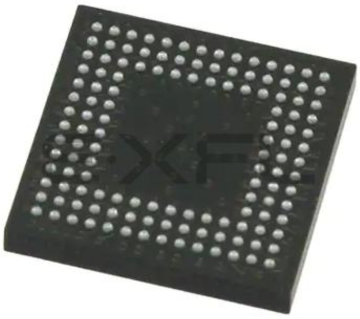
Lattice Semiconductor Corporation - LFXP2-5E-5MN132I Datasheet



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Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Active
Number of LABs/CLBs	625
Number of Logic Elements/Cells	5000
Total RAM Bits	169984
Number of I/O	86
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	132-LFBGA, CSPBGA
Supplier Device Package	132-CSBGA (8x8)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfxp2-5e-5mn132i

Introduction

LatticeXP2 devices combine a Look-up Table (LUT) based FPGA fabric with non-volatile Flash cells in an architecture referred to as flexiFLASH.

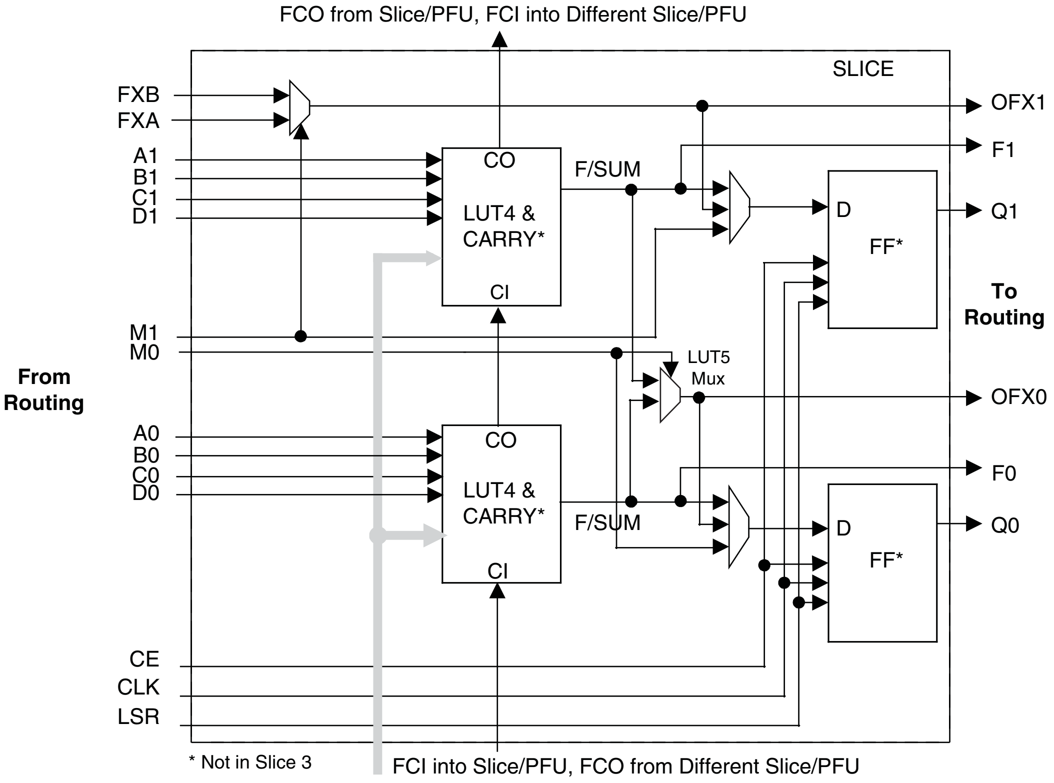
The flexiFLASH approach provides benefits including instant-on, infinite reconfigurability, on chip storage with FlashBAK embedded block memory and Serial TAG memory and design security. The parts also support Live Update technology with TransFR, 128-bit AES Encryption and Dual-boot technologies.

The LatticeXP2 FPGA fabric was optimized for the new technology from the outset with high performance and low cost in mind. LatticeXP2 devices include LUT-based logic, distributed and embedded memory, Phase Locked Loops (PLLs), pre-engineered source synchronous I/O support and enhanced sysDSP blocks.

Lattice Diamond® design software allows large and complex designs to be efficiently implemented using the LatticeXP2 family of FPGA devices. Synthesis library support for LatticeXP2 is available for popular logic synthesis tools. The Diamond software uses the synthesis tool output along with the constraints from its floor planning tools to place and route the design in the LatticeXP2 device. The Diamond tool extracts the timing from the routing and back-annotates it into the design for timing verification.

Lattice provides many pre-designed Intellectual Property (IP) LatticeCORE™ modules for the LatticeXP2 family. By using these IPs as standardized blocks, designers are free to concentrate on the unique aspects of their design, increasing their productivity.

Figure 2-3. Slice Diagram



* Not in Slice 3

For Slices 0 and 2, memory control signals are generated from Slice 1 as follows:
WCK is CLK
WRE is from LSR
DI[3:2] for Slice 2 and DI[1:0] for Slice 0 data
WAD [A:D] is a 4bit address from slice 1 LUT input

Table 2-2. Slice Signal Descriptions

Function	Type	Signal Names	Description
Input	Data signal	A0, B0, C0, D0	Inputs to LUT4
Input	Data signal	A1, B1, C1, D1	Inputs to LUT4
Input	Multi-purpose	M0	Multipurpose Input
Input	Multi-purpose	M1	Multipurpose Input
Input	Control signal	CE	Clock Enable
Input	Control signal	LSR	Local Set/Reset
Input	Control signal	CLK	System Clock
Input	Inter-PFU signal	FCI	Fast Carry-In[1]
Input	Inter-slice signal	FXA	Intermediate signal to generate LUT6 and LUT7
Input	Inter-slice signal	FXB	Intermediate signal to generate LUT6 and LUT7
Output	Data signals	F0, F1	LUT4 output register bypass signals
Output	Data signals	Q0, Q1	Register outputs
Output	Data signals	OFX0	Output of a LUT5 MUX
Output	Data signals	OFX1	Output of a LUT6, LUT7, LUT8[2] MUX depending on the slice
Output	Inter-PFU signal	FCO	Slice 2 of each PFU is the fast carry chain output[1]

1. See Figure 2-3 for connection details.
2. Requires two PFUs.

Modes of Operation

Each slice has up to four potential modes of operation: Logic, Ripple, RAM and ROM.

Logic Mode

In this mode, the LUTs in each slice are configured as LUT4s. A LUT4 has 16 possible input combinations. Four-input logic functions are generated by programming the LUT4. Since there are two LUT4s per slice, a LUT5 can be constructed within one slice. Larger LUTs such as LUT6, LUT7 and LUT8, can be constructed by concatenating two or more slices. Note that a LUT8 requires more than four slices.

