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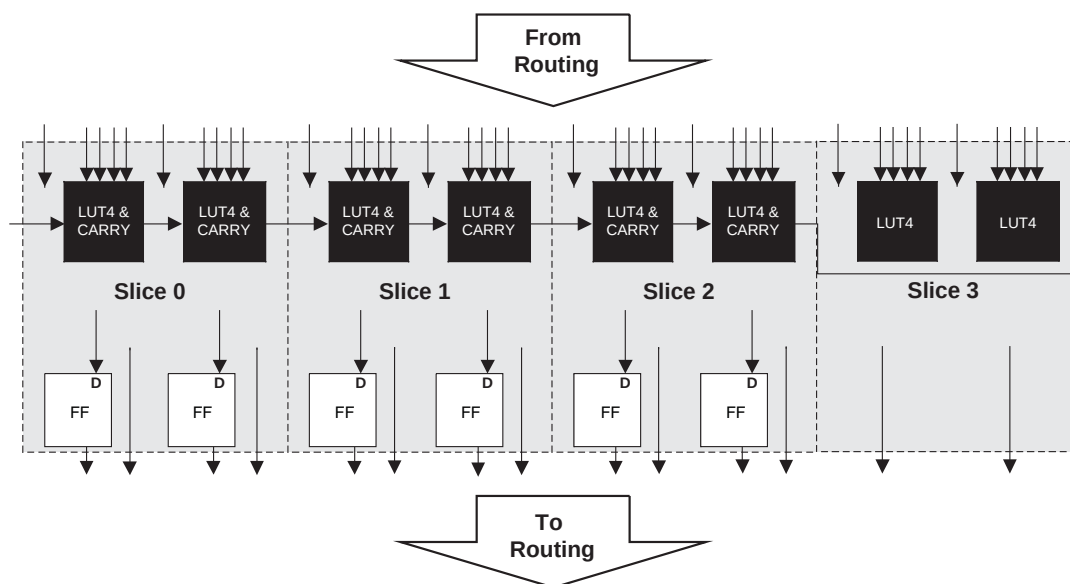
Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Active
Number of LABs/CLBs	625
Number of Logic Elements/Cells	5000
Total RAM Bits	169984
Number of I/O	146
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lfxp2-5e-6qn208i

Figure 2-2. PFU Diagram



Slice

Slice 0 through Slice 2 contain two 4-input combinatorial Look-Up Tables (LUT4), which feed two registers. Slice 3 contains two LUT4s and no registers. For PFUs, Slice 0 and Slice 2 can also be configured as distributed memory, a capability not available in PFF blocks. Table 2-1 shows the capability of the slices in both PFF and PFU blocks along with the operation modes they enable. In addition, each PFU contains logic that allows the LUTs to be combined to perform functions such as LUT5, LUT6, LUT7 and LUT8. There is control logic to perform set/reset functions (programmable as synchronous/asynchronous), clock select, chip-select and wider RAM/ROM functions. Figure 2-3 shows an overview of the internal logic of the slice. The registers in the slice can be configured as positive/negative edge triggered or level sensitive clocks.

Table 2-1. Resources and Modes Available per Slice

Slice	PFU BBlock		PFF Block	
	Resources	Modes	Resources	Modes
Slice 0	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 1	2 LUT4s and 2 Registers	Logic, Ripple, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 2	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 3	2 LUT4s	Logic, ROM	2 LUT4s	Logic, ROM

Slice 0 through Slice 2 have 14 input signals: 13 signals from routing and one from the carry-chain (from the adjacent slice or PFU). There are seven outputs: six to routing and one to carry-chain (to the adjacent PFU). Slice 3 has 13 input signals from routing and four signals to routing. Table 2-2 lists the signals associated with Slice 0 to Slice 2.

Routing

There are many resources provided in the LatticeXP2 devices to route signals individually or as busses with related control signals. The routing resources consist of switching circuitry, buffers and metal interconnect (routing) segments.

The inter-PFU connections are made with x1 (spans two PFU), x2 (spans three PFU) or x6 (spans seven PFU) connections. The x1 and x2 connections provide fast and efficient connections in horizontal and vertical directions. The x2 and x6 resources are buffered to allow both short and long connections routing between PFUs.

The LatticeXP2 family has an enhanced routing architecture to produce a compact design. The Diamond design tool takes the output of the synthesis tool and places and routes the design. Generally, the place and route tool is completely automatic, although an interactive routing editor is available to optimize the design.

sysCLOCK Phase Locked Loops (PLL)

The sysCLOCK PLLs provide the ability to synthesize clock frequencies. The LatticeXP2 family supports between two and four full featured General Purpose PLLs (GPLL). The architecture of the GPLL is shown in Figure 2-4.

CLKI, the PLL reference frequency, is provided either from the pin or from routing; it feeds into the Input Clock Divider block. CLKFB, the feedback signal, is generated from CLKOP (the primary clock output) or from a user clock pin/logic. CLKFB feeds into the Feedback Divider and is used to multiply the reference frequency.

Both the input path and feedback signals enter the Voltage Controlled Oscillator (VCO) block. The phase and frequency of the VCO are determined from the input path and feedback signals. A LOCK signal is generated by the VCO to indicate that the VCO is locked with the input clock signal.

The output of the VCO feeds into the CLKOP Divider, a post-scalar divider. The duty cycle of the CLKOP Divider output can be fine tuned using the Duty Trim block, which creates the CLKOP signal. By allowing the VCO to operate at higher frequencies than CLKOP, the frequency range of the GPLL is expanded. The output of the CLKOP Divider is passed through the CLKOK Divider, a secondary clock divider, to generate lower frequencies for the CLKOK output. For applications that require even lower frequencies, the CLKOP signal is passed through a divide-by-three divider to produce the CLKOK2 output. The CLKOK2 output is provided for applications that use source synchronous logic. The Phase/Duty Cycle/Duty Trim block is used to adjust the phase and duty cycle of the CLKOP Divider output to generate the CLKOS signal. The phase/duty cycle setting can be pre-programmed or dynamically adjusted.

