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Details	
Product Status	Active
Number of LABs/CLBs	5000
Number of Logic Elements/Cells	40000
Total RAM Bits	906240
Number of I/O	363
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	484-BBGA
Supplier Device Package	484-FPBGA (23x23)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfxp2-40e-7fn484c

LatticeXP2 Family Data Sheet
Architecture

August 2014 **Data Sheet DS1009**

Architecture Overview

Each LatticeXP2 device contains an array of logic blocks surrounded by Programmable I/O Cells (PIC). Interspersed between the rows of logic blocks are rows of sysMEM™ Embedded Block RAM (EBR) and a row of sysDSP™ Digital Signal Processing blocks as shown in Figure 2-1.

On the left and right sides of the Programmable Functional Unit (PFU) array, there are Non-volatile Memory Blocks. In configuration mode the nonvolatile memory is programmed via the IEEE 1149.1 TAP port or the sysCONFIG™ peripheral port. On power up, the configuration data is transferred from the Non-volatile Memory Blocks to the configuration SRAM. With this technology, expensive external configuration memory is not required, and designs are secured from unauthorized read-back. This transfer of data from non-volatile memory to configuration SRAM via wide busses happens in microseconds, providing an “instant-on” capability that allows easy interfacing in many applications. LatticeXP2 devices can also transfer data from the sysMEM EBR blocks to the Non-volatile Memory Blocks at user request.

There are two kinds of logic blocks, the PFU and the PFU without RAM (PFF). The PFU contains the building blocks for logic, arithmetic, RAM and ROM functions. The PFF block contains building blocks for logic, arithmetic and ROM functions. Both PFU and PFF blocks are optimized for flexibility allowing complex designs to be implemented quickly and efficiently. Logic Blocks are arranged in a two-dimensional array. Only one type of block is used per row.

LatticeXP2 devices contain one or more rows of sysMEM EBR blocks. sysMEM EBRs are large dedicated 18Kbit memory blocks. Each sysMEM block can be configured in a variety of depths and widths of RAM or ROM. In addition, LatticeXP2 devices contain up to two rows of DSP Blocks. Each DSP block has multipliers and adder/accumulators, which are the building blocks for complex signal processing capabilities.

Each PIC block encompasses two PIOs (PIO pairs) with their respective sysIO buffers. The sysIO buffers of the LatticeXP2 devices are arranged into eight banks, allowing the implementation of a wide variety of I/O standards. PIO pairs on the left and right edges of the device can be configured as LVDS transmit/receive pairs. The PIC logic also includes pre-engineered support to aid in the implementation of high speed source synchronous standards such as 7:1 LVDS interfaces, found in many display applications, and memory interfaces including DDR and DDR2.

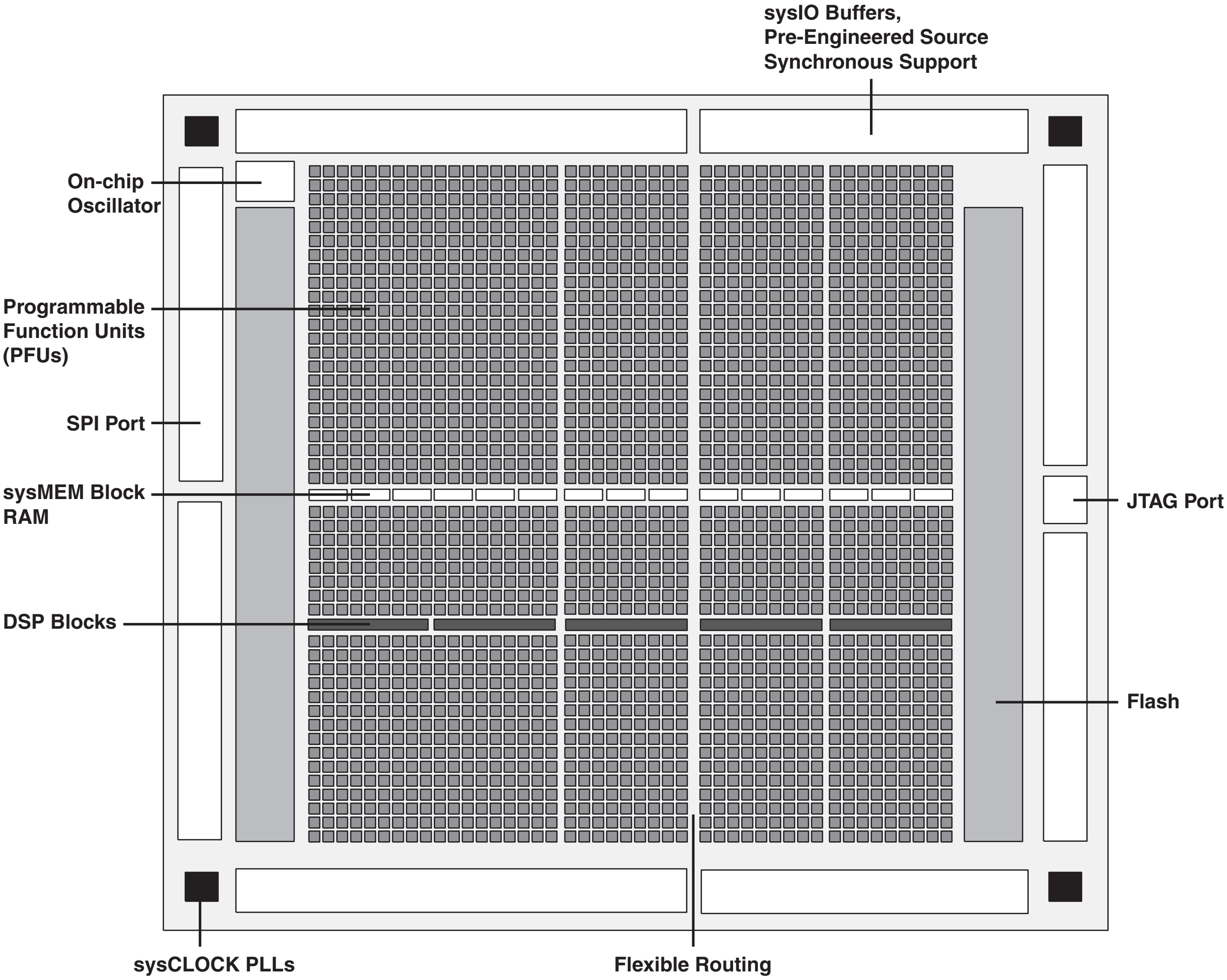
The LatticeXP2 registers in PFU and sysI/O can be configured to be SET or RESET. After power up and device is configured, the device enters into user mode with these registers SET/RESET according to the configuration setting, allowing device entering to a known state for predictable system function.

Other blocks provided include PLLs and configuration functions. The LatticeXP2 architecture provides up to four General Purpose PLLs (GPLL) per device. The GPLL blocks are located in the corners of the device.

The configuration block that supports features such as configuration bit-stream de-encryption, transparent updates and dual boot support is located between banks two and three. Every device in the LatticeXP2 family supports a sysCONFIG port, muxed with bank seven I/Os, which supports serial device configuration. A JTAG port is provided between banks two and three.

This family also provides an on-chip oscillator. LatticeXP2 devices use 1.2V as their core voltage.