Ripple Mode

Ripple mode allows efficient implementation of small arithmetic functions. In ripple mode, the following functions can be implemented by each slice:

- Addition 2-bit
- Subtraction 2-bit
- Add/Subtract 2-bit using dynamic control
- Up counter 2-bit
- Down counter 2-bit
- Up/Down counter with async clear
- Up/Down counter with preload (sync)
- Ripple mode multiplier building block
- Multiplier support
- Comparator functions of A and B inputs
  - A greater-than-or-equal-to B
  - A not-equal-to B
  - A less-than-or-equal-to B

Two carry signals, FCI and FCO, are generated per slice in this mode, allowing fast arithmetic functions to be constructed by concatenating slices.

RAM Mode

In this mode, a 16x4-bit distributed Single Port RAM (SPR) can be constructed using each LUT block in Slice 0 and Slice 2 as a 16x1-bit memory. Slice 1 is used to provide memory address and control signals. A 16x2-bit Pseudo Dual Port RAM (PDPR) memory is created by using one slice as the read-write port and the other companion slice as the read-only port.

The Lattice design tools support the creation of a variety of different size memories. Where appropriate, the software will construct these using distributed memory primitives that represent the capabilities of the PFU. Table 2-3 shows the number of slices required to implement different distributed RAM primitives. For more information on using RAM in LatticeXP2 devices, please see TN1137, LatticeXP2 Memory Usage Guide.

Table 2-3. Number of Slices Required For Implementing Distributed RAM

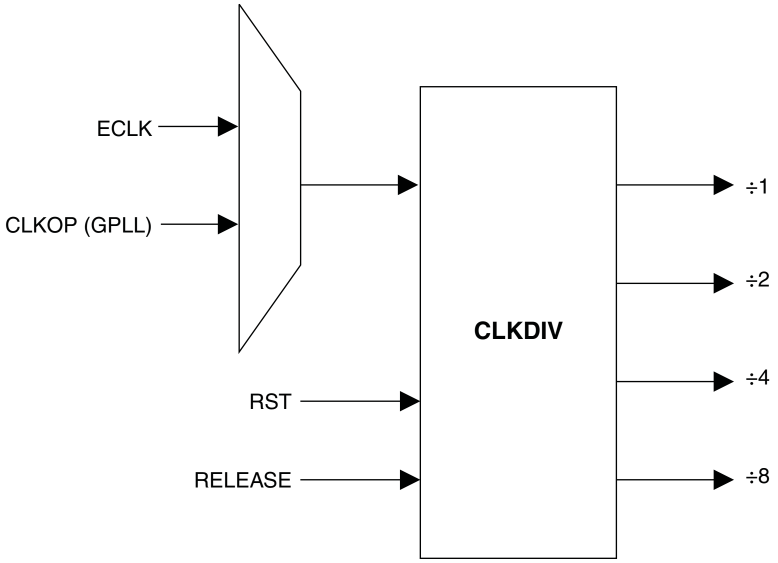
	SPR 16X4	PDPR 16X4
Number of slices	3	3

Note: SPR = Single Port RAM, PDPR = Pseudo Dual Port RAM

ROM Mode

ROM mode uses the LUT logic; hence, Slices 0 through 3 can be used in the ROM mode. Preloading is accomplished through the programming interface during PFU configuration.

Figure 2-5. Clock Divider Connections



Clock Distribution Network

LatticeXP2 devices have eight quadrant-based primary clocks and between six and eight flexible region-based secondary clocks/control signals. Two high performance edge clocks are available on each edge of the device to support high speed interfaces. The clock inputs are selected from external I/Os, the sysCLOCK PLLs, or routing. Clock inputs are fed throughout the chip via the primary, secondary and edge clock networks.

