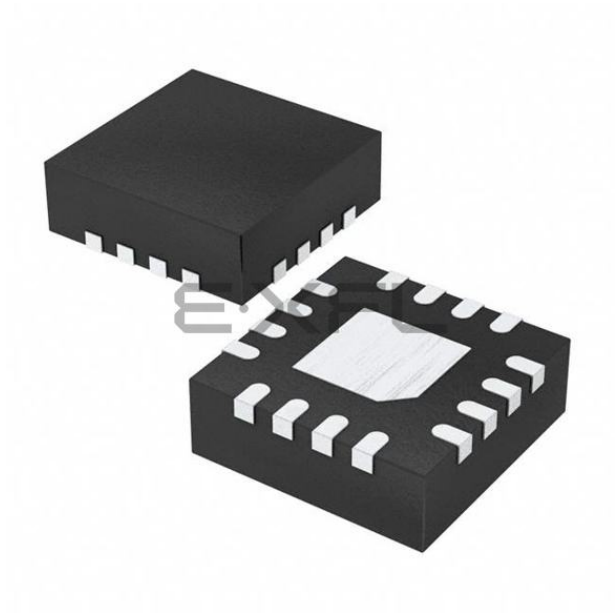

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	11
Program Memory Size	7KB (4K x 14)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	256 x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 11x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	16-VQFN Exposed Pad
Supplier Device Package	16-QFN (4x4)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf1554t-i-ml

TABLE 3-6: PIC16LF1554/1559 MEMORY MAP, BANKS 16-23

BANK 16		BANK 17		BANK 18		BANK 19		BANK 20		BANK 21		BANK 22		BANK 23	
800h	INDF0	880h	INDF0	900h	INDF0	980h	INDF0	A00h	INDF0	A80h	INDF0	B00h	INDF0	B80h	INDF0
801h	INDF1	881h	INDF1	901h	INDF1	981h	INDF1	A01h	INDF1	A81h	INDF1	B01h	INDF1	B81h	INDF1
802h	PCL	882h	PCL	902h	PCL	982h	PCL	A02h	PCL	A82h	PCL	B02h	PCL	B82h	PCL
803h	STATUS	883h	STATUS	903h	STATUS	983h	STATUS	A03h	STATUS	A83h	STATUS	B03h	STATUS	B83h	STATUS
804h	FSR0L	884h	FSR0L	904h	FSR0L	984h	FSR0L	A04h	FSR0L	A84h	FSR0L	B04h	FSR0L	B84h	FSR0L
805h	FSR0H	885h	FSR0H	905h	FSR0H	985h	FSR0H	A05h	FSR0H	A85h	FSR0H	B05h	FSR0H	B85h	FSR0H
806h	FSR1L	886h	FSR1L	906h	FSR1L	986h	FSR1L	A06h	FSR1L	A86h	FSR1L	B06h	FSR1L	B86h	FSR1L
807h	FSR1H	887h	FSR1H	907h	FSR1H	987h	FSR1H	A07h	FSR1H	A87h	FSR1H	B07h	FSR1H	B87h	FSR1H
808h	BSR	888h	BSR	908h	BSR	988h	BSR	A08h	BSR	A88h	BSR	B08h	BSR	B88h	BSR
809h	WREG	889h	WREG	909h	WREG	989h	WREG	A09h	WREG	A89h	WREG	B09h	WREG	B89h	WREG
80Ah	PCLATH	88Ah	PCLATH	90Ah	PCLATH	98Ah	PCLATH	A0Ah	PCLATH	A8Ah	PCLATH	B0Ah	PCLATH	B8Ah	PCLATH
80Bh	INTCON	88Bh	INTCON	90Bh	INTCON	98Bh	INTCON	A0Bh	INTCON	A8Bh	INTCON	B0Bh	INTCON	B8Bh	INTCON
80Ch	—	88Ch	—	90Ch	—	98Ch	—	A0Ch	—	A8Ch	—	B0Ch	—	B8Ch	—
80Dh	—	88Dh	—	90Dh	—	98Dh	—	A0Dh	—	A8Dh	—	B0Dh	—	B8Dh	—
80Eh	—	88Eh	—	90Eh	—	98Eh	—	A0Eh	—	A8Eh	—	B0Eh	—	B8Eh	—
80Fh	—	88Fh	—	90Fh	—	98Fh	—	A0Fh	—	A8Fh	—	B0Fh	—	B8Fh	—
810h	—	890h	—	910h	—	990h	—	A10h	—	A90h	—	B10h	—	B90h	—
811h	—	891h	—	911h	—	991h	—	A11h	—	A91h	—	B11h	—	B91h	—
812h	—	892h	—	912h	—	992h	—	A12h	—	A92h	—	B12h	—	B92h	—
813h	—	893h	—	913h	—	993h	—	A13h	—	A93h	—	B13h	—	B93h	—
814h	—	894h	—	914h	—	994h	—	A14h	—	A94h	—	B14h	—	B94h	—
815h	—	895h	—	915h	—	995h	—	A15h	—	A95h	—	B15h	—	B95h	—
816h	—	896h	—	916h	—	996h	—	A16h	—	A96h	—	B16h	—	B96h	—
817h	—	897h	—	917h	—	997h	—	A17h	—	A97h	—	B17h	—	B97h	—
818h	—	898h	—	918h	—	998h	—	A18h	—	A98h	—	B18h	—	B98h	—
819h	—	899h	—	919h	—	999h	—	A19h	—	A99h	—	B19h	—	B99h	—
81Ah	—	89Ah	—	91Ah	—	99Ah	—	A1Ah	—	A9Ah	—	B1Ah	—	B9Ah	—
81Bh	—	89Bh	—	91Bh	—	99Bh	—	A1Bh	—	A9Bh	—	B1Bh	—	B9Bh	—
81Ch	—	89Ch	—	91Ch	—	99Ch	—	A1Ch	—	A9Ch	—	B1Ch	—	B9Ch	—
81Dh	—	89Dh	—	91Dh	—	99Dh	—	A1Dh	—	A9Dh	—	B1Dh	—	B9Dh	—
81Eh	—	89Eh	—	91Eh	—	99Eh	—	A1Eh	—	A9Eh	—	B1Eh	—	B9Eh	—
81Fh	—	89Fh	—	91Fh	—	99Fh	—	A1Fh	—	A9Fh	—	B1Fh	—	B9Fh	—
820h Unimplemented Read as '0' 86Fh		8A0h Unimplemented Read as '0' 8EFh		920h Unimplemented Read as '0' 96Fh		9A0h Unimplemented Read as '0' 9EFh		A20h Unimplemented Read as '0' A6Fh		AA0h Unimplemented Read as '0' AEFh		B20h Unimplemented Read as '0' B6Fh		BA0h Unimplemented Read as '0' BEFh	
870h Accesses 70h – 7Fh 87Fh		8F0h Accesses 70h – 7Fh 8FFh		970h Accesses 70h – 7Fh 97Fh		9F0h Accesses 70h – 7Fh 9FFh		A70h Accesses 70h – 7Fh A7Fh		AF0h Accesses 70h – 7Fh AFFh		B70h Accesses 70h – 7Fh B7Fh		BF0h Accesses 70h – 7Fh BFFh	