Figure 2-1. Simplified Block Diagram, LatticeXP2-17 Device (Top Level)



PFU Blocks

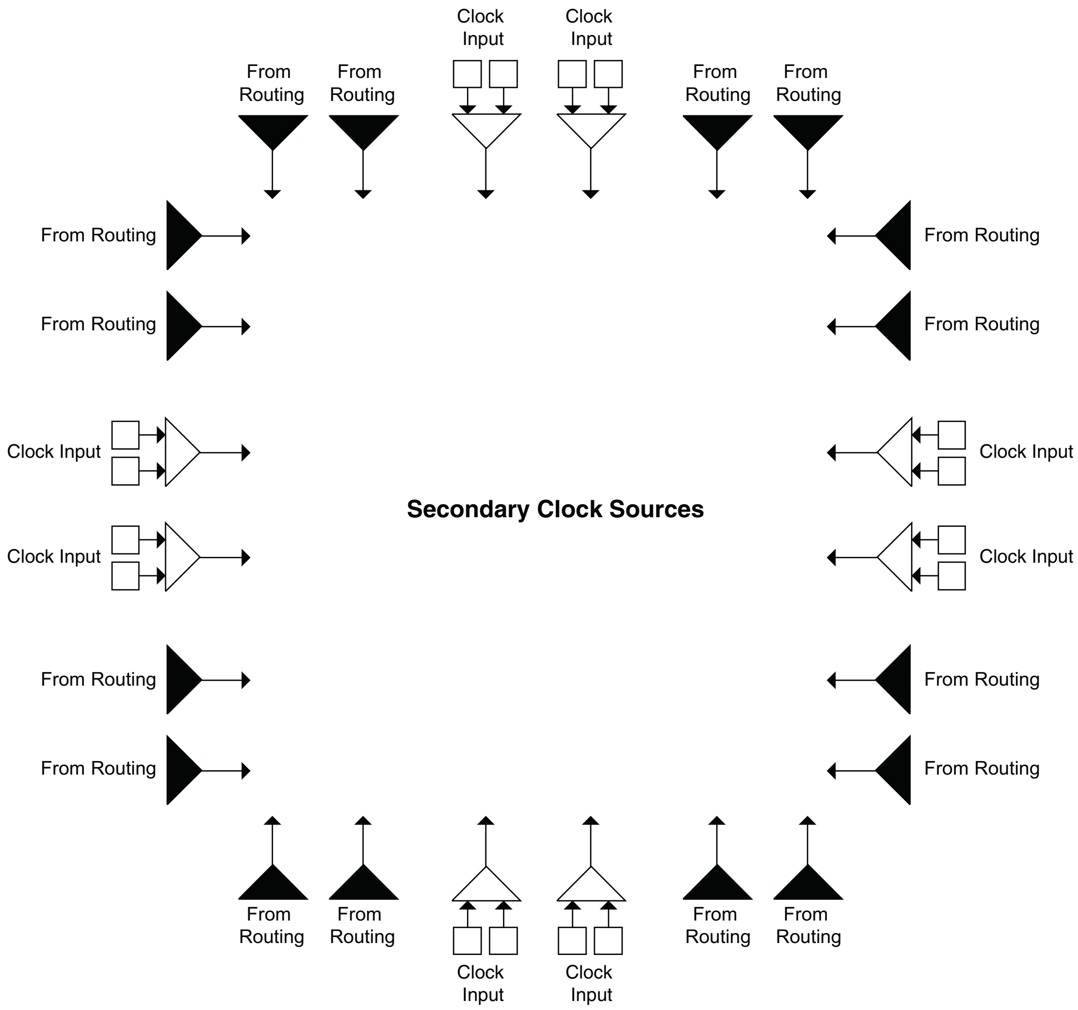
The core of the LatticeXP2 device is made up of logic blocks in two forms, PFUs and PFFs. PFUs can be programmed to perform logic, arithmetic, distributed RAM and distributed ROM functions. PFF blocks can be programmed to perform logic, arithmetic and ROM functions. Except where necessary, the remainder of this data sheet will use the term PFU to refer to both PFU and PFF blocks.

Each PFU block consists of four interconnected slices, numbered Slice 0 through Slice 3, as shown in Figure 2-2. All the interconnections to and from PFU blocks are from routing. There are 50 inputs and 23 outputs associated with each PFU block.

Secondary Clock/Control Sources

LatticeXP2 devices derive secondary clocks (SC0 through SC7) from eight dedicated clock input pads and the rest from routing. Figure 2-7 shows the secondary clock sources.

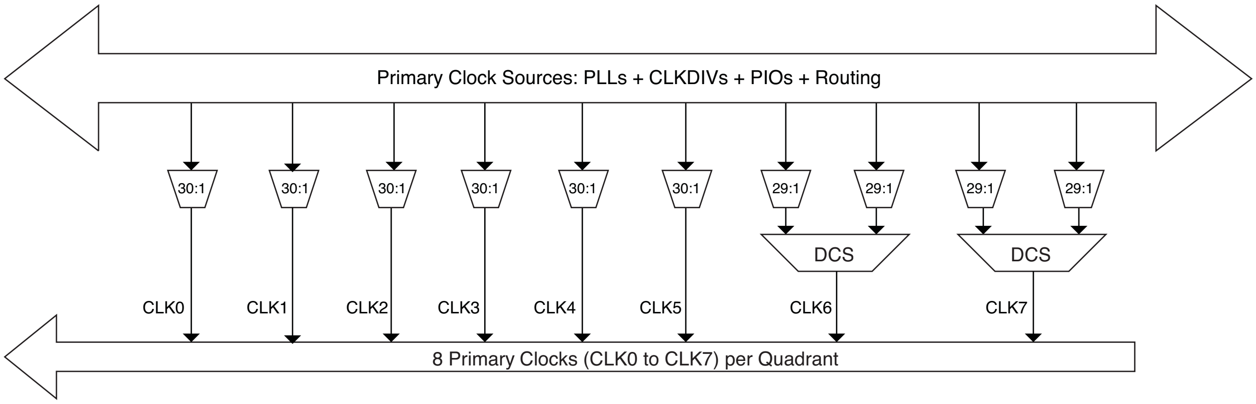
Figure 2-7. Secondary Clock Sources



Primary Clock Routing

The clock routing structure in LatticeXP2 devices consists of a network of eight primary clock lines (CLK0 through CLK7) per quadrant. The primary clocks of each quadrant are generated from muxes located in the center of the device. All the clock sources are connected to these muxes. Figure 2-9 shows the clock routing for one quadrant. Each quadrant mux is identical. If desired, any clock can be routed globally.

Figure 2-9. Per Quadrant Primary Clock Selection

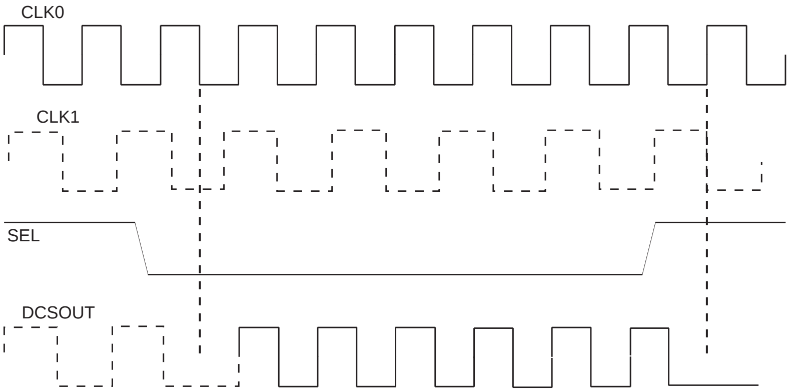


Dynamic Clock Select (DCS)

The DCS is a smart multiplexer function available in the primary clock routing. It switches between two independent input clock sources without any glitches or runt pulses. This is achieved irrespective of when the select signal is toggled. There are two DCS blocks per quadrant; in total, eight DCS blocks per device. The inputs to the DCS block come from the center muxes. The output of the DCS is connected to primary clocks CLK6 and CLK7 (see Figure 2-9).

Figure 2-10 shows the timing waveforms of the default DCS operating mode. The DCS block can be programmed to other modes. For more information on the DCS, please see TN1126, LatticeXP2 sysCLOCK PLL Design and Usage Guide.

