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## Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

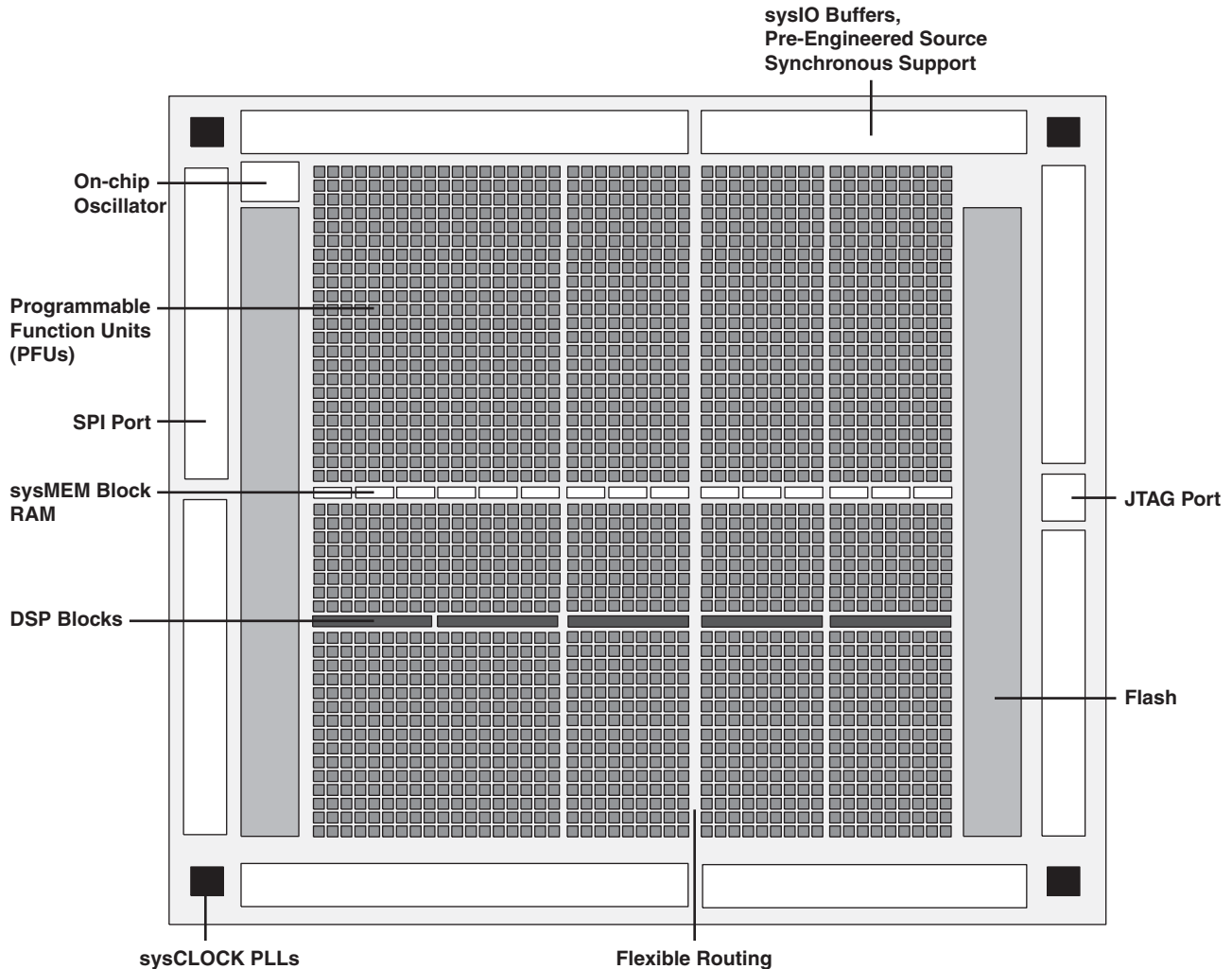
## Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

### Details

Product Status	Active
Number of LABs/CLBs	625
Number of Logic Elements/Cells	5000
Total RAM Bits	169984
Number of I/O	172
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FTBGA (17x17)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/lattice-semiconductor/lfxp2-5e-5ftn256c">https://www.e-xfl.com/product-detail/lattice-semiconductor/lfxp2-5e-5ftn256c</a>

**Figure 2-1. Simplified Block Diagram, LatticeXP2-17 Device (Top Level)**



## PFU Blocks

The core of the LatticeXP2 device is made up of logic blocks in two forms, PFUs and PFFs. PFUs can be programmed to perform logic, arithmetic, distributed RAM and distributed ROM functions. PFF blocks can be programmed to perform logic, arithmetic and ROM functions. Except where necessary, the remainder of this data sheet will use the term PFU to refer to both PFU and PFF blocks.

Each PFU block consists of four interconnected slices, numbered Slice 0 through Slice 3, as shown in Figure 2-2. All the interconnections to and from PFU blocks are from routing. There are 50 inputs and 23 outputs associated with each PFU block.

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## Routing

There are many resources provided in the LatticeXP2 devices to route signals individually or as busses with related control signals. The routing resources consist of switching circuitry, buffers and metal interconnect (routing) segments.

The inter-PFU connections are made with x1 (spans two PFU), x2 (spans three PFU) or x6 (spans seven PFU) connections. The x1 and x2 connections provide fast and efficient connections in horizontal and vertical directions. The x2 and x6 resources are buffered to allow both short and long connections routing between PFUs.

The LatticeXP2 family has an enhanced routing architecture to produce a compact design. The Diamond design tool takes the output of the synthesis tool and places and routes the design. Generally, the place and route tool is completely automatic, although an interactive routing editor is available to optimize the design.

## sysCLOCK Phase Locked Loops (PLL)

The sysCLOCK PLLs provide the ability to synthesize clock frequencies. The LatticeXP2 family supports between two and four full featured General Purpose PLLs (GPLL). The architecture of the GPLL is shown in Figure 2-4.

CLKI, the PLL reference frequency, is provided either from the pin or from routing; it feeds into the Input Clock Divider block. CLKFB, the feedback signal, is generated from CLKOP (the primary clock output) or from a user clock pin/logic. CLKFB feeds into the Feedback Divider and is used to multiply the reference frequency.

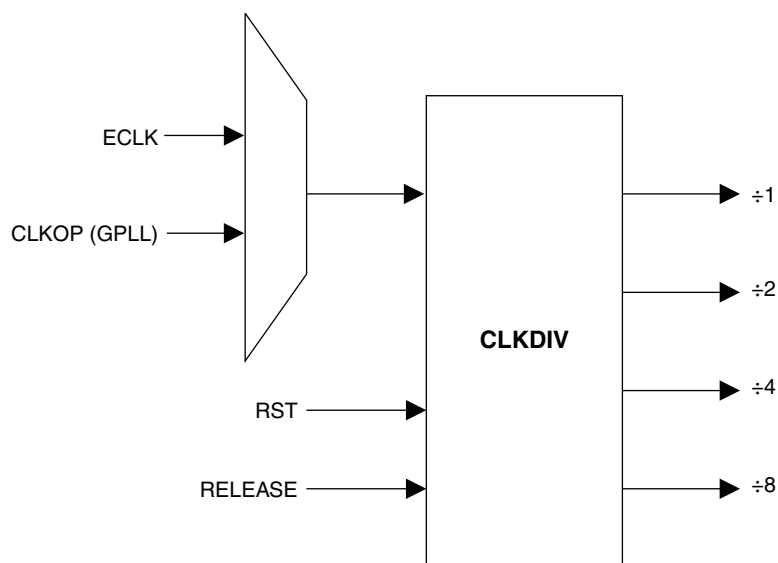
Both the input path and feedback signals enter the Voltage Controlled Oscillator (VCO) block. The phase and frequency of the VCO are determined from the input path and feedback signals. A LOCK signal is generated by the VCO to indicate that the VCO is locked with the input clock signal.