TABLE 3-9: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets
Bank 6											
300h	INDF0(1)	Addressing this location uses contents of FSR0H/FSR0L to address data memory (not a physical register)								xxxx xxxx	uuuu uuuu
301h	INDF1(1)	Addressing this location uses contents of FSR1H/FSR1L to address data memory (not a physical register)								xxxx xxxx	uuuu uuuu
302h	PCL(1)	Program Counter (PC) Least Significant Byte								0000 0000	0000 0000
303h	STATUS(1)	—	—	—	$\overline{TO}$	$\overline{PD}$	Z	DC	C	---1 1000	---q quuu
304h	FSR0L(1)	Indirect Data Memory Address 0 Low Pointer								0000 0000	uuuu uuuu
305h	FSR0H(1)	Indirect Data Memory Address 0 High Pointer								0000 0000	0000 0000
306h	FSR1L(1)	Indirect Data Memory Address 1 Low Pointer								0000 0000	uuuu uuuu
307h	FSR1H(1)	Indirect Data Memory Address 1 High Pointer								0000 0000	0000 0000
308h	BSR(1)	—	—	—	BSR<4:0>					---0 0000	---0 0000
309h	WREG(1)	Working Register								0000 0000	uuuu uuuu
30Ah	PCLATH(1)	—	Write Buffer for the upper 7 bits of the Program Counter							-000 0000	-000 0000
30Bh	INTCON(1)	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	0000 0000	0000 0000
30Ch	—	Unimplemented								—	—
30Dh	—	Unimplemented								—	—
30Eh	—	Unimplemented								—	—
30Fh	—	Unimplemented								—	—
310h	—	Unimplemented								—	—
311h	—	Unimplemented								—	—
312h	—	Unimplemented								—	—
313h	—	Unimplemented								—	—
314h	—	Unimplemented								—	—
315h	—	Unimplemented								—	—
316h	—	Unimplemented								—	—
317h	—	Unimplemented								—	—
318h	—	Unimplemented								—	—
319h	—	Unimplemented								—	—
31Ah	—	Unimplemented								—	—
31Bh	—	Unimplemented								—	—
31Ch	—	Unimplemented								—	—
31Dh	—	Unimplemented								—	—
31Eh	—	Unimplemented								—	—
31Fh	—	Unimplemented								—	—

Legend: x = unknown, u = unchanged, q = depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations unimplemented, read as '0'.

Note 1: These registers can be accessed from any bank.

2: PIC16LF1554.

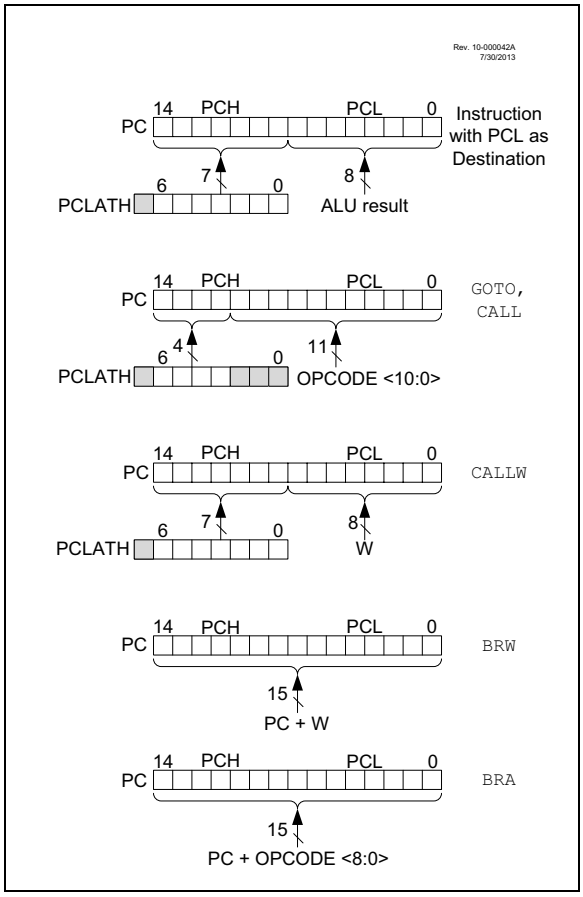
3: PIC16LF1559.

4: These registers/bits are available at two address locations, in Bank 1 and Bank 14.

3.3 PCL and PCLATH

The Program Counter (PC) is 15 bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<14:8>) is not directly readable or writable and comes from PCLATH. On any Reset, the PC is cleared. Figure 3-4 shows the five situations for the loading of the PC.

FIGURE 3-4: LOADING OF PC IN DIFFERENT SITUATIONS



3.3.1 MODIFYING PCL

Executing any instruction with the PCL register as the destination simultaneously causes the Program Counter PC<14:8> bits (PCH) to be replaced by the contents of the PCLATH register. This allows the entire contents of the program counter to be changed by writing the desired upper seven bits to the PCLATH register. When the lower eight bits are written to the PCL register, all 15 bits of the program counter will change to the values contained in the PCLATH register and those being written to the PCL register.

3.3.2 COMPUTED GOTO

A computed GOTO is accomplished by adding an offset to the program counter (ADDWF PCL). When performing a table read using a computed GOTO method, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block). Refer to Application Note AN556, *"Implementing a Table Read"* (DS00556).

3.3.3 COMPUTED FUNCTION CALLS

A computed function CALL allows programs to maintain tables of functions and provide another way to execute state machines or look-up tables. When performing a table read using a computed function CALL, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block).

If using the CALL instruction, the PCH<2:0> and PCL registers are loaded with the operand of the CALL instruction. PCH<6:3> is loaded with PCLATH<6:3>.

The CALLW instruction enables computed calls by combining PCLATH and W to form the destination address. A computed CALLW is accomplished by loading the W register with the desired address and executing CALLW. The PCL register is loaded with the value of W and PCH is loaded with PCLATH.

3.3.4 BRANCHING

The branching instructions add an offset to the PC. This allows relocatable code and code that crosses page boundaries. There are two forms of branching, BRW and BRA. The PC will have incremented to fetch the next instruction in both cases. When using either branching instruction, a PCL memory boundary may be crossed.

If using BRW, load the W register with the desired unsigned address and execute BRW. The entire PC will be loaded with the address PC + 1 + W.

If using BRA, the entire PC will be loaded with PC + 1 + the signed value of the operand of the BRA instruction.

10.2.2 FLASH MEMORY UNLOCK SEQUENCE

The unlock sequence is a mechanism that protects the Flash program memory from unintended self-write programming or erasing. The sequence must be executed and completed without interruption to successfully complete any of the following operations:

- Row Erase
- Load program memory write latches
- Write of program memory write latches to program memory
- Write of program memory write latches to User IDs

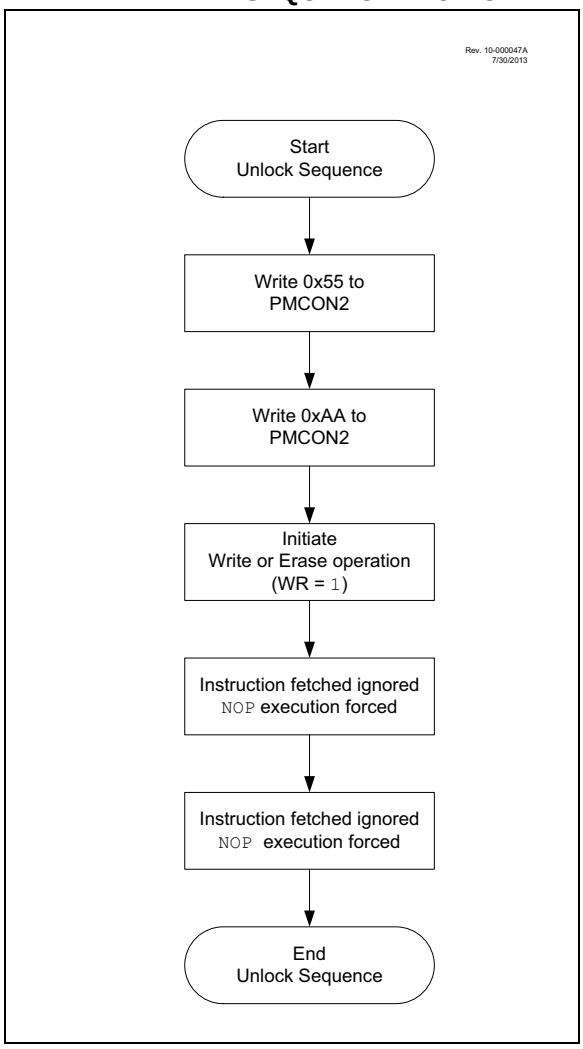
The unlock sequence consists of the following steps:

1. Write 55h to PMCON2
2. Write AAh to PMCON2
3. Set the WR bit in PMCON1
4. NOP instruction
5. NOP instruction

Once the WR bit is set, the processor will always force two NOP instructions. When an Erase Row or Program Row operation is being performed, the processor will stall internal operations (typical 2 ms), until the operation is complete and then resume with the next instruction. When the operation is loading the program memory write latches, the processor will always force the two NOP instructions and continue uninterrupted with the next instruction.

