through or Normal, EBR Output Registers)	315	MHz
1024x18 True-Dual Port RAM (Write Through or Normal, PLC Output Registers)	231	MHz
Distributed Memory Functions		
16x4 Pseudo-Dual Port RAM (One PFU)	760	MHz
32x2 Pseudo-Dual Port RAM	455	MHz
64x1 Pseudo-Dual Port RAM	351	MHz
DSP Functions		
18x18 Multiplier (All Registers)	342	MHz
9x9 Multiplier (All Registers)	342	MHz
36x36 Multiply (All Registers)	330	MHz
18x18 Multiply/Accumulate (Input and Output Registers)	218	MHz
18x18 Multiply-Add/Sub-Sum (All Registers)	292	MHz

LatticeXP2 External Switching Characteristics (Continued)

Over Recommended Operating Conditions

Parameter	Description	Device	-7		-6		-5		Units
			Min.	Max.	Min.	Max.	Min.	Max.	
t _{HE}	Clock to Data Hold - PIO Input Register	XP2-5	1.00	—	1.30	—	1.60	—	ns
		XP2-8	1.00	—	1.30	—	1.60	—	ns
		XP2-17	1.00	—	1.30	—	1.60	—	ns
		XP2-30	1.20	—	1.60	—	1.90	—	ns
		XP2-40	1.20	—	1.60	—	1.90	—	ns
t _{SU_DELE}	Clock to Data Setup - PIO Input Register with Data Input Delay	XP2-5	1.00	—	1.30	—	1.60	—	ns
		XP2-8	1.00	—	1.30	—	1.60	—	ns
		XP2-17	1.00	—	1.30	—	1.60	—	ns
		XP2-30	1.20	—	1.60	—	1.90	—	ns
		XP2-40	1.20	—	1.60	—	1.90	—	ns
t _{H_DELE}	Clock to Data Hold - PIO Input Register with Input Data Delay	XP2-5	0.00	—	0.00	—	0.00	—	ns
		XP2-8	0.00	—	0.00	—	0.00	—	ns
		XP2-17	0.00	—	0.00	—	0.00	—	ns
		XP2-30	0.00	—	0.00	—	0.00	—	ns
		XP2-40	0.00	—	0.00	—	0.00	—	ns
f _{MAX_IOE}	Clock Frequency of I/O and PFU Register	XP2	—	420	—	357	—	311	MHz
General I/O Pin Parameters (using Primary Clock with PLL)¹									
t _{COPLL}	Clock to Output - PIO Output Register	XP2-5	—	3.00	—	3.30	—	3.70	ns
		XP2-8	—	3.00	—	3.30	—	3.70	ns
		XP2-17	—	3.00	—	3.30	—	3.70	ns
		XP2-30	—	3.00	—	3.30	—	3.70	ns
		XP2-40	—	3.00	—	3.30	—	3.70	ns
t _{SUPLL}	Clock to Data Setup - PIO Input Register	XP2-5	1.00	—	1.20	—	1.40	—	ns
		XP2-8	1.00	—	1.20	—	1.40	—	ns
		XP2-17	1.00	—	1.20	—	1.40	—	ns
		XP2-30	1.00	—	1.20	—	1.40	—	ns
		XP2-40	1.00	—	1.20	—	1.40	—	ns
t _{HPLL}	Clock to Data Hold - PIO Input Register	XP2-5	0.90	—	1.10	—	1.30	—	ns
		XP2-8	0.90	—	1.10	—	1.30	—	ns
		XP2-17	0.90	—	1.10	—	1.30	—	ns
		XP2-30	1.00	—	1.20	—	1.40	—	ns
		XP2-40	1.00	—	1.20	—	1.40	—	ns
t _{SU_DELPLL}	Clock to Data Setup - PIO Input Register with Data Input Delay	XP2-5	1.90	—	2.10	—	2.30	—	ns
		XP2-8	1.90	—	2.10	—	2.30	—	ns
		XP2-17	1.90	—	2.10	—	2.30	—	ns
		XP2-30	2.00	—	2.20	—	2.40	—	ns
		XP2-40	2.00	—	2.20	—	2.40	—	ns

LatticeXP2 Internal Switching Characteristics¹ (Continued)
Over Recommended Operating Conditions