The output of the VCO feeds into the CLKOP Divider, a post-scalar divider. The duty cycle of the CLKOP Divider output can be fine tuned using the Duty Trim block, which creates the CLKOP signal. By allowing the VCO to operate at higher frequencies than CLKOP, the frequency range of the GPLL is expanded. The output of the CLKOP Divider is passed through the CLKOK Divider, a secondary clock divider, to generate lower frequencies for the CLKOK output. For applications that require even lower frequencies, the CLKOP signal is passed through a divide-by-three divider to produce the CLKOK2 output. The CLKOK2 output is provided for applications that use source synchronous logic. The Phase/Duty Cycle/Duty Trim block is used to adjust the phase and duty cycle of the CLKOP Divider output to generate the CLKOS signal. The phase/duty cycle setting can be pre-programmed or dynamically adjusted.

The clock outputs from the GPLL; CLKOP, CLKOK, CLKOK2 and CLKOS, are fed to the clock distribution network.

For further information on the GPLL please see TN1126, [LatticeXP2 sysCLOCK PLL Design and Usage Guide](#).

**Figure 2-5. Clock Divider Connections**



## Clock Distribution Network

LatticeXP2 devices have eight quadrant-based primary clocks and between six and eight flexible region-based secondary clocks/control signals. Two high performance edge clocks are available on each edge of the device to support high speed interfaces. The clock inputs are selected from external I/Os, the sysCLOCK PLLs, or routing. Clock inputs are fed throughout the chip via the primary, secondary and edge clock networks.

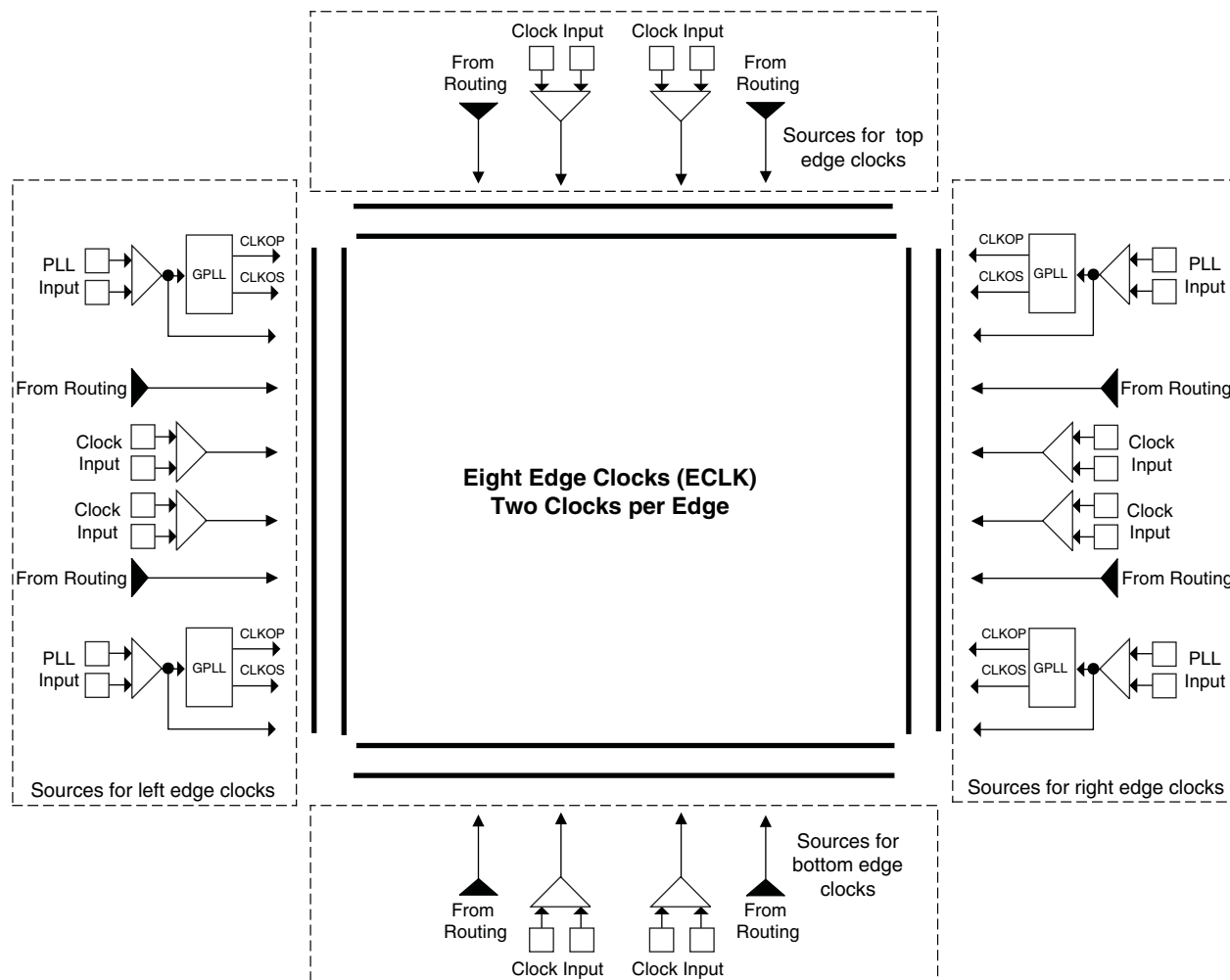
## Primary Clock Sources

LatticeXP2 devices derive primary clocks from four sources: PLL outputs, CLKDIV outputs, dedicated clock inputs and routing. LatticeXP2 devices have two to four sysCLOCK PLLs, located in the four corners of the device. There are eight dedicated clock inputs, two on each side of the device. Figure 2-6 shows the primary clock sources.

## Edge Clock Sources

Edge clock resources can be driven from a variety of sources at the same edge. Edge clock resources can be driven from adjacent edge clock PIOs, primary clock PIOs, PLLs and clock dividers as shown in Figure 2-8.

**Figure 2-8. Edge Clock Sources**

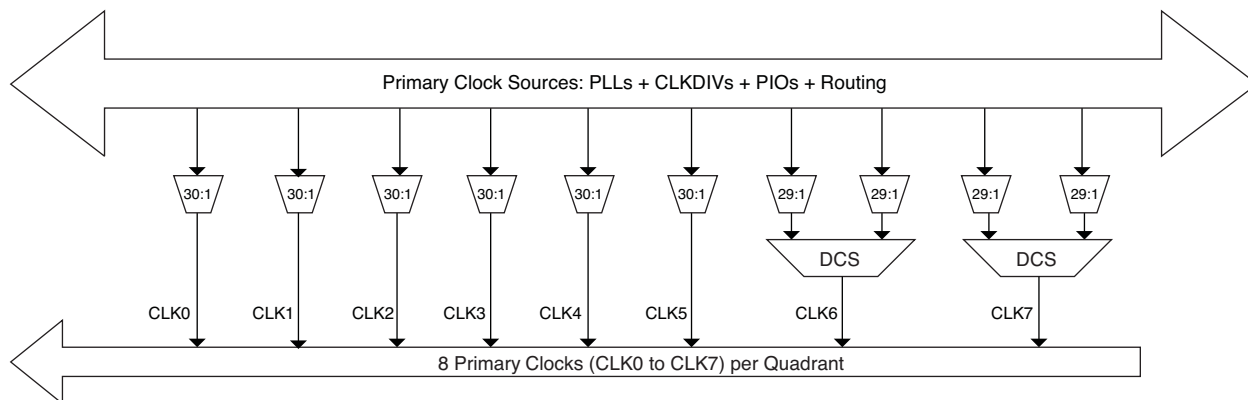


Note: This diagram shows sources for the XP2-17 device. Smaller LatticeXP2 devices have two GPLLs.

