Welcome to E-XFL.COM

What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	ARM® Cortex®-M3
Core Size	32-Bit Single-Core
Speed	72MHz
Connectivity	CANbus, I²C, IrDA, LINbus, SPI, UART/USART, USB
Peripherals	DMA, Motor Control PWM, PDR, POR, PVD, PWM, Temp Sensor, WDT
Number of I/O	80
Program Memory Size	256KB (256K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	48K x 8
Voltage - Supply (Vcc/Vdd)	2V ~ 3.6V
Data Converters	A/D 16x12b; D/A 2x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP
Supplier Device Package	100-LQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/stmicroelectronics/stm32f103vct6tr

2.3.14 Low-power modes

The STM32F103xC, STM32F103xD and STM32F103xE performance line supports three low-power modes to achieve the best compromise between low-power consumption, short startup time and available wakeup sources:

- **Sleep** mode

  In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

- **Stop** mode

  Stop mode achieves the lowest power consumption while retaining the content of SRAM and registers. All clocks in the 1.8 V domain are stopped, the PLL, the HSI RC and the HSE crystal oscillators are disabled. The voltage regulator can also be put either in normal or in low-power mode.
  The device can be woken up from Stop mode by any of the EXTI line. The EXTI line source can be one of the 16 external lines, the PVD output, the RTC alarm or the USB wakeup.

- **Standby** mode

  The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire 1.8 V domain is powered off. The PLL, the HSI RC and the HSE crystal oscillators are also switched off. After entering Standby mode, SRAM and register contents are lost except for registers in the Backup domain and Standby circuitry.

  The device exits Standby mode when an external reset (NRST pin), an IWDG reset, a rising edge on the WKUP pin, or an RTC alarm occurs.

Note: The RTC, the IWDG, and the corresponding clock sources are not stopped by entering Stop or Standby mode.

2.3.15 DMA

The flexible 12-channel general-purpose DMAs (7 channels for DMA1 and 5 channels for DMA2) are able to manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers. The two DMA controllers support circular buffer management, removing the need for user code intervention when the controller reaches the end of the buffer.

Each channel is connected to dedicated hardware DMA requests, with support for software trigger on each channel. Configuration is made by software and transfer sizes between source and destination are independent.

The DMA can be used with the main peripherals: SPI, I^2C, USART, general-purpose, basic and advanced-control timers TIMx, DAC, I^2S, SDIO and ADC.

2.3.16 RTC (real-time clock) and backup registers

The RTC and the backup registers are supplied through a switch that takes power either on V_{DD} supply when present or through the V_{BAT} pin. The backup registers are forty-two 16-bit registers used to store 84 bytes of user application data when V_{DD} power is not present. They are not reset by a system or power reset, and they are not reset when the device wakes up from the Standby mode.

The real-time clock provides a set of continuously running counters which can be used with suitable software to provide a clock calendar function, and provides an alarm interrupt and a

mode, the master clock can be output to the external DAC/CODEC at 256 times the sampling frequency.

2.3.22 SDIO

An SD/SDIO/MMC host interface is available, that supports MultiMediaCard System Specification Version 4.2 in three different databus modes: 1-bit (default), 4-bit and 8-bit. The interface allows data transfer at up to 48 MHz in 8-bit mode, and is compliant with SD Memory Card Specifications Version 2.0.
The SDIO Card Specification Version 2.0 is also supported with two different databus modes: 1-bit (default) and 4-bit.

The current version supports only one SD/SDIO/MMC4.2 card at any one time and a stack of MMC4.1 or previous.

In addition to SD/SDIO/MMC, this interface is also fully compliant with the CE-ATA digital protocol Rev1.1.

2.3.23 Controller area network (CAN)

The CAN is compliant with specifications 2.0A and B (active) with a bit rate up to 1 Mbit/s. It can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers. It has three transmit mailboxes, two receive FIFOs with 3 stages and 14 scalable filter banks.

2.3.24 Universal serial bus (USB)

The STM32F103xC, STM32F103xD and STM32F103xE performance line embed a USB device peripheral compatible with the USB full-speed 12 Mbs. The USB interface implements a full-speed (12 Mbit/s) function interface. It has software-configurable endpoint setting and suspend/resume support. The dedicated 48 MHz clock is generated from the internal main PLL (the clock source must use a HSE crystal oscillator).

2.3.25 GPIOs (general-purpose inputs/outputs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. All GPIOs are high current-capable.