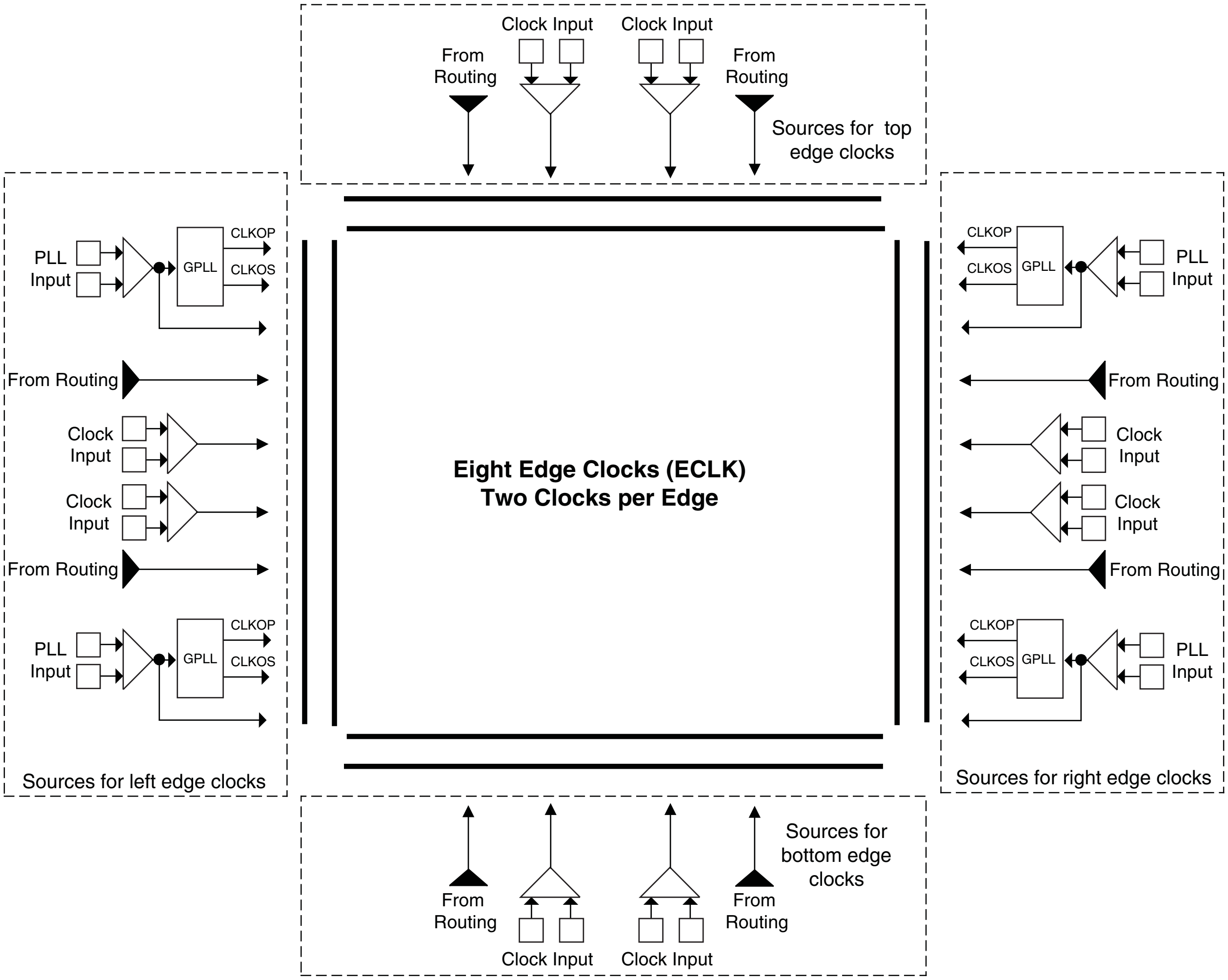
Primary Clock Sources

LatticeXP2 devices derive primary clocks from four sources: PLL outputs, CLKDIV outputs, dedicated clock inputs and routing. LatticeXP2 devices have two to four sysCLOCK PLLs, located in the four corners of the device. There are eight dedicated clock inputs, two on each side of the device. Figure 2-6 shows the primary clock sources.

Edge Clock Sources

Edge clock resources can be driven from a variety of sources at the same edge. Edge clock resources can be driven from adjacent edge clock PIOs, primary clock PIOs, PLLs and clock dividers as shown in Figure 2-8.

Figure 2-8. Edge Clock Sources

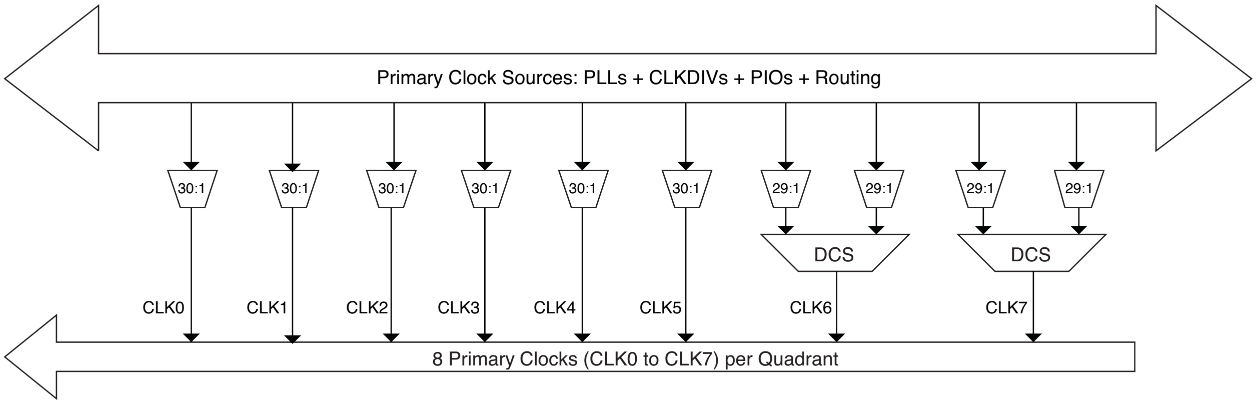


Note: This diagram shows sources for the XP2-17 device. Smaller LatticeXP2 devices have two GPLLs.

Primary Clock Routing

The clock routing structure in LatticeXP2 devices consists of a network of eight primary clock lines (CLK0 through CLK7) per quadrant. The primary clocks of each quadrant are generated from muxes located in the center of the device. All the clock sources are connected to these muxes. Figure 2-9 shows the clock routing for one quadrant. Each quadrant mux is identical. If desired, any clock can be routed globally.

Figure 2-9. Per Quadrant Primary Clock Selection

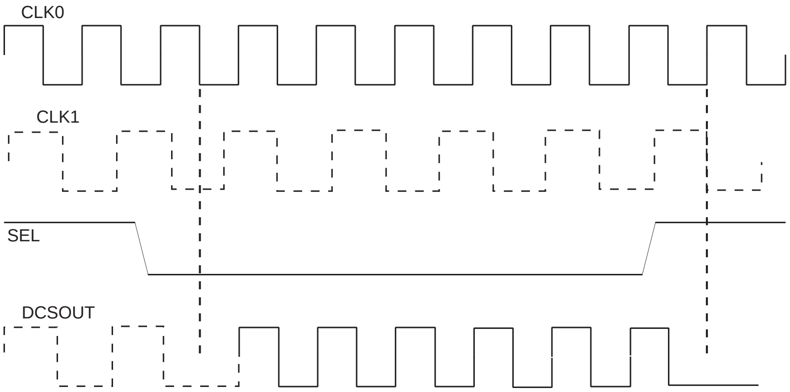


Dynamic Clock Select (DCS)

The DCS is a smart multiplexer function available in the primary clock routing. It switches between two independent input clock sources without any glitches or runt pulses. This is achieved irrespective of when the select signal is toggled. There are two DCS blocks per quadrant; in total, eight DCS blocks per device. The inputs to the DCS block come from the center muxes. The output of the DCS is connected to primary clocks CLK6 and CLK7 (see Figure 2-9).

Figure 2-10 shows the timing waveforms of the default DCS operating mode. The DCS block can be programmed to other modes. For more information on the DCS, please see TN1126, LatticeXP2 sysCLOCK PLL Design and Usage Guide.

Figure 2-10. DCS Waveforms



Secondary Clock/Control Routing

Secondary clocks in the LatticeXP2 devices are region-based resources. The benefit of region-based resources is the relatively low injection delay and skew within the region, as compared to primary clocks. EBR rows, DSP rows and a special vertical routing channel bound the secondary clock regions. This special vertical routing channel aligns with either the left edge of the center DSP block in the DSP row or the center of the DSP row. Figure 2-11 shows this special vertical routing channel and the eight secondary clock regions for the LatticeXP2-40.

sysMEM Memory

LatticeXP2 devices contains a number of sysMEM Embedded Block RAM (EBR). The EBR consists of 18 Kbit RAM with dedicated input and output registers.

sysMEM Memory Block

The sysMEM block can implement single port, dual port or pseudo dual port memories. Each block can be used in a variety of depths and widths as shown in Table 2-5. FIFOs can be implemented in sysMEM EBR blocks by using support logic with PFUs. The EBR block supports an optional parity bit for each data byte to facilitate parity checking. EBR blocks provide byte-enable support for configurations with18-bit and 36-bit data widths.