## Primary Clock Routing

The clock routing structure in LatticeXP2 devices consists of a network of eight primary clock lines (CLK0 through CLK7) per quadrant. The primary clocks of each quadrant are generated from muxes located in the center of the device. All the clock sources are connected to these muxes. Figure 2-9 shows the clock routing for one quadrant. Each quadrant mux is identical. If desired, any clock can be routed globally.

**Figure 2-9. Per Quadrant Primary Clock Selection**

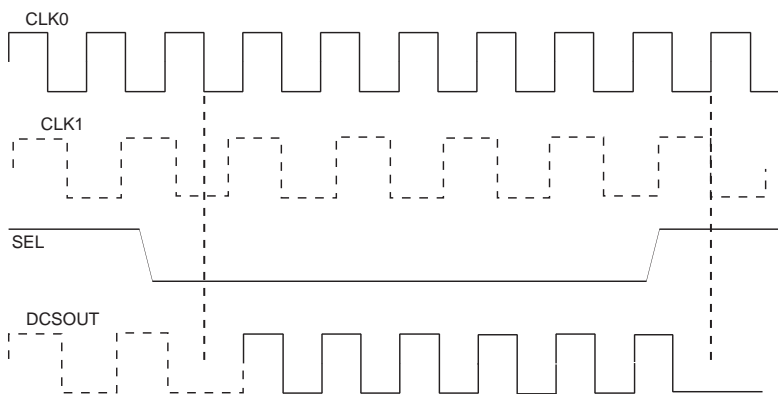


## Dynamic Clock Select (DCS)

The DCS is a smart multiplexer function available in the primary clock routing. It switches between two independent input clock sources without any glitches or runt pulses. This is achieved irrespective of when the select signal is toggled. There are two DCS blocks per quadrant; in total, eight DCS blocks per device. The inputs to the DCS block come from the center muxes. The output of the DCS is connected to primary clocks CLK6 and CLK7 (see Figure 2-9).

Figure 2-10 shows the timing waveforms of the default DCS operating mode. The DCS block can be programmed to other modes. For more information on the DCS, please see TN1126, [LatticeXP2 sysCLOCK PLL Design and Usage Guide](#).

**Figure 2-10. DCS Waveforms**



## Secondary Clock/Control Routing

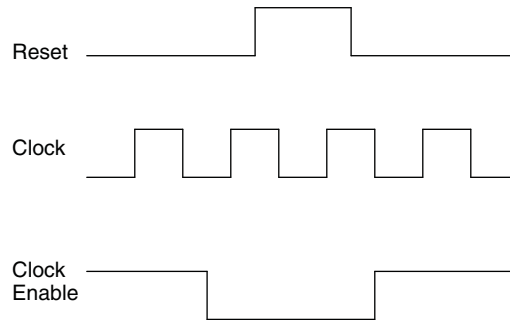
Secondary clocks in the LatticeXP2 devices are region-based resources. The benefit of region-based resources is the relatively low injection delay and skew within the region, as compared to primary clocks. EBR rows, DSP rows and a special vertical routing channel bound the secondary clock regions. This special vertical routing channel aligns with either the left edge of the center DSP block in the DSP row or the center of the DSP row. Figure 2-11 shows this special vertical routing channel and the eight secondary clock regions for the LatticeXP2-40.

For further information on the sysMEM EBR block, please see TN1137, [LatticeXP2 Memory Usage Guide](#).

### EBR Asynchronous Reset

EBR asynchronous reset or GSR (if used) can only be applied if all clock enables are low for a clock cycle before the reset is applied and released a clock cycle after the low-to-high transition of the reset signal, as shown in Figure 2-18. The GSR input to the EBR is always asynchronous.

**Figure 2-18. EBR Asynchronous Reset (Including GSR) Timing Diagram**



If all clock enables remain enabled, the EBR asynchronous reset or GSR may only be applied and released after the EBR read and write clock inputs are in a steady state condition for a minimum of  $1/f_{MAX}$  (EBR clock). The reset release must adhere to the EBR synchronous reset setup time before the next active read or write clock edge.

If an EBR is pre-loaded during configuration, the GSR input must be disabled or the release of the GSR during device Wake Up must occur before the release of the device I/Os becoming active.

These instructions apply to all EBR RAM and ROM implementations.

Note that there are no reset restrictions if the EBR synchronous reset is used and the EBR GSR input is disabled.

### sysDSP™ Block

The LatticeXP2 family provides a sysDSP block making it ideally suited for low cost, high performance Digital Signal Processing (DSP) applications. Typical functions used in these applications include Bit Correlators, Fast Fourier Transform (FFT) functions, Finite Impulse Response (FIR) Filter, Reed-Solomon Encoder/Decoder, Turbo Encoder/Decoder and Convolutional Encoder/Decoder. These complex signal processing functions use similar building blocks such as multiply-adders and multiply-accumulators.

### sysDSP Block Approach Compare to General DSP

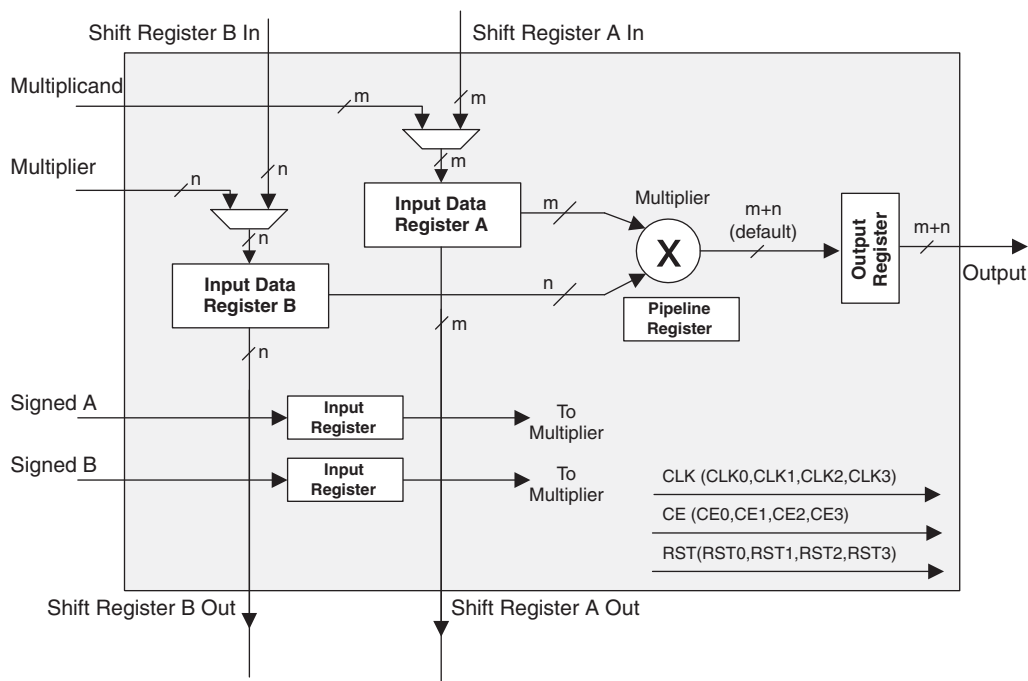
Conventional general-purpose DSP chips typically contain one to four (Multiply and Accumulate) MAC units with fixed data-width multipliers; this leads to limited parallelism and limited throughput. Their throughput is increased by higher clock speeds. The LatticeXP2 family, on the other hand, has many DSP blocks that support different data-widths. This allows the designer to use highly parallel implementations of DSP functions. The designer can optimize the DSP performance vs. area by choosing appropriate levels of parallelism. Figure 2-19 compares the fully serial and the mixed parallel and serial implementations.

- In the 'Signed/Unsigned' options the operands can be switched between signed and unsigned on every cycle.
- In the 'Add/Sub' option the Accumulator can be switched between addition and subtraction on every cycle.
- The loading of operands can switch between parallel and serial operations.