The I/Os alternate function configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

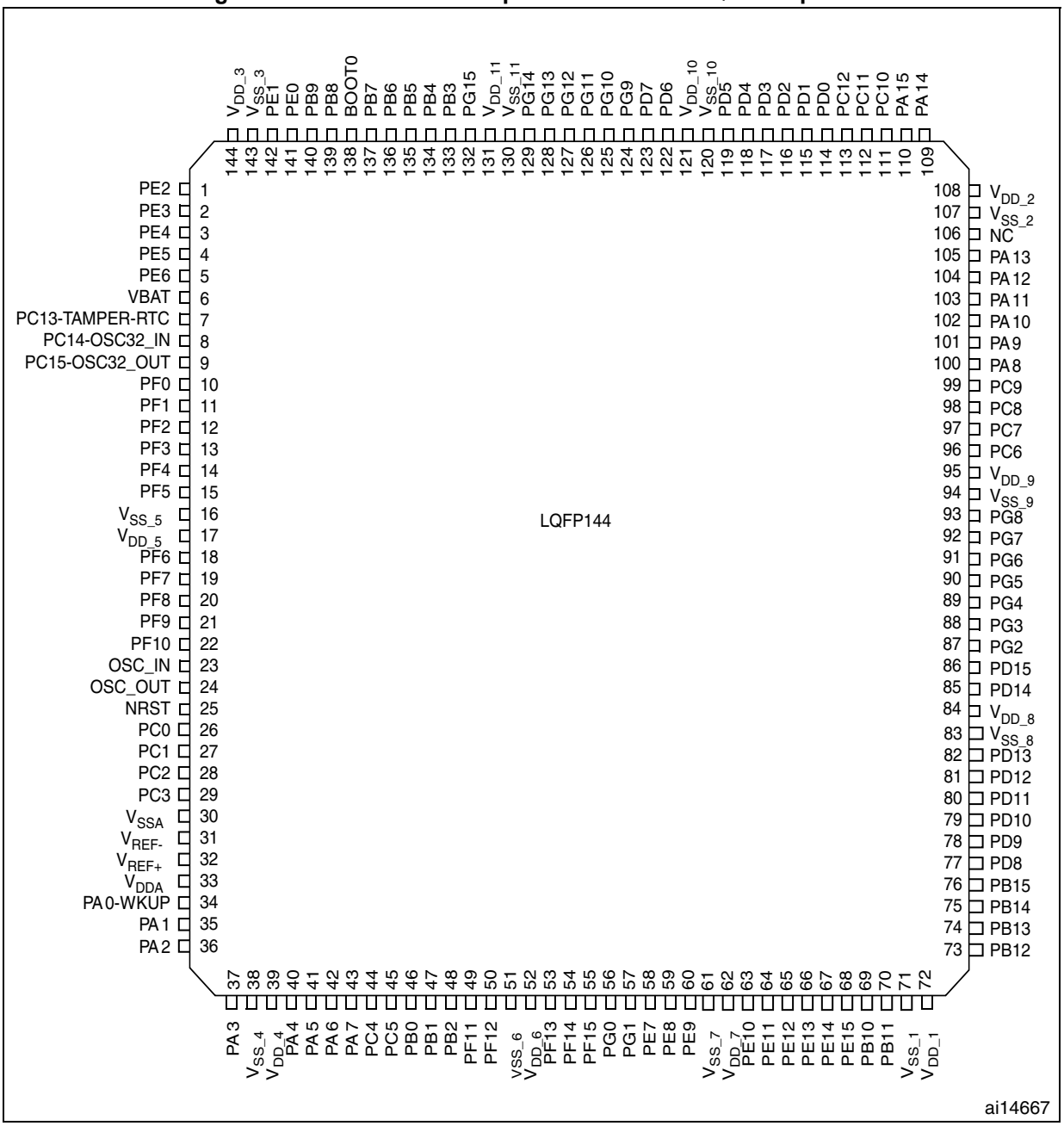
2.3.26 ADC (analog to digital converter)

Three 12-bit analog-to-digital converters are embedded into STM32F103xC, STM32F103xD and STM32F103xE performance line devices and each ADC shares up to 21 external channels, performing conversions in single-shot or scan modes. In scan mode, automatic conversion is performed on a selected group of analog inputs.

Additional logic functions embedded in the ADC interface allow:

- Simultaneous sample and hold
- Interleaved sample and hold
- Single shunt

Figure 5. STM32F103xC/D/E performance line LQFP144 pinout



1. The above figure shows the package top view.

Table 6. FSMC pin definition

Pins	FSMC					LQFP100 BGA100[1]
	CF	CF/IDE	NOR/PSRAM/ SRAM	NOR/PSRAM Mux	NAND 16 bit	
PE2	-	-	A23	A23	-	Yes
PE3	-	-	A19	A19	-	Yes
PE4	-	-	A20	A20	-	Yes
PE5	-	-	A21	A21	-	Yes
PE6	-	-	A22	A22	-	Yes
PF0	A0	A0	A0	-	-	-
PF1	A1	A1	A1	-	-	-
PF2	A2	A2	A2	-	-	-
PF3	A3	-	A3	-	-	-
PF4	A4	-	A4	-	-	-
PF5	A5	-	A5	-	-	-
PF6	NIORD	NIORD	-	-	-	-
PF7	NREG	NREG	-	-	-	-
PF8	NIOWR	NIOWR	-	-	-	-
PF9	CD	CD	-	-	-	-
PF10	INTR	INTR	-	-	-	-
PF11	NIOS16	NIOS16	-	-	-	-
PF12	A6	-	A6	-	-	-
PF13	A7	-	A7	-	-	-
PF14	A8	-	A8	-	-	-
PF15	A9	-	A9	-	-	-
PG0	A10	-	A10	-	-	-
PG1	-	-	A11	-	-	-
PE7	D4	D4	D4	DA4	D4	Yes
PE8	D5	D5	D5	DA5	D5	Yes
PE9	D6	D6	D6	DA6	D6	Yes
PE10	D7	D7	D7	DA7	D7	Yes
PE11	D8	D8	D8	DA8	D8	Yes
PE12	D9	D9	D9	DA9	D9	Yes
PE13	D10	D10	D10	DA10	D10	Yes
PE14	D11	D11	D11	DA11	D11	Yes
PE15	D12	D12	D12	DA12	D12	Yes
PD8	D13	D13	D13	DA13	D13	Yes

Table 6. FSMC pin definition (continued)

Pins	FSMC					LQFP100 BGA100(1)
	CF	CF/IDE	NOR/PSRAM/ SRAM	NOR/PSRAM Mux	NAND 16 bit	
PD9	D14	D14	D14	DA14	D14	Yes
PD10	D15	D15	D15	DA15	D15	Yes
PD11	-	-	A16	A16	CLE	Yes
PD12	-	-	A17	A17	ALE	Yes
PD13	-	-	A18	A18	-	Yes
PD14	D0	D0	D0	DA0	D0	Yes
PD15	D1	D1	D1	DA1	D1	Yes
PG2	-	-	A12	-	-	-
PG3	-	-	A13	-	-	-
PG4	-	-	A14	-	-	-
PG5	-	-	A15	-	-	-
PG6	-	-	-	-	INT2	-
PG7	-	-	-	-	INT3	-
PD0	D2	D2	D2	DA2	D2	Yes
PD1	D3	D3	D3	DA3	D3	Yes
PD3	-	-	CLK	CLK	-	Yes
PD4	NOE	NOE	NOE	NOE	NOE	Yes
PD5	NWE	NWE	NWE	NWE	NWE	Yes
PD6	NWAIT	NWAIT	NWAIT	NWAIT	NWAIT	Yes
PD7	-	-	NE1	NE1	NCE2	Yes
PG9	-	-	NE2	NE2	NCE3	-
PG10	NCE4_1	NCE4_1	NE3	NE3	-	-
PG11	NCE4_2	NCE4_2	-	-	-	-
PG12	-	-	NE4	NE4	-	-
PG13	-	-	A24	A24	-	-
PG14	-	-	A25	A25	-	-
PB7	-	-	NADV	NADV	-	Yes
PE0	-	-	NBL0	NBL0	-	Yes
PE1	-	-	NBL1	NBL1	-	Yes

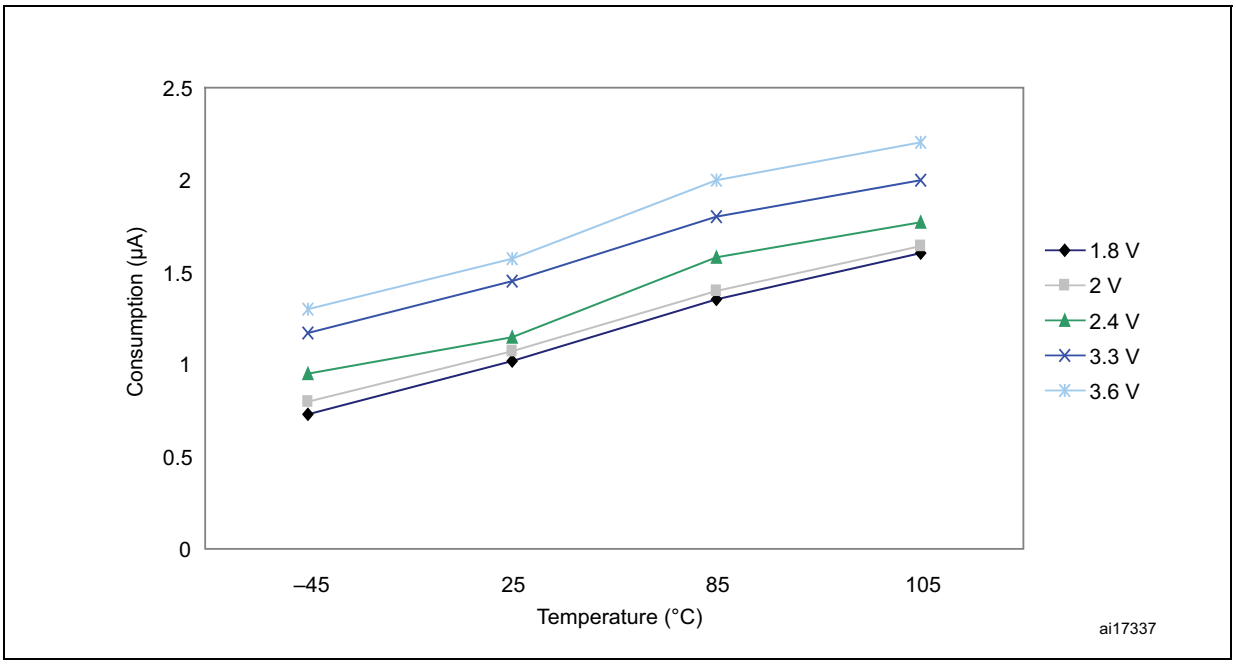
1. Ports F and G are not available in devices delivered in 100-pin packages.