Table 2-5. sysMEM Block Configurations

Memory Mode	Configurations
Single Port	16,384 x 1 8,192 x 2 4,096 x 4 2,048 x 9 1,024 x 18 512 x 36
True Dual Port	16,384 x 1 8,192 x 2 4,096 x 4 2,048 x 9 1,024 x 18
Pseudo Dual Port	16,384 x 1 8,192 x 2 4,096 x 4 2,048 x 9 1,024 x 18 512 x 36

Bus Size Matching

All of the multi-port memory modes support different widths on each of the ports. The RAM bits are mapped LSB word 0 to MSB word 0, LSB word 1 to MSB word 1, and so on. Although the word size and number of words for each port varies, this mapping scheme applies to each port.

FlashBAK EBR Content Storage

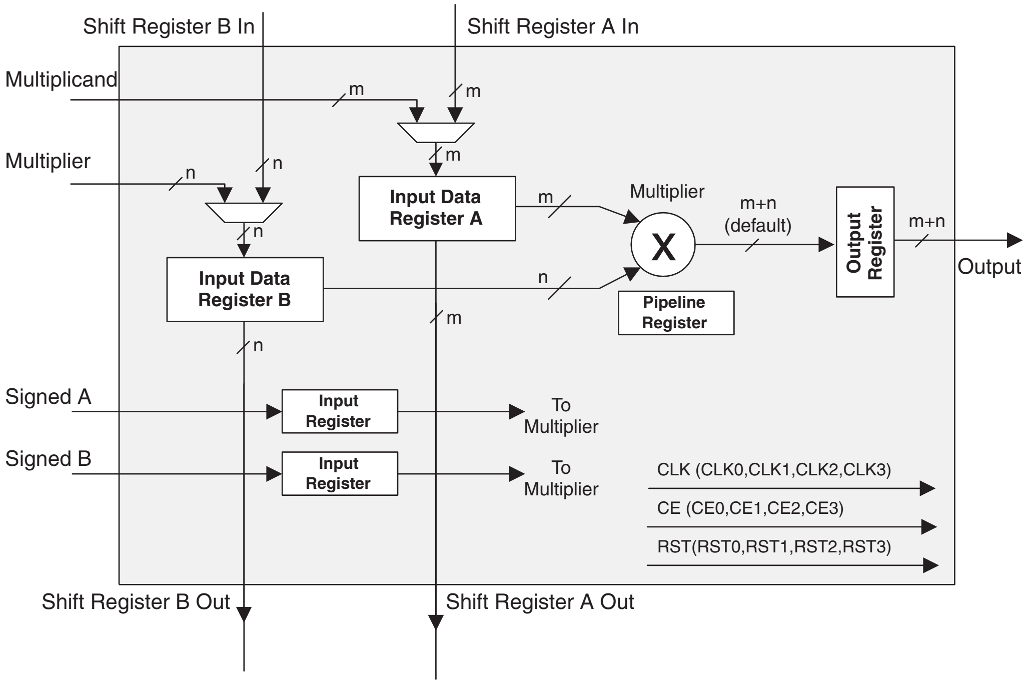
All the EBR memory in the LatticeXP2 is shadowed by Flash memory. Optionally, initialization values for the memory blocks can be defined using the Lattice Diamond design tools. The initialization values are loaded into the Flash memory during device programming and into the SRAM at power up or whenever the device is reconfigured. This feature is ideal for the storage of a variety of information such as look-up tables and microprocessor code. It is also possible to write the current contents of the EBR memory back to Flash memory. This capability is useful for the storage of data such as error codes and calibration information. For additional information on the FlashBAK capability see TN1137, LatticeXP2 Memory Usage Guide.

- In the ‘Signed/Unsigned’ options the operands can be switched between signed and unsigned on every cycle.
- In the ‘Add/Sub’ option the Accumulator can be switched between addition and subtraction on every cycle.
- The loading of operands can switch between parallel and serial operations.

MULT sysDSP Element

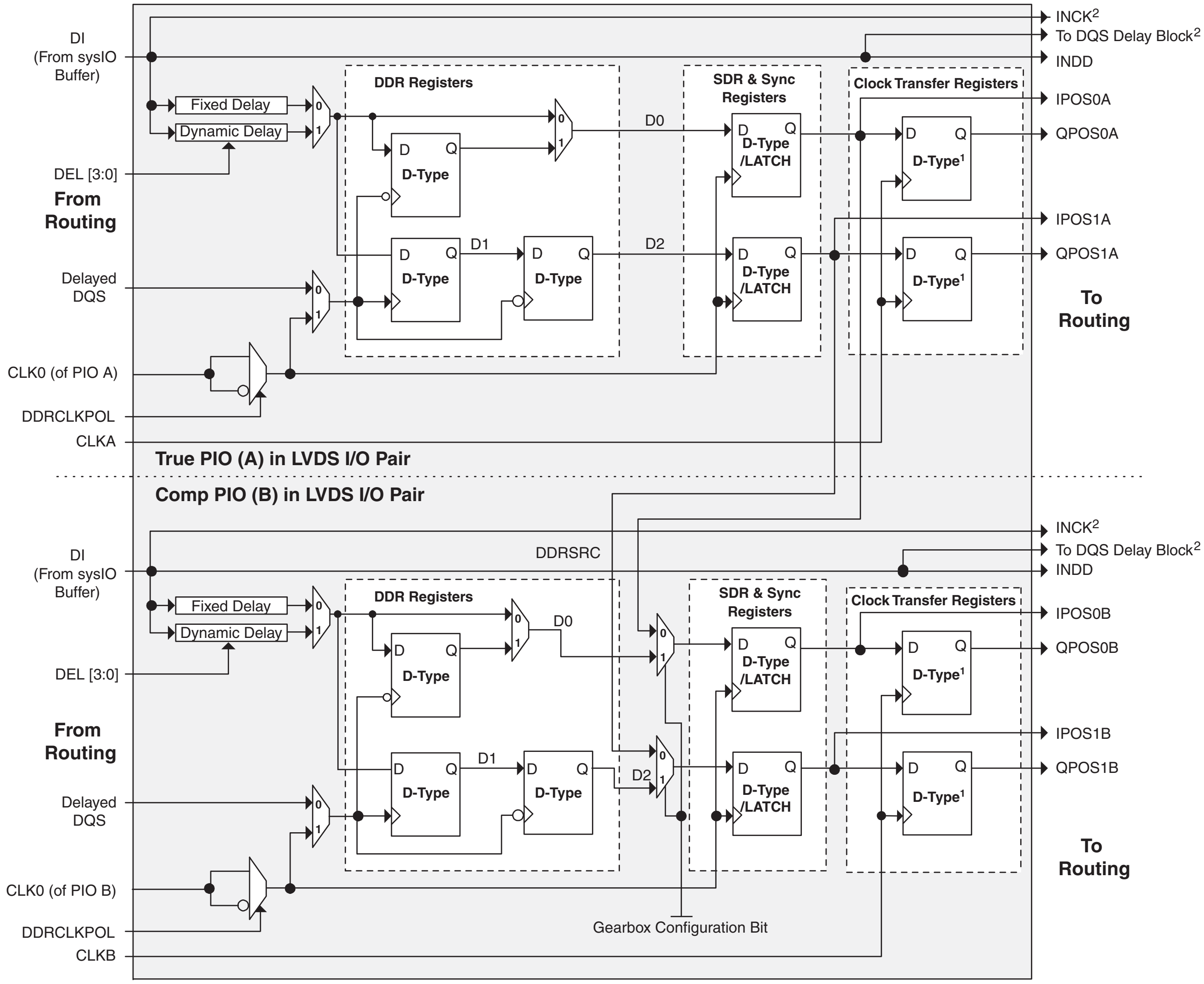
This multiplier element implements a multiply with no addition or accumulator nodes. The two operands, A and B, are multiplied and the result is available at the output. The user can enable the input/output and pipeline registers. Figure 2-20 shows the MULT sysDSP element.