Parameter	Description	-7		-6		-5		Units
		Min.	Max.	Min.	Max.	Min.	Max.	
t _{HP_DSP}	Pipeline Register Hold Time	-0.787	—	-0.890	—	-0.994	—	ns
t _{SUO_DSP}	Output Register Setup Time	4.896	—	5.413	—	5.931	—	ns
t _{HO_DSP}	Output Register Hold Time	-1.439	—	-1.604	—	-1.770	—	ns
t _{COI_DSP} ³	Input Register Clock to Output Time	—	4.513	—	4.947	—	5.382	ns
t _{COP_DSP} ³	Pipeline Register Clock to Output Time	—	2.153	—	2.272	—	2.391	ns
t _{COO_DSP} ³	Output Register Clock to Output Time	—	0.569	—	0.600	—	0.631	ns
t _{SUADSUB}	AdSub Input Register Setup Time	-0.270	—	-0.298	—	-0.327	—	ns
t _{HADSUB}	AdSub Input Register Hold Time	0.306	—	0.338	—	0.371	—	ns

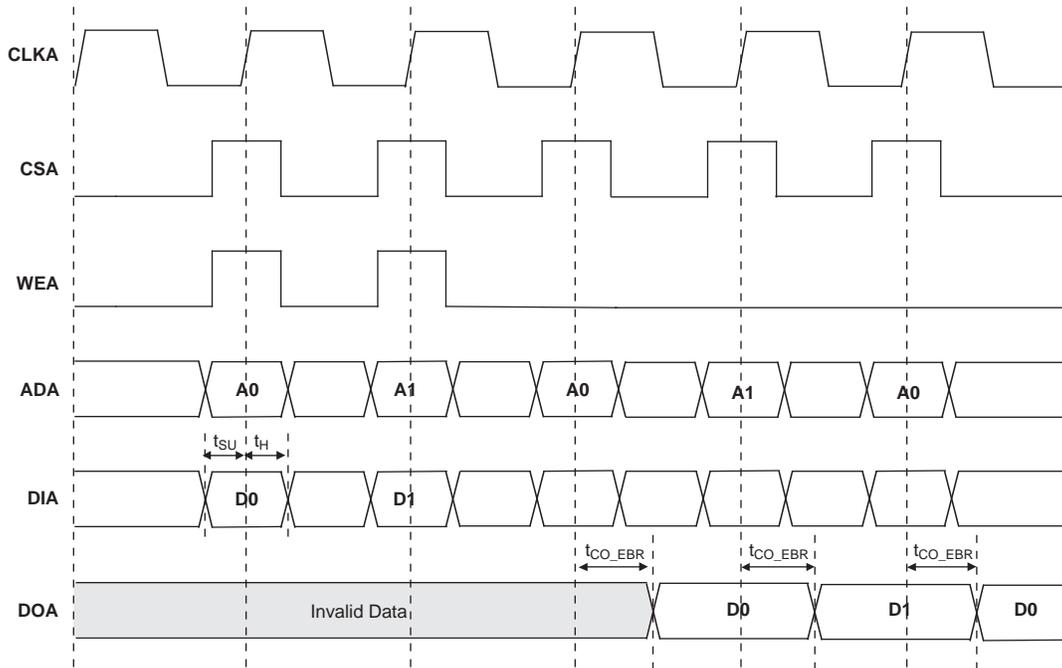
1. Internal parameters are characterized, but not tested on every device.

2. RST resets VCO and all counters in PLL.

3. These parameters include the Adder Subtractor block in the path.

EBR Timing Diagrams

Figure 3-6. Read/Write Mode (Normal)



Note: Input data and address are registered at the positive edge of the clock and output data appears after the positive edge of the clock.