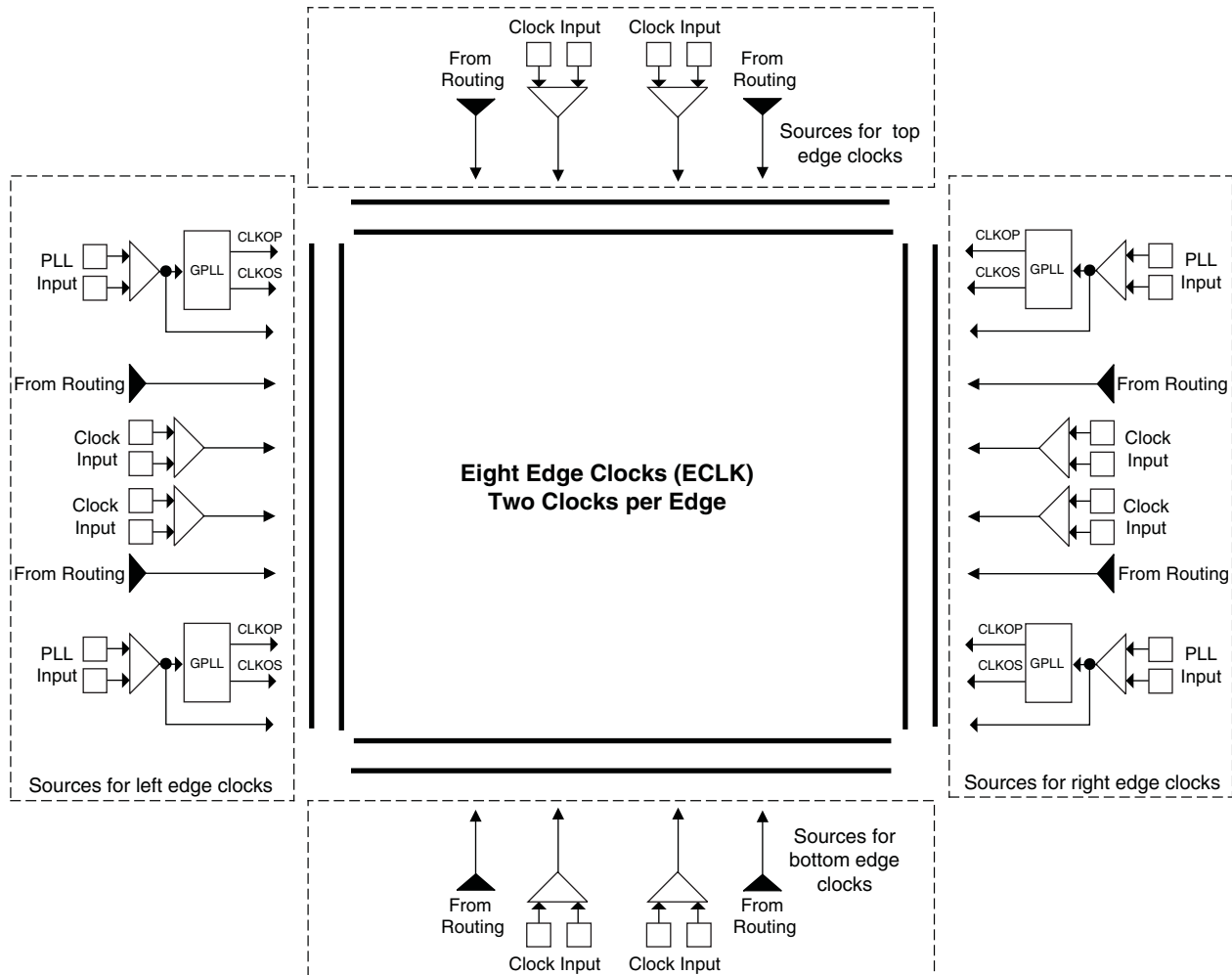
The clock outputs from the GPLL; CLKOP, CLKOK, CLKOK2 and CLKOS, are fed to the clock distribution network.

For further information on the GPLL please see TN1126, [LatticeXP2 sysCLOCK PLL Design and Usage Guide](#).

Edge Clock Sources

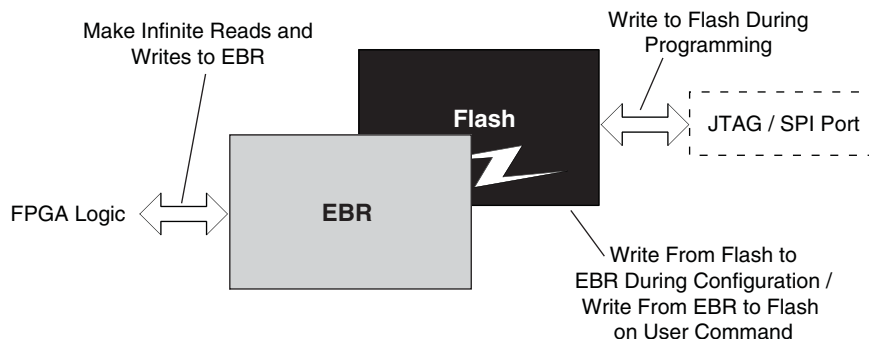
Edge clock resources can be driven from a variety of sources at the same edge. Edge clock resources can be driven from adjacent edge clock PIOs, primary clock PIOs, PLLs and clock dividers as shown in Figure 2-8.

Figure 2-8. Edge Clock Sources



Note: This diagram shows sources for the XP2-17 device. Smaller LatticeXP2 devices have two GPLLs.

Figure 2-16. FlashBAK Technology



Memory Cascading

Larger and deeper blocks of RAMs can be created using EBR sysMEM Blocks. Typically, the Lattice design tools cascade memory transparently, based on specific design inputs.

Single, Dual and Pseudo-Dual Port Modes

In all the sysMEM RAM modes the input data and address for the ports are registered at the input of the memory array. The output data of the memory is optionally registered at the output.

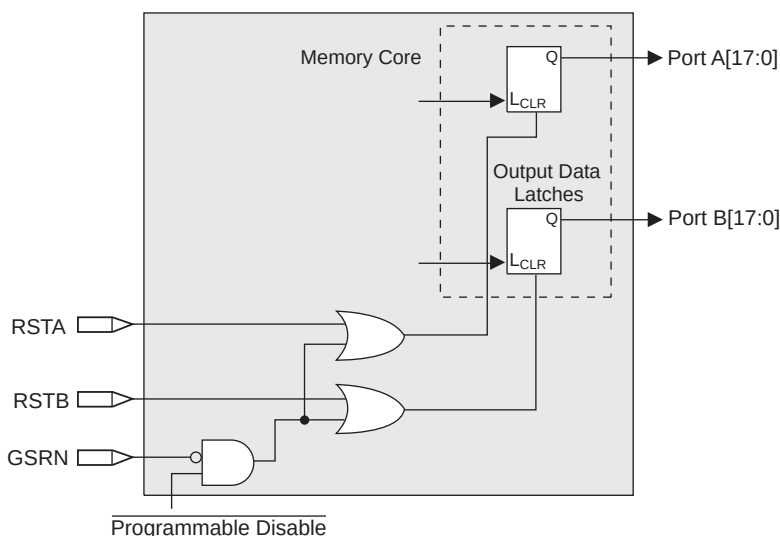
EBR memory supports two forms of write behavior for single port or dual port operation:

1. Normal – Data on the output appears only during a read cycle. During a write cycle, the data (at the current address) does not appear on the output. This mode is supported for all data widths.
2. Write Through – A copy of the input data appears at the output of the same port during a write cycle. This mode is supported for all data widths.

Memory Core Reset

The memory array in the EBR utilizes latches at the A and B output ports. These latches can be reset asynchronously or synchronously. RSTA and RSTB are local signals, which reset the output latches associated with Port A and Port B respectively. GSRN, the global reset signal, resets both ports. The output data latches and associated resets for both ports are as shown in Figure 2-17.