Figure 2-10. DCS Waveforms



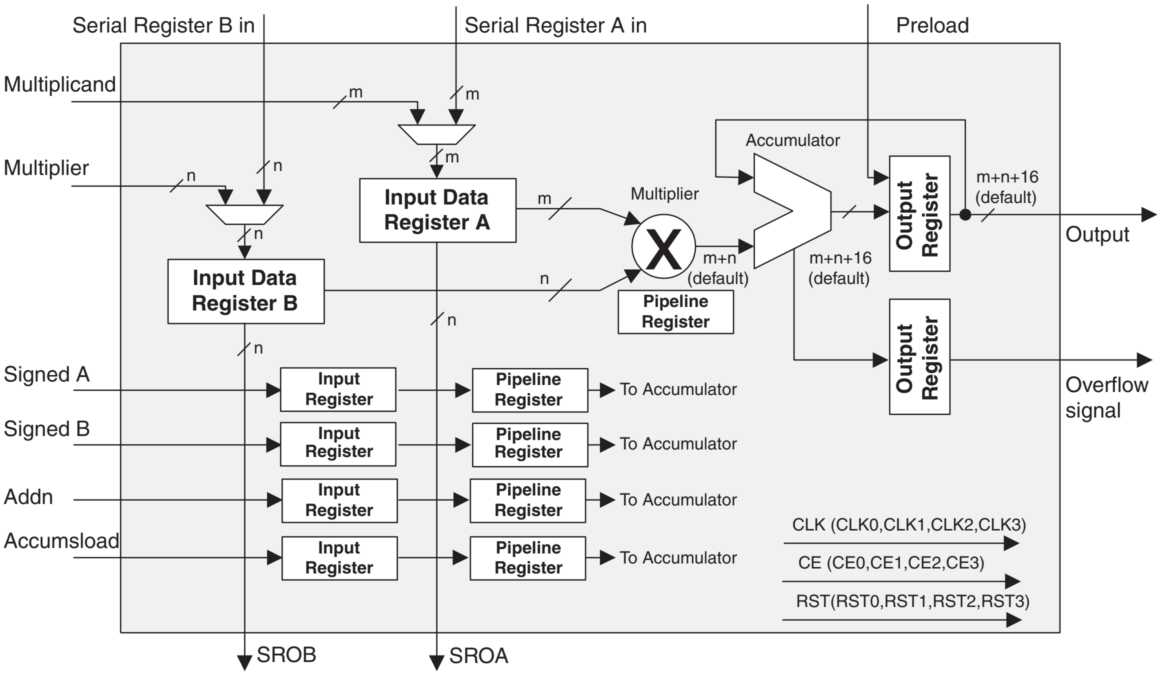
Secondary Clock/Control Routing

Secondary clocks in the LatticeXP2 devices are region-based resources. The benefit of region-based resources is the relatively low injection delay and skew within the region, as compared to primary clocks. EBR rows, DSP rows and a special vertical routing channel bound the secondary clock regions. This special vertical routing channel aligns with either the left edge of the center DSP block in the DSP row or the center of the DSP row. Figure 2-11 shows this special vertical routing channel and the eight secondary clock regions for the LatticeXP2-40.

MAC sysDSP Element

In this case, the two operands, A and B, are multiplied and the result is added with the previous accumulated value. This accumulated value is available at the output. The user can enable the input and pipeline registers but the output register is always enabled. The output register is used to store the accumulated value. The Accumulators in the DSP blocks in LatticeXP2 family can be initialized dynamically. A registered overflow signal is also available. The overflow conditions are provided later in this document. Figure 2-21 shows the MAC sysDSP element.

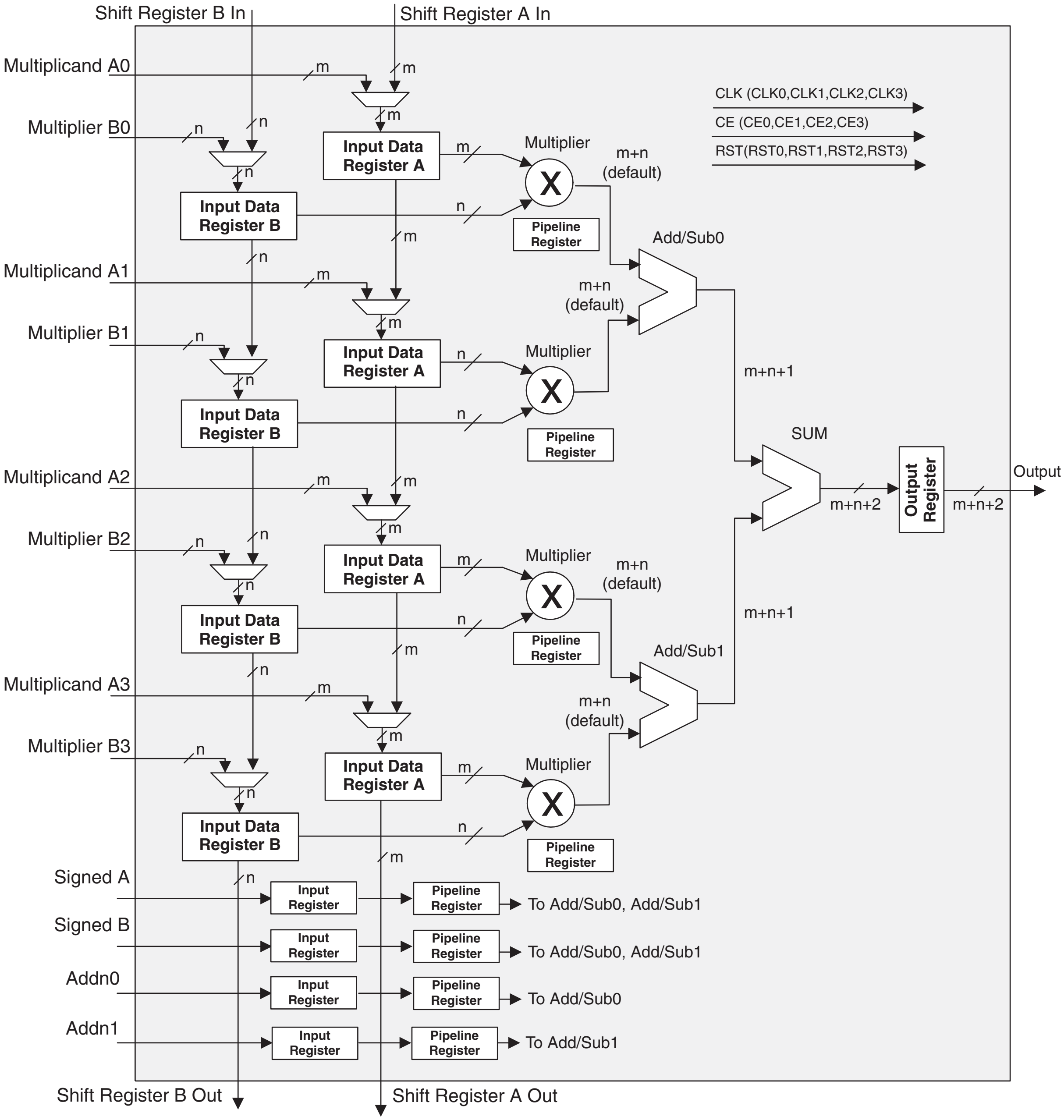
Figure 2-21. MAC sysDSP



MULTADDSUBSUM sysDSP Element

In this case, the operands A0 and B0 are multiplied and the result is added/subtracted with the result of the multiplier operation of operands A1 and B1. Additionally the operands A2 and B2 are multiplied and the result is added/ subtracted with the result of the multiplier operation of operands A3 and B3. The result of both addition/subtraction are added in a summation block. The user can enable the input, output and pipeline registers. Figure 2-23 shows the MULTADDSUBSUM sysDSP element.

Figure 2-23. MULTADDSUBSUM



Clock, Clock Enable and Reset Resources

Global Clock, Clock Enable (CE) and Reset (RST) signals from routing are available to every DSP block. From four clock sources (CLK0, CLK1, CLK2, CLK3) one clock is selected for each input register, pipeline register and output register. Similarly, CE and RST are selected from their four respective sources (CE0, CE1, CE2, CE3 and RST0, RST1, RST2, RST3) at each input register, pipeline register and output register.

Signed and Unsigned with Different Widths

The DSP block supports other widths, in addition to x9, x18 and x36 widths, of signed and unsigned multipliers. For unsigned operands, unused upper data bits should be filled to create a valid x9, x18 or x36 operand. For signed two's complement operands, sign extension of the most significant bit should be performed until x9, x18 or x36 width is reached. Table 2-7 provides an example of this.