## MULT sysDSP Element

This multiplier element implements a multiply with no addition or accumulator nodes. The two operands, A and B, are multiplied and the result is available at the output. The user can enable the input/output and pipeline registers. Figure 2-20 shows the MULT sysDSP element.

**Figure 2-20. MULT sysDSP Element**

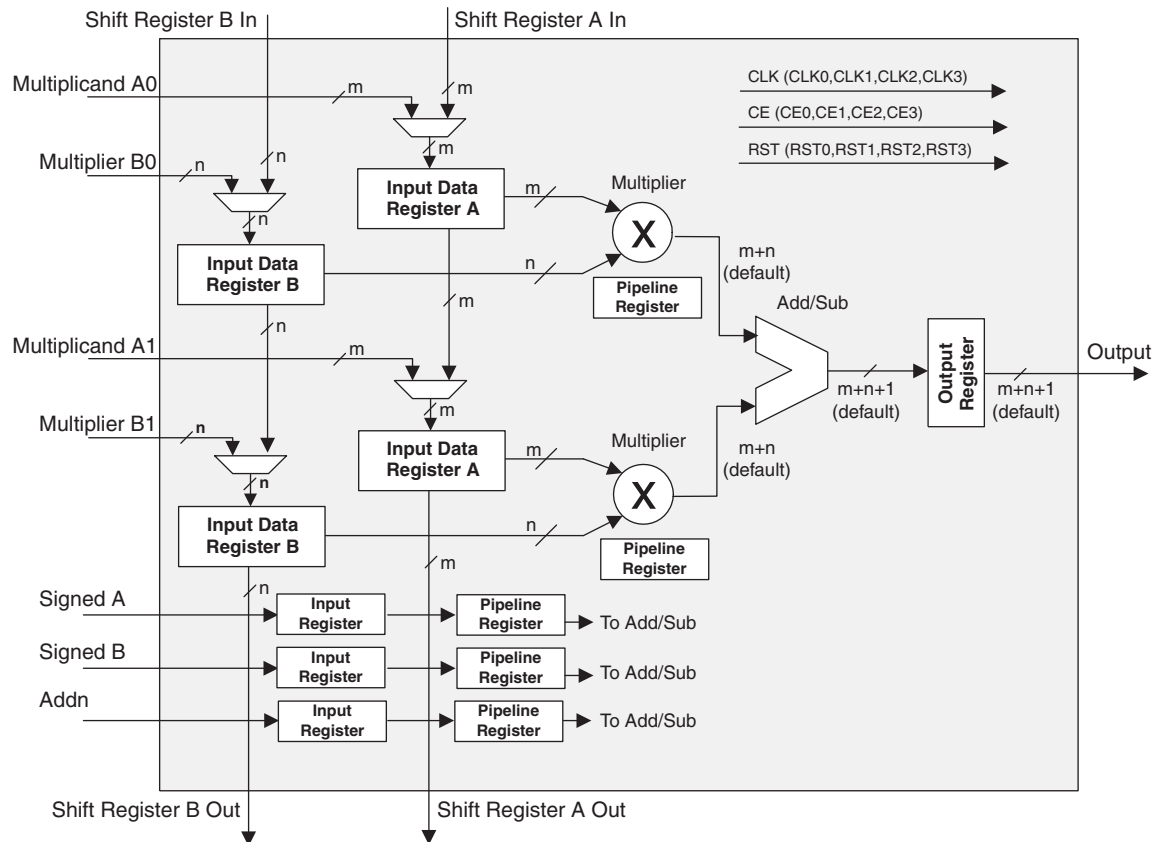




## MULTADDSUB sysDSP Element

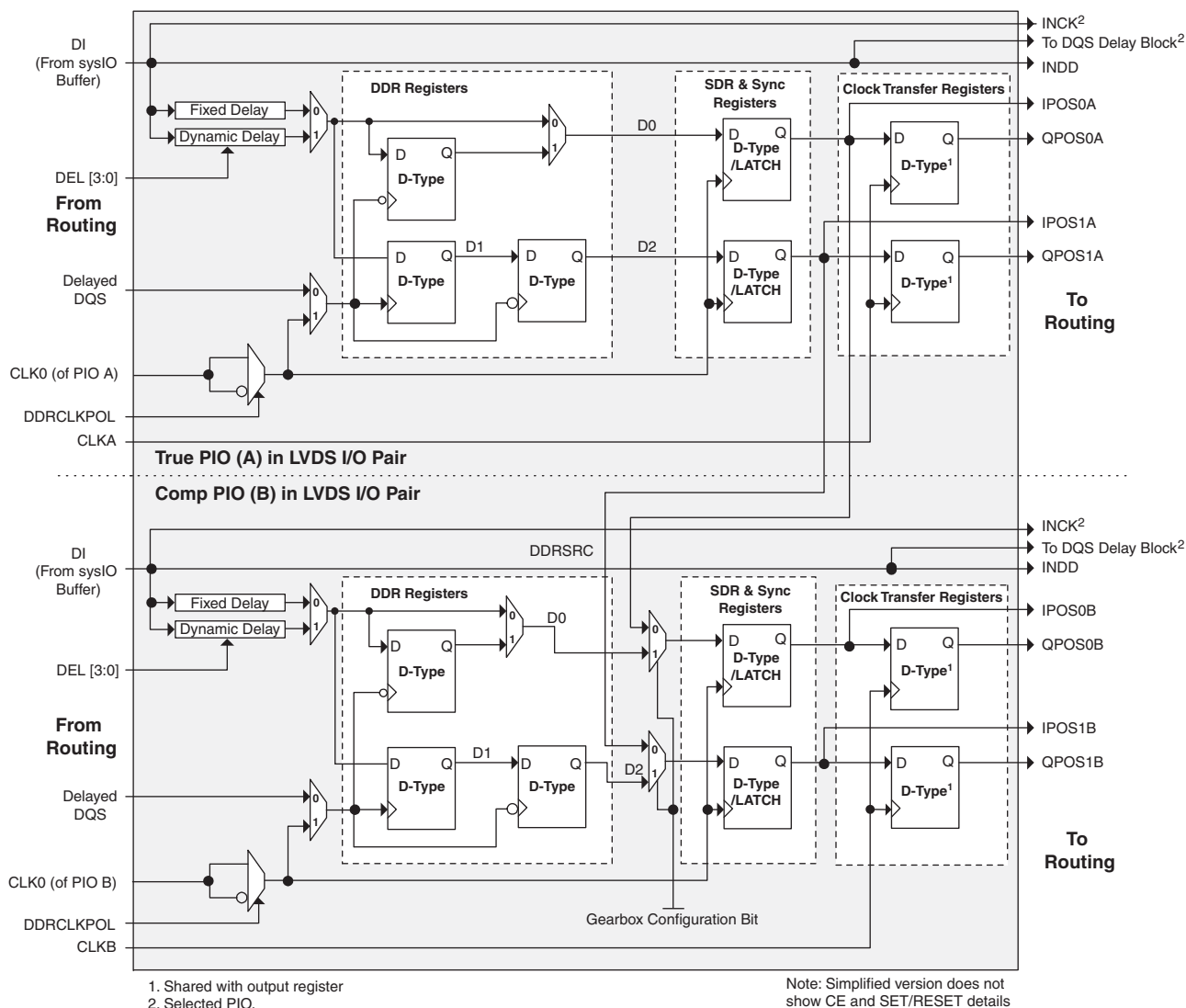
In this case, the operands A0 and B0 are multiplied and the result is added/subtracted with the result of the multiplier operation of operands A1 and B1. The user can enable the input, output and pipeline registers. Figure 2-22 shows the MULTADDSUB sysDSP element.

**Figure 2-22. MULTADDSUB**



The signal DDRCLKPOL controls the polarity of the clock used in the synchronization registers. It ensures adequate timing when data is transferred from the DQS to system clock domain. For further discussion on this topic, see the DDR Memory section of this data sheet.