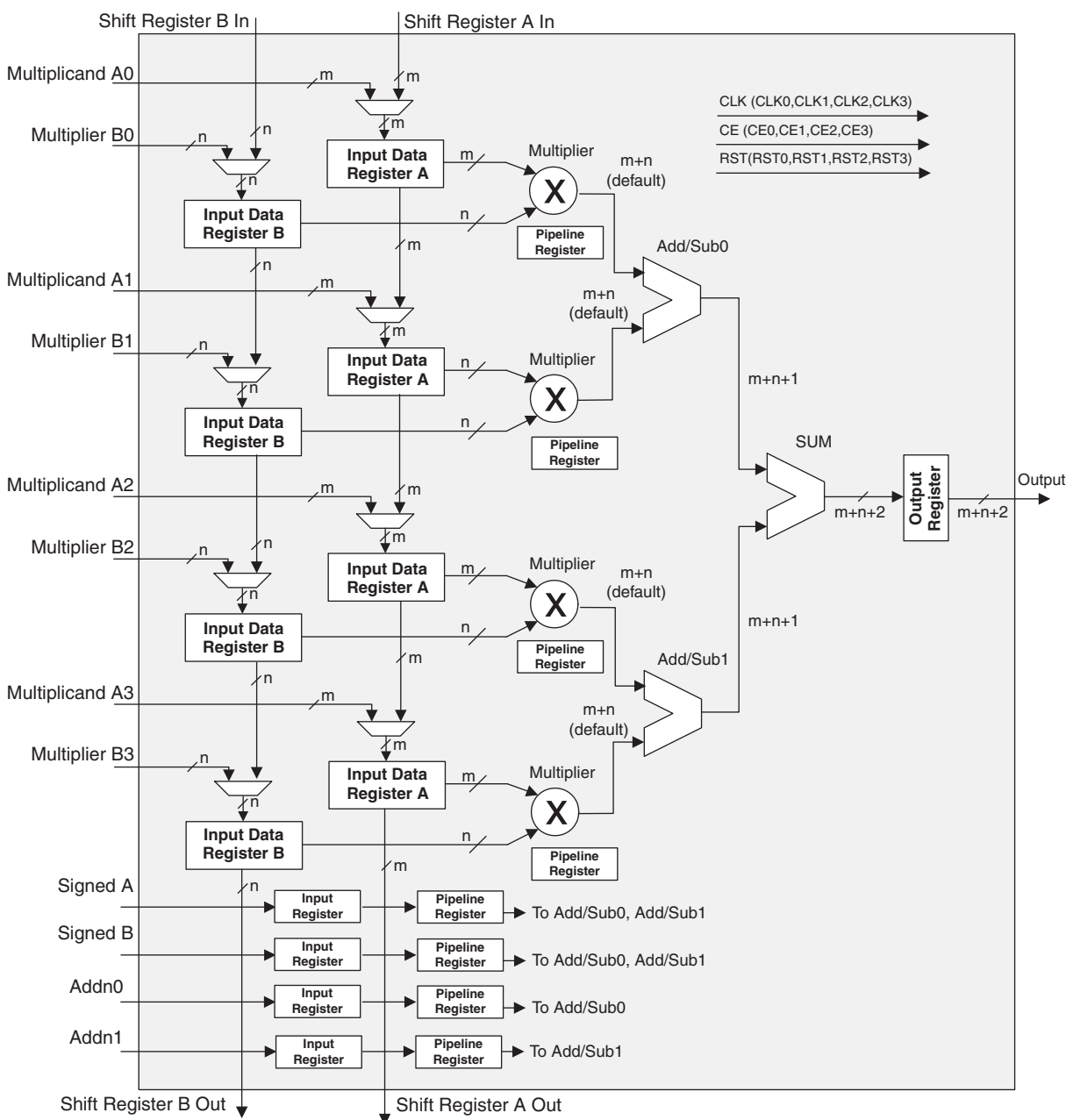
Figure 2-17. Memory Core Reset



MULTADDSUBSUM sysDSP Element

In this case, the operands A0 and B0 are multiplied and the result is added/subtracted with the result of the multiplier operation of operands A1 and B1. Additionally the operands A2 and B2 are multiplied and the result is added/subtracted with the result of the multiplier operation of operands A3 and B3. The result of both addition/subtraction are added in a summation block. The user can enable the input, output and pipeline registers. Figure 2-23 shows the MULTADDSUBSUM sysDSP element.

Figure 2-23. MULTADDSUBSUM



Clock, Clock Enable and Reset Resources

Global Clock, Clock Enable (CE) and Reset (RST) signals from routing are available to every DSP block. From four clock sources (CLK0, CLK1, CLK2, CLK3) one clock is selected for each input register, pipeline register and output

IPexpress™

The user can access the sysDSP block via the Lattice IPexpress tool, which provides the option to configure each DSP module (or group of modules), or by direct HDL instantiation. In addition, Lattice has partnered with The MathWorks® to support instantiation in the Simulink® tool, a graphical simulation environment. Simulink works with Diamond to dramatically shorten the DSP design cycle in Lattice FPGAs.

Optimized DSP Functions

Lattice provides a library of optimized DSP IP functions. Some of the IP cores planned for the LatticeXP2 DSP include the Bit Correlator, FFT functions, FIR Filter, Reed-Solomon Encoder/Decoder, Turbo Encoder/Decoder and Convolutional Encoder/Decoder. Please contact Lattice to obtain the latest list of available DSP IP cores.

Resources Available in the LatticeXP2 Family

Table 2-8 shows the maximum number of multipliers for each member of the LatticeXP2 family. Table 2-9 shows the maximum available EBR RAM Blocks and Serial TAG Memory bits in each LatticeXP2 device. EBR blocks, together with Distributed RAM can be used to store variables locally for fast DSP operations.

Table 2-8. Maximum Number of DSP Blocks in the LatticeXP2 Family

Device	DSP Block	9x9 Multiplier	18x18 Multiplier	36x36 Multiplier
XP2-5	3	24	12	3
XP2-8	4	32	16	4
XP2-17	5	40	20	5
XP2-30	7	56	28	7
XP2-40	8	64	32	8

Table 2-9. Embedded SRAM/TAG Memory in the LatticeXP2 Family

Device	EBR SRAM Block	Total EBR SRAM (Kbits)	TAG Memory (Bits)
XP2-5	9	166	632
XP2-8	12	221	768
XP2-17	15	276	2184
XP2-30	21	387	2640
XP2-40	48	885	3384

LatticeXP2 DSP Performance

Table 2-10 lists the maximum performance in Millions of MAC (MMAC) operations per second for each member of the LatticeXP2 family.

Table 2-10. DSP Performance

Device	DSP Block	DSP Performance MMAC
XP2-5	3	3,900
XP2-8	4	5,200
XP2-17	5	6,500
XP2-30	7	9,100
XP2-40	8	10,400

For further information on the sysDSP block, please see TN1140, [LatticeXP2 sysDSP Usage Guide](#).

shows the diagram using this gearbox function. For more information on this topic, see TN1138, [LatticeXP2 High Speed I/O Interface](#).