Table 17. Typical and maximum current consumptions in Stop and Standby modes

Symbol	Parameter	Conditions	Typ(1)			Max		Unit
			V_{DD}/V_{BAT} = 2.0 V	V_{DD}/V_{BAT} = 2.4 V	V_{DD}/V_{BAT} = 3.3 V	T_A = 85 °C	T_A = 105 °C	
I_{DD}	Supply current in Stop mode	Regulator in run mode, low-speed and high-speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	-	34.5	35	379	1130	µA
		Regulator in low-power mode, low-speed and high-speed internal RC oscillators and high-speed oscillator OFF (no independent watchdog)	-	24.5	25	365	1110	
	Supply current in Standby mode	Low-speed internal RC oscillator and independent watchdog ON	-	3	3.8	-	-	
		Low-speed internal RC oscillator ON, independent watchdog OFF	-	2.8	3.6	-	-	
		Low-speed internal RC oscillator and independent watchdog OFF, low-speed oscillator and RTC OFF	-	1.9	2.1	5(2)	6.5(2)	
I_{DD_VBAT}	Backup domain supply current	Low-speed oscillator and RTC ON	1.05	1.1	1.4	2(2)	2.3(2)	

1. Typical values are measured at T_A = 25 °C.
2. Guaranteed by characterization results.

Figure 16. Typical current consumption on V_{BAT} with RTC on vs. temperature at different V_{BAT} values



On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in *Table 20*. The MCU is placed under the following conditions:

- all I/O pins are in input mode with a static value at V_{DD} or V_{SS} (no load)
- all peripherals are disabled unless otherwise mentioned
- the given value is calculated by measuring the current consumption
  - with all peripherals clocked off
  - with only one peripheral clocked on
- ambient operating temperature and V_{DD} supply voltage conditions summarized in *Table 7*

Table 20. Peripheral current consumption

Peripheral		Current consumption	Unit
AHB (up to 72 MHz)	DMA1	20,42	µA/MHz
	DMA2	19,03	
	FSMC	52,36	
	CRC	2,36	
	SDIO	33,33	
	BusMatrix(1)	9,72	

Table 20. Peripheral current consumption (continued)

Peripheral		Current consumption	Unit
APB1 (up to 36 MHz)	APB1-Bridge	7,78	µA/MHz
	TIM2	33,06	
	TIM3	31,94	
	TIM4	31,67	
	TIM5	31,94	
	TIM6	8,06	
	TIM7	8,06	
	SPI2/I2S2(2)	8,33	
	SPI3/I2S3(2)	8,33	
	USART2	12,22	
	USART3	12,22	
	UART4	12,22	
	UART5	12,22	
	I2C1	10,28	
	I2C2	10,00	
	USB	18,06	
	CAN1	18,33	
	DAC(3)	8,06	
	WWDG	3,89	
	PWR	1,11	
	BKP	1,11	
	IWDG	5,28	

High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 16 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in *Table 23*. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