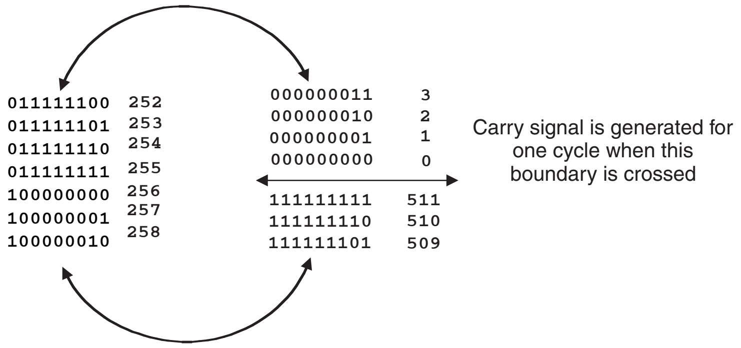
Table 2-7. Sign Extension Example

Number	Unsigned	Unsigned 9-bit	Unsigned 18-bit	Signed	Two's Complement Signed 9 Bits	Two's Complement Signed 18 Bits
+5	0101	000000101	000000000000000101	0101	000000101	000000000000000101
-6	N/A	N/A	N/A	1010	111111010	111111111111111010

OVERFLOW Flag from MAC

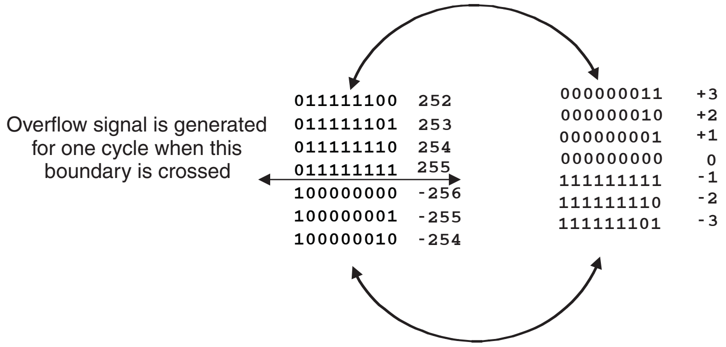
The sysDSP block provides an overflow output to indicate that the accumulator has overflowed. "Roll-over" occurs and an overflow signal is indicated when any of the following is true: two unsigned numbers are added and the result is a smaller number than the accumulator, two positive numbers are added with a negative sum or two negative numbers are added with a positive sum. Note that when overflow occurs the overflow flag is present for only one cycle. By counting these overflow pulses in FPGA logic, larger accumulators can be constructed. The conditions for the overflow signal for signed and unsigned operands are listed in Figure 2-24.

Figure 2-24. Accumulator Overflow/Underflow



Carry signal is generated for one cycle when this boundary is crossed

Unsigned Operation



Overflow signal is generated for one cycle when this boundary is crossed

Signed Operation

Table 2-11. PIO Signal List

Name	Type	Description
CE	Control from the core	Clock enables for input and output block flip-flops
CLK	Control from the core	System clocks for input and output blocks
ECLK1, ECLK2	Control from the core	Fast edge clocks
LSR	Control from the core	Local Set/Reset
GSRN	Control from routing	Global Set/Reset (active low)
INCK[2]	Input to the core	Input to Primary Clock Network or PLL reference inputs
DQS	Input to PIO	DQS signal from logic (routing) to PIO
INDD	Input to the core	Unregistered data input to core
INFF	Input to the core	Registered input on positive edge of the clock (CLK0)
IPOS0, IPOS1	Input to the core	Double data rate registered inputs to the core
QPOS0[1], QPOS1[1]	Input to the core	Gearbox pipelined inputs to the core
QNEG0[1], QNEG1[1]	Input to the core	Gearbox pipelined inputs to the core
OPOS0, ONEG0, OPOS2, ONEG2	Output data from the core	Output signals from the core for SDR and DDR operation
OPOS1 ONEG1	Tristate control from the core	Signals to Tristate Register block for DDR operation
DEL[3:0]	Control from the core	Dynamic input delay control bits
TD	Tristate control from the core	Tristate signal from the core used in SDR operation
DDRCLKPOL	Control from clock polarity bus	Controls the polarity of the clock (CLK0) that feed the DDR input block
DQSXFER	Control from core	Controls signal to the Output block

1. Signals available on left/right/bottom only.
2. Selected I/O.

PIO

The PIO contains four blocks: an input register block, output register block, tristate register block and a control logic block. These blocks contain registers for operating in a variety of modes along with necessary clock and selection logic.

Input Register Block

The input register blocks for PIOs contain delay elements and registers that can be used to condition high-speed interface signals, such as DDR memory interfaces and source synchronous interfaces, before they are passed to the device core. Figure 2-26 shows the diagram of the input register block.

Input signals are fed from the sysIO buffer to the input register block (as signal DI). If desired, the input signal can bypass the register and delay elements and be used directly as a combinatorial signal (INDD), a clock (INCK) and, in selected blocks, the input to the DQS delay block. If an input delay is desired, designers can select either a fixed delay or a dynamic delay DEL[3:0]. The delay, if selected, reduces input register hold time requirements when using a global clock.

The input block allows three modes of operation. In the Single Data Rate (SDR) mode, the data is registered, by one of the registers in the SDR Sync register block, with the system clock. In DDR mode two registers are used to sample the data on the positive and negative edges of the DQS signal which creates two data streams, D0 and D2. D0 and D2 are synchronized with the system clock before entering the core. Further information on this topic can be found in the DDR Memory Support section of this data sheet.

By combining input blocks of the complementary PIOs and sharing registers from output blocks, a gearbox function can be implemented, that takes a double data rate signal applied to PIOA and converts it as four data streams, IPOS0A, IPOS1A, IPOS0B and IPOS1B. Figure 2-26 shows the diagram using this gearbox function. For more information on this topic, please see TN1138, LatticeXP2 High Speed I/O Interface.

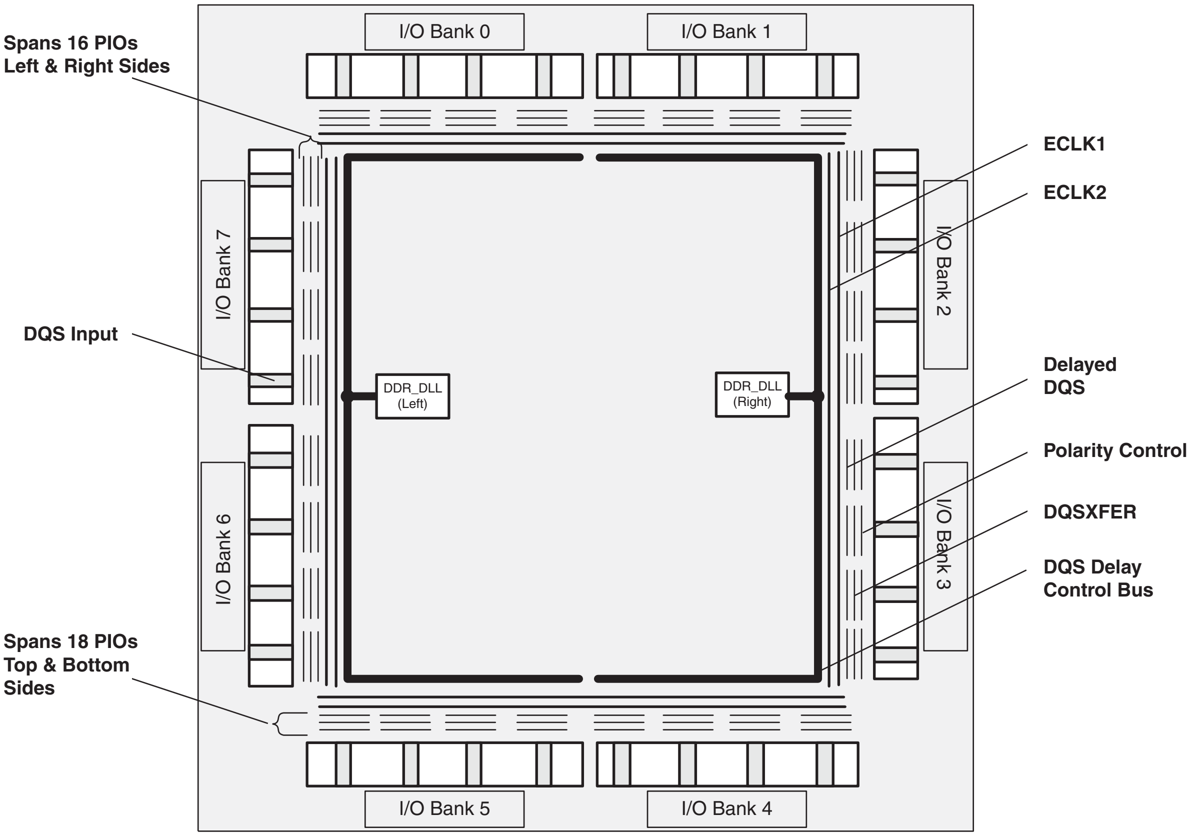
DLL Calibrated DQS Delay Block

Source synchronous interfaces generally require the input clock to be adjusted in order to correctly capture data at the input register. For most interfaces a PLL is used for this adjustment. However, in DDR memories the clock, referred to as DQS, is not free-running, and this approach cannot be used. The DQS Delay block provides the required clock alignment for DDR memory interfaces.