Figure 2-27. Output and Tristate Block

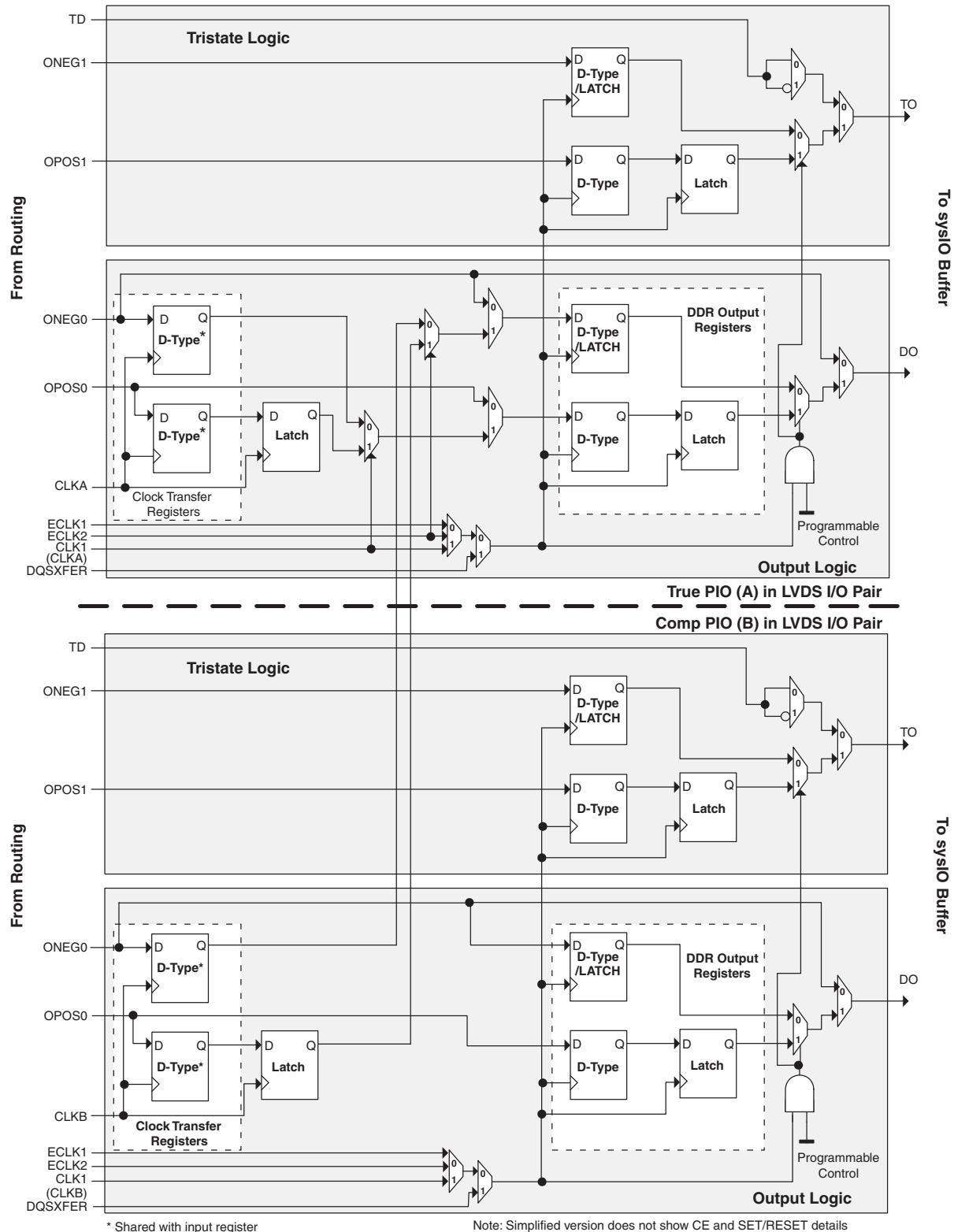
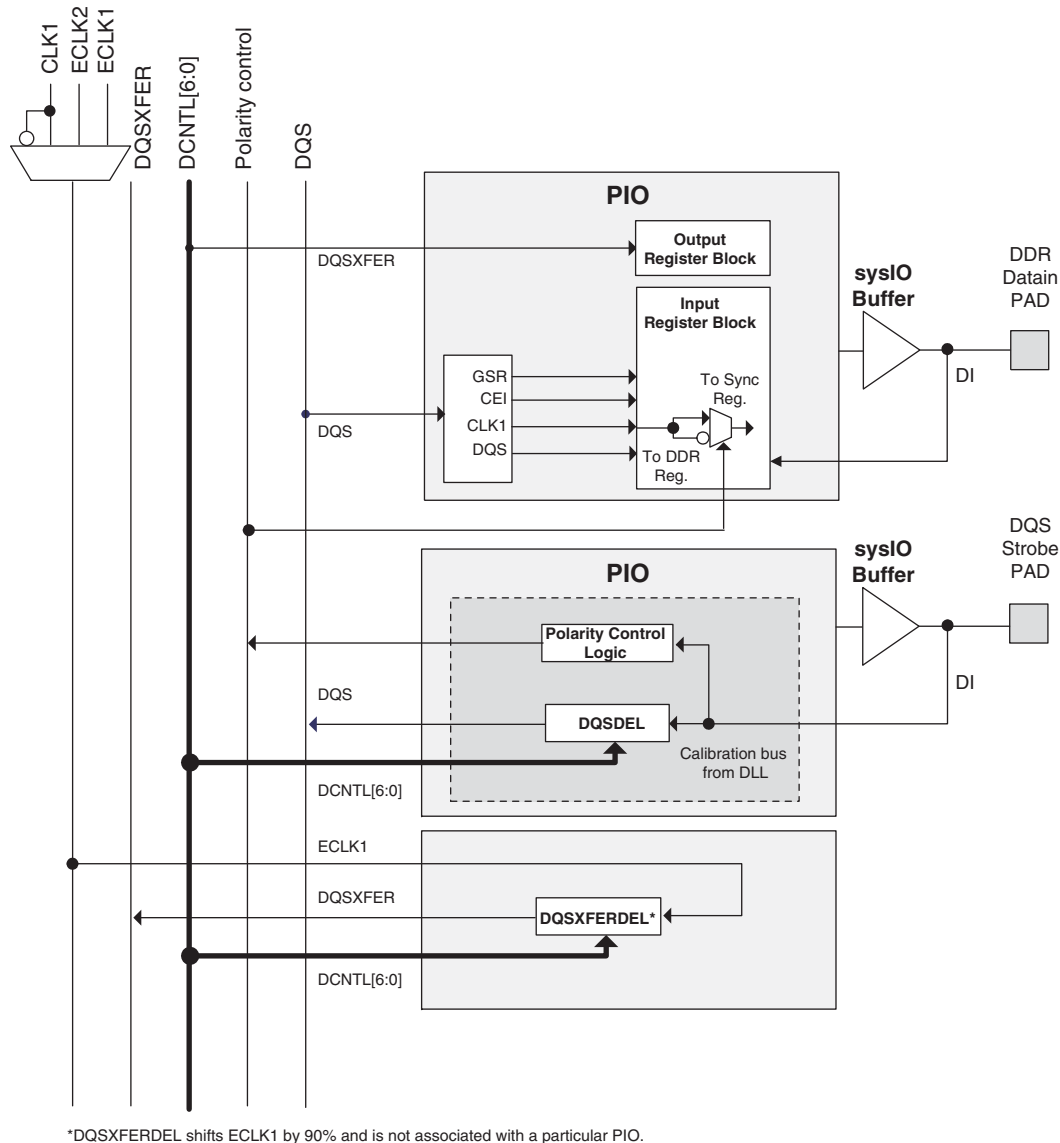


Figure 2-31. DQS Local Bus



*DQSXFERDEL shifts ECLK1 by 90° and is not associated with a particular PIO.

Polarity Control Logic

In a typical DDR memory interface design, the phase relationship between the incoming delayed DQS strobe and the internal system clock (during the READ cycle) is unknown. The LatticeXP2 family contains dedicated circuits to transfer data between these domains. To prevent set-up and hold violations, at the domain transfer between DQS (delayed) and the system clock, a clock polarity selector is used. This changes the edge on which the data is registered in the synchronizing registers in the input register block and requires evaluation at the start of each READ cycle for the correct clock polarity.

Prior to the READ operation in DDR memories, DQS is in tristate (pulled by termination). The DDR memory device drives DQS low at the start of the preamble state. A dedicated circuit detects this transition. This signal is used to control the polarity of the clock to the synchronizing registers.

Table 2-13. Supported Output Standards

Output Standard	Drive	V _{CCIO} (Nom.)
Single-ended Interfaces		
LVTTL	4mA, 8mA, 12mA, 16mA, 20mA	3.3
LVC MOS33	4mA, 8mA, 12mA 16mA, 20mA	3.3
LVC MOS25	4mA, 8mA, 12mA, 16mA, 20mA	2.5
LVC MOS18	4mA, 8mA, 12mA, 16mA	1.8
LVC MOS15	4mA, 8mA	1.5
LVC MOS12	2mA, 6mA	1.2
LVC MOS33, Open Drain	4mA, 8mA, 12mA 16mA, 20mA	—
LVC MOS25, Open Drain	4mA, 8mA, 12mA 16mA, 20mA	—
LVC MOS18, Open Drain	4mA, 8mA, 12mA 16mA	—
LVC MOS15, Open Drain	4mA, 8mA	—
LVC MOS12, Open Drain	2mA, 6mA	—
PCI33	N/A	3.3
HSTL18 Class I, II	N/A	1.8
HSTL15 Class I	N/A	1.5
SSTL33 Class I, II	N/A	3.3
SSTL25 Class I, II	N/A	2.5
SSTL18 Class I, II	N/A	1.8
Differential Interfaces		
Differential SSTL33, Class I, II	N/A	3.3
Differential SSTL25, Class I, II	N/A	2.5
Differential SSTL18, Class I, II	N/A	1.8
Differential HSTL18, Class I, II	N/A	1.8
Differential HSTL15, Class I	N/A	1.5
LVDS ^{1, 2}	N/A	2.5
MLVDS ¹	N/A	2.5
BLVDS ¹	N/A	2.5
LVPECL ¹	N/A	3.3
RSDS ¹	N/A	2.5
LVC MOS33D ¹	4mA, 8mA, 12mA, 16mA, 20mA	3.3

1. Emulated with external resistors.

2. On the left and right edges, LVDS outputs are supported with a dedicated differential output driver on 50% of the I/Os. This solution does not require external resistors at the driver.