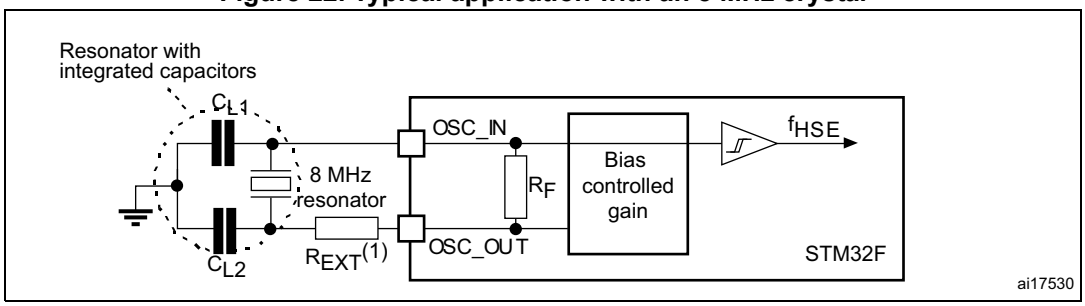
Table 23. HSE 4-16 MHz oscillator characteristics(1)(2)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{OSC_IN}	Oscillator frequency	-	4	8	16	MHz
R_F	Feedback resistor	-	-	200	-	kΩ
C	Recommended load capacitance versus equivalent serial resistance of the crystal (R_S)(3)	R_S = 30 Ω	-	30	-	pF
i_2	HSE driving current	V_{DD}= 3.3 V, V_{IN} = V_{SS} with 30 pF load	-	-	1	mA
g_m	Oscillator transconductance	Startup	25	-	-	mA/V
$t_{SU(HSE)}$(4)	Startup time	V_{DD} is stabilized	-	2	-	ms

1. Resonator characteristics given by the crystal/ceramic resonator manufacturer.
2. Guaranteed by characterization results.
3. The relatively low value of the RF resistor offers a good protection against issues resulting from use in a humid environment, due to the induced leakage and the bias condition change. However, it is recommended to take this point into account if the MCU is used in tough humidity conditions.
4. $t_{SU(HSE)}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer

For C_{L1} and C_{L2}, it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see *Figure 22*). C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2}. PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing C_{L1} and C_{L2}. Refer to the application note AN2867 “Oscillator design guide for ST microcontrollers” available from the ST website *www.st.com*.

Figure 22. Typical application with an 8 MHz crystal



1. R_{EXT} value depends on the crystal characteristics.

5.3.8 PLL characteristics

The parameters given in *Table 28* are derived from tests performed under ambient temperature and V_{DD} supply voltage conditions summarized in *Table 10*.

Table 28. PLL characteristics

Symbol	Parameter	Value			Unit
		Min	Typ	Max(1)	
f_{PLL_IN}	PLL input clock(2)	1	8.0	25	MHz
	PLL input clock duty cycle	40	-	60	%
f_{PLL_OUT}	PLL multiplier output clock	16	-	72	MHz
t_{LOCK}	PLL lock time	-	-	200	µs
Jitter	Cycle-to-cycle jitter	-	-	300	ps

1. Guaranteed by characterization results.
2. Take care of using the appropriate multiplier factors so as to have PLL input clock values compatible with the range defined by f_{PLL_OUT}.

5.3.9 Memory characteristics

Flash memory

The characteristics are given at T_A = –40 to 105 °C unless otherwise specified.

Table 29. Flash memory characteristics

Symbol	Parameter	Conditions	Min	Typ	Max(1)	Unit
t_{prog}	16-bit programming time	T_A = –40 to +105 °C	40	52.5	70	µs
t_{ERASE}	Page (2 KB) erase time	T_A = –40 to +105 °C	20	-	40	ms
t_{ME}	Mass erase time	T_A = –40 to +105 °C	20	-	40	ms
I_{DD}	Supply current	Read mode f_{HCLK} = 72 MHz with 2 wait states, V_{DD} = 3.3 V	-	-	28	mA
		Write mode f_{HCLK} = 72 MHz, V_{DD} = 3.3 V	-	-	7	mA
		Erase mode f_{HCLK} = 72 MHz, V_{DD} = 3.3 V	-	-	5	mA
		Power-down mode / Halt, V_{DD} = 3.0 to 3.6 V	-	-	50	µA
V_{prog}	Programming voltage	-	2	-	3.6	V

1. Guaranteed by design.

Table 30. Flash memory endurance and data retention

Symbol	Parameter	Conditions	Value	Unit
			Min(1)	
N_{END}	Endurance	T_A = –40 to +85 °C (6 suffix versions) T_A = –40 to +105 °C (7 suffix versions)	10	kcycles
t_{RET}	Data retention	1 kcycle(2) at T_A = 85 °C	30	Years
		1 kcycle(2) at T_A = 105 °C	10	
		10 kcycles(2) at T_A = 55 °C	20	