Figure 3-7. Read/Write Mode with Input and Output Registers

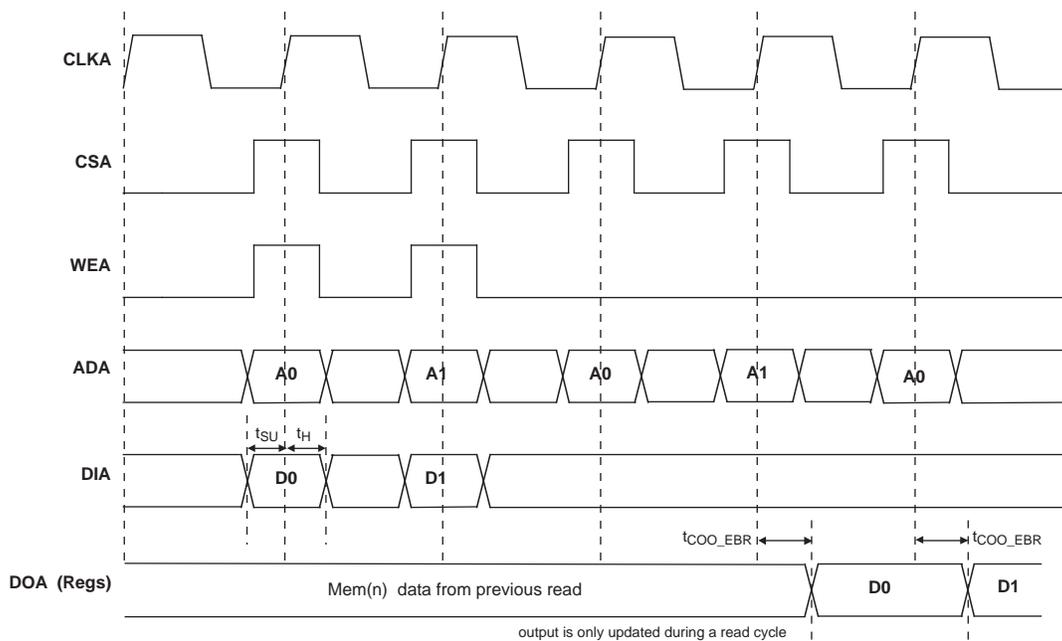
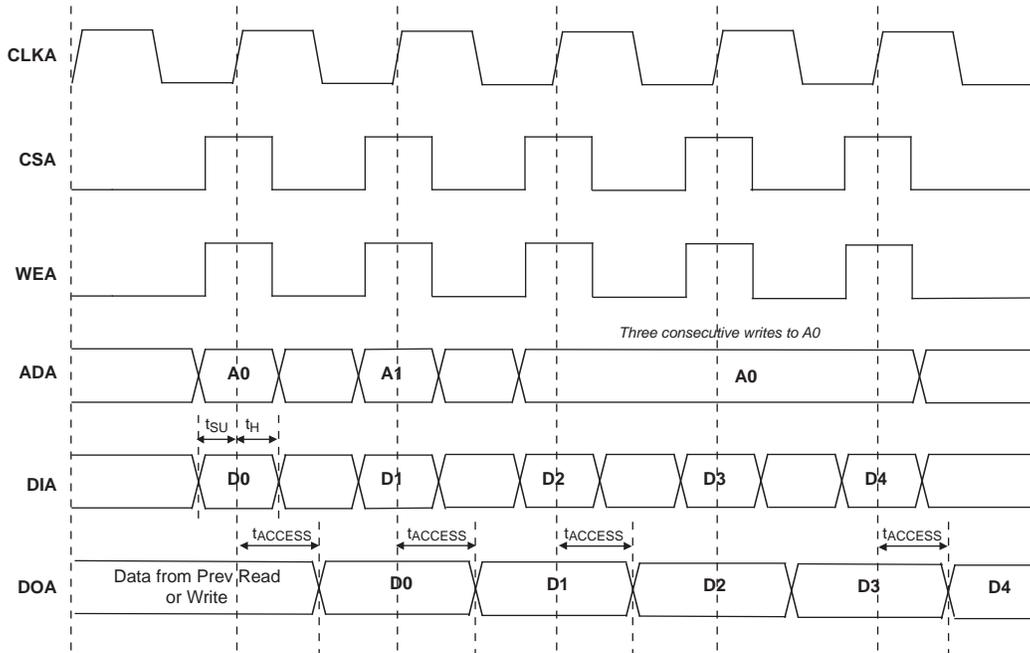


Figure 3-8. Write Through (SP Read/Write on Port A, Input Registers Only)



Note: Input data and address are registered at the positive edge of the clock and output data appears after the positive edge of the clock.