Hot Socketing

LatticeXP2 devices have been carefully designed to ensure predictable behavior during power-up and power-down. Power supplies can be sequenced in any order. During power-up and power-down sequences, the I/Os remain in tri-state until the power supply voltage is high enough to ensure reliable operation. In addition, leakage into I/O pins is controlled to within specified limits. This allows for easy integration with the rest of the system. These capabilities make the LatticeXP2 ideal for many multiple power supply and hot-swap applications.

IEEE 1149.1-Compliant Boundary Scan Testability

All LatticeXP2 devices have boundary scan cells that are accessed through an IEEE 1149.1 compliant Test Access Port (TAP). This allows functional testing of the circuit board, on which the device is mounted, through a serial scan path that can access all critical logic nodes. Internal registers are linked internally, allowing test data to be shifted in

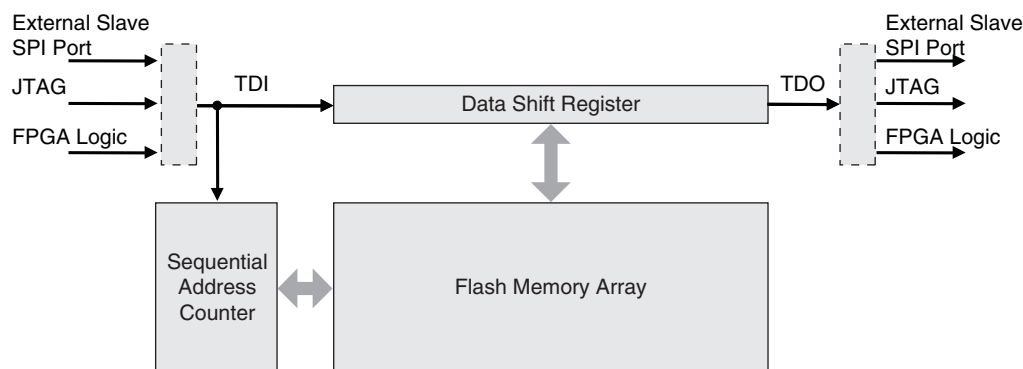
1. Unlocked
2. Key Locked – Presenting the key through the programming interface allows the device to be unlocked.
3. Permanently Locked – The device is permanently locked.

To further complement the security of the device a One Time Programmable (OTP) mode is available. Once the device is set in this mode it is not possible to erase or re-program the Flash portion of the device.

Serial TAG Memory

LatticeXP2 devices offer 0.6 to 3.3kbits of Flash memory in the form of Serial TAG memory. The TAG memory is an area of the on-chip Flash that can be used for non-volatile storage including electronic ID codes, version codes, date stamps, asset IDs and calibration settings. A block diagram of the TAG memory is shown in Figure 2-34. The TAG memory is accessed in the same way as external SPI Flash and it can be read or programmed either through JTAG, an external Slave SPI Port, or directly from FPGA logic. To read the TAG memory, a start address is specified and the entire TAG memory contents are read sequentially in a first-in-first-out manner. The TAG memory is independent of the Flash used for device configuration and given its use for general-purpose storage functions is always accessible regardless of the device security settings. For more information, see TN1137, [LatticeXP2 Memory Usage Guide](#) and TN1141, [LatticeXP2 sysCONFIG Usage Guide](#).

Figure 2-34. Serial TAG Memory Diagram



Live Update Technology

Many applications require field updates of the FPGA. LatticeXP2 devices provide three features that enable this configuration to be done in a secure and failsafe manner while minimizing impact on system operation.

1. **Decryption Support**
LatticeXP2 devices provide on-chip, non-volatile key storage to support decryption of a 128-bit AES encrypted bitstream, securing designs and deterring design piracy.
2. **TransFR (Transparent Field Reconfiguration)**
TransFR I/O (TFR) is a unique Lattice technology that allows users to update their logic in the field without interrupting system operation using a single ispVM command. TransFR I/O allows I/O states to be frozen during device configuration. This allows the device to be field updated with a minimum of system disruption and downtime. For more information please see TN1087, [Minimizing System Interruption During Configuration Using TransFR Technology](#).
3. **Dual Boot Image Support**
Dual boot images are supported for applications requiring reliable remote updates of configuration data for the system FPGA. After the system is running with a basic configuration, a new boot image can be downloaded remotely and stored in a separate location in the configuration storage device. Any time after the update the LatticeXP2 can be re-booted from this new configuration file. If there is a problem such as corrupt data during download or incorrect version number with this new boot image, the LatticeXP2 device can revert back to the

Density Shifting

The LatticeXP2 family is designed to ensure that different density devices in the same family and in the same package have the same pinout. Furthermore, the architecture ensures a high success rate when performing design migration from lower density devices to higher density devices. In many cases, it is also possible to shift a lower utilization design targeted for a high-density device to a lower density device. However, the exact details of the final resource utilization will impact the likely success in each case.

Hot Socketing Specifications^{1, 2, 3, 4}

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
I_{DK}	Input or I/O Leakage Current	$0 \leq V_{IN} \leq V_{IH} \text{ (MAX.)}$	—	—	+/-1	mA

1. Insensitive to sequence of V_{CC} , V_{CCAUX} and V_{CCIO} . However, assumes monotonic rise/fall rates for V_{CC} , V_{CCAUX} and V_{CCIO} .
2. $0 \leq V_{CC} \leq V_{CC} \text{ (MAX)}$, $0 \leq V_{CCIO} \leq V_{CCIO} \text{ (MAX)}$ or $0 \leq V_{CCAUX} \leq V_{CCAUX} \text{ (MAX)}$.
3. I_{DK} is additive to I_{PU} , I_{PW} or I_{BH} .
4. LVCMOS and LVTTL only.

ESD Performance

Please refer to the [LatticeXP2 Product Family Qualification Summary](#) for complete qualification data, including ESD performance.