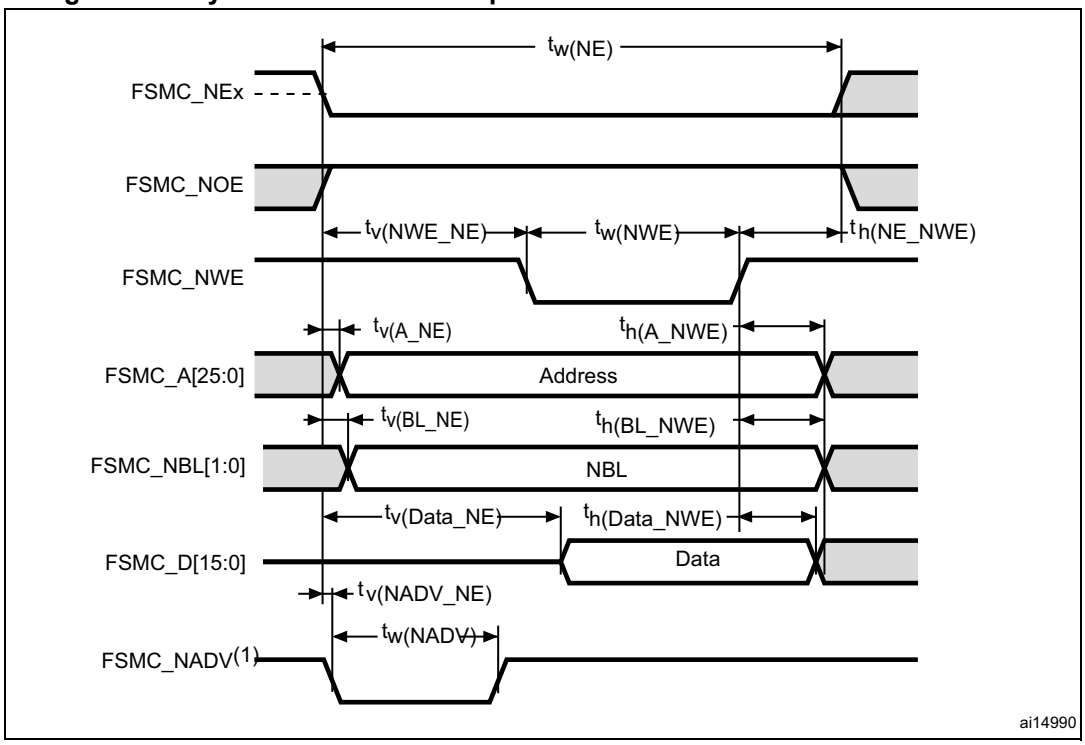
1. Guaranteed by characterization results.
2. Cycling performed over the whole temperature range.

Table 31. Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings(1)

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FSMC_NE low time	$5t_{HCLK} - 1.5$	$5t_{HCLK} + 2$	ns
$t_{v(NOE\_NE)}$	FSMC_NEx low to FSMC_NOE low	0.5	1.5	ns
$t_{w(NOE)}$	FSMC_NOE low time	$5t_{HCLK} - 1.5$	$5t_{HCLK} + 1.5$	ns
$t_{h(NE\_NOE)}$	FSMC_NOE high to FSMC_NE high hold time	–1.5	-	ns
$t_{v(A\_NE)}$	FSMC_NEx low to FSMC_A valid	-	0	ns
$t_{h(A\_NOE)}$	Address hold time after FSMC_NOE high	0.1	-	ns
$t_{v(BL\_NE)}$	FSMC_NEx low to FSMC_BL valid	-	0	ns
$t_{h(BL\_NOE)}$	FSMC_BL hold time after FSMC_NOE high	0	-	ns
$t_{su(Data\_NE)}$	Data to FSMC_NEx high setup time	$2t_{HCLK} + 25$	-	ns
$t_{su(Data\_NOE)}$	Data to FSMC_NOEx high setup time	$2t_{HCLK} + 25$	-	ns
$t_{h(Data\_NOE)}$	Data hold time after FSMC_NOE high	0	-	ns
$t_{h(Data\_NE)}$	Data hold time after FSMC_NEx high	0	-	ns
$t_{v(NADV\_NE)}$	FSMC_NEx low to FSMC_NADV low	-	5	ns
$t_{w(NADV)}$	FSMC_NADV low time	-	$t_{HCLK} + 1.5$	ns

1. C_L = 15 pF.

Figure 25. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms



1. Mode 2/B, C and D only. In Mode 1, FSMC_NADV is not used.

Table 35. Synchronous multiplexed NOR/PSRAM read timings(1)(2)