Figure 2-20. MULT sysDSP Element



The signal DDRCLKPOL controls the polarity of the clock used in the synchronization registers. It ensures adequate timing when data is transferred from the DQS to system clock domain. For further discussion on this topic, see the DDR Memory section of this data sheet.

Figure 2-26. Input Register Block



1. Shared with output register
2. Selected PIO.

Note: Simplified version does not show CE and SET/RESET details

Output Register Block

The output register block provides the ability to register signals from the core of the device before they are passed to the sysIO buffers. The blocks on the PIOs on the left, right and bottom contain registers for SDR operation that are combined with an additional latch for DDR operation. Figure 2-27 shows the diagram of the Output Register Block for PIOs.

In SDR mode, ONEG0 feeds one of the flip-flops that then feeds the output. The flip-flop can be configured as a D-type or latch. In DDR mode, ONEG0 and OPOS0 are fed into registers on the positive edge of the clock. At the next clock cycle the registered OPOS0 is latched. A multiplexer running off the same clock cycle selects the correct register to feed the output (D0).

By combining output blocks of the complementary PIOs and sharing some registers from input blocks, a gearbox function can be implemented, to take four data streams ONEG0A, ONEG1A, ONEG1B and ONEG1B. Figure 2-27 original backup configuration and try again. This all can be done without power cycling the system. For more information please see TN1220, LatticeXP2 Dual Boot Feature.

For more information on device configuration, please see TN1141, LatticeXP2 sysCONFIG Usage Guide.

Soft Error Detect (SED) Support

LatticeXP2 devices have dedicated logic to perform Cyclic Redundancy Code (CRC) checks. During configuration, the configuration data bitstream can be checked with the CRC logic block. In addition, LatticeXP2 devices can be programmed for checking soft errors in SRAM. SED can be run on a programmed device when the user logic is not active. In the event a soft error occurs, the device can be programmed to either reload from a known good boot image (from internal Flash or external SPI memory) or generate an error signal.

For further information on SED support, please see TN1130, LatticeXP2 Soft Error Detection (SED) Usage Guide.

On-Chip Oscillator

Every LatticeXP2 device has an internal CMOS oscillator that is used to derive a Master Clock (CCLK) for configuration. The oscillator and CCLK run continuously and are available to user logic after configuration is complete. The available CCLK frequencies are listed in Table 2-14. When a different CCLK frequency is selected during the design process, the following sequence takes place:

1. Device powers up with the default CCLK frequency.
2. During configuration, users select a different CCLK frequency.
3. CCLK frequency changes to the selected frequency after clock configuration bits are received.

This internal CMOS oscillator is available to the user by routing it as an input clock to the clock tree. For further information on the use of this oscillator for configuration or user mode, please see TN1141, LatticeXP2 sysCONFIG Usage Guide.

Table 2-14. Selectable CCLKs and Oscillator Frequencies During Configuration and User Mode

CCLK/Oscillator (MHz)
2.5[1]
3.1[2]
4.3
5.4
6.9
8.1
9.2
10
13
15
20
26
32
40
54
80[3]
163[3]

1. Software default oscillator frequency.
2. Software default CCLK frequency.
3. Frequency not valid for CCLK.

Hot Socketing Specifications[1, 2, 3, 4]

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
I_{DK}	Input or I/O Leakage Current	$0 \le V_{IN} \le V_{IH}$ (MAX.)	—	—	+/-1	mA

1. Insensitive to sequence of V_{CC}, V_{CCAUX} and V_{CCIO}. However, assumes monotonic rise/fall rates for V_{CC}, V_{CCAUX} and V_{CCIO}.
2. $0 \le V_{CC} \le V_{CC}$ (MAX), $0 \le V_{CCIO} \le V_{CCIO}$ (MAX) or $0 \le V_{CCAUX} \le V_{CCAUX}$ (MAX).
3. I_{DK} is additive to I_{PU}, I_{PW} or I_{BH}.
4. LVCMOS and LVTTL only.

ESD Performance

Please refer to the LatticeXP2 Product Family Qualification Summary for complete qualification data, including ESD performance.

DC Electrical Characteristics

Over Recommended Operating Conditions

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
I_{IL}, I_{IH}[1]	Input or I/O Low Leakage	$0 \le V_{IN} \le V_{CCIO}$	—	—	10	µA
		$V_{CCIO} \le V_{IN} \le V_{IH}$ (MAX)	—	—	150	µA
I_{PU}	I/O Active Pull-up Current	$0 \le V_{IN} \le 0.7\ V_{CCIO}$	-30	—	-150	µA
I_{PD}	I/O Active Pull-down Current	V_{IL} (MAX) $\le V_{IN} \le V_{CCIO}$	30	—	210	µA
I_{BHLS}	Bus Hold Low Sustaining Current	$V_{IN} = V_{IL}$ (MAX)	30	—	—	µA
I_{BHHS}	Bus Hold High Sustaining Current	$V_{IN} = 0.7\ V_{CCIO}$	-30	—	—	µA
I_{BHLO}	Bus Hold Low Overdrive Current	$0 \le V_{IN} \le V_{CCIO}$	—	—	210	µA
I_{BHHO}	Bus Hold High Overdrive Current	$0 \le V_{IN} \le V_{CCIO}$	—	—	-150	µA
V_{BHT}	Bus Hold Trip Points		V_{IL} (MAX)	—	V_{IH} (MIN)	V
C1	I/O Capacitance[2]	V_{CCIO} = 3.3V, 2.5V, 1.8V, 1.5V, 1.2V, V_{CC} = 1.2V, V_{IO} = 0 to V_{IH} (MAX)	—	8	—	pf
C2	Dedicated Input Capacitance	V_{CCIO} = 3.3V, 2.5V, 1.8V, 1.5V, 1.2V, V_{CC} = 1.2V, V_{IO} = 0 to V_{IH} (MAX)	—	6	—	pf

1. Input or I/O leakage current is measured with the pin configured as an input or as an I/O with the output driver tri-stated. It is not measured with the output driver active. Bus maintenance circuits are disabled.
2. T_A 25°C, f = 1.0 MHz.