DC Electrical Characteristics

Over Recommended Operating Conditions

Symbol	Parameter	Condition	Min.	Typ.	Max.	Units
I_{IL} , I_{IH} ¹	Input or I/O Low Leakage	$0 \leq V_{IN} \leq V_{CCIO}$	—	—	10	μA
		$V_{CCIO} \leq V_{IN} \leq V_{IH} \text{ (MAX)}$	—	—	150	μA
I_{PU}	I/O Active Pull-up Current	$0 \leq V_{IN} \leq 0.7 V_{CCIO}$	-30	—	-150	μA
I_{PD}	I/O Active Pull-down Current	$V_{IL} \text{ (MAX)} \leq V_{IN} \leq V_{CCIO}$	30	—	210	μA
I_{BHLS}	Bus Hold Low Sustaining Current	$V_{IN} = V_{IL} \text{ (MAX)}$	30	—	—	μA
I_{BHHS}	Bus Hold High Sustaining Current	$V_{IN} = 0.7 V_{CCIO}$	-30	—	—	μA
I_{BHLO}	Bus Hold Low Overdrive Current	$0 \leq V_{IN} \leq V_{CCIO}$	—	—	210	μA
I_{BHHO}	Bus Hold High Overdrive Current	$0 \leq V_{IN} \leq V_{CCIO}$	—	—	-150	μA
V_{BHT}	Bus Hold Trip Points		$V_{IL} \text{ (MAX)}$	—	$V_{IH} \text{ (MIN)}$	V
C1	I/O Capacitance ²	$V_{CCIO} = 3.3V, 2.5V, 1.8V, 1.5V, 1.2V$, $V_{CC} = 1.2V$, $V_{IO} = 0$ to $V_{IH} \text{ (MAX)}$	—	8	—	pf
C2	Dedicated Input Capacitance	$V_{CCIO} = 3.3V, 2.5V, 1.8V, 1.5V, 1.2V$, $V_{CC} = 1.2V$, $V_{IO} = 0$ to $V_{IH} \text{ (MAX)}$	—	6	—	pf

1. Input or I/O leakage current is measured with the pin configured as an input or as an I/O with the output driver tri-stated. It is not measured with the output driver active. Bus maintenance circuits are disabled.
2. T_A 25°C, $f = 1.0$ MHz.

Programming and Erase Flash Supply Current^{1, 2, 3, 4, 5}

Over Recommended Operating Conditions

Symbol	Parameter	Device	Typical (25°C, Max. Supply) ⁶	Units
I_{CC}	Core Power Supply Current	XP2-5	17	mA
		XP2-8	21	mA
		XP2-17	28	mA
		XP2-30	36	mA
		XP2-40	50	mA
I_{CCAUX}	Auxiliary Power Supply Current ⁷	XP2-5	64	mA
		XP2-8	66	mA
		XP2-17	83	mA
		XP2-30	87	mA
		XP2-40	88	mA
I_{CCPLL}	PLL Power Supply Current (per PLL)		0.1	mA
I_{CCIO}	Bank Power Supply Current (per Bank)		5	mA
I_{CCJ}	V_{CCJ} Power Supply Current ⁸		14	mA

1. For further information on supply current, please see TN1139, [Power Estimation and Management for LatticeXP2 Devices](#).
2. Assumes all outputs are tristated, all inputs are configured as LVCMOS and held at the V_{CCIO} or GND.
3. Frequency 0 MHz (excludes dynamic power from FPGA operation).
4. A specific configuration pattern is used that scales with the size of the device; consists of 75% PFU utilization, 50% EBR, and 25% I/O configuration.
5. Bypass or decoupling capacitor across the supply.
6. $T_J = 25^\circ\text{C}$, power supplies at nominal voltage.
7. In fpBGA and ftBGA packages the PLLs are connected to and powered from the auxiliary power supply. For these packages, the actual auxiliary supply current is the sum of I_{CCAUX} and I_{CCPLL} . For csBGA, PQFP and TQFP packages the PLLs are powered independent of the auxiliary power supply.
8. When programming via JTAG.

sysIO Recommended Operating Conditions

Over Recommended Operating Conditions

Standard	V _{CCIO}			V _{REF} (V)		
	Min.	Typ.	Max.	Min.	Typ.	Max.
LVC MOS33 ²	3.135	3.3	3.465	—	—	—
LVC MOS25 ²	2.375	2.5	2.625	—	—	—
LVC MOS18	1.71	1.8	1.89	—	—	—
LVC MOS15	1.425	1.5	1.575	—	—	—
LVC MOS12 ²	1.14	1.2	1.26	—	—	—
LV TTL33 ²	3.135	3.3	3.465	—	—	—
PCI33	3.135	3.3	3.465	—	—	—
SSTL18_I ² , SSTL18_II ²	1.71	1.8	1.89	0.833	0.9	0.969
SSTL25_I ² , SSTL25_II ²	2.375	2.5	2.625	1.15	1.25	1.35
SSTL33_I ² , SSTL33_II ²	3.135	3.3	3.465	1.3	1.5	1.7
HSTL15_I ²	1.425	1.5	1.575	0.68	0.75	0.9
HSTL18_I ² , HSTL18_II ²	1.71	1.8	1.89	0.816	0.9	1.08
LVDS25 ²	2.375	2.5	2.625	—	—	—
MLVDS25 ¹	2.375	2.5	2.625	—	—	—
LVPECL33 ^{1, 2}	3.135	3.3	3.465	—	—	—
BLVDS25 ^{1, 2}	2.375	2.5	2.625	—	—	—
RSDS ^{1, 2}	2.375	2.5	2.625	—	—	—
SSTL18D_I ² , SSTL18D_II ²	1.71	1.8	1.89	—	—	—
SSTL25D_I ² , SSTL25D_II ²	2.375	2.5	2.625	—	—	—
SSTL33D_I ² , SSTL33D_II ²	3.135	3.3	3.465	—	—	—
HSTL15D_I ²	1.425	1.5	1.575	—	—	—
HSTL18D_I ² , HSTL18D_II ²	1.71	1.8	1.89	—	—	—

1. Inputs on chip. Outputs are implemented with the addition of external resistors.

2. Input on this standard does not depend on the value of V_{CCIO}.

MLVDS

The LatticeXP2 devices support the differential MLVDS standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The MLVDS input standard is supported by the LVDS differential input buffer. The scheme shown in Figure 3-5 is one possible solution for MLVDS standard implementation. Resistor values in Figure 3-5 are industry standard values for 1% resistors.

Figure 3-5. MLVDS (Reduced Swing Differential Standard)

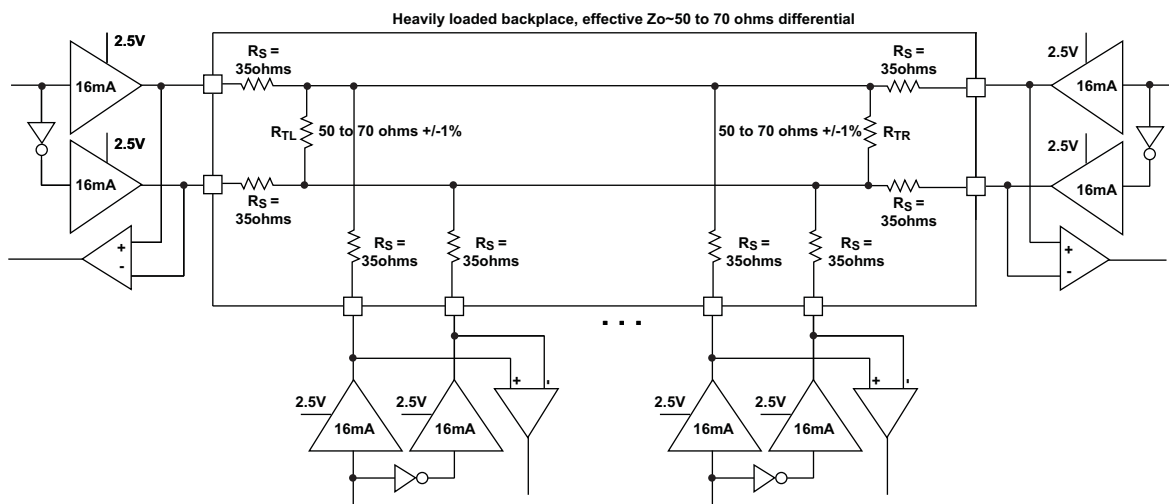


Table 3-5. MLVDS DC Conditions¹

Parameter	Description	Typical		Units
		$Z_o=50\Omega$	$Z_o=70\Omega$	
V_{CCIO}	Output Driver Supply (+/-5%)	2.50	2.50	V
Z_{OUT}	Driver Impedance	10.00	10.00	Ω
R_S	Driver Series Resistor (+/-1%)	35.00	35.00	Ω
R_{TL}	Driver Parallel Resistor (+/-1%)	50.00	70.00	Ω
R_{TR}	Receiver Termination (+/-1%)	50.00	70.00	Ω
V_{OH}	Output High Voltage (After R_{TL})	1.52	1.60	V
V_{OL}	Output Low Voltage (After R_{TL})	0.98	0.90	V
V_{OD}	Output Differential Voltage (After R_{TL})	0.54	0.70	V
V_{CM}	Output Common Mode Voltage	1.25	1.25	V
I_{DC}	DC Output Current	21.74	20.00	mA

1. For input buffer, see LVDS table.

For further information on LVPECL, RSDS, MLVDS, BLVDS and other differential interfaces please see details of additional technical information at the end of this data sheet.