**Figure 2-26. Input Register Block**



## Output Register Block

The output register block provides the ability to register signals from the core of the device before they are passed to the sysIO buffers. The blocks on the PIOs on the left, right and bottom contain registers for SDR operation that are combined with an additional latch for DDR operation. Figure 2-27 shows the diagram of the Output Register Block for PIOs.

In SDR mode, ONEG0 feeds one of the flip-flops that then feeds the output. The flip-flop can be configured as a D-type or latch. In DDR mode, ONEG0 and OPOS0 are fed into registers on the positive edge of the clock. At the next clock cycle the registered OPOS0 is latched. A multiplexer running off the same clock cycle selects the correct register to feed the output (D0).

By combining output blocks of the complementary PIOs and sharing some registers from input blocks, a gearbox function can be implemented, to take four data streams ONEG0A, ONEG1A, ONEG1B and ONEG1B. Figure 2-27

## DQSXFER

LatticeXP2 devices provide a DQSXFER signal to the output buffer to assist it in data transfer to DDR memories that require DQS strobe be shifted 90°. This shifted DQS strobe is generated by the DQSDEL block. The DQSXFER signal runs the span of the data bus.

## sysIO Buffer

Each I/O is associated with a flexible buffer referred to as a sysIO buffer. These buffers are arranged around the periphery of the device in groups referred to as banks. The sysIO buffers allow users to implement the wide variety of standards that are found in today's systems including LVCMOS, SSTL, HSTL, LVDS and LVPECL.

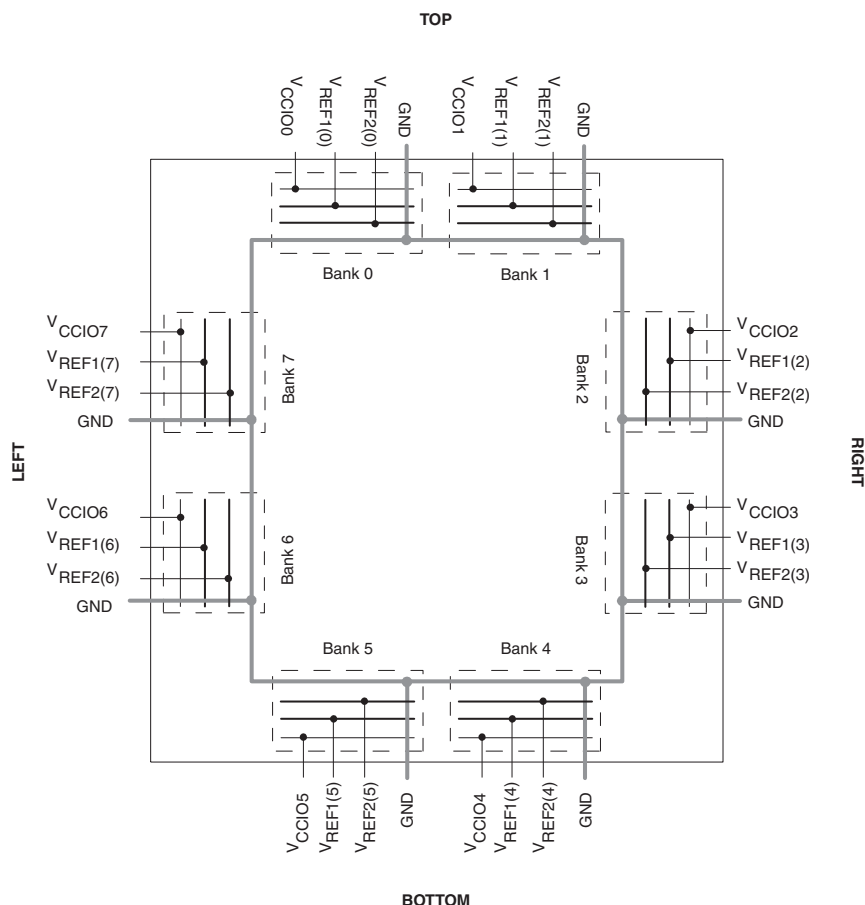
## sysIO Buffer Banks

LatticeXP2 devices have eight sysIO buffer banks for user I/Os arranged two per side. Each bank is capable of supporting multiple I/O standards. Each sysIO bank has its own I/O supply voltage ( $V_{CCIO}$ ). In addition, each bank has voltage references,  $V_{REF1}$  and  $V_{REF2}$ , that allow it to be completely independent from the others. Figure 2-32 shows the eight banks and their associated supplies.

In LatticeXP2 devices, single-ended output buffers and ratioed input buffers (LVTTL, LVCMOS and PCI) are powered using  $V_{CCIO}$ . LVTTL, LVCMOS33, LVCMOS25 and LVCMOS12 can also be set as fixed threshold inputs independent of  $V_{CCIO}$ .

Each bank can support up to two separate  $V_{REF}$  voltages,  $V_{REF1}$  and  $V_{REF2}$ , that set the threshold for the referenced input buffers. Some dedicated I/O pins in a bank can be configured to be a reference voltage supply pin. Each I/O is individually configurable based on the bank's supply and reference voltages.

**Figure 2-32. LatticeXP2 Banks**



**Supply Current (Standby)<sup>1, 2, 3, 4</sup>**
**Over Recommended Operating Conditions**

Symbol	Parameter	Device	Typical <sup>5</sup>	Units
$I_{CC}$	Core Power Supply Current	XP2-5	14	mA
		XP2-8	18	mA
		XP2-17	24	mA
		XP2-30	35	mA
		XP2-40	45	mA
$I_{CCAUX}$	Auxiliary Power Supply Current <sup>6</sup>	XP2-5	15	mA
		XP2-8	15	mA
		XP2-17	15	mA
		XP2-30	16	mA
		XP2-40	16	mA
$I_{CCPLL}$	PLL Power Supply Current (per PLL)		0.1	mA
$I_{CCIO}$	Bank Power Supply Current (per bank)		2	mA
$I_{CCJ}$	$V_{CCJ}$ Power Supply Current		0.25	mA

1. For further information on supply current, please see TN1139, [Power Estimation and Management for LatticeXP2 Devices](#).
2. Assumes all outputs are tristated, all inputs are configured as LVCMOS and held at the  $V_{CCIO}$  or GND.
3. Frequency 0 MHz.
4. Pattern represents a "blank" configuration data file.
5.  $T_J = 25^\circ\text{C}$ , power supplies at nominal voltage.
6. In fpBGA and ftBGA packages the PLLs are connected to and powered from the auxiliary power supply. For these packages, the actual auxiliary supply current is the sum of  $I_{CCAUX}$  and  $I_{CCPLL}$ . For csBGA, PQFP and TQFP packages the PLLs are powered independent of the auxiliary power supply.