BLVDS

The LatticeXP2 devices support the BLVDS standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel external resistor across the driver outputs. BLVDS is intended for use when multi-drop and bi-directional multi-point differential signaling is required. The scheme shown in Figure 3-2 is one possible solution for bi-directional multi-point differential signals.

Figure 3-2. BLVDS Multi-point Output Example

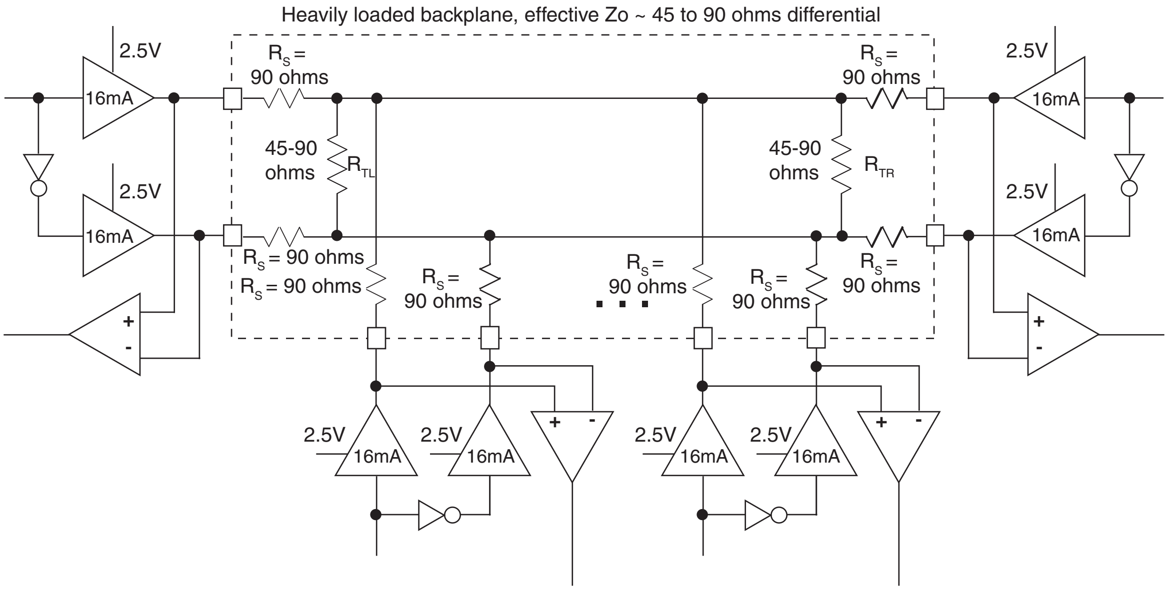


Table 3-2. BLVDS DC Conditions[1]

Over Recommended Operating Conditions

Parameter	Description	Typical		Units
		Zo = 45Ω	Zo = 90Ω	
V_{CCIO}	Output Driver Supply (+/- 5%)	2.50	2.50	V
Z_{OUT}	Driver Impedance	10.00	10.00	Ω
R_S	Driver Series Resistor (+/- 1%)	90.00	90.00	Ω
R_{TL}	Driver Parallel Resistor (+/- 1%)	45.00	90.00	Ω
R_{TR}	Receiver Termination (+/- 1%)	45.00	90.00	Ω
V_{OH}	Output High Voltage (After R_{TL})	1.38	1.48	V
V_{OL}	Output Low Voltage (After R_{TL})	1.12	1.02	V
V_{OD}	Output Differential Voltage (After R_{TL})	0.25	0.46	V
V_{CM}	Output Common Mode Voltage	1.25	1.25	V
I_{DC}	DC Output Current	11.24	10.20	mA

1. For input buffer, see LVDS table.

Register-to-Register Performance (Continued)

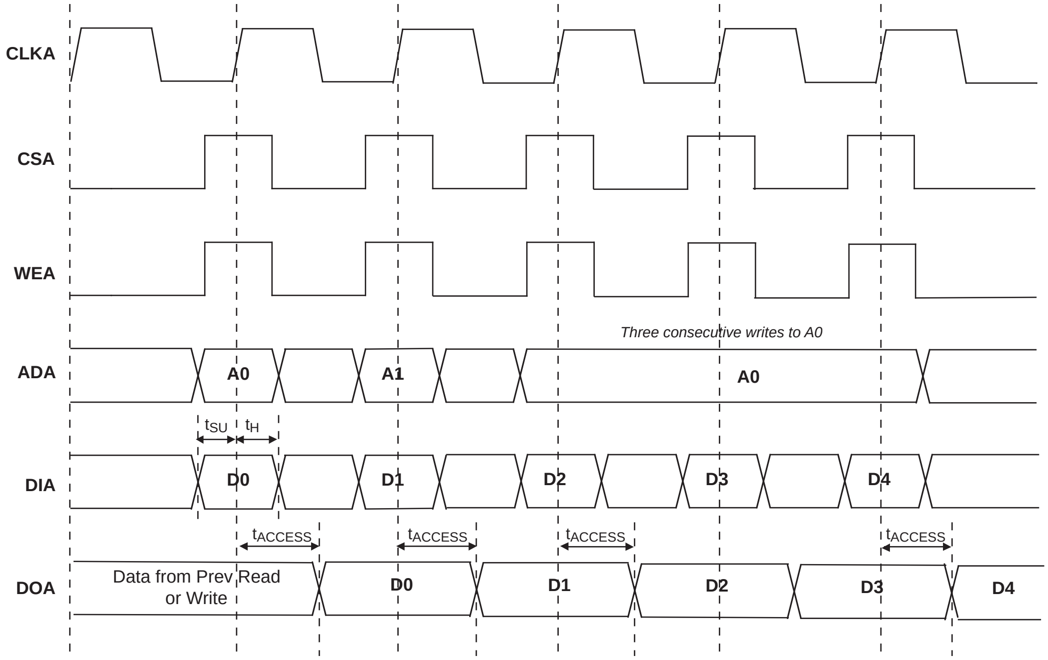
Function	-7 Timing	Units
DSP IP Functions		
16-Tap Fully-Parallel FIR Filter	198	MHz
1024-pt FFT	221	MHz
8X8 Matrix Multiplication	196	MHz

1. These timing numbers were generated using the ispLEVER design tool. Exact performance may vary with device, design and tool version. The tool uses internal parameters that have been characterized but are not tested on every device.