The DQS signal (selected PIOs only, as shown in Figure 2-30) feeds from the PAD through a DQS delay element to a dedicated DQS routing resource. The DQS signal also feeds polarity control logic which controls the polarity of the clock to the sync registers in the input register blocks. Figure 2-30 and Figure 2-31 show how the DQS transition signals are routed to the PIOs.

The temperature, voltage and process variations of the DQS delay block are compensated by a set of 6-bit bus calibration signals from two dedicated DLLs (DDR_DLL) on opposite sides of the device. Each DLL compensates DQS delays in its half of the device as shown in Figure 2-30. The DLL loop is compensated for temperature, voltage and process variations by the system clock and feedback loop.

Figure 2-30. Edge Clock, DLL Calibration and DQS Local Bus Distribution



DQSXFER

LatticeXP2 devices provide a DQSXFER signal to the output buffer to assist it in data transfer to DDR memories that require DQS strobe be shifted 90°. This shifted DQS strobe is generated by the DQSDEL block. The DQSXFER signal runs the span of the data bus.

sysIO Buffer

Each I/O is associated with a flexible buffer referred to as a sysIO buffer. These buffers are arranged around the periphery of the device in groups referred to as banks. The sysIO buffers allow users to implement the wide variety of standards that are found in today's systems including LVCMOS, SSTL, HSTL, LVDS and LVPECL.

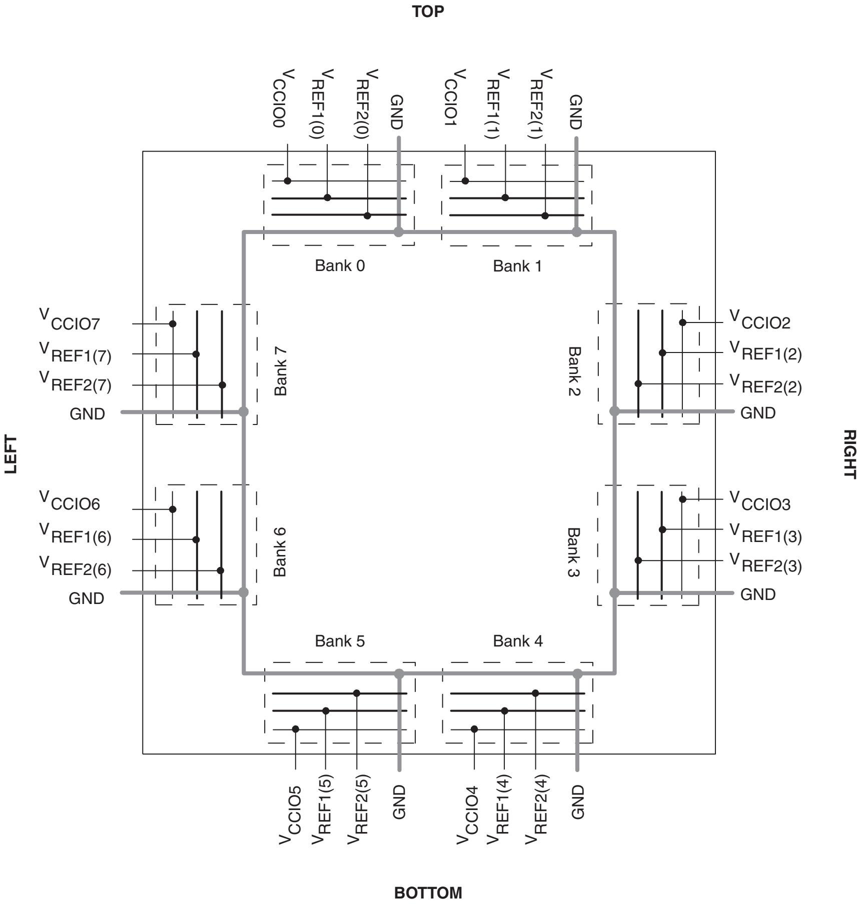
sysIO Buffer Banks

LatticeXP2 devices have eight sysIO buffer banks for user I/Os arranged two per side. Each bank is capable of supporting multiple I/O standards. Each sysIO bank has its own I/O supply voltage (V_{CCIO}). In addition, each bank has voltage references, V_{REF1} and V_{REF2}, that allow it to be completely independent from the others. Figure 2-32 shows the eight banks and their associated supplies.

In LatticeXP2 devices, single-ended output buffers and ratioed input buffers (LVTTL, LVCMOS and PCI) are powered using V_{CCIO}. LVTTL, LVCMOS33, LVCMOS25 and LVCMOS12 can also be set as fixed threshold inputs independent of V_{CCIO}.

Each bank can support up to two separate V_{REF} voltages, V_{REF1} and V_{REF2}, that set the threshold for the referenced input buffers. Some dedicated I/O pins in a bank can be configured to be a reference voltage supply pin. Each I/O is individually configurable based on the bank's supply and reference voltages.

Figure 2-32. LatticeXP2 Banks



LatticeXP2 devices contain two types of sysIO buffer pairs.

1. **Top and Bottom (Banks 0, 1, 4 and 5) sysIO Buffer Pairs (Single-Ended Outputs Only)**
   The sysIO buffer pairs in the top banks of the device consist of two single-ended output drivers and two sets of single-ended input buffers (both ratioed and referenced). One of the referenced input buffers can also be configured as a differential input.

   The two pads in the pair are described as “true” and “comp”, where the true pad is associated with the positive side of the differential input buffer and the comp (complementary) pad is associated with the negative side of the differential input buffer.

   Only the I/Os on the top and bottom banks have programmable PCI clamps.

2. **Left and Right (Banks 2, 3, 6 and 7) sysIO Buffer Pairs (50% Differential and 100% Single-Ended Outputs)**
   The sysIO buffer pairs in the left and right banks of the device consist of two single-ended output drivers, two sets of single-ended input buffers (both ratioed and referenced) and one differential output driver. One of the referenced input buffers can also be configured as a differential input.

   The two pads in the pair are described as “true” and “comp”, where the true pad is associated with the positive side of the differential I/O, and the comp pad is associated with the negative side of the differential I/O.

   LVDS differential output drivers are available on 50% of the buffer pairs on the left and right banks.

Typical sysIO I/O Behavior During Power-up

The internal power-on-reset (POR) signal is deactivated when $V_{CC,}$ $V_{CCCONFIG}$ (V_{CCIO7}) and V_{CCAUX} have reached satisfactory levels. After the POR signal is deactivated, the FPGA core logic becomes active. It is the user’s responsibility to ensure that all other V_{CCIO} banks are active with valid input logic levels to properly control the output logic states of all the I/O banks that are critical to the application. During power up and before the FPGA core logic becomes active, all user I/Os will be high-impedance with weak pull-up. Please refer to TN1136, LatticeXP2 sysIO Usage Guide for additional information.

The V_{CC} and V_{CCAUX} supply the power to the FPGA core fabric, whereas the V_{CCIO} supplies power to the I/O buffers. In order to simplify system design while providing consistent and predictable I/O behavior, it is recommended that the I/O buffers be powered-up prior to the FPGA core fabric. V_{CCIO} supplies should be powered-up before or together with the V_{CC} and V_{CCAUX} supplies.