Since the unlock sequence must not be interrupted, global interrupts should be disabled prior to the unlock sequence and re-enabled after the unlock sequence is completed.

FIGURE 10-3: FLASH PROGRAM MEMORY UNLOCK SEQUENCE FLOWCHART

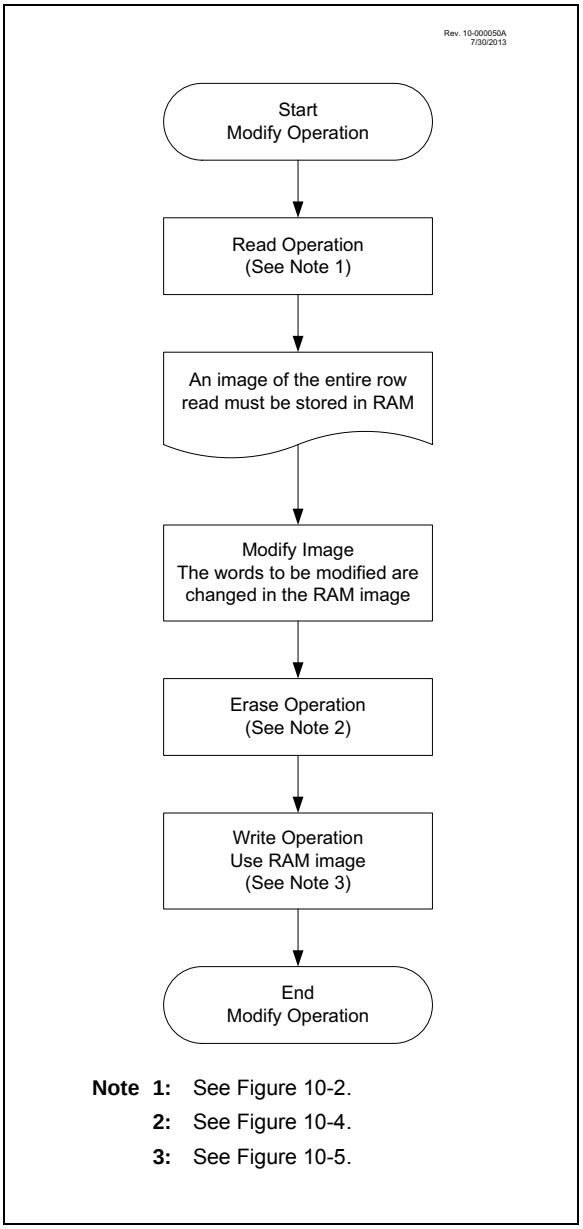


10.3 Modifying Flash Program Memory

When modifying existing data in a program memory row, and data within that row must be preserved, it must first be read and saved in a RAM image. Program memory is modified using the following steps:

1. Load the starting address of the row to be modified.
2. Read the existing data from the row into a RAM image.
3. Modify the RAM image to contain the new data to be written into program memory.
4. Load the starting address of the row to be rewritten.
5. Erase the program memory row.
6. Load the write latches with data from the RAM image.
7. Initiate a programming operation.

FIGURE 10-7: FLASH PROGRAM MEMORY MODIFY FLOWCHART



Note 1: See Figure 10-2.
2: See Figure 10-4.
3: See Figure 10-5.

11.5 PORTB Registers (PIC16LF1559 Only)

11.5.1 DATA REGISTER

PORTB is a 4-bit wide, bidirectional port. The corresponding data direction register is TRISB (Register 11-8). Setting a TRISB bit (= 1) will make the corresponding PORTB pin an input (i.e., disable the output driver). Clearing a TRISB bit (= 0) will make the corresponding PORTB pin an output (i.e., enables output driver and puts the contents of the output latch on the selected pin). Example 11-1 shows how to initialize an I/O port.

Reading the PORTB register (Register 11-7) reads the status of the pins, whereas writing to it will write to the PORT latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, this value is modified and then written to the PORT data latch (LATB).

11.5.2 DIRECTION CONTROL

The TRISB register (Register 11-8) controls the PORTB pin output drivers, even when they are being used as analog inputs. The user should ensure the bits in the TRISB register are maintained set when using them as analog inputs. I/O pins configured as analog input always read '0'.

11.5.3 ANALOG CONTROL

The ANSELB register (Register 11-10) is used to configure the Input mode of an I/O pin to analog. Setting the appropriate ANSELB bit high will cause all digital reads on the pin to be read as '0' and allow analog functions on the pin to operate correctly.

The state of the ANSELB bits has no effect on digital output functions. A pin with TRIS clear and ANSEL set will still operate as a digital output, but the Input mode will be analog. This can cause unexpected behavior when executing read-modify-write instructions on the affected port.

Note:	The ANSELB bits default to the Analog mode after Reset. To use any pins as digital general purpose or peripheral inputs, the corresponding ANSELx bits must be initialized to '0' by user software.

11.5.4 PORTB FUNCTIONS AND OUTPUT PRIORITIES

Each PORTB pin is multiplexed with other functions. The pins, their combined functions and their output priorities are shown in Table 11-5.

When multiple outputs are enabled, the actual pin control goes to the peripheral with the highest priority.

Analog input functions, such as ADC and comparator inputs, are not shown in the priority lists. These inputs are active when the I/O pin is set for Analog mode using the ANSELx registers. Digital output functions may control the pin when it is in Analog mode with the priority shown below in Table 11-5.

TABLE 11-5: PORTB OUTPUT PRIORITY

Pin Name	Function Priority(1)
RB4	SDA RB4
RB5	RB5
RB6	SCL SCK RB6
RB7	TX RB7

Note 1: Priority listed from highest to lowest.

15.4 ADC Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The Analog Input model is shown in Figure 15-4. The source impedance (RS) and the internal sampling switch (RSS) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (RSS) impedance varies over the device voltage (VDD), refer to Figure 15-4. **The maximum recommended impedance for analog sources is 10 kΩ.**

As the source impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed), an ADC acquisition must be done before the conversion can be started. To calculate the minimum acquisition time, Equation 15-1 may be used. This equation assumes that 1/2 LSb error is used (1,024 steps for the ADC). The 1/2 LSb error is the maximum error allowed for the ADC to meet its specified resolution.

EQUATION 15-1: ACQUISITION TIME EXAMPLE

Assumptions: Temperature = 50°C and external impedance of 10 kΩ 3.3V VDD

$$T_{ACQ} = \text{Amplifier Settling Time} + \text{Hold Capacitor Charging Time} + \text{Temperature Coefficient}$$
$$= T_{AMP} + T_C + T_{COFF}$$
$$= 2\ \mu s + T_C + [(\text{Temperature} - 25°C)(0.05\ \mu s/°C)]$$

The value for TC can be approximated with the following equations:

$$V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) = V_{CHOLD} \quad ;[1]\ V_{CHOLD}\ \text{charged to within 1/2 lsb}$$

$$V_{APPLIED}\left(1 - e^{\frac{-T_C}{RC}}\right) = V_{CHOLD} \quad ;[2]\ V_{CHOLD}\ \text{charge response to}\ V_{APPLIED}$$

$$V_{APPLIED}\left(1 - e^{\frac{-T_C}{RC}}\right) = V_{APPLIED}\left(1 - \frac{1}{(2^{n+1}) - 1}\right) \quad ;\text{combining [1] and [2]}$$

Note: Where n = number of bits of the ADC.