Derating Timing Tables

Logic timing provided in the following sections of this data sheet and the Diamond design tools are worst case numbers in the operating range. Actual delays at nominal temperature and voltage for best case process, can be much better than the values given in the tables. The Diamond design tool can provide logic timing numbers at a particular temperature and voltage.

Figure 3-8. Write Through (SP Read/Write on Port A, Input Registers Only)



Note: Input data and address are registered at the positive edge of the clock and output data appears after the positive edge of the clock.

sysCLOCK PLL Timing

Over Recommended Operating Conditions

Parameter	Description	Conditions	Min.	Typ.	Max.	Units
f_{IN}	Input Clock Frequency (CLKI, CLKFB)		10	—	435	MHz
f_{OUT}	Output Clock Frequency (CLKOP, CLKOS)		10	—	435	MHz
f_{OUT2}	K-Divider Output Frequency	CLKOK	0.078	—	217.5	MHz
		CLKOK2	3.3	—	145	MHz
f_{VCO}	PLL VCO Frequency		435	—	870	MHz
f_{PFD}	Phase Detector Input Frequency		10	—	435	MHz
AC Characteristics						
t_{DT}	Output Clock Duty Cycle	Default duty cycle selected [3]	45	50	55	%
t_{CPA}	Coarse Phase Adjust		-5	0	5	%
t_{PH}[4]	Output Phase Accuracy		-5	0	5	%
t_{OPJIT}[1]	Output Clock Period Jitter	$f_{OUT} > 400$ MHz	—	—	±50	ps
		100 MHz $< f_{OUT} <$ 400 MHz	—	—	±125	ps
		$f_{OUT} < 100$ MHz	—	—	0.025	UIPP
t_{SK}	Input Clock to Output Clock Skew	N/M = integer	—	—	±240	ps
t_{OPW}	Output Clock Pulse Width	At 90% or 10%	1	—	—	ns
t_{LOCK}[2]	PLL Lock-in Time	25 to 435 MHz	—	—	50	µs
		10 to 25 MHz	—	—	100	µs
t_{IPJIT}	Input Clock Period Jitter		—	—	±200	ps
t_{FBKDLY}	External Feedback Delay		—	—	10	ns
t_{HI}	Input Clock High Time	90% to 90%	0.5	—	—	ns
t_{LO}	Input Clock Low Time	10% to 10%	0.5	—	—	ns
t_{RSTKW}	Reset Signal Pulse Width (RSTK)		10	—	—	ns
t_{RSTW}	Reset Signal Pulse Width (RST)		500	—	—	ns

1. Jitter sample is taken over 10,000 samples of the primary PLL output with clean reference clock.
2. Output clock is valid after t_{LOCK} for PLL reset and dynamic delay adjustment.
3. Using LVDS output buffers.
4. Relative to CLKOP.

FlashBAK Time (from EBR to Flash)

Over Recommended Operating Conditions

Device	EBR Density (Bits)	Time (Typ.)	Units
XP2-5	166K	1.5	s
XP2-8	221K	1.5	s
XP2-17	276K	1.5	s
XP2-30	387K	2.0	s
XP2-40	885K	3.0	s

JTAG Port Timing Specifications

Over Recommended Operating Conditions

Symbol	Parameter	Min.	Max.	Units
f_{MAX}	TCK Clock Frequency	—	25	MHz
t_{BTCP}	TCK [BSCAN] clock pulse width	40	—	ns
t_{BTCPH}	TCK [BSCAN] clock pulse width high	20	—	ns
t_{BTCPL}	TCK [BSCAN] clock pulse width low	20	—	ns
t_{BTS}	TCK [BSCAN] setup time	8	—	ns
t_{BTH}	TCK [BSCAN] hold time	10	—	ns
t_{BTRF}	TCK [BSCAN] rise/fall time	50	—	mV/ns
t_{BTCO}	TAP controller falling edge of clock to valid output	—	10	ns
$t_{BTCODIS}$	TAP controller falling edge of clock to valid disable	—	10	ns
t_{BTCOEN}	TAP controller falling edge of clock to valid enable	—	10	ns
t_{BTCRS}	BSCAN test capture register setup time	8	—	ns
t_{BTCRH}	BSCAN test capture register hold time	25	—	ns
t_{BUTCO}	BSCAN test update register, falling edge of clock to valid output	—	25	ns
$t_{BTUODIS}$	BSCAN test update register, falling edge of clock to valid disable	—	25	ns
$t_{BTUPOEN}$	BSCAN test update register, falling edge of clock to valid enable	—	25	ns

PICs and DDR Data (DQ) Pins Associated with the DDR Strobe (DQS) Pin

PICs Associated with DQS Strobe	PIO Within PIC	DDR Strobe (DQS) and Data (DQ) Pins
For Left and Right Edges of the Device		
P[Edge] [n-4]	A	DQ
	B	DQ
P[Edge] [n-3]	A	DQ
	B	DQ
P[Edge] [n-2]	A	DQ
	B	DQ
P[Edge] [n-1]	A	DQ
	B	DQ
P[Edge] [n]	A	[Edge]DQSn
	B	DQ
P[Edge] [n+1]	A	DQ
	B	DQ
P[Edge] [n+2]	A	DQ
	B	DQ
P[Edge] [n+3]	A	DQ
	B	DQ
For Top and Bottom Edges of the Device		
P[Edge] [n-4]	A	DQ
	B	DQ
P[Edge] [n-3]	A	DQ
	B	DQ
P[Edge] [n-2]	A	DQ
	B	DQ
P[Edge] [n-1]	A	DQ
	B	DQ
P[Edge] [n]	A	[Edge]DQSn
	B	DQ
P[Edge] [n+1]	A	DQ
	B	DQ
P[Edge] [n+2]	A	DQ
	B	DQ
P[Edge] [n+3]	A	DQ
	B	DQ
P[Edge] [n+4]	A	DQ
	B	DQ

Notes:

1. “n” is a row PIC number.
2. The DDR interface is designed for memories that support one DQS strobe up to 16 bits of data for the left and right edges and up to 18 bits of data for the top and bottom edges. In some packages, all the potential DDR data (DQ) pins may not be available. PIC numbering definitions are provided in the “Signal Names” column of the Signal Descriptions table.