Supported sysIO Standards

The LatticeXP2 sysIO buffer supports both single-ended and differential standards. Single-ended standards can be further subdivided into LVCMOS, LVTTL and other standards. The buffers support the LVTTL, LVCMOS 1.2V, 1.5V, 1.8V, 2.5V and 3.3V standards. In the LVCMOS and LVTTL modes, the buffer has individual configuration options for drive strength, bus maintenance (weak pull-up, weak pull-down, or a bus-keeper latch) and open drain. Other single-ended standards supported include SSTL and HSTL. Differential standards supported include LVDS, MLVDS, BLVDS, LVPECL, RSDS, differential SSTL and differential HSTL. Tables 2-12 and 2-13 show the I/O standards (together with their supply and reference voltages) supported by LatticeXP2 devices. For further information on utilizing the sysIO buffer to support a variety of standards please see TN1136, LatticeXP2 sysIO Usage Guide.

Table 2-13. Supported Output Standards

Output Standard	Drive	V_{CCIO} (Nom.)
Single-ended Interfaces		
LVTTL	4mA, 8mA, 12mA, 16mA, 20mA	3.3
LVCMOS33	4mA, 8mA, 12mA 16mA, 20mA	3.3
LVCMOS25	4mA, 8mA, 12mA, 16mA, 20mA	2.5
LVCMOS18	4mA, 8mA, 12mA, 16mA	1.8
LVCMOS15	4mA, 8mA	1.5
LVCMOS12	2mA, 6mA	1.2
LVCMOS33, Open Drain	4mA, 8mA, 12mA 16mA, 20mA	—
LVCMOS25, Open Drain	4mA, 8mA, 12mA 16mA, 20mA	—
LVCMOS18, Open Drain	4mA, 8mA, 12mA 16mA	—
LVCMOS15, Open Drain	4mA, 8mA	—
LVCMOS12, Open Drain	2mA, 6mA	—
PCI33	N/A	3.3
HSTL18 Class I, II	N/A	1.8
HSTL15 Class I	N/A	1.5
SSTL33 Class I, II	N/A	3.3
SSTL25 Class I, II	N/A	2.5
SSTL18 Class I, II	N/A	1.8
Differential Interfaces		
Differential SSTL33, Class I, II	N/A	3.3
Differential SSTL25, Class I, II	N/A	2.5
Differential SSTL18, Class I, II	N/A	1.8
Differential HSTL18, Class I, II	N/A	1.8
Differential HSTL15, Class I	N/A	1.5
LVDS[1, 2]	N/A	2.5
MLVDS[1]	N/A	2.5
BLVDS[1]	N/A	2.5
LVPECL[1]	N/A	3.3
RSDS[1]	N/A	2.5
LVCMOS33D[1]	4mA, 8mA, 12mA, 16mA, 20mA	3.3

1. Emulated with external resistors.
2. On the left and right edges, LVDS outputs are supported with a dedicated differential output driver on 50% of the I/Os. This solution does not require external resistors at the driver.

Hot Socketing

LatticeXP2 devices have been carefully designed to ensure predictable behavior during power-up and power-down. Power supplies can be sequenced in any order. During power-up and power-down sequences, the I/Os remain in tri-state until the power supply voltage is high enough to ensure reliable operation. In addition, leakage into I/O pins is controlled to within specified limits. This allows for easy integration with the rest of the system. These capabilities make the LatticeXP2 ideal for many multiple power supply and hot-swap applications.

IEEE 1149.1-Compliant Boundary Scan Testability

All LatticeXP2 devices have boundary scan cells that are accessed through an IEEE 1149.1 compliant Test Access Port (TAP). This allows functional testing of the circuit board, on which the device is mounted, through a serial scan path that can access all critical logic nodes. Internal registers are linked internally, allowing test data to be shifted in

Programming and Erase Flash Supply Current[1, 2, 3, 4, 5]

Over Recommended Operating Conditions

Symbol	Parameter	Device	Typical (25°C, Max. Supply)[6]	Units
I_{CC}	Core Power Supply Current	XP2-5	17	mA
		XP2-8	21	mA
		XP2-17	28	mA
		XP2-30	36	mA
		XP2-40	50	mA
I_{CCAUX}	Auxiliary Power Supply Current[7]	XP2-5	64	mA
		XP2-8	66	mA
		XP2-17	83	mA
		XP2-30	87	mA
		XP2-40	88	mA
I_{CCPLL}	PLL Power Supply Current (per PLL)		0.1	mA
I_{CCIO}	Bank Power Supply Current (per Bank)		5	mA
I_{CCJ}	V_{CCJ} Power Supply Current[8]		14	mA

1. For further information on supply current, please see TN1139, Power Estimation and Management for LatticeXP2 Devices.
2. Assumes all outputs are tristated, all inputs are configured as LVCMOS and held at the V_{CCIO} or GND.
3. Frequency 0 MHz (excludes dynamic power from FPGA operation).
4. A specific configuration pattern is used that scales with the size of the device; consists of 75% PFU utilization, 50% EBR, and 25% I/O configuration.
5. Bypass or decoupling capacitor across the supply.
6. T_J = 25°C, power supplies at nominal voltage.
7. In fpBGA and ftBGA packages the PLLs are connected to and powered from the auxiliary power supply. For these packages, the actual auxiliary supply current is the sum of I_{CCAUX} and I_{CCPLL}. For csBGA, PQFP and TQFP packages the PLLs are powered independent of the auxiliary power supply.
8. When programming via JTAG.

BLVDS

The LatticeXP2 devices support the BLVDS standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel external resistor across the driver outputs. BLVDS is intended for use when multi-drop and bi-directional multi-point differential signaling is required. The scheme shown in Figure 3-2 is one possible solution for bi-directional multi-point differential signals.

Figure 3-2. BLVDS Multi-point Output Example

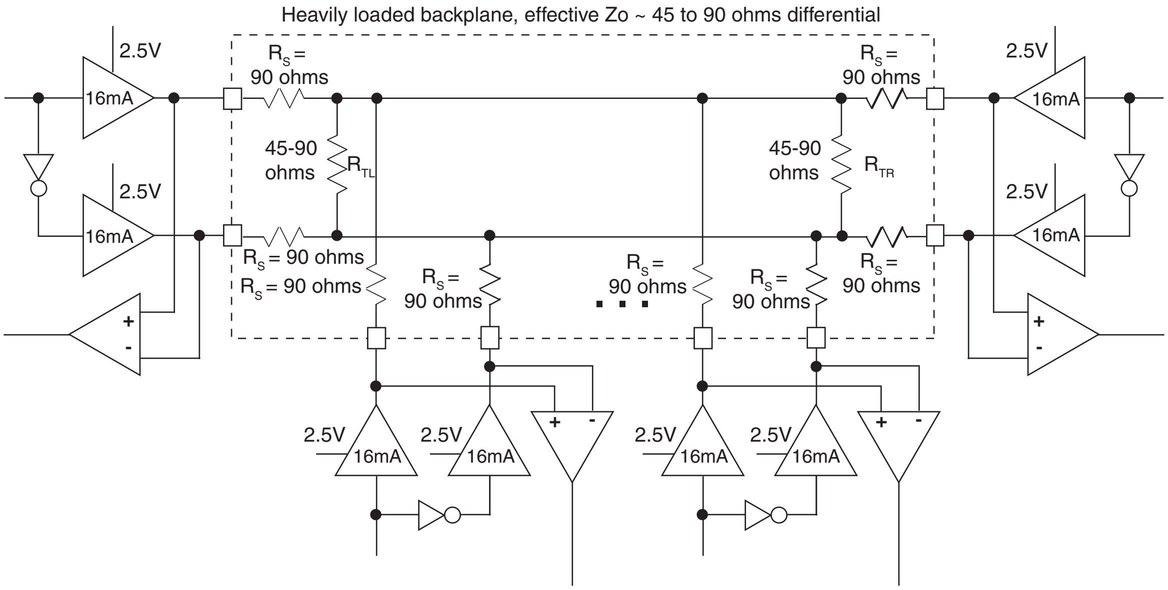


Table 3-2. BLVDS DC Conditions[1]