Solving for TC:

$$T_C = -C_{HOLD}(R_{IC} + R_{SS} + R_S)\ ln(1/2047)$$
$$= -15\ pF(1k\Omega + 7k\Omega + 10k\Omega)\ ln(0.0004885)$$
$$= 2.06\ \mu s$$

Therefore: $T_{ACQ} = 2\mu s + 2.06\mu s + [(50°C - 25°C)(0.05\mu s/°C)]$
$= 5.31\ \mu s$

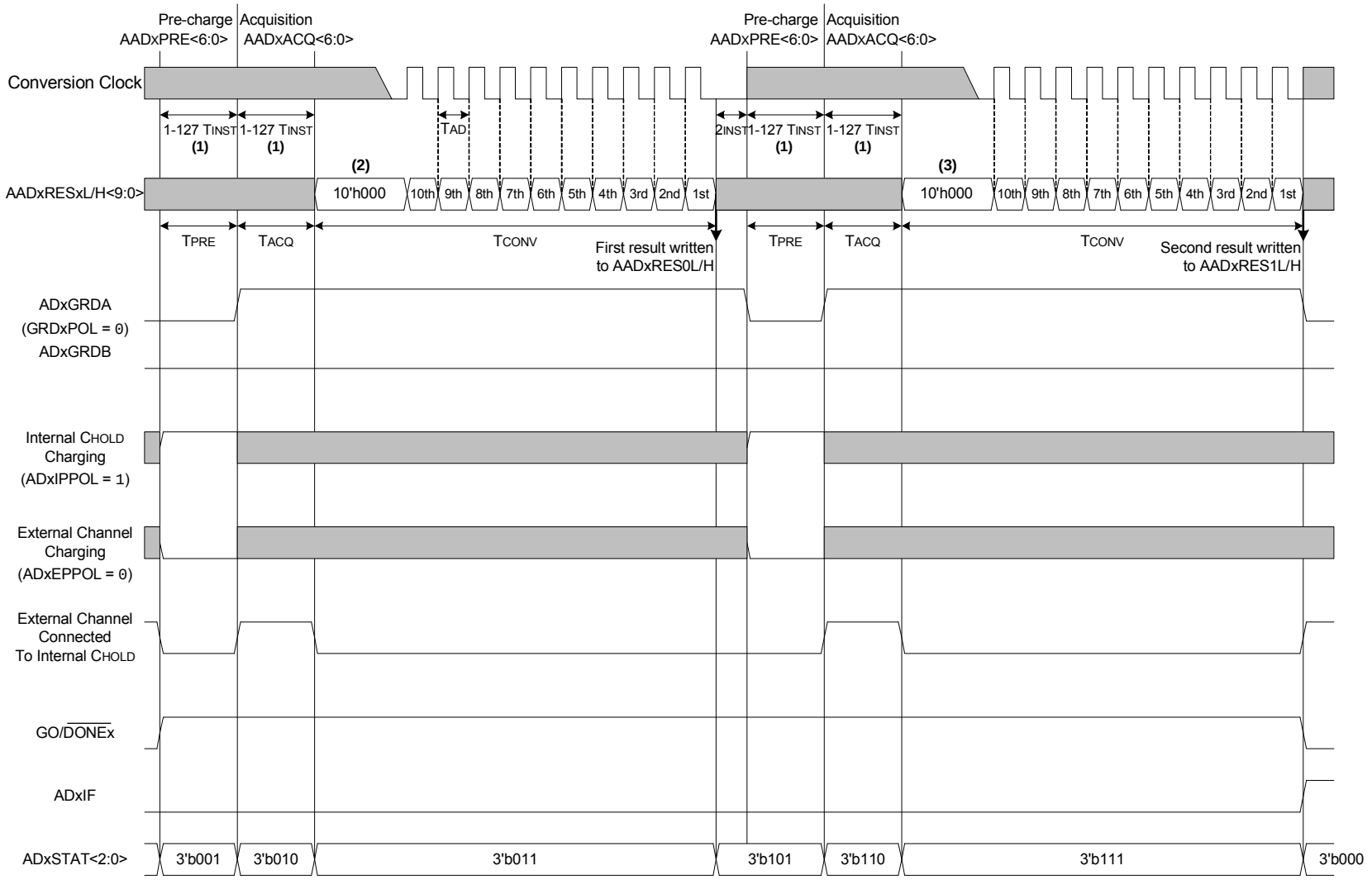
Note 1: The reference voltage (VRPOS) has no effect on the equation, since it cancels itself out.

2: The charge holding capacitor (CHOLD) is not discharged after each conversion.

3: The maximum recommended impedance for analog sources is 10 kΩ. This is required to meet the pin leakage specification.

4: The calculation above assumed CHOLD = 15pF. This value can be larger than 15pF by setting the AADxCAP register.

FIGURE 16-6: DOUBLE SAMPLE CONVERSION SEQUENCE (ADDSEN = 1 AND ADIPEN = 0)



Note 1: When the conversion clock is ADCRC, the pre-charge and acquisition timers are clocked by ADCRC.
2: The AADxRES0L/H registers are set to zero during this period.
3: The AADxRES1L/H registers are set to zero during this period.

REGISTER 16-5: AADCON1(1)/ADCOMCON(2): HARDWARE CVD CONTROL REGISTER 1(1,2)

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	U-0	U-0	R/W-0/0	R/W-0/0
ADFM	ADCS<2:0>			—	GO/$\overline{\text{DONE}}$_ALL	ADPREF<1:0>	
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7 **ADFM:** ADC Result Format Select bit
- 1 = Right justified. Six Most Significant bits of AADxRESxH are set to '0' when the conversion result is loaded.
- 0 = Left justified. Six Least Significant bits of AADxRESxL are set to '0' when the conversion result is loaded.

bit 6-4 **ADCS<2:0>:** ADC Conversion Clock Select bits
- 111 = FRC (clock supplied from a dedicated RC oscillator)
- 110 = FOSC/64
- 101 = FOSC/16
- 100 = FOSC/4
- 011 = FRC (clock supplied from a dedicated RC oscillator)
- 010 = FOSC/32
- 001 = FOSC/8
- 000 = FOSC/2

bit 3 **Unimplemented**: Read as '0'

bit 2 **GO/$\overline{\text{DONE}}$_ALL(3):** Synchronized ADC Conversion Status bit
- 1 = Synchronized ADC conversion in progress. Setting this bit starts conversion in any ADC with ADxON = 1.
- 0 = Synchronized ADC conversion completed/ not in progress.

bit 1-0 **ADPREF<1:0>:** ADC Positive Voltage Reference Configuration bits
- 00 = VREFH is connected to VDD
- 01 = Reserved
- 10 = VREFH is connected to external VREF+ pin(4)
- 11 = VREFH is connected to internal Fixed Voltage Reference

Note 1: Bank 1 name is ADCON1.

2: Bank 14 name is AADCON1/ADCOMCON.

3: Setting this bit triggers the GO/$\overline{\text{DONE}}$x bits in both ADCs. Each ADC will run a conversion according to its control register settings. This bit reads as an OR of the individual GO/$\overline{\text{DONE}}$x bits.

4: When selecting the VREF+ pin as the source of the positive reference, be aware that a minimum voltage specification exists. See **Section 25.0 "Electrical Specifications"** for details.