LatticeXP2 Family Timing Adders^{1, 2, 3, 4} (Continued)

Over Recommended Operating Conditions

Buffer Type	Description	-7	-6	-5	Units
HSTL15_I	HSTL_15 class I 4mA drive	0.32	0.69	1.06	ns
HSTL15D_I	Differential HSTL 15 class I 4mA drive	0.32	0.69	1.06	ns
SSTL33_I	SSTL_3 class I	-0.25	0.05	0.35	ns
SSTL33_II	SSTL_3 class II	-0.31	-0.02	0.27	ns
SSTL33D_I	Differential SSTL_3 class I	-0.25	0.05	0.35	ns
SSTL33D_II	Differential SSTL_3 class II	-0.31	-0.02	0.27	ns
SSTL25_I	SSTL_2 class I 8mA drive	-0.25	0.02	0.30	ns
SSTL25_II	SSTL_2 class II 16mA drive	-0.28	0.00	0.28	ns
SSTL25D_I	Differential SSTL_2 class I 8mA drive	-0.25	0.02	0.30	ns
SSTL25D_II	Differential SSTL_2 class II 16mA drive	-0.28	0.00	0.28	ns
SSTL18_I	SSTL_1.8 class I	-0.17	0.13	0.43	ns
SSTL18_II	SSTL_1.8 class II 8mA drive	-0.18	0.12	0.42	ns
SSTL18D_I	Differential SSTL_1.8 class I	-0.17	0.13	0.43	ns
SSTL18D_II	Differential SSTL_1.8 class II 8mA drive	-0.18	0.12	0.42	ns
LVTTTL33_4mA	LVTTTL 4mA drive	-0.37	-0.05	0.26	ns
LVTTTL33_8mA	LVTTTL 8mA drive	-0.45	-0.18	0.10	ns
LVTTTL33_12mA	LVTTTL 12mA drive	-0.52	-0.24	0.04	ns
LVTTTL33_16mA	LVTTTL 16mA drive	-0.43	-0.14	0.14	ns
LVTTTL33_20mA	LVTTTL 20mA drive	-0.46	-0.18	0.09	ns
LVCMOS33_4mA	LVCMOS 3.3 4mA drive, fast slew rate	-0.37	-0.05	0.26	ns
LVCMOS33_8mA	LVCMOS 3.3 8mA drive, fast slew rate	-0.45	-0.18	0.10	ns
LVCMOS33_12mA	LVCMOS 3.3 12mA drive, fast slew rate	-0.52	-0.24	0.04	ns
LVCMOS33_16mA	LVCMOS 3.3 16mA drive, fast slew rate	-0.43	-0.14	0.14	ns
LVCMOS33_20mA	LVCMOS 3.3 20mA drive, fast slew rate	-0.46	-0.18	0.09	ns
LVCMOS25_4mA	LVCMOS 2.5 4mA drive, fast slew rate	-0.42	-0.15	0.13	ns
LVCMOS25_8mA	LVCMOS 2.5 8mA drive, fast slew rate	-0.48	-0.21	0.05	ns
LVCMOS25_12mA	LVCMOS 2.5 12mA drive, fast slew rate	0.00	0.00	0.00	ns
LVCMOS25_16mA	LVCMOS 2.5 16mA drive, fast slew rate	-0.45	-0.18	0.08	ns
LVCMOS25_20mA	LVCMOS 2.5 20mA drive, fast slew rate	-0.49	-0.22	0.04	ns
LVCMOS18_4mA	LVCMOS 1.8 4mA drive, fast slew rate	-0.46	-0.18	0.10	ns
LVCMOS18_8mA	LVCMOS 1.8 8mA drive, fast slew rate	-0.52	-0.25	0.02	ns
LVCMOS18_12mA	LVCMOS 1.8 12mA drive, fast slew rate	-0.56	-0.30	-0.03	ns
LVCMOS18_16mA	LVCMOS 1.8 16mA drive, fast slew rate	-0.50	-0.24	0.03	ns
LVCMOS15_4mA	LVCMOS 1.5 4mA drive, fast slew rate	-0.45	-0.17	0.11	ns
LVCMOS15_8mA	LVCMOS 1.5 8mA drive, fast slew rate	-0.53	-0.26	0.00	ns
LVCMOS12_2mA	LVCMOS 1.2 2mA drive, fast slew rate	-0.46	-0.19	0.08	ns
LVCMOS12_6mA	LVCMOS 1.2 6mA drive, fast slew rate	-0.55	-0.29	-0.02	ns
LVCMOS33_4mA	LVCMOS 3.3 4mA drive, slow slew rate	0.98	1.41	1.84	ns
LVCMOS33_8mA	LVCMOS 3.3 8mA drive, slow slew rate	0.74	1.16	1.58	ns
LVCMOS33_12mA	LVCMOS 3.3 12mA drive, slow slew rate	0.56	0.97	1.38	ns
LVCMOS33_16mA	LVCMOS 3.3 16mA drive, slow slew rate	0.77	1.19	1.61	ns
LVCMOS33_20mA	LVCMOS 3.3 20mA drive, slow slew rate	0.57	0.98	1.40	ns