## sysIO Recommended Operating Conditions

### Over Recommended Operating Conditions

Standard	V <sub>CCIO</sub>			V <sub>REF</sub> (V)		
	Min.	Typ.	Max.	Min.	Typ.	Max.
LVC MOS33 <sup>2</sup>	3.135	3.3	3.465	—	—	—
LVC MOS25 <sup>2</sup>	2.375	2.5	2.625	—	—	—
LVC MOS18	1.71	1.8	1.89	—	—	—
LVC MOS15	1.425	1.5	1.575	—	—	—
LVC MOS12 <sup>2</sup>	1.14	1.2	1.26	—	—	—
LV TTL33 <sup>2</sup>	3.135	3.3	3.465	—	—	—
PCI33	3.135	3.3	3.465	—	—	—
SSTL18_I <sup>2</sup> , SSTL18_II <sup>2</sup>	1.71	1.8	1.89	0.833	0.9	0.969
SSTL25_I <sup>2</sup> , SSTL25_II <sup>2</sup>	2.375	2.5	2.625	1.15	1.25	1.35
SSTL33_I <sup>2</sup> , SSTL33_II <sup>2</sup>	3.135	3.3	3.465	1.3	1.5	1.7
HSTL15_I <sup>2</sup>	1.425	1.5	1.575	0.68	0.75	0.9
HSTL18_I <sup>2</sup> , HSTL18_II <sup>2</sup>	1.71	1.8	1.89	0.816	0.9	1.08
LVDS25 <sup>2</sup>	2.375	2.5	2.625	—	—	—
MLVDS25 <sup>1</sup>	2.375	2.5	2.625	—	—	—
LVPECL33 <sup>1, 2</sup>	3.135	3.3	3.465	—	—	—
BLVDS25 <sup>1, 2</sup>	2.375	2.5	2.625	—	—	—
RSDS <sup>1, 2</sup>	2.375	2.5	2.625	—	—	—
SSTL18D_I <sup>2</sup> , SSTL18D_II <sup>2</sup>	1.71	1.8	1.89	—	—	—
SSTL25D_I <sup>2</sup> , SSTL25D_II <sup>2</sup>	2.375	2.5	2.625	—	—	—
SSTL33D_I <sup>2</sup> , SSTL33D_II <sup>2</sup>	3.135	3.3	3.465	—	—	—
HSTL15D_I <sup>2</sup>	1.425	1.5	1.575	—	—	—
HSTL18D_I <sup>2</sup> , HSTL18D_II <sup>2</sup>	1.71	1.8	1.89	—	—	—

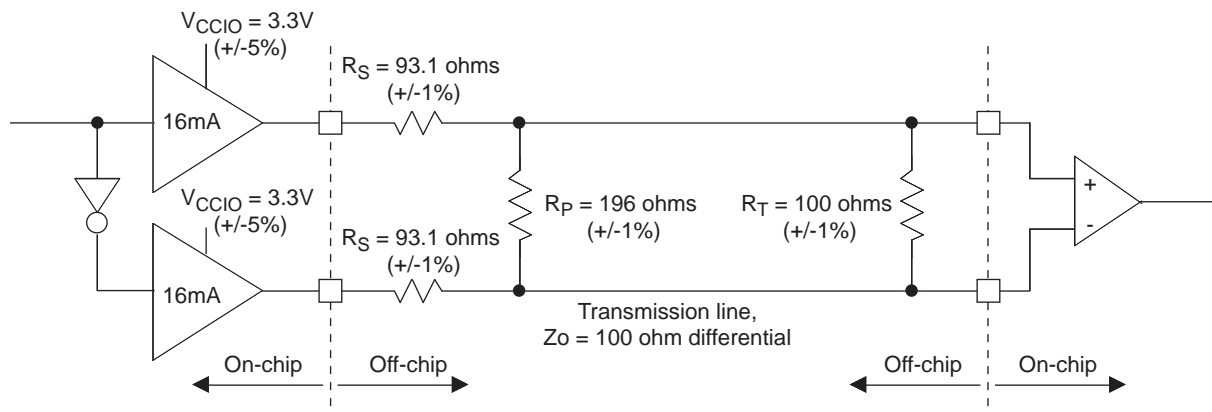
1. Inputs on chip. Outputs are implemented with the addition of external resistors.

2. Input on this standard does not depend on the value of V<sub>CCIO</sub>.

### LVPECL

The LatticeXP2 devices support the differential LVPECL standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The LVPECL input standard is supported by the LVDS differential input buffer. The scheme shown in Figure 3-3 is one possible solution for point-to-point signals.

**Figure 3-3. Differential LVPECL**



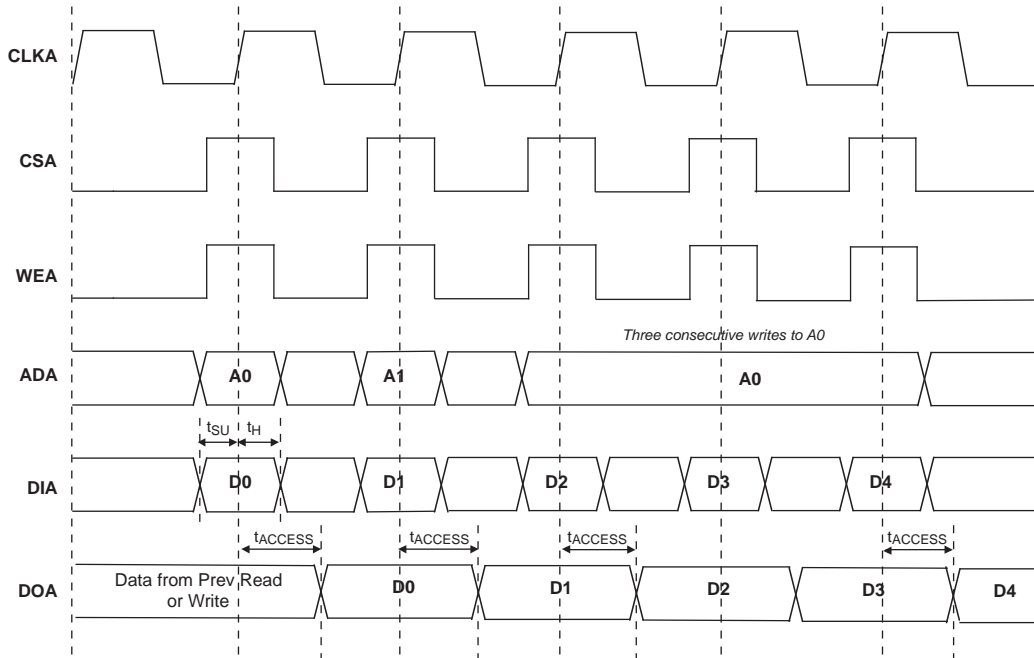
**Table 3-3. LVPECL DC Conditions<sup>1</sup>**

#### Over Recommended Operating Conditions

Parameter	Description	Typical	Units
$V_{CCIO}$	Output Driver Supply (+/-5%)	3.30	V
$Z_{OUT}$	Driver Impedance	10	$\Omega$
$R_S$	Driver Series Resistor (+/-1%)	93	$\Omega$
$R_P$	Driver Parallel Resistor (+/-1%)	196	$\Omega$
$R_T$	Receiver Termination (+/-1%)	100	$\Omega$
$V_{OH}$	Output High Voltage (After $R_P$ )	2.05	V
$V_{OL}$	Output Low Voltage (After $R_P$ )	1.25	V
$V_{OD}$	Output Differential Voltage (After $R_P$ )	0.80	V
$V_{CM}$	Output Common Mode Voltage	1.65	V
$Z_{BACK}$	Back Impedance	100.5	$\Omega$
$I_{DC}$	DC Output Current	12.11	mA

1. For input buffer, see LVDS table.

**Figure 3-8. Write Through (SP Read/Write on Port A, Input Registers Only)**



Note: Input data and address are registered at the positive edge of the clock and output data appears after the positive edge of the clock.