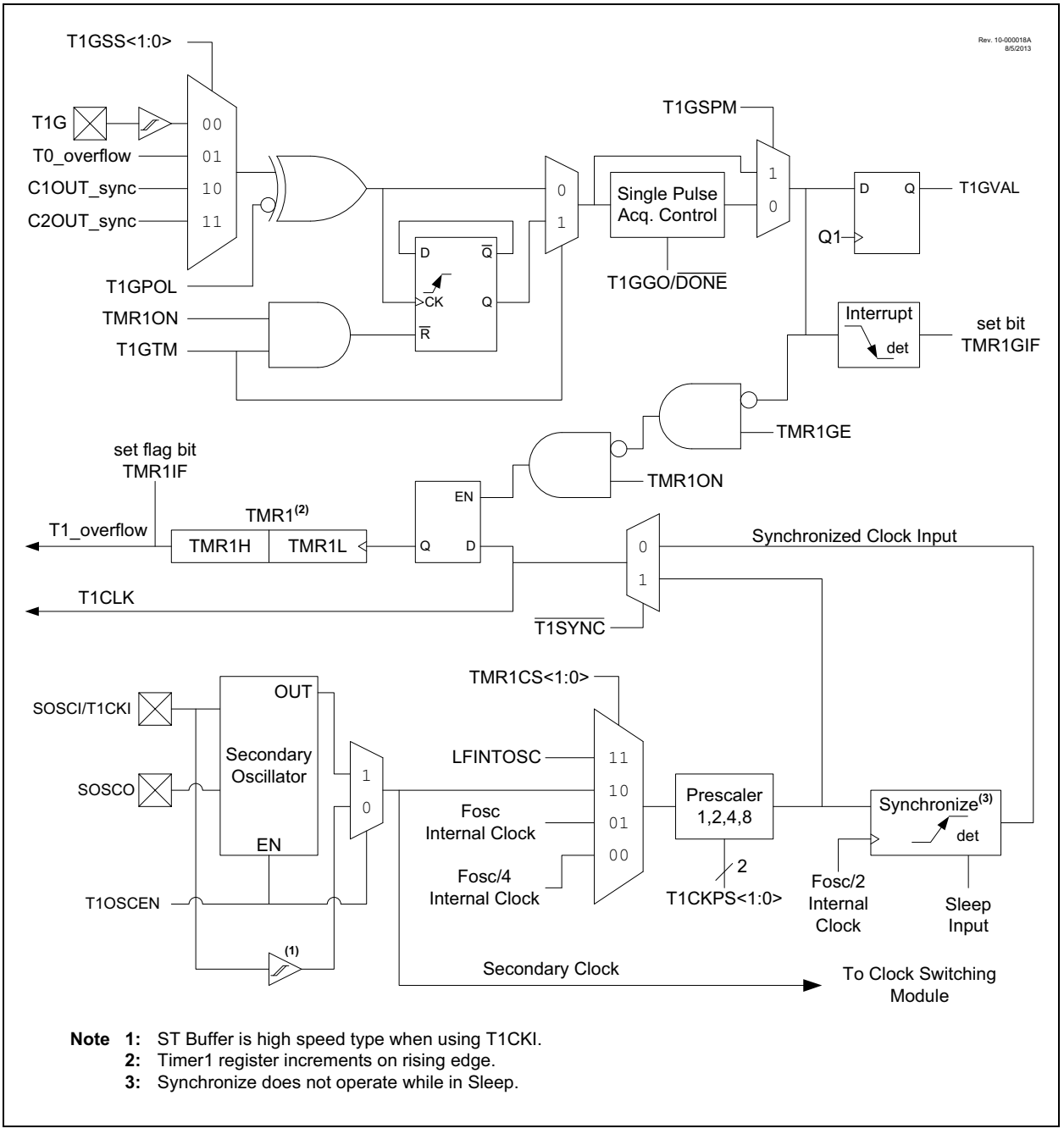
18.0 TIMER1 MODULE WITH GATE CONTROL

The Timer1 module is a 16-bit timer/counter with the following features:

- 16-bit timer/counter register pair (TMR1H:TMR1L)
- Programmable internal or external clock source
- 2-bit prescaler
- Optionally synchronized comparator out
- Multiple Timer1 gate (count enable) sources
- Interrupt on overflow
- Wake-up on overflow (external clock, Asynchronous mode only)
- ADC Auto-Conversion Trigger(s)
- Selectable Gate Source Polarity
- Gate Toggle mode
- Gate Single-Pulse mode
- Gate Value Status
- Gate Event Interrupt

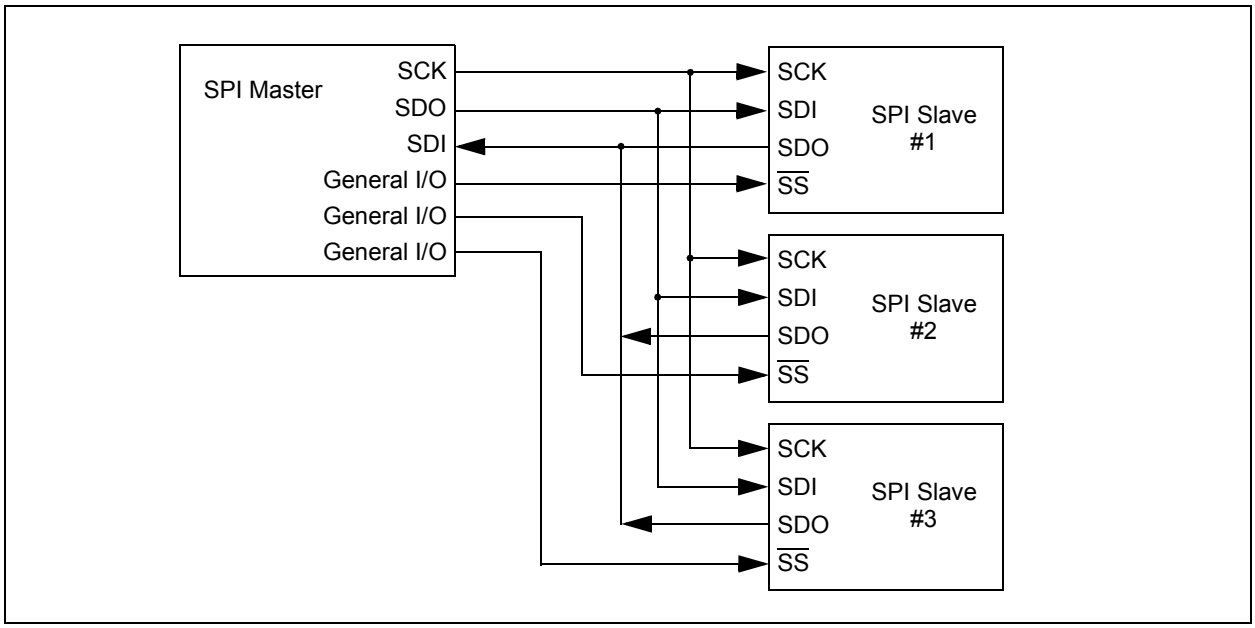
Figure 18-1 is a block diagram of the Timer1 module.

FIGURE 18-1: TIMER1 BLOCK DIAGRAM



Note 1: ST Buffer is high speed type when using T1CKI.
2: Timer1 register increments on rising edge.
3: Synchronize does not operate while in Sleep.

FIGURE 20-4: SPI MASTER AND MULTIPLE SLAVE CONNECTION



20.2.1 SPI MODE REGISTERS

The MSSP module has five registers for SPI mode operation. These are:

- MSSP STATUS register (SSPSTAT)
- MSSP Control Register 1 (SSPCON1)
- MSSP Control Register 3 (SSPCON3)
- MSSP Data Buffer register (SSPBUF)
- MSSP Address register (SSPADD)
- MSSP Shift register (SSPSR) (Not directly accessible)

SSPCON1 and SSPSTAT are the control and STATUS registers in SPI mode operation. The SSPCON1 register is readable and writable. The lower six bits of the SSPSTAT are read-only. The upper two bits of the SSPSTAT are read/write.

In SPI master mode, SSPADD can be loaded with a value used in the Baud Rate Generator. More information on the Baud Rate Generator is available in **Section 20.7 “Baud Rate Generator”**

SSPSR is the shift register used for shifting data in and out. SSPBUF provides indirect access to the SSPSR register. SSPBUF is the buffer register to which data bytes are written, and from which data bytes are read.

In receive operations, SSPSR and SSPBUF together create a buffered receiver. When SSPSR receives a complete byte, it is transferred to SSPBUF and the SSP1IF interrupt is set.

During transmission, the SSPBUF is not buffered. A write to SSPBUF will write to both SSPBUF and SSPSR.

20.3 I²C MODE OVERVIEW

The Inter-Integrated Circuit Bus (I²C) is a multi-master serial data communication bus. Devices communicate in a master/slave environment where the master devices initiate the communication. A Slave device is controlled through addressing.

The I²C bus specifies two signal connections:

- Serial Clock (SCL)
- Serial Data (SDA)

Figure 20-2 and Figure 20-3 show the block diagrams of the MSSP module when operating in I²C mode.

Both the SCL and SDA connections are bidirectional open-drain lines, each requiring pull-up resistors for the supply voltage. Pulling the line to ground is considered a logical zero and letting the line float is considered a logical one.

Figure 20-11 shows a typical connection between two processors configured as master and slave devices.

The I²C bus can operate with one or more master devices and one or more slave devices.