LatticeXP2 External Switching Characteristics (Continued)

Over Recommended Operating Conditions

Parameter	Description	Device	-7		-6		-5		Units
			Min.	Max.	Min.	Max.	Min.	Max.	
t _{H_DELPLL}	Clock to Data Hold - PIO Input Register with Input Data Delay	XP2-5	0.00	—	0.00	—	0.00	—	ns
		XP2-8	0.00	—	0.00	—	0.00	—	ns
		XP2-17	0.00	—	0.00	—	0.00	—	ns
		XP2-30	0.00	—	0.00	—	0.00	—	ns
		XP2-40	0.00	—	0.00	—	0.00	—	ns
DDR ² and DDR2 ³ I/O Pin Parameters									
t _{DVADQ}	Data Valid After DQS (DDR Read)	XP2	—	0.29	—	0.29	—	0.29	UI
t _{DVEDQ}	Data Hold After DQS (DDR Read)	XP2	0.71	—	0.71	—	0.71	—	UI
t _{DQVBS}	Data Valid Before DQS	XP2	0.25	—	0.25	—	0.25	—	UI
t _{DQVAS}	Data Valid After DQS	XP2	0.25	—	0.25	—	0.25	—	UI
f _{MAX_DDR}	DDR Clock Frequency	XP2	95	200	95	166	95	133	MHz
f _{MAX_DDR2}	DDR Clock Frequency	XP2	133	200	133	200	133	166	MHz
Primary Clock									
f _{MAX_PRI}	Frequency for Primary Clock Tree	XP2	—	420	—	357	—	311	MHz
t _{W_PRI}	Clock Pulse Width for Primary Clock	XP2	1	—	1	—	1	—	ns
t _{SKEW_PRI}	Primary Clock Skew Within a Bank	XP2	—	160	—	160	—	160	ps
Edge Clock (ECLK1 and ECLK2)									
f _{MAX_ECLK}	Frequency for Edge Clock	XP2	—	420	—	357	—	311	MHz
t _{W_ECLK}	Clock Pulse Width for Edge Clock	XP2	1	—	1	—	1	—	ns
t _{SKEW_ECLK}	Edge Clock Skew Within an Edge of the Device	XP2	—	130	—	130	—	130	ps

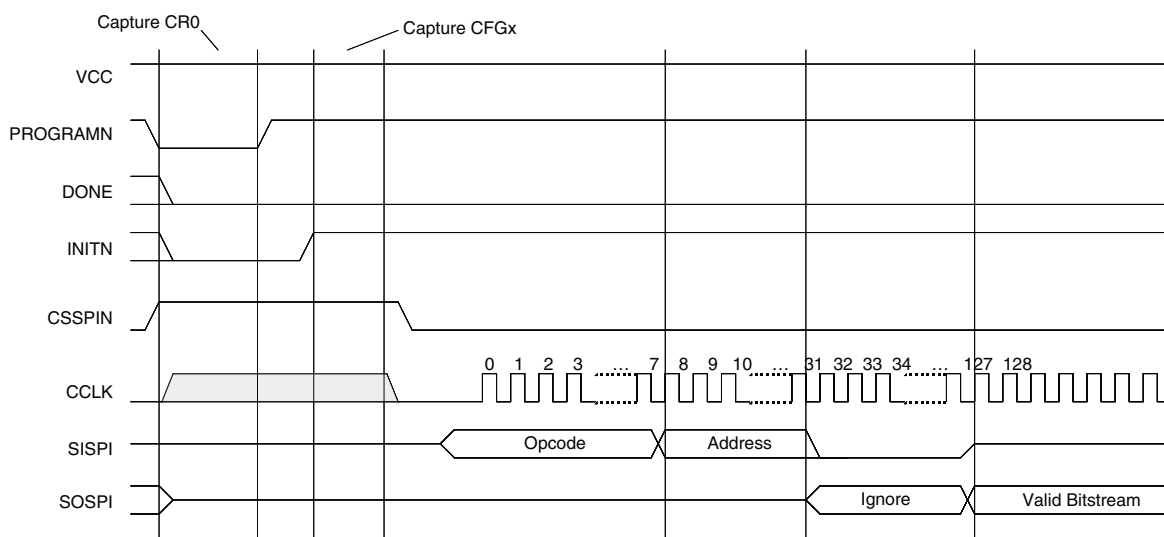
1. General timing numbers based on LVCMOS 2.5, 12mA, 0pf load.
2. DDR timing numbers based on SSTL25.
3. DDR2 timing numbers based on SSTL18.

On-Chip Oscillator and Configuration Master Clock Characteristics

Over Recommended Operating Conditions

Parameter	Min.	Max.	Units
Master Clock Frequency	Selected value -30%	Selected value +30%	MHz
Duty Cycle	40	60	%

Figure 3-9. Master SPI Configuration Waveforms



FlashBAK Time (from EBR to Flash)

Over Recommended Operating Conditions

Device	EBR Density (Bits)	Time (Typ.)	Units
XP2-5	166K	1.5	s
XP2-8	221K	1.5	s
XP2-17	276K	1.5	s
XP2-30	387K	2.0	s
XP2-40	885K	3.0	s

JTAG Port Timing Specifications

Over Recommended Operating Conditions

Symbol	Parameter	Min.	Max.	Units
f_{MAX}	TCK Clock Frequency	—	25	MHz
t_{BTCP}	TCK [BSCAN] clock pulse width	40	—	ns
t_{BTCPH}	TCK [BSCAN] clock pulse width high	20	—	ns
t_{BTCPL}	TCK [BSCAN] clock pulse width low	20	—	ns
t_{BTS}	TCK [BSCAN] setup time	8	—	ns
t_{BTH}	TCK [BSCAN] hold time	10	—	ns
t_{BTRF}	TCK [BSCAN] rise/fall time	50	—	mV/ns
t_{BTCO}	TAP controller falling edge of clock to valid output	—	10	ns
$t_{BTCODIS}$	TAP controller falling edge of clock to valid disable	—	10	ns
t_{BTCOEN}	TAP controller falling edge of clock to valid enable	—	10	ns
t_{BTCRS}	BSCAN test capture register setup time	8	—	ns
t_{BTCRH}	BSCAN test capture register hold time	25	—	ns
t_{BUTCO}	BSCAN test update register, falling edge of clock to valid output	—	25	ns
$t_{BTUODIS}$	BSCAN test update register, falling edge of clock to valid disable	—	25	ns
$t_{BTUPOEN}$	BSCAN test update register, falling edge of clock to valid enable	—	25	ns