LatticeXP2 sysCONFIG Port Timing Specifications

Over Recommended Operating Conditions

Parameter	Description	Min	Max	Units
sysCONFIG POR, Initialization and Wake Up				
t_{ICFG}	Minimum Vcc to INITN High	—	50	ms
t_{VMC}	Time from t_{ICFG} to valid Master CCLK	—	2	μ s
t_{PRGMRJ}	PROGRAMN Pin Pulse Rejection	—	12	ns
t_{PRGM}	PROGRAMN Low Time to Start Configuration	50	—	ns
t_{DINIT}^1	PROGRAMN High to INITN High Delay	—	1	ms
$t_{DPPINIT}$	Delay Time from PROGRAMN Low to INITN Low	—	50	ns
$t_{DPPDONE}$	Delay Time from PROGRAMN Low to DONE Low	—	50	ns
t_{IODISS}	User I/O Disable from PROGRAMN Low	—	35	ns
t_{IOENSS}	User I/O Enabled Time from CCLK Edge During Wake-up Sequence	—	25	ns
t_{MWC}	Additional Wake Master Clock Signals after DONE Pin High	0	—	Cycles
sysCONFIG SPI Port (Master)				
t_{CFGX}	INITN High to CCLK Low	—	1	μ s
t_{CSSPI}	INITN High to CSSPIN Low	—	2	μ s
t_{CSCCLK}	CCLK Low before CSSPIN Low	0	—	ns
t_{SOCDO}	CCLK Low to Output Valid	—	15	ns
t_{CSPID}	CSSPIN[0:1] Low to First CCLK Edge Setup Time	2cyc	600+6cyc	ns
f_{MAXSPI}	Max CCLK Frequency	—	20	MHz
t_{SUSPI}	SOSPI Data Setup Time Before CCLK	7	—	ns
t_{HSPI}	SOSPI Data Hold Time After CCLK	10	—	ns
sysCONFIG SPI Port (Slave)				
$f_{MAXSPIS}$	Slave CCLK Frequency	—	25	MHz
t_{RF}	Rise and Fall Time	50	—	mV/ns
t_{STCO}	Falling Edge of CCLK to SOSPI Active	—	20	ns
t_{STOZ}	Falling Edge of CCLK to SOSPI Disable	—	20	ns
t_{STSU}	Data Setup Time (SISPI)	8	—	ns
t_{STH}	Data Hold Time (SISPI)	10	—	ns
t_{STCKH}	CCLK Clock Pulse Width, High	0.02	200	μ s
t_{STCKL}	CCLK Clock Pulse Width, Low	0.02	200	μ s
t_{STVO}	Falling Edge of CCLK to Valid SOSPI Output	—	20	ns
t_{SCS}	CSSPISN High Time	25	—	ns
t_{SCSS}	CSSPISN Setup Time	25	—	ns
t_{SCSH}	CSSPISN Hold Time	25	—	ns

1. Re-toggling the PROGRAMN pin is not permitted until the INITN pin is high. Avoid consecutive toggling of PROGRAMN.

Pin Information Summary (Cont.)

Pin Type		XP2-5				XP2-8				XP2-17			XP2-30			XP2-40	
		132 csBGA	144 TQFP	208 PQFP	256 ftBGA	132 csBGA	144 TQFP	208 PQFP	256 ftBGA	208 PQFP	256 ftBGA	484 fpBGA	256 ftBGA	484 fpBGA	672 fpBGA	484 fpBGA	672 fpBGA
PCI capable I/Os Bonding Out per Bank	Bank0	18	20	20	26	18	20	20	28	20	28	52	28	52	70	52	70
	Bank1	4	6	18	18	4	6	18	22	18	22	36	22	36	54	36	70
	Bank2	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	Bank3	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	Bank4	8	8	18	18	8	8	18	26	18	26	36	26	38	54	38	70
	Bank5	14	18	20	24	14	18	20	24	20	24	52	24	53	70	53	70
	Bank6	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	Bank7	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0

1. Minimum requirement to implement a fully functional 8-bit wide DDR bus. Available DDR interface consists of at least 12 I/Os (1 DQS + 1 DQSB + 8 DQs + 1 DM + Bank VREF1).