There are four potential modes of operation for a given device:

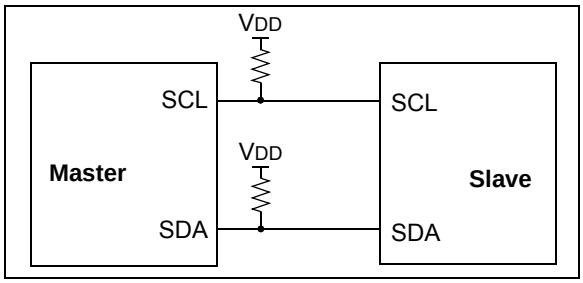
- Master Transmit mode
  (master is transmitting data to a slave)
- Master Receive mode
  (master is receiving data from a slave)
- Slave Transmit mode
  (slave is transmitting data to a master)
- Slave Receive mode
  (slave is receiving data from the master)

To begin communication, a master device starts out in Master Transmit mode. The master device sends out a Start bit followed by the address byte of the slave it intends to communicate with. This is followed by a single Read/Write bit, which determines whether the master intends to transmit to or receive data from the slave device.

If the requested slave exists on the bus, it will respond with an Acknowledge bit, otherwise known as an $\overline{ACK}$. The master then continues in either Transmit mode or Receive mode and the slave continues in the complement, either in Receive mode or Transmit mode, respectively.

A Start bit is indicated by a high-to-low transition of the SDA line while the SCL line is held high. Address and data bytes are sent out, Most Significant bit (MSb) first. The Read/Write bit is sent out as a logical one when the master intends to read data from the slave, and is sent out as a logical zero when it intends to write data to the slave.

FIGURE 20-11: I²C MASTER/ SLAVE CONNECTION



The Acknowledge bit ($\overline{ACK}$) is an active-low signal, which holds the SDA line low to indicate to the transmitter that the slave device has received the transmitted data and is ready to receive more.

The transition of a data bit is always performed while the SCL line is held low. Transitions that occur while the SCL line is held high are used to indicate Start and Stop bits.

If the master intends to write to the slave, then it repeatedly sends out a byte of data, with the slave responding after each byte with an $\overline{ACK}$ bit. In this example, the master device is in Master Transmit mode and the slave is in Slave Receive mode.

If the master intends to read from the slave, then it repeatedly receives a byte of data from the slave, and responds after each byte with an $\overline{ACK}$ bit. In this example, the master device is in Master Receive mode and the slave is Slave Transmit mode.

On the last byte of data communicated, the master device may end the transmission by sending a Stop bit. If the master device is in Receive mode, it first sends a not $\overline{ACK}$ bit in place of an $\overline{ACK}$ and then terminates the transfer with a Stop bit.

In some cases, the master may want to maintain control of the bus and re-initiate another transmission. If so, the master device may send another Start bit in place of the Stop bit or last $\overline{ACK}$ bit when it is in receive mode.

The I²C bus specifies three message protocols;

- Single message where a master writes data to a slave.
- Single message where a master reads data from a slave.
- Combined message where a master initiates a minimum of two writes, or two reads, or a combination of writes and reads, to one or more slaves.

When one device is transmitting a logical one, or letting the line float, and a second device is transmitting a logical zero, or holding the line low, the first device can detect that the line is not a logical one. This detection, when used on the SCL line, is called clock stretching. Clock stretching gives slave devices a mechanism to control the flow of data. When this detection is used on

FIGURE 20-14: I²C SLAVE, 7-BIT ADDRESS, RECEPTION (SEN = 0, AHEN = 0, DHEN = 0)

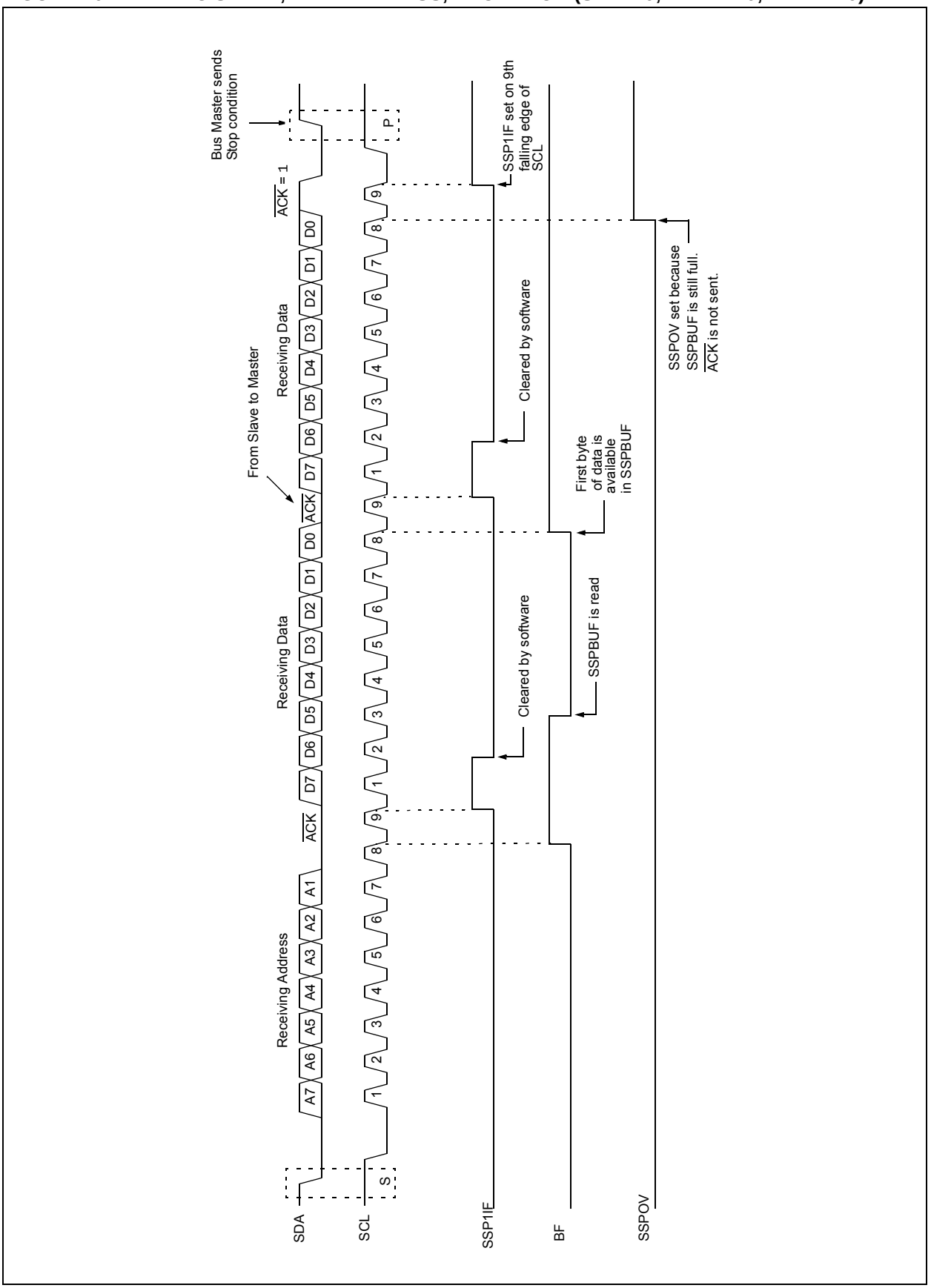


TABLE 21-1: SUMMARY OF REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
BAUDCON	ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	245
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	77
PIE1	TMR1GIE	AD1IE	RCIE	TXIE	SSP1IE	—	TMR2IE	TMR1IE	78
PIR1	TMR1GIF	AD1IF	RCIF	TXIF	SSP1IF	—	TMR2IF	TMR1IF	80
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	244
SPBRGL	BRG<7:0>								246*
SPBRGH	BRG<15:8>								246*
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	—	—	—	—	112
TXREG	EUSART Transmit Data Register								236
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	243

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for asynchronous transmission.

* Page provides register information.

21.1.2 EUSART ASYNCHRONOUS RECEIVER

The Asynchronous mode is typically used in RS-232 systems. The receiver block diagram is shown in Figure 21-2. The data is received on the RX/DT pin and drives the data recovery block. The data recovery block is actually a high-speed shifter operating at 16 times the baud rate, whereas the serial Receive Shift Register (RSR) operates at the bit rate. When all eight or nine bits of the character have been shifted in, they are immediately transferred to a two character First-In-First-Out (FIFO) memory. The FIFO buffering allows reception of two complete characters and the start of a third character before software must start servicing the EUSART receiver. The FIFO and RSR registers are not directly accessible by software. Access to the received data is via the RCREG register.