Symbol	Parameter	Min	Max	Unit
$t_{w(CLK)}$	FSMC_CLK period	27.7	-	ns
$t_{d(CLKL-NExL)}$	FSMC_CLK low to FSMC_NEx low (x = 0...2)	-	1.5	ns
$t_{d(CLKL-NExH)}$	FSMC_CLK low to FSMC_NEx high (x = 0...2)	2	-	ns
$t_{d(CLKL-NADVL)}$	FSMC_CLK low to FSMC_NADV low	-	4	ns
$t_{d(CLKL-NADVH)}$	FSMC_CLK low to FSMC_NADV high	5	-	ns
$t_{d(CLKL-AV)}$	FSMC_CLK low to FSMC_Ax valid (x = 16...25)	-	0	ns
$t_{d(CLKL-AIV)}$	FSMC_CLK low to FSMC_Ax invalid (x = 16...25)	2	-	ns
$t_{d(CLKL-NOEL)}$	FSMC_CLK low to FSMC_NOE low	-	1	ns
$t_{d(CLKL-NOEH)}$	FSMC_CLK low to FSMC_NOE high	1.5	-	ns
$t_{d(CLKL-ADV)}$	FSMC_CLK low to FSMC_AD[15:0] valid	-	12	ns
$t_{d(CLKL-ADIV)}$	FSMC_CLK low to FSMC_AD[15:0] invalid	0	-	ns
$t_{su(ADV-CLKH)}$	FSMC_A/D[15:0] valid data before FSMC_CLK high	6	-	ns
$t_{h(CLKH-ADV)}$	FSMC_A/D[15:0] valid data after FSMC_CLK high	0	-	ns
$t_{su(NWAITV-CLKH)}$	FSMC_NWAIT valid before FSMC_CLK high	8	-	ns
$t_{h(CLKH-NWAITV)}$	FSMC_NWAIT valid after FSMC_CLK high	2	-	ns

1. C_L = 15 pF.
2. Guaranteed by characterization results.

Table 36. Synchronous multiplexed PSRAM write timings(1)(2)

Symbol	Parameter	Min	Max	Unit
$t_{w(CLK)}$	FSMC_CLK period	27.7	-	ns
$t_{d(CLKL-NExL)}$	FSMC_CLK low to FSMC_Nex low (x = 0...2)	-	2	ns
$t_{d(CLKL-NExH)}$	FSMC_CLK low to FSMC_NEx high (x = 0...2)	2	-	ns
$t_{d(CLKL-NADVL)}$	FSMC_CLK low to FSMC_NADV low	-	4	ns
$t_{d(CLKL-NADVH)}$	FSMC_CLK low to FSMC_NADV high	5	-	ns
$t_{d(CLKL-AV)}$	FSMC_CLK low to FSMC_Ax valid (x = 16...25)	-	0	ns
$t_{d(CLKL-AIV)}$	FSMC_CLK low to FSMC_Ax invalid (x = 16...25)	2	-	ns
$t_{d(CLKL-NWEL)}$	FSMC_CLK low to FSMC_NWE low	-	1	ns
$t_{d(CLKL-NWEH)}$	FSMC_CLK low to FSMC_NWE high	1	-	ns
$t_{d(CLKL-ADV)}$	FSMC_CLK low to FSMC_AD[15:0] valid	-	12	ns
$t_{d(CLKL-ADIV)}$	FSMC_CLK low to FSMC_AD[15:0] invalid	3	-	ns
$t_{d(CLKL-Data)}$	FSMC_A/D[15:0] valid after FSMC_CLK low	-	6	ns
$t_{d(CLKL-NBLH)}$	FSMC_CLK low to FSMC_NBL high	1	-	ns
$t_{su(NWAITV-CLKH)}$	FSMC_NWAIT valid before FSMC_CLK high	7	-	ns
$t_{h(CLKH-NWAITV)}$	FSMC_NWAIT valid after FSMC_CLK high	2	-	ns

1. C_L = 15 pF.
2. Guaranteed by characterization results.

Figure 41. NAND controller waveforms for common memory write access

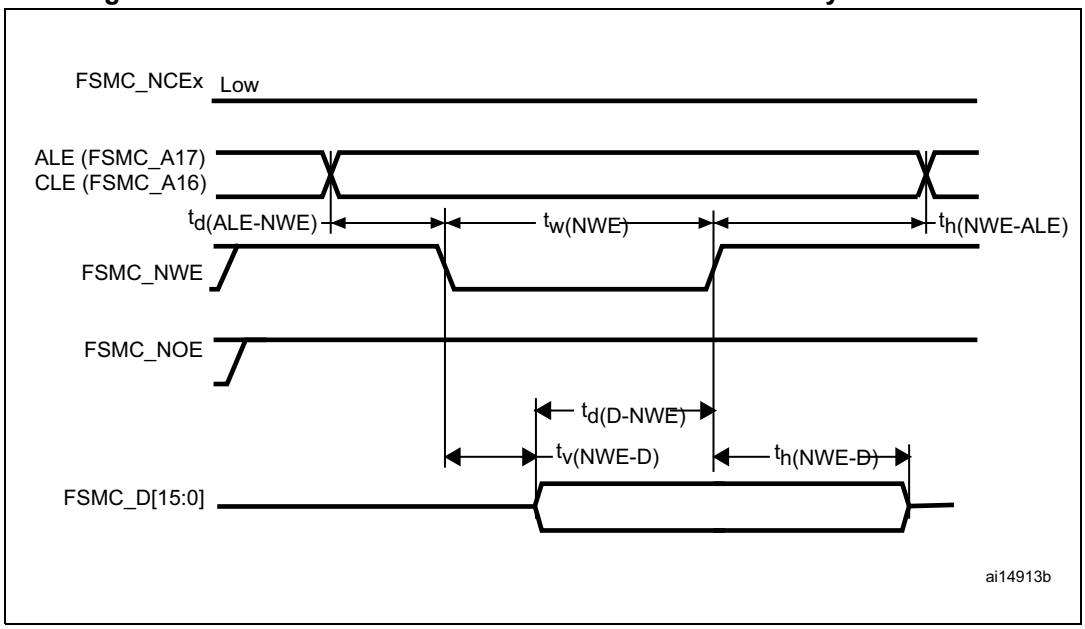


Table 40. Switching characteristics for NAND Flash read and write cycles(1)