## sysCLOCK PLL Timing

### Over Recommended Operating Conditions

Parameter	Description	Conditions	Min.	Typ.	Max.	Units
$f_{IN}$	Input Clock Frequency (CLKI, CLKFB)		10	—	435	MHz
$f_{OUT}$	Output Clock Frequency (CLKOP, CLKOS)		10	—	435	MHz
$f_{OUT2}$	K-Divider Output Frequency	CLKOK	0.078	—	217.5	MHz
		CLKOK2	3.3	—	145	MHz
$f_{VCO}$	PLL VCO Frequency		435	—	870	MHz
$f_{PFD}$	Phase Detector Input Frequency		10	—	435	MHz
<b>AC Characteristics</b>						
$t_{DT}$	Output Clock Duty Cycle	Default duty cycle selected <sup>3</sup>	45	50	55	%
$t_{CPA}$	Coarse Phase Adjust		-5	0	5	%
$t_{PH}$ <sup>4</sup>	Output Phase Accuracy		-5	0	5	%
$t_{OPJIT}$ <sup>1</sup>	Output Clock Period Jitter	$f_{OUT} > 400$ MHz	—	—	±50	ps
		$100 \text{ MHz} < f_{OUT} < 400$ MHz	—	—	±125	ps
		$f_{OUT} < 100$ MHz	—	—	0.025	UIPP
$t_{SK}$	Input Clock to Output Clock Skew	N/M = integer	—	—	±240	ps
$t_{OPW}$	Output Clock Pulse Width	At 90% or 10%	1	—	—	ns
$t_{LOCK}$ <sup>2</sup>	PLL Lock-in Time	25 to 435 MHz	—	—	50	μs
		10 to 25 MHz	—	—	100	μs
$t_{IPJIT}$	Input Clock Period Jitter		—	—	±200	ps
$t_{FBKDL}$	External Feedback Delay		—	—	10	ns
$t_{HI}$	Input Clock High Time	90% to 90%	0.5	—	—	ns
$t_{LO}$	Input Clock Low Time	10% to 10%	0.5	—	—	ns
$t_{RSTKW}$	Reset Signal Pulse Width (RSTK)		10	—	—	ns
$t_{RSTW}$	Reset Signal Pulse Width (RST)		500	—	—	ns

1. Jitter sample is taken over 10,000 samples of the primary PLL output with clean reference clock.

2. Output clock is valid after  $t_{LOCK}$  for PLL reset and dynamic delay adjustment.

3. Using LVDS output buffers.

4. Relative to CLKOP.



## Signal Descriptions (Cont.)

Signal Name	I/O	Description
TDO	O	Output pin. Test Data Out pin used to shift data out of a device using 1149.1.
VCCJ	—	Power supply pin for JTAG Test Access Port.
<b>Configuration Pads</b> (Used during sysCONFIG)		
CFG[1:0]	I	Mode pins used to specify configuration mode values latched on rising edge of INITN. During configuration, an internal pull-up is enabled.
INITN <sup>1</sup>	I/O	Open Drain pin. Indicates the FPGA is ready to be configured. During configuration, a pull-up is enabled.
PROGRAMN	I	Initiates configuration sequence when asserted low. This pin always has an active pull-up.
DONE	I/O	Open Drain pin. Indicates that the configuration sequence is complete, and the startup sequence is in progress.
CCLK	I/O	Configuration Clock for configuring an FPGA in sysCONFIG mode.
SISPI <sup>2</sup>	I/O	Input data pin in slave SPI mode and Output data pin in Master SPI mode.
SOSPI <sup>2</sup>	I/O	Output data pin in slave SPI mode and Input data pin in Master SPI mode.
CSSPIN <sup>2</sup>	O	Chip select for external SPI Flash memory in Master SPI mode. This pin has a weak internal pull-up.
CSSPISN	I	Chip select in Slave SPI mode. This pin has a weak internal pull-up.
TOE	I	Test Output Enable tristates all I/O pins when driven low. This pin has a weak internal pull-up, but when not used an external pull-up to V <sub>CC</sub> is recommended.

1. If not actively driven, the internal pull-up may not be sufficient. An external pull-up resistor of 4.7k to 10k $\Omega$  is recommended.
2. When using the device in Master SPI mode, it must be mutually exclusive from JTAG operations (i.e. TCK tied to GND) or the JTAG TCK must be free-running when used in a system JTAG test environment. If Master SPI mode is used in conjunction with a JTAG download cable, the device power cycle is required after the cable is unplugged.

### Pin Information Summary

Pin Type		XP2-5				XP2-8				XP2-17			XP2-30			XP2-40	
		132 csBGA	144 TQFP	208 PQFP	256 ftBGA	132 csBGA	144 TQFP	208 PQFP	256 ftBGA	208 PQFP	256 ftBGA	484 fpBGA	256 ftBGA	484 fpBGA	672 fpBGA	484 fpBGA	672 fpBGA
Single Ended User I/O		86	100	146	172	86	100	146	201	146	201	358	201	363	472	363	540
Differential Pair User I/O	Normal	35	39	57	66	35	39	57	77	57	77	135	77	137	180	137	204
	Highspeed	8	11	16	20	8	11	16	23	16	23	44	23	44	56	44	66
Configuration	TAP	5	5	5	5	5	5	5	5	5	5	5	5	5	5	5	5
	Muxed	9	9	9	9	9	9	9	9	9	9	9	9	9	9	9	9
	Dedicated	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
Non Configuration	Muxed	5	5	7	7	7	7	9	9	11	11	21	7	11	13	11	13
	Dedicated	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
Vcc		6	4	9	6	6	4	9	6	9	6	16	6	16	20	16	20
Vccaux		4	4	4	4	4	4	4	4	4	4	8	4	8	8	8	8
VCCPLL		2	2	2	-	2	2	2	-	4	-	-	-	-	-	-	-
VCCIO	Bank0	2	2	2	2	2	2	2	2	2	2	4	2	4	4	4	4
	Bank1	1	1	2	2	1	1	2	2	2	2	4	2	4	4	4	4
	Bank2	2	2	2	2	2	2	2	2	2	2	4	2	4	4	4	4
	Bank3	1	1	2	2	1	1	2	2	2	2	4	2	4	4	4	4
	Bank4	1	1	2	2	1	1	2	2	2	2	4	2	4	4	4	4
	Bank5	2	2	2	2	2	2	2	2	2	2	4	2	4	4	4	4
	Bank6	1	1	2	2	1	1	2	2	2	2	4	2	4	4	4	4
	Bank7	2	2	2	2	2	2	2	2	2	2	4	2	4	4	4	4
GND, GND0-GND7		15	15	20	20	15	15	22	20	22	20	56	20	56	64	56	64
NC		-	-	4	31	-	-	2	2	-	2	7	2	2	69	2	1
Single Ended/ Differential I/O per Bank	Bank0	18/9	20/10	20/10	26/13	18/9	20/10	20/10	28/14	20/10	28/14	52/26	28/14	52/26	70/35	52/26	70/35
	Bank1	4/2	6/3	18/9	18/9	4/2	6/3	18/9	22/11	18/9	22/11	36/18	22/11	36/18	54/27	36/18	70/35
	Bank2	16/8	18/9	18/9	22/11	16/8	18/9	18/9	26/13	18/9	26/13	46/23	26/13	46/23	56/28	46/23	64/32
	Bank3	4/2	4/2	16/8	20/10	4/2	4/2	16/8	24/12	16/8	24/12	44/22	24/12	46/23	56/28	46/23	66/33
	Bank4	8/4	8/4	18/9	18/9	8/4	8/4	18/9	26/13	18/9	26/13	36/18	26/13	38/19	54/27	38/19	70/35
	Bank5	14/7	18/9	20/10	24/12	14/7	18/9	20/10	24/12	20/10	24/12	52/26	24/12	53/26	70/35	53/26	70/35
	Bank6	6/3	8/4	18/9	22/11	6/3	8/4	18/9	27/13	18/9	27/13	46/23	27/13	46/23	56/28	46/23	66/33
	Bank7	16/8	18/9	18/9	22/11	16/8	18/9	18/9	24/12	18/9	24/12	46/23	24/12	46/23	56/28	46/23	64/32
True LVDS Pairs Bonding Out per Bank	Bank0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	Bank1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	Bank2	3	4	4	5	3	4	4	6	4	6	11	6	11	14	11	16
	Bank3	1	1	4	5	1	1	4	6	4	6	11	6	11	14	11	17
	Bank4	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	Bank5	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	Bank6	1	2	4	5	1	2	4	6	4	6	11	6	11	14	11	17
	Bank7	3	4	4	5	3	4	4	5	4	5	11	5	11	14	11	16
DDR Banks Bonding Out per I/O Bank <sup>1</sup>	Bank0	1	1	1	1	1	1	1	1	1	1	3	1	2	4	2	4
	Bank1	0	0	1	1	0	0	1	1	1	1	2	1	2	3	2	4
	Bank2	1	1	1	1	1	1	1	1	1	1	2	1	3	3	3	4
	Bank3	0	0	1	1	0	0	1	1	1	1	2	1	3	3	3	4
	Bank4	0	0	1	1	0	0	1	1	1	1	2	1	2	3	2	4
	Bank5	1	1	1	1	1	1	1	1	1	1	3	1	2	4	2	4
	Bank6	0	0	1	1	0	0	1	1	1	1	2	1	3	3	3	4
	Bank7	1	1	1	1	1	1	1	1	1	1	2	1	3	3	3	4