21.1.2.1 Enabling the Receiver

The EUSART receiver is enabled for asynchronous operation by configuring the following three control bits:

- CREN = 1
- SYNC = 0
- SPEN = 1

All other EUSART control bits are assumed to be in their default state.

Setting the CREN bit of the RCSTA register enables the receiver circuitry of the EUSART. Clearing the SYNC bit of the TXSTA register configures the EUSART for asynchronous operation. Setting the SPEN bit of the RCSTA register enables the EUSART. The programmer must set the corresponding TRISx bit to configure the RX/DT I/O pin as an input.

Note:	If the RX/DT function is on an analog pin, the corresponding ANSELx bit must be cleared for the receiver to function.

21.1.2.2 Receiving Data

The receiver data recovery circuit initiates character reception on the falling edge of the first bit. The first bit, also known as the Start bit, is always a zero. The data recovery circuit counts one-half bit time to the center of the Start bit and verifies that the bit is still a zero. If it is not a zero then the data recovery circuit aborts character reception, without generating an error, and resumes looking for the falling edge of the Start bit. If the Start bit zero verification succeeds then the data recovery circuit counts a full bit time to the center of the next bit. The bit is then sampled by a majority detect circuit and the resulting '0' or '1' is shifted into the RSR. This repeats until all data bits have been sampled and shifted into the RSR. One final bit time is measured and the level sampled. This is the Stop bit, which is always a '1'. If the data recovery circuit samples a '0' in the Stop bit position then a framing error is set for this character, otherwise the framing error is cleared for this character. See **Section 21.1.2.4 "Receive Framing Error"** for more information on framing errors.

Immediately after all data bits and the Stop bit have been received, the character in the RSR is transferred to the EUSART receive FIFO and the RCIF interrupt flag bit of the PIR1 register is set. The top character in the FIFO is transferred out of the FIFO by reading the RCREG register.

Note:	If the receive FIFO is overrun, no additional characters will be received until the overrun condition is cleared. See **Section 21.1.2.5 "Receive Overrun Error"** for more information on overrun errors.

21.1.2.3 Receive Interrupts

The RCIF interrupt flag bit of the PIR1 register is set whenever the EUSART receiver is enabled and there is an unread character in the receive FIFO. The RCIF interrupt flag bit is read-only, it cannot be set or cleared by software.

RCIF interrupts are enabled by setting all of the following bits:

- RCIE, Interrupt Enable bit of the PIE1 register
- PEIE, Peripheral Interrupt Enable bit of the INTCON register
- GIE, Global Interrupt Enable bit of the INTCON register

The RCIF interrupt flag bit will be set when there is an unread character in the FIFO, regardless of the state of interrupt enable bits.

TABLE 21-3: BAUD RATE FORMULAS

Configuration Bits			BRG/EUSART Mode	Baud Rate Formula
SYNC	BRG16	BRGH		
0	0	0	8-bit/Asynchronous	FOSC/[64 (n+1)]
0	0	1	8-bit/Asynchronous	FOSC/[16 (n+1)]
0	1	0	16-bit/Asynchronous	
0	1	1	16-bit/Asynchronous	FOSC/[4 (n+1)]
1	0	x	8-bit/Synchronous	
1	1	x	16-bit/Synchronous	

Legend: x = Don't care, n = value of SPBRGH, SPBRGL register pair.

TABLE 21-4: SUMMARY OF REGISTERS ASSOCIATED WITH THE BAUD RATE GENERATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
BAUDCON	ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	245
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	244
SPBRGL	BRG<7:0>								246*
SPBRGH	BRG<15:8>								246*
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	243

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for the Baud Rate Generator.

* Page provides register information.

21.5.2.3 EUSART Synchronous Slave Reception

The operation of the Synchronous Master and Slave modes is identical (**Section 21.5.1.5 "Synchronous Master Reception"**), with the following exceptions:

- Sleep
- CREN bit is always set, therefore the receiver is never idle
- SREN bit, which is a "don't care" in Slave mode

A character may be received while in Sleep mode by setting the CREN bit prior to entering Sleep. Once the word is received, the RSR register will transfer the data to the RCREG register. If the RCIE enable bit is set, the interrupt generated will wake the device from Sleep and execute the next instruction. If the GIE bit is also set, the program will branch to the interrupt vector.

21.5.2.4 Synchronous Slave Reception Set-up:

1. Set the SYNC and SPEN bits and clear the CSRC bit.
2. Clear the ANSELx bit for both the CK and DT pins (if applicable).
3. If interrupts are desired, set the RCIE bit of the PIE1 register and the GIE and PEIE bits of the INTCON register.
4. If 9-bit reception is desired, set the RX9 bit.
5. Set the CREN bit to enable reception.
6. The RCIF bit will be set when reception is complete. An interrupt will be generated if the RCIE bit was set.
7. If 9-bit mode is enabled, retrieve the Most Significant bit from the RX9D bit of the RCSTA register.
8. Retrieve the eight Least Significant bits from the receive FIFO by reading the RCREG register.
9. If an overrun error occurs, clear the error by either clearing the CREN bit of the RCSTA register or by clearing the SPEN bit which resets the EUSART.

TABLE 21-10: SUMMARY OF REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
BAUDCON	ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	245
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	77
PIE1	TMR1GIE	AD1IE	RCIE	TXIE	SSP1IE	—	TMR2IE	TMR1IE	78
PIR1	TMR1GIF	AD1IF	RCIF	TXIF	SSP1IF	—	TMR2IF	TMR1IF	80
RCREG	EUSART Receive Data Register								239*
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	244
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	—	—	—	—	112
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	243

Legend: — = unimplemented location, read as '0'. Shaded cells are not used for synchronous slave reception.

* Page provides register information.

BCF **Bit Clear f**

Syntax:	[*label*] BCF f,b
Operands:	0 ≤ f ≤ 127 0 ≤ b ≤ 7
Operation:	0 → (f<b>)
Status Affected:	None
Description:	Bit 'b' in register 'f' is cleared.

BRA **Relative Branch**

Syntax:	[*label*] BRA label [*label*] BRA $+k
Operands:	-256 ≤ label - PC + 1 ≤ 255 -256 ≤ k ≤ 255
Operation:	(PC) + 1 + k → PC
Status Affected:	None
Description:	Add the signed 9-bit literal 'k' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 1 + k. This instruction is a 2-cycle instruction. This branch has a limited range.

BRW **Relative Branch with W**

Syntax:	[*label*] BRW
Operands:	None
Operation:	(PC) + (W) → PC
Status Affected:	None
Description:	Add the contents of W (unsigned) to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 1 + (W). This instruction is a 2-cycle instruction.

BSF **Bit Set f**

Syntax:	[*label*] BSF f,b
Operands:	0 ≤ f ≤ 127 0 ≤ b ≤ 7
Operation:	1 → (f<b>)
Status Affected:	None
Description:	Bit 'b' in register 'f' is set.

BTFSC **Bit Test f, Skip if Clear**

Syntax:	[*label*] BTFSC f,b
Operands:	0 ≤ f ≤ 127 0 ≤ b ≤ 7
Operation:	skip if (f<b>) = 0
Status Affected:	None
Description:	If bit 'b' in register 'f' is '1', the next instruction is executed. If bit 'b', in register 'f', is '0', the next instruction is discarded, and a NOP is executed instead, making this a 2-cycle instruction.

BTFSS **Bit Test f, Skip if Set**

Syntax:	[*label*] BTFSS f,b
Operands:	0 ≤ f ≤ 127 0 ≤ b < 7
Operation:	skip if (f<b>) = 1
Status Affected:	None
Description:	If bit 'b' in register 'f' is '0', the next instruction is executed. If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a 2-cycle instruction.

Note: Unless otherwise noted C_{IN} = 0.1 µF and T_A = 25°C.