Lead-Free Packaging

Commercial

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-5E-5MN132C	1.2V	-5	Lead-Free csBGA	132	COM	5
LFXP2-5E-6MN132C	1.2V	-6	Lead-Free csBGA	132	COM	5
LFXP2-5E-7MN132C	1.2V	-7	Lead-Free csBGA	132	COM	5
LFXP2-5E-5TN144C	1.2V	-5	Lead-Free TQFP	144	COM	5
LFXP2-5E-6TN144C	1.2V	-6	Lead-Free TQFP	144	COM	5
LFXP2-5E-7TN144C	1.2V	-7	Lead-Free TQFP	144	COM	5
LFXP2-5E-5QN208C	1.2V	-5	Lead-Free PQFP	208	COM	5
LFXP2-5E-6QN208C	1.2V	-6	Lead-Free PQFP	208	COM	5
LFXP2-5E-7QN208C	1.2V	-7	Lead-Free PQFP	208	COM	5
LFXP2-5E-5FTN256C	1.2V	-5	Lead-Free ftBGA	256	COM	5
LFXP2-5E-6FTN256C	1.2V	-6	Lead-Free ftBGA	256	COM	5
LFXP2-5E-7FTN256C	1.2V	-7	Lead-Free ftBGA	256	COM	5

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-8E-5MN132C	1.2V	-5	Lead-Free csBGA	132	COM	8
LFXP2-8E-6MN132C	1.2V	-6	Lead-Free csBGA	132	COM	8
LFXP2-8E-7MN132C	1.2V	-7	Lead-Free csBGA	132	COM	8
LFXP2-8E-5TN144C	1.2V	-5	Lead-Free TQFP	144	COM	8
LFXP2-8E-6TN144C	1.2V	-6	Lead-Free TQFP	144	COM	8
LFXP2-8E-7TN144C	1.2V	-7	Lead-Free TQFP	144	COM	8
LFXP2-8E-5QN208C	1.2V	-5	Lead-Free PQFP	208	COM	8
LFXP2-8E-6QN208C	1.2V	-6	Lead-Free PQFP	208	COM	8
LFXP2-8E-7QN208C	1.2V	-7	Lead-Free PQFP	208	COM	8
LFXP2-8E-5FTN256C	1.2V	-5	Lead-Free ftBGA	256	COM	8
LFXP2-8E-6FTN256C	1.2V	-6	Lead-Free ftBGA	256	COM	8
LFXP2-8E-7FTN256C	1.2V	-7	Lead-Free ftBGA	256	COM	8

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-17E-5QN208C	1.2V	-5	Lead-Free PQFP	208	COM	17
LFXP2-17E-6QN208C	1.2V	-6	Lead-Free PQFP	208	COM	17
LFXP2-17E-7QN208C	1.2V	-7	Lead-Free PQFP	208	COM	17
LFXP2-17E-5FTN256C	1.2V	-5	Lead-Free ftBGA	256	COM	17
LFXP2-17E-6FTN256C	1.2V	-6	Lead-Free ftBGA	256	COM	17
LFXP2-17E-7FTN256C	1.2V	-7	Lead-Free ftBGA	256	COM	17
LFXP2-17E-5FN484C	1.2V	-5	Lead-Free fpBGA	484	COM	17
LFXP2-17E-6FN484C	1.2V	-6	Lead-Free fpBGA	484	COM	17
LFXP2-17E-7FN484C	1.2V	-7	Lead-Free fpBGA	484	COM	17

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-40E-5F484C	1.2V	-5	fpBGA	484	COM	40
LFXP2-40E-6F484C	1.2V	-6	fpBGA	484	COM	40
LFXP2-40E-7F484C	1.2V	-7	fpBGA	484	COM	40
LFXP2-40E-5F672C	1.2V	-5	fpBGA	672	COM	40
LFXP2-40E-6F672C	1.2V	-6	fpBGA	672	COM	40
LFXP2-40E-7F672C	1.2V	-7	fpBGA	672	COM	40

Industrial

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-5E-5M132I	1.2V	-5	csBGA	132	IND	5
LFXP2-5E-6M132I	1.2V	-6	csBGA	132	IND	5
LFXP2-5E-6FT256I	1.2V	-6	ftBGA	256	IND	5

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-8E-5M132I	1.2V	-5	csBGA	132	IND	8
LFXP2-8E-6M132I	1.2V	-6	csBGA	132	IND	8
LFXP2-5E-5FT256I	1.2V	-5	ftBGA	256	IND	5
LFXP2-8E-5FT256I	1.2V	-5	ftBGA	256	IND	8
LFXP2-8E-6FT256I	1.2V	-6	ftBGA	256	IND	8

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-17E-5FT256I	1.2V	-5	ftBGA	256	IND	17
LFXP2-17E-6FT256I	1.2V	-6	ftBGA	256	IND	17
LFXP2-17E-5F484I	1.2V	-5	fpBGA	484	IND	17
LFXP2-17E-6F484I	1.2V	-6	fpBGA	484	IND	17

Part Number	Voltage	Grade	Package	Pins	Temp.	LUTs (k)
LFXP2-30E-5FT256I	1.2V	-5	ftBGA	256	IND	30
LFXP2-30E-6FT256I	1.2V	-6	ftBGA	256	IND	30
LFXP2-30E-5F484I	1.2V	-5	fpBGA	484	IND	30
LFXP2-30E-6F484I	1.2V	-6	fpBGA	484	IND	30
LFXP2-30E-5F672I	1.2V	-5	fpBGA	672	IND	30
LFXP2-30E-6F672I	1.2V	-6	fpBGA	672	IND	30

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For Further Information

A variety of technical notes for the LatticeXP2 FPGA family are available on the Lattice Semiconductor web site at www.latticesemi.com.

- TN1136, LatticeXP2 sysIO Usage Guide
- TN1137, LatticeXP2 Memory Usage Guide
- TN1138, LatticeXP2 High Speed I/O Interface
- TN1126, LatticeXP2 sysCLOCK PLL Design and Usage Guide
- TN1139, Power Estimation and Management for LatticeXP2 Devices
- TN1140, LatticeXP2 sysDSP Usage Guide
- TN1141, LatticeXP2 sysCONFIG Usage Guide
- TN1142, LatticeXP2 Configuration Encryption and Security Usage Guide
- TN1087, Minimizing System Interruption During Configuration Using TransFR Technology
- TN1220, LatticeXP2 Dual Boot Feature
- TN1130, LatticeXP2 Soft Error Detection (SED) Usage Guide
- TN1143, LatticeXP2 Hardware Checklist

For further information on interface standards refer to the following websites:

- JEDEC Standards (LVTTL, LVCMOS, SSTL, HSTL): www.jedec.org
- PCI: www.pcisig.com