Logic Signal Connections

Package pinout information can be found under “Data Sheets” on the LatticeXP2 product page of the Lattice website at www.latticesemi.com/products/fpga/xp2 and in the Lattice Diamond design software.

Thermal Management

Thermal management is recommended as part of any sound FPGA design methodology. To assess the thermal characteristics of a system, Lattice specifies a maximum allowable junction temperature in all device data sheets. Designers must complete a thermal analysis of their specific design to ensure that the device and package do not exceed the junction temperature limits. Refer to the Lattice [Thermal Management](#) document to find the device/package specific thermal values.

For Further Information

- TN1139, [Power Estimation and Management for LatticeXP2 Devices](#)
- Power Calculator tool is included with the Lattice Diamond design tool or as a standalone download from www.latticesemi.com/products/designsoftware

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-17E-5QN208I	1.2V	-5	Lead-Free PQFP	208	IND	17
LFXP2-17E-6QN208I	1.2V	-6	Lead-Free PQFP	208	IND	17
LFXP2-17E-5FTN256I	1.2V	-5	Lead-Free ftBGA	256	IND	17
LFXP2-17E-6FTN256I	1.2V	-6	Lead-Free ftBGA	256	IND	17
LFXP2-17E-5FN484I	1.2V	-5	Lead-Free fpBGA	484	IND	17
LFXP2-17E-6FN484I	1.2V	-6	Lead-Free fpBGA	484	IND	17

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-30E-5FTN256I	1.2V	-5	Lead-Free ftBGA	256	IND	30
LFXP2-30E-6FTN256I	1.2V	-6	Lead-Free ftBGA	256	IND	30
LFXP2-30E-5FN484I	1.2V	-5	Lead-Free fpBGA	484	IND	30
LFXP2-30E-6FN484I	1.2V	-6	Lead-Free fpBGA	484	IND	30
LFXP2-30E-5FN672I	1.2V	-5	Lead-Free fpBGA	672	IND	30
LFXP2-30E-6FN672I	1.2V	-6	Lead-Free fpBGA	672	IND	30

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-40E-5FN484I	1.2V	-5	Lead-Free fpBGA	484	IND	40
LFXP2-40E-6FN484I	1.2V	-6	Lead-Free fpBGA	484	IND	40
LFXP2-40E-5FN672I	1.2V	-5	Lead-Free fpBGA	672	IND	40
LFXP2-40E-6FN672I	1.2V	-6	Lead-Free fpBGA	672	IND	40

For Further Information

A variety of technical notes for the LatticeXP2 FPGA family are available on the Lattice Semiconductor web site at www.latticesemi.com.

- TN1136, [LatticeXP2 sysIO Usage Guide](#)
- TN1137, [LatticeXP2 Memory Usage Guide](#)
- TN1138, [LatticeXP2 High Speed I/O Interface](#)
- TN1126, [LatticeXP2 sysCLOCK PLL Design and Usage Guide](#)
- TN1139, [Power Estimation and Management for LatticeXP2 Devices](#)
- TN1140, [LatticeXP2 sysDSP Usage Guide](#)
- TN1141, [LatticeXP2 sysCONFIG Usage Guide](#)
- TN1142, [LatticeXP2 Configuration Encryption and Security Usage Guide](#)
- TN1087, [Minimizing System Interruption During Configuration Using TransFR Technology](#)
- TN1220, [LatticeXP2 Dual Boot Feature](#)
- TN1130, [LatticeXP2 Soft Error Detection \(SED\) Usage Guide](#)
- TN1143, [LatticeXP2 Hardware Checklist](#)

For further information on interface standards refer to the following websites:

- JEDEC Standards (LVTTTL, LVCMOS, SSTL, HSTL): www.jedec.org
- PCI: www.pcisig.com