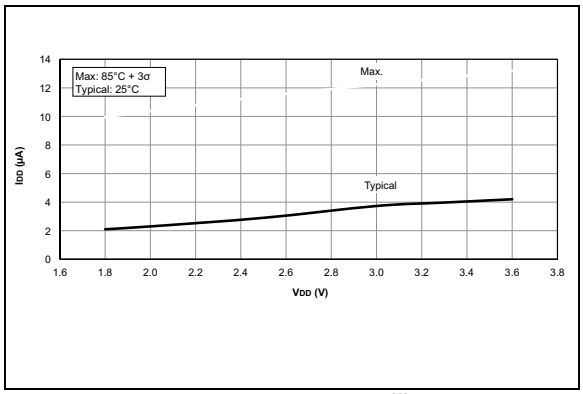


FIGURE 26-1: *IDD, EC Oscillator, Low-Power mode, FOSC=32 kHz*

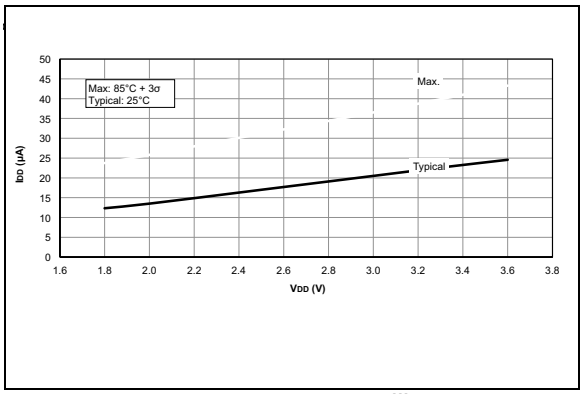


FIGURE 26-4: *IDD, EC Oscillator, Low-Power mode, FOSC = 500 kHz*

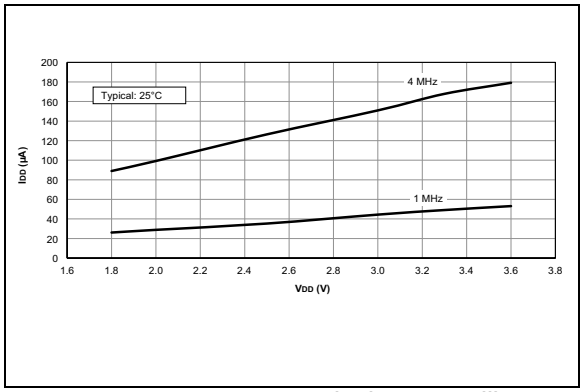


FIGURE 26-2: *IDD Typical, EC Oscillator, Medium-Power Mode*

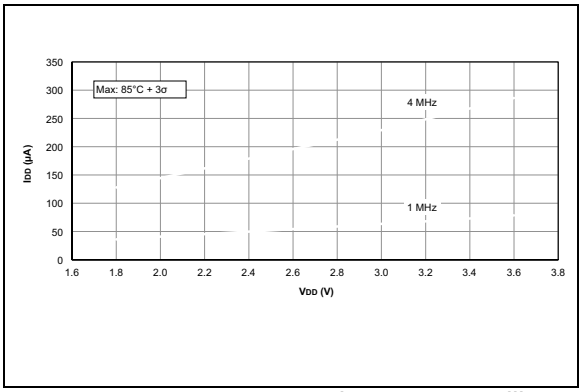


FIGURE 26-5: *IDD Maximum, EC Oscillator, Medium-Power Mode*

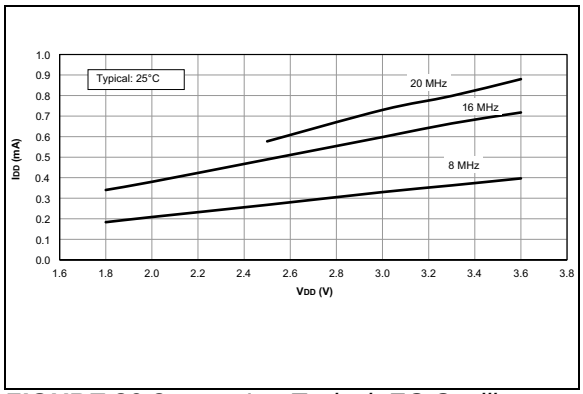


FIGURE 26-3: *IDD Typical, EC Oscillator, High-Power Mode*

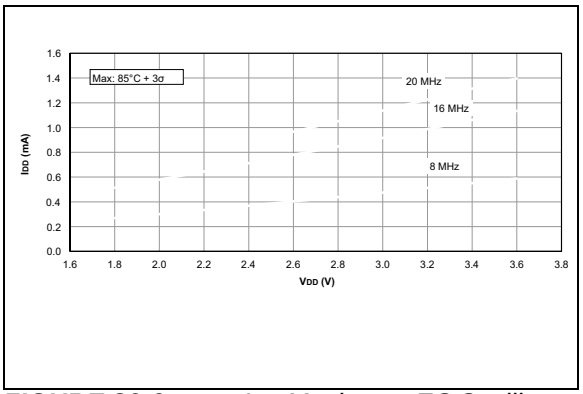
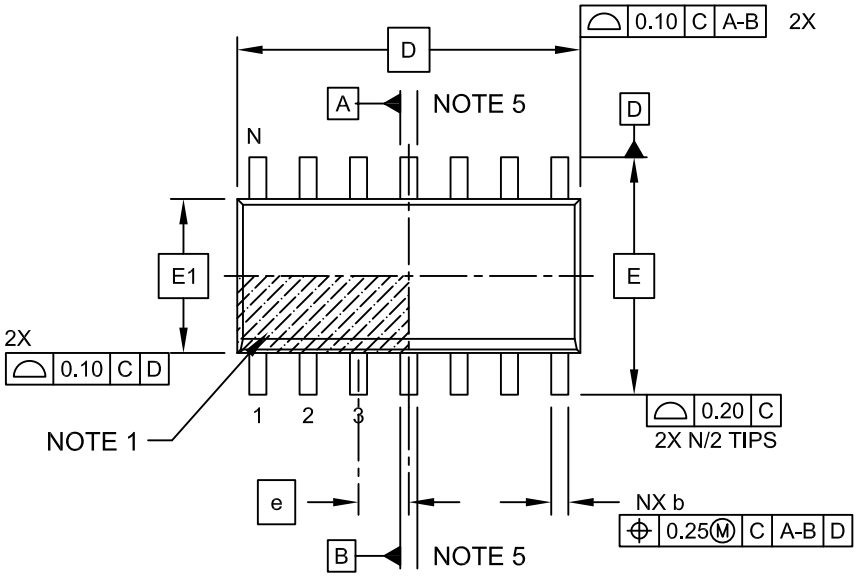


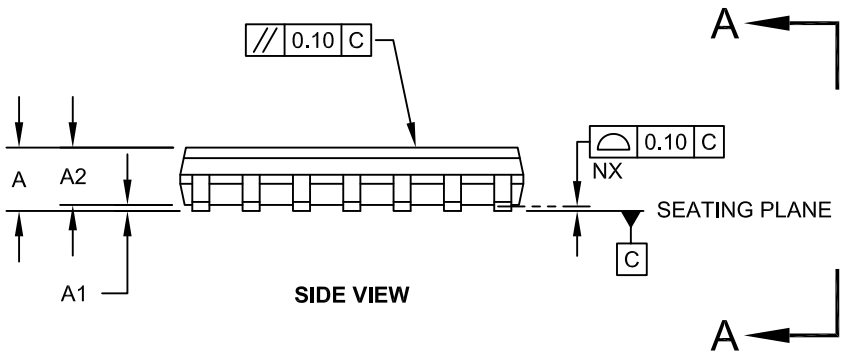
FIGURE 26-6: *IDD Maximum, EC Oscillator, High-Power Mode*

14-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

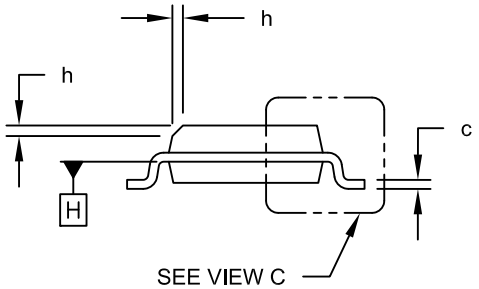
Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



TOP VIEW



SIDE VIEW



VIEW A-A

Microchip Technology Drawing No. C04-065C Sheet 1 of 2