Over Recommended Operating Conditions

Parameter	Description	Typical		Units
		Zo = 45Ω	Zo = 90Ω	
V_{CCIO}	Output Driver Supply (+/- 5%)	2.50	2.50	V
Z_{OUT}	Driver Impedance	10.00	10.00	Ω
R_S	Driver Series Resistor (+/- 1%)	90.00	90.00	Ω
R_{TL}	Driver Parallel Resistor (+/- 1%)	45.00	90.00	Ω
R_{TR}	Receiver Termination (+/- 1%)	45.00	90.00	Ω
V_{OH}	Output High Voltage (After R_{TL})	1.38	1.48	V
V_{OL}	Output Low Voltage (After R_{TL})	1.12	1.02	V
V_{OD}	Output Differential Voltage (After R_{TL})	0.25	0.46	V
V_{CM}	Output Common Mode Voltage	1.25	1.25	V
I_{DC}	DC Output Current	11.24	10.20	mA

1. For input buffer, see LVDS table.

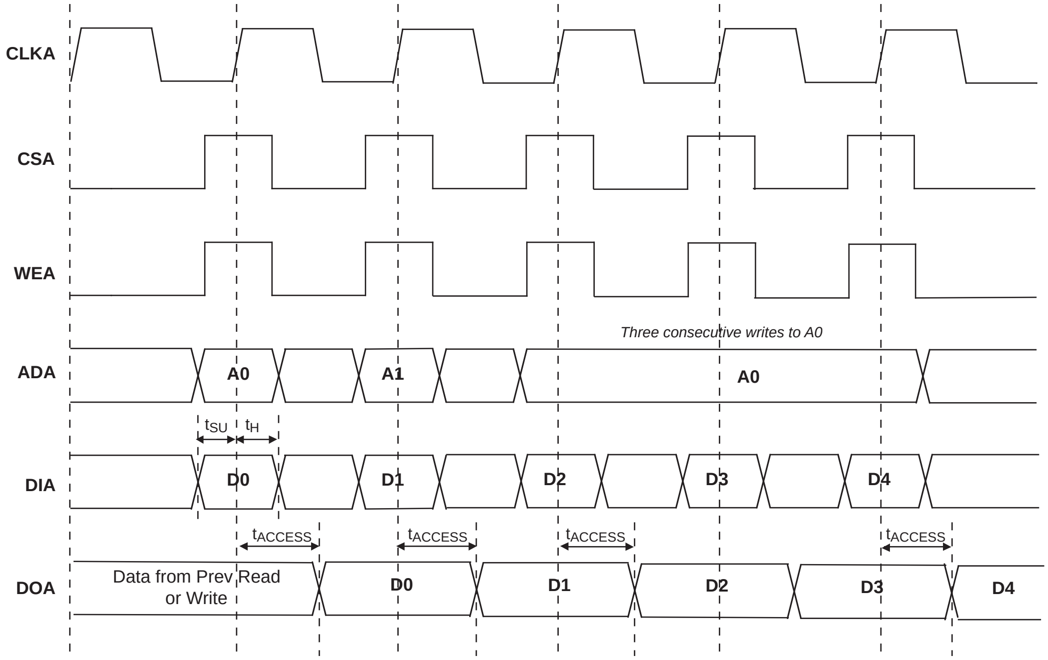
LatticeXP2 External Switching Characteristics (Continued)

Over Recommended Operating Conditions

Parameter	Description	Device	-7		-6		-5		Units
			Min.	Max.	Min.	Max.	Min.	Max.	
t_{H_DELPLL}	Clock to Data Hold - PIO Input Register with Input Data Delay	XP2-5	0.00	—	0.00	—	0.00	—	ns
		XP2-8	0.00	—	0.00	—	0.00	—	ns
		XP2-17	0.00	—	0.00	—	0.00	—	ns
		XP2-30	0.00	—	0.00	—	0.00	—	ns
		XP2-40	0.00	—	0.00	—	0.00	—	ns
DDR[2] and DDR2[3] I/O Pin Parameters									
t_{DVADQ}	Data Valid After DQS (DDR Read)	XP2	—	0.29	—	0.29	—	0.29	UI
t_{DVEDQ}	Data Hold After DQS (DDR Read)	XP2	0.71	—	0.71	—	0.71	—	UI
t_{DQVBS}	Data Valid Before DQS	XP2	0.25	—	0.25	—	0.25	—	UI
t_{DQVAS}	Data Valid After DQS	XP2	0.25	—	0.25	—	0.25	—	UI
f_{MAX_DDR}	DDR Clock Frequency	XP2	95	200	95	166	95	133	MHz
f_{MAX_DDR2}	DDR Clock Frequency	XP2	133	200	133	200	133	166	MHz
Primary Clock									
f_{MAX_PRI}	Frequency for Primary Clock Tree	XP2	—	420	—	357	—	311	MHz
t_{W_PRI}	Clock Pulse Width for Primary Clock	XP2	1	—	1	—	1	—	ns
t_{SKEW_PRI}	Primary Clock Skew Within a Bank	XP2	—	160	—	160	—	160	ps
Edge Clock (ECLK1 and ECLK2)									
f_{MAX_ECLK}	Frequency for Edge Clock	XP2	—	420	—	357	—	311	MHz
t_{W_ECLK}	Clock Pulse Width for Edge Clock	XP2	1	—	1	—	1	—	ns
t_{SKEW_ECLK}	Edge Clock Skew Within an Edge of the Device	XP2	—	130	—	130	—	130	ps

1. General timing numbers based on LVCMOS 2.5, 12mA, 0pf load.
2. DDR timing numbers based on SSTL25.
3. DDR2 timing numbers based on SSTL18.

Figure 3-8. Write Through (SP Read/Write on Port A, Input Registers Only)



Note: Input data and address are registered at the positive edge of the clock and output data appears after the positive edge of the clock.

LatticeXP2 sysCONFIG Port Timing Specifications

Over Recommended Operating Conditions

Parameter	Description	Min	Max	Units
sysCONFIG POR, Initialization and Wake Up				
t_{ICFG}	Minimum Vcc to INITN High	—	50	ms
t_{VMC}	Time from t_{ICFG} to valid Master CCLK	—	2	µs
t_{PRGMRJ}	PROGRAMN Pin Pulse Rejection	—	12	ns
t_{PRGM}	PROGRAMN Low Time to Start Configuration	50	—	ns
t_{DINIT}[1]	PROGRAMN High to INITN High Delay	—	1	ms
$t_{DPPINIT}$	Delay Time from PROGRAMN Low to INITN Low	—	50	ns
$t_{DPPDONE}$	Delay Time from PROGRAMN Low to DONE Low	—	50	ns
t_{IODISS}	User I/O Disable from PROGRAMN Low	—	35	ns
t_{IOENSS}	User I/O Enabled Time from CCLK Edge During Wake-up Sequence	—	25	ns
t_{MWC}	Additional Wake Master Clock Signals after DONE Pin High	0	—	Cycles
sysCONFIG SPI Port (Master)				
t_{CFGX}	INITN High to CCLK Low	—	1	µs
t_{CSSPI}	INITN High to CSSPIN Low	—	2	µs
t_{CSCCLK}	CCLK Low before CSSPIN Low	0	—	ns
t_{SOCDO}	CCLK Low to Output Valid	—	15	ns
t_{CSPID}	CSSPIN[0:1] Low to First CCLK Edge Setup Time	2cyc	600+6cyc	ns
f_{MAXSPI}	Max CCLK Frequency	—	20	MHz
t_{SUSPI}	SOSPI Data Setup Time Before CCLK	7	—	ns
t_{HSPI}	SOSPI Data Hold Time After CCLK	10	—	ns
sysCONFIG SPI Port (Slave)				
$f_{MAXSPIS}$	Slave CCLK Frequency	—	25	MHz
t_{RF}	Rise and Fall Time	50	—	mV/ns
t_{STCO}	Falling Edge of CCLK to SOSPI Active	—	20	ns
t_{STOZ}	Falling Edge of CCLK to SOSPI Disable	—	20	ns
t_{STSU}	Data Setup Time (SISPI)	8	—	ns
t_{STH}	Data Hold Time (SISPI)	10	—	ns
t_{STCKH}	CCLK Clock Pulse Width, High	0.02	200	µs
t_{STCKL}	CCLK Clock Pulse Width, Low	0.02	200	µs
t_{STVO}	Falling Edge of CCLK to Valid SOSPI Output	—	20	ns
t_{SCS}	CSSPISN High Time	25	—	ns
t_{SCSS}	CSSPISN Setup Time	25	—	ns
t_{SCSH}	CSSPISN Hold Time	25	—	ns