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-40E-5F484C	1.2V	-5	fpBGA	484	COM	40
LFXP2-40E-6F484C	1.2V	-6	fpBGA	484	COM	40
LFXP2-40E-7F484C	1.2V	-7	fpBGA	484	COM	40
LFXP2-40E-5F672C	1.2V	-5	fpBGA	672	COM	40
LFXP2-40E-6F672C	1.2V	-6	fpBGA	672	COM	40
LFXP2-40E-7F672C	1.2V	-7	fpBGA	672	COM	40

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Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-5E-5M132I	1.2V	-5	csBGA	132	IND	5
LFXP2-5E-6M132I	1.2V	-6	csBGA	132	IND	5
LFXP2-5E-6FT256I	1.2V	-6	ftBGA	256	IND	5

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-8E-5M132I	1.2V	-5	csBGA	132	IND	8
LFXP2-8E-6M132I	1.2V	-6	csBGA	132	IND	8
LFXP2-5E-5FT256I	1.2V	-5	ftBGA	256	IND	5
LFXP2-8E-5FT256I	1.2V	-5	ftBGA	256	IND	8
LFXP2-8E-6FT256I	1.2V	-6	ftBGA	256	IND	8

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-17E-5FT256I	1.2V	-5	ftBGA	256	IND	17
LFXP2-17E-6FT256I	1.2V	-6	ftBGA	256	IND	17
LFXP2-17E-5F484I	1.2V	-5	fpBGA	484	IND	17
LFXP2-17E-6F484I	1.2V	-6	fpBGA	484	IND	17

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-30E-5FT256I	1.2V	-5	ftBGA	256	IND	30
LFXP2-30E-6FT256I	1.2V	-6	ftBGA	256	IND	30
LFXP2-30E-5F484I	1.2V	-5	fpBGA	484	IND	30
LFXP2-30E-6F484I	1.2V	-6	fpBGA	484	IND	30
LFXP2-30E-5F672I	1.2V	-5	fpBGA	672	IND	30
LFXP2-30E-6F672I	1.2V	-6	fpBGA	672	IND	30

### Revision History

Date	Version	Section	Change Summary
May 2007	01.1	—	Initial release.
September 2007	01.2	DC and Switching Characteristics	Added JTAG Port Timing Waveforms diagram.
			Updated sysCLOCK PLL Timing table.
		Pinout Information	Added Thermal Management text section.
February 2008	01.3	Architecture	Added LVC MOS33D to Supported Output Standards table.
			Clarified: "This Flash can be programmed through either the JTAG or Slave SPI ports of the device. The SRAM configuration space can also be infinitely reconfigured through the JTAG and Master SPI ports."
			Added External Slave SPI Port to Serial TAG Memory section. Updated Serial TAG Memory diagram.
		DC and Switching Characteristics	Updated Flash Programming Specifications table.
			Added "8W" specification to Hot Socketing Specifications table.
			Updated Timing Tables
			Clarifications for IIH in DC Electrical Characteristics table.
			Added LVC MOS33D section
			Updated DOA and DOA (Regs) to EBR Timing diagrams.
			Removed Master Clock Frequency and Duty Cycle sections from the LatticeXP2 sysCONFIG Port Timing Specifications table. These are listed on the On-chip Oscillator and Configuration Master Clock Characteristics table.
			Changed CSSPIN to CSSPISN in description of $t_{SCS}$ , $t_{SCSS}$ , and $t_{SCSH}$ parameters. Removed $t_{SOE}$ parameter.
			Clarified On-chip Oscillator documentation
			Added Switching Test Conditions
		Pinout Information	Added "True LVDS Pairs Bonding Out per Bank," "DDR Banks Bonding Out per I/O Bank," and "PCI capable I/Os Bonding Out per Bank" to Pin Information Summary in place of previous blank table "PCI and DDR Capabilities of the Device-Package Combinations"
			Removed pinout listing. This information is available on the LatticeXP2 product web pages
		Ordering Information	Added XP2-17 "8W" and all other family OPNs.
April 2008	01.4	DC and Switching Characteristics	Updated Absolute Maximum Ratings footnotes.
			Updated Recommended Operating Conditions Table footnotes.
			Updated Supply Current (Standby) Table
			Updated Initialization Supply Current Table
			Updated Programming and Erase Flash Supply Current Table
			Updated Register to Register Performance Table
			Updated LatticeXP2 External Switching Characteristics Table
			Updated LatticeXP2 Internal Switching Characteristics Table
			Updated sysCLOCK PLL Timing Table

Date	Version	Section	Change Summary
April 2008 (cont.)	01.4 (cont.)	DC and Switching Characteristics (cont.)	Updated Flash Download Time (From On-Chip Flash to SRAM) Table
			Updated Flash Program Time Table
			Updated Flash Erase Time Table
			Updated FlashBAK (from EBR to Flash) Table
			Updated Hot Socketing Specifications Table footnotes
June 2008	01.5	Pinout Information	Updated Signal Descriptions Table
		Architecture	Removed Read-Before-Write sysMEM EBR mode.
			Clarification of the operation of the secondary clock regions.
		DC and Switching Characteristics	Removed Read-Before-Write sysMEM EBR mode.
August 2008	01.6	Pinout Information	Updated DDR Banks Bonding Out per I/O Bank section of Pin Information Summary Table.
		—	Data sheet status changed from preliminary to final.
		Architecture	Clarification of the operation of the secondary clock regions.
		DC and Switching Characteristics	Removed “8W” specification from Hot Socketing Specifications table.
			Removed “8W” footnote from DC Electrical Characteristics table.
			Updated Register-to-Register Performance table.
		Ordering Information	Removed “8W” option from Part Number Description.
			Removed XP2-17 “8W” OPNs.
April 2011	01.7	DC and Switching Characteristics	Recommended Operating Conditions table, added footnote 5.
			On-Chip Flash Memory Specifications table, added footnote 1.
			BLVDS DC Conditions, corrected column title to be Z0 = 90 ohms.
			sysCONFIG Port Timing Specifications table, added footnote 1 for $t_{DINIT}$ .
January 2012	01.8	Multiple	Added support for Lattice Diamond design software.
		Architecture	Corrected information regarding SED support.
		DC and Switching Characteristics	Added reference to ESD Performance Qualification Summary information.
May 2013	01.9	All	Updated document with new corporate logo.
		Architecture	Architecture Overview – Added information on the state of the register on power up and after configuration.
			Added information regarding SED support.
		DC and Switching Characteristics	Removed Input Clock Rise/Fall Time 1ns max from the sysCLOCK PLL Timing table.
March 2014	02.0	Architecture	Updated topside mark in Ordering Information diagram.
			Updated Typical sysIO I/O Behavior During Power-up section. Added information on POR signal deactivation.
August 2014	02.1	Architecture	Updated Typical sysIO I/O Behavior During Power-up section. Described user I/Os during power up and before FPGA core logic is active.
September 2014	2.2	DC and Switching Characteristics	Updated <a href="#">Switching Test Conditions</a> section. Re-linked missing figure.