Signal Descriptions

Signal Name	I/O	Description
General Purpose		
P[Edge] [Row/Column Number*]_[A/B]	I/O	[Edge] indicates the edge of the device on which the pad is located. Valid edge designations are L (Left), B (Bottom), R (Right), T (Top). [Row/Column Number] indicates the PFU row or the column of the device on which the PIC exists. When Edge is T (Top) or B (Bottom), only need to specify Row Number. When Edge is L (Left) or R (Right), only need to specify Column Number. [A/B] indicates the PIO within the PIC to which the pad is connected. Some of these user-programmable pins are shared with special function pins. These pins, when not used as special purpose pins, can be programmed as I/Os for user logic. During configuration the user-programmable I/Os are tri-stated with an internal pull-up resistor enabled. If any pin is not used (or not bonded to a package pin), it is also tri-stated with an internal pull-up resistor enabled after configuration.
GSRN	I	Global RESET signal (active low). Any I/O pin can be GSRN.
NC	—	No connect.
GND	—	Ground. Dedicated pins.
V _{CC}	—	Power supply pins for core logic. Dedicated pins.
V _{CCAUX}	—	Auxiliary power supply pin. This dedicated pin powers all the differential and referenced input buffers.
V _{CCPLL}	—	PLL supply pins. csBGA, PQFP and TQFP packages only.
V _{CCIOx}	—	Dedicated power supply pins for I/O bank x.
V _{REF1_x} , V _{REF2_x}	—	Reference supply pins for I/O bank x. Pre-determined pins in each bank are assigned as V _{REF} inputs. When not used, they may be used as I/O pins.
PLL and Clock Functions (Used as user programmable I/O pins when not in use for PLL or clock pins)		
[LOC][num]_V _{CCPLL}	—	Power supply pin for PLL: LLC, LRC, URC, ULC, num = row from center.
[LOC][num]_GPLL[T, C]_IN_A	I	General Purpose PLL (GPLL) input pads: LLC, LRC, URC, ULC, num = row from center, T = true and C = complement, index A,B,C...at each side.
[LOC][num]_GPLL[T, C]_FB_A	I	Optional feedback GPLL input pads: LLC, LRC, URC, ULC, num = row from center, T = true and C = complement, index A,B,C...at each side.
PCLK[T, C]_[n:0]_[3:0]	I	Primary Clock pads, T = true and C = complement, n per side, indexed by bank and 0,1,2,3 within bank.
[LOC]DQS[num]	I	DQS input pads: T (Top), R (Right), B (Bottom), L (Left), DQS, num = ball function number. Any pad can be configured to be output.
Test and Programming (Dedicated Pins)		
TMS	I	Test Mode Select input, used to control the 1149.1 state machine. Pull-up is enabled during configuration.
TCK	I	Test Clock input pin, used to clock the 1149.1 state machine. No pull-up enabled.
TDI	I	Test Data in pin. Used to load data into device using 1149.1 state machine. After power-up, this TAP port can be activated for configuration by sending appropriate command. (Note: once a configuration port is selected it is locked. Another configuration port cannot be selected until the power-up sequence). Pull-up is enabled during configuration.

Pin Information Summary

Pin Type		XP2-5				XP2-8				XP2-17			XP2-30			XP2-40	
		132 csBGA	144 TQFP	208 PQFP	256 ftBGA	132 csBGA	144 TQFP	208 PQFP	256 ftBGA	208 PQFP	256 ftBGA	484 fpBGA	256 ftBGA	484 fpBGA	672 fpBGA	484 fpBGA	672 fpBGA
Single Ended User I/O		86	100	146	172	86	100	146	201	146	201	358	201	363	472	363	540
Differential Pair User I/O	Normal	35	39	57	66	35	39	57	77	57	77	135	77	137	180	137	204
	Highspeed	8	11	16	20	8	11	16	23	16	23	44	23	44	56	44	66
Configuration	TAP	5	5	5	5	5	5	5	5	5	5	5	5	5	5	5	5
	Muxed	9	9	9	9	9	9	9	9	9	9	9	9	9	9	9	9
	Dedicated	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
Non Configuration	Muxed	5	5	7	7	7	7	9	9	11	11	21	7	11	13	11	13
	Dedicated	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
Vcc		6	4	9	6	6	4	9	6	9	6	16	6	16	20	16	20
Vccaux		4	4	4	4	4	4	4	4	4	4	8	4	8	8	8	8
VCCPLL		2	2	2	-	2	2	2	-	4	-	-	-	-	-	-	-
VCCIO	Bank0	2	2	2	2	2	2	2	2	2	2	4	2	4	4	4	4
	Bank1	1	1	2	2	1	1	2	2	2	2	4	2	4	4	4	4
	Bank2	2	2	2	2	2	2	2	2	2	2	4	2	4	4	4	4
	Bank3	1	1	2	2	1	1	2	2	2	2	4	2	4	4	4	4
	Bank4	1	1	2	2	1	1	2	2	2	2	4	2	4	4	4	4
	Bank5	2	2	2	2	2	2	2	2	2	2	4	2	4	4	4	4
	Bank6	1	1	2	2	1	1	2	2	2	2	4	2	4	4	4	4
	Bank7	2	2	2	2	2	2	2	2	2	2	4	2	4	4	4	4
GND, GND0-GND7		15	15	20	20	15	15	22	20	22	20	56	20	56	64	56	64
NC		-	-	4	31	-	-	2	2	-	2	7	2	2	69	2	1
Single Ended/ Differential I/O per Bank	Bank0	18/9	20/10	20/10	26/13	18/9	20/10	20/10	28/14	20/10	28/14	52/26	28/14	52/26	70/35	52/26	70/35
	Bank1	4/2	6/3	18/9	18/9	4/2	6/3	18/9	22/11	18/9	22/11	36/18	22/11	36/18	54/27	36/18	70/35
	Bank2	16/8	18/9	18/9	22/11	16/8	18/9	18/9	26/13	18/9	26/13	46/23	26/13	46/23	56/28	46/23	64/32
	Bank3	4/2	4/2	16/8	20/10	4/2	4/2	16/8	24/12	16/8	24/12	44/22	24/12	46/23	56/28	46/23	66/33
	Bank4	8/4	8/4	18/9	18/9	8/4	8/4	18/9	26/13	18/9	26/13	36/18	26/13	38/19	54/27	38/19	70/35
	Bank5	14/7	18/9	20/10	24/12	14/7	18/9	20/10	24/12	20/10	24/12	52/26	24/12	53/26	70/35	53/26	70/35
	Bank6	6/3	8/4	18/9	22/11	6/3	8/4	18/9	27/13	18/9	27/13	46/23	27/13	46/23	56/28	46/23	66/33
	Bank7	16/8	18/9	18/9	22/11	16/8	18/9	18/9	24/12	18/9	24/12	46/23	24/12	46/23	56/28	46/23	64/32
True LVDS Pairs Bonding Out per Bank	Bank0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	Bank1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	Bank2	3	4	4	5	3	4	4	6	4	6	11	6	11	14	11	16
	Bank3	1	1	4	5	1	1	4	6	4	6	11	6	11	14	11	17
	Bank4	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	Bank5	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	Bank6	1	2	4	5	1	2	4	6	4	6	11	6	11	14	11	17
	Bank7	3	4	4	5	3	4	4	5	4	5	11	5	11	14	11	16
DDR Banks Bonding Out per I/O Bank ¹	Bank0	1	1	1	1	1	1	1	1	1	1	3	1	2	4	2	4
	Bank1	0	0	1	1	0	0	1	1	1	1	2	1	2	3	2	4
	Bank2	1	1	1	1	1	1	1	1	1	1	2	1	3	3	3	4
	Bank3	0	0	1	1	0	0	1	1	1	1	2	1	3	3	3	4
	Bank4	0	0	1	1	0	0	1	1	1	1	2	1	2	3	2	4
	Bank5	1	1	1	1	1	1	1	1	1	1	3	1	2	4	2	4
	Bank6	0	0	1	1	0	0	1	1	1	1	2	1	3	3	3	4
	Bank7	1	1	1	1	1	1	1	1	1	1	2	1	3	3	3	4