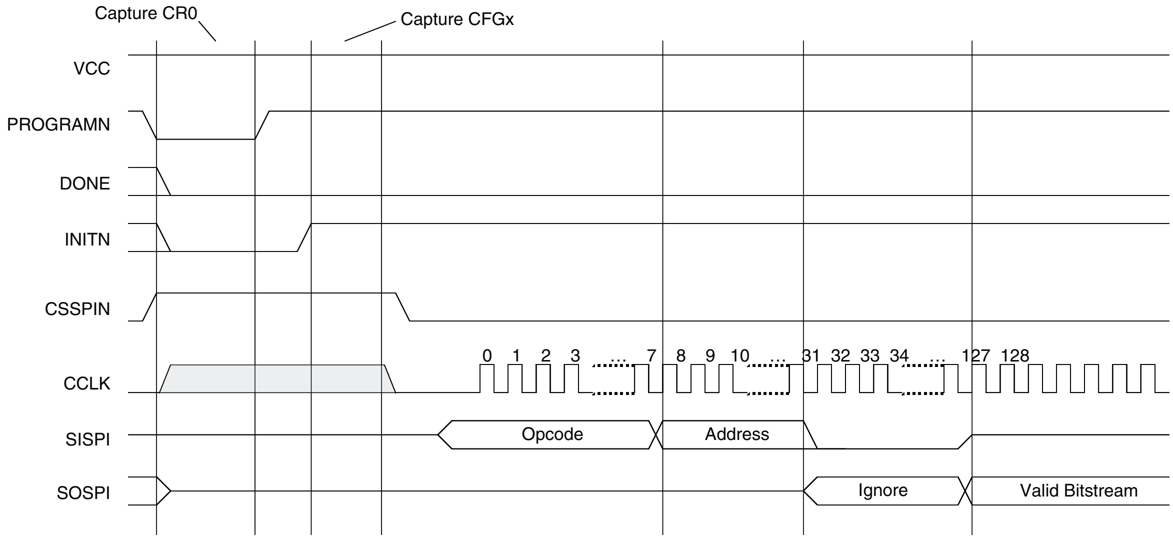
1. Re-toggling the PROGRAMN pin is not permitted until the INITN pin is high. Avoid consecutive toggling of PROGRAMN.

On-Chip Oscillator and Configuration Master Clock Characteristics

Over Recommended Operating Conditions

Parameter	Min.	Max.	Units
Master Clock Frequency	Selected value -30%	Selected value +30%	MHz
Duty Cycle	40	60	%

Figure 3-9. Master SPI Configuration Waveforms



Lead-Free Packaging

Commercial

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-5E-5MN132C	1.2V	-5	Lead-Free csBGA	132	COM	5
LFXP2-5E-6MN132C	1.2V	-6	Lead-Free csBGA	132	COM	5
LFXP2-5E-7MN132C	1.2V	-7	Lead-Free csBGA	132	COM	5
LFXP2-5E-5TN144C	1.2V	-5	Lead-Free TQFP	144	COM	5
LFXP2-5E-6TN144C	1.2V	-6	Lead-Free TQFP	144	COM	5
LFXP2-5E-7TN144C	1.2V	-7	Lead-Free TQFP	144	COM	5
LFXP2-5E-5QN208C	1.2V	-5	Lead-Free PQFP	208	COM	5
LFXP2-5E-6QN208C	1.2V	-6	Lead-Free PQFP	208	COM	5
LFXP2-5E-7QN208C	1.2V	-7	Lead-Free PQFP	208	COM	5
LFXP2-5E-5FTN256C	1.2V	-5	Lead-Free ftBGA	256	COM	5
LFXP2-5E-6FTN256C	1.2V	-6	Lead-Free ftBGA	256	COM	5
LFXP2-5E-7FTN256C	1.2V	-7	Lead-Free ftBGA	256	COM	5

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-8E-5MN132C	1.2V	-5	Lead-Free csBGA	132	COM	8
LFXP2-8E-6MN132C	1.2V	-6	Lead-Free csBGA	132	COM	8
LFXP2-8E-7MN132C	1.2V	-7	Lead-Free csBGA	132	COM	8
LFXP2-8E-5TN144C	1.2V	-5	Lead-Free TQFP	144	COM	8
LFXP2-8E-6TN144C	1.2V	-6	Lead-Free TQFP	144	COM	8
LFXP2-8E-7TN144C	1.2V	-7	Lead-Free TQFP	144	COM	8
LFXP2-8E-5QN208C	1.2V	-5	Lead-Free PQFP	208	COM	8
LFXP2-8E-6QN208C	1.2V	-6	Lead-Free PQFP	208	COM	8
LFXP2-8E-7QN208C	1.2V	-7	Lead-Free PQFP	208	COM	8
LFXP2-8E-5FTN256C	1.2V	-5	Lead-Free ftBGA	256	COM	8
LFXP2-8E-6FTN256C	1.2V	-6	Lead-Free ftBGA	256	COM	8
LFXP2-8E-7FTN256C	1.2V	-7	Lead-Free ftBGA	256	COM	8

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-17E-5QN208C	1.2V	-5	Lead-Free PQFP	208	COM	17
LFXP2-17E-6QN208C	1.2V	-6	Lead-Free PQFP	208	COM	17
LFXP2-17E-7QN208C	1.2V	-7	Lead-Free PQFP	208	COM	17
LFXP2-17E-5FTN256C	1.2V	-5	Lead-Free ftBGA	256	COM	17
LFXP2-17E-6FTN256C	1.2V	-6	Lead-Free ftBGA	256	COM	17
LFXP2-17E-7FTN256C	1.2V	-7	Lead-Free ftBGA	256	COM	17
LFXP2-17E-5FN484C	1.2V	-5	Lead-Free fpBGA	484	COM	17
LFXP2-17E-6FN484C	1.2V	-6	Lead-Free fpBGA	484	COM	17
LFXP2-17E-7FN484C	1.2V	-7	Lead-Free fpBGA	484	COM	17

Conventional Packaging

Commercial

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-5E-5M132C	1.2V	-5	csBGA	132	COM	5
LFXP2-5E-6M132C	1.2V	-6	csBGA	132	COM	5
LFXP2-5E-7M132C	1.2V	-7	csBGA	132	COM	5
LFXP2-5E-5FT256C	1.2V	-5	ftBGA	256	COM	5
LFXP2-5E-6FT256C	1.2V	-6	ftBGA	256	COM	5
LFXP2-5E-7FT256C	1.2V	-7	ftBGA	256	COM	5

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-8E-5M132C	1.2V	-5	csBGA	132	COM	8
LFXP2-8E-6M132C	1.2V	-6	csBGA	132	COM	8
LFXP2-8E-7M132C	1.2V	-7	csBGA	132	COM	8
LFXP2-8E-5FT256C	1.2V	-5	ftBGA	256	COM	8
LFXP2-8E-6FT256C	1.2V	-6	ftBGA	256	COM	8
LFXP2-8E-7FT256C	1.2V	-7	ftBGA	256	COM	8

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-17E-5FT256C	1.2V	-5	ftBGA	256	COM	17
LFXP2-17E-6FT256C	1.2V	-6	ftBGA	256	COM	17
LFXP2-17E-7FT256C	1.2V	-7	ftBGA	256	COM	17
LFXP2-17E-5F484C	1.2V	-5	fpBGA	484	COM	17
LFXP2-17E-6F484C	1.2V	-6	fpBGA	484	COM	17
LFXP2-17E-7F484C	1.2V	-7	fpBGA	484	COM	17

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-30E-5FT256C	1.2V	-5	ftBGA	256	COM	30
LFXP2-30E-6FT256C	1.2V	-6	ftBGA	256	COM	30
LFXP2-30E-7FT256C	1.2V	-7	ftBGA	256	COM	30
LFXP2-30E-5F484C	1.2V	-5	fpBGA	484	COM	30
LFXP2-30E-6F484C	1.2V	-6	fpBGA	484	COM	30
LFXP2-30E-7F484C	1.2V	-7	fpBGA	484	COM	30
LFXP2-30E-5F672C	1.2V	-5	fpBGA	672	COM	30
LFXP2-30E-6F672C	1.2V	-6	fpBGA	672	COM	30
LFXP2-30E-7F672C	1.2V	-7	fpBGA	672	COM	30