Symbol	Parameter	Min	Max	Unit
$t_{d(D-NWE)}$(2)	FSMC_D[15:0] valid before FSMC_NWE high	$5t_{HCLK}$ + 12	-	ns
$t_{w(NOE)}$(2)	FSMC_NWE low width	$4t_{HCLK-1.5}$	$4t_{HCLK+1.5}$	ns
$t_{su(D-NOE)}$(2)	FSMC_D[15:0] valid data before FSMC_NOE high	25	-	ns
$t_{h(NOE-D)}$(2)	FSMC_D[15:0] valid data after FSMC_NOE high	7	-	-
$t_{w(NWE)}$(2)	FSMC_NWE low width	$4t_{HCLK-1}$	$4t_{HCLK+1}$	ns
$t_{v(NWE-D)}$(2)	FSMC_NWE low to FSMC_D[15:0] valid	-	0	ns
$t_{h(NWE-D)}$(2)	FSMC_NWE high to FSMC_D[15:0] invalid	$2t_{HCLK}$ + 4	-	ns
$t_{d(ALE-NWE)}$(3)	FSMC_ALE valid before FSMC_NWE low	-	$3t_{HCLK}$ + 1.5	ns
$t_{h(NWE-ALE)}$(3)	FSMC_NWE high to FSMC_ALE invalid	$3t_{HCLK}$ + 4.5	-	ns
$t_{d(ALE-NOE)}$(3)	FSMC_ALE valid before FSMC_NOE low	-	$3t_{HCLK}$+ 2	ns
$t_{h(NOE-ALE)}$(3)	FSMC_NWE high to FSMC_ALE invalid	$3t_{HCLK}$+ 4.5	-	ns

1. C_L = 15 pF.
2. Guaranteed by characterization results.
3. Guaranteed by design.

5.3.14 I/O port characteristics

General input/output characteristics

Unless otherwise specified, the parameters given in *Table 46* are derived from tests performed under the conditions summarized in *Table 10*. All I/Os are CMOS and TTL compliant.

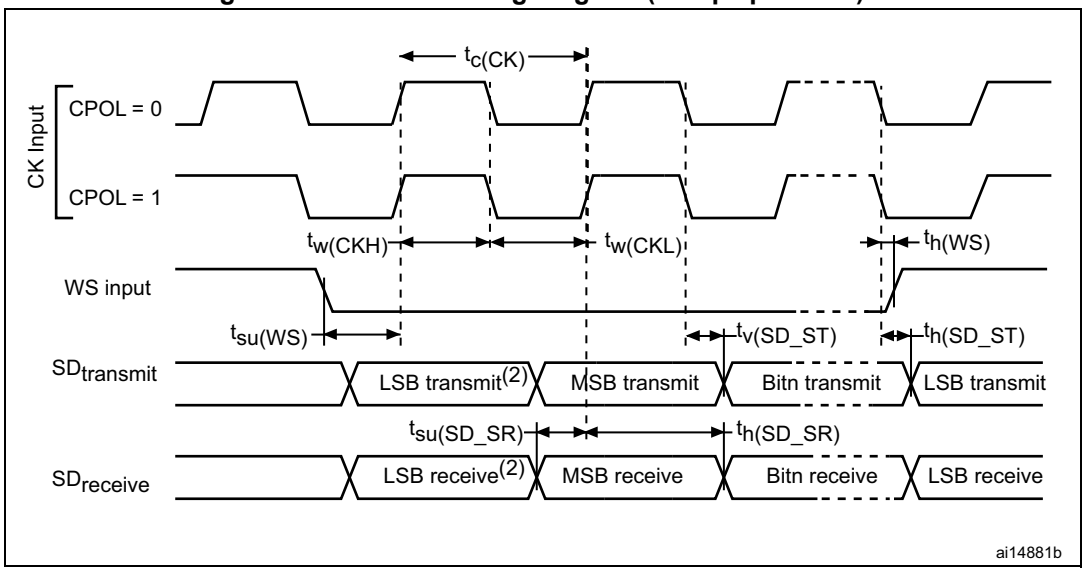
Table 46. I/O static characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{IL}	Standard IO input low level voltage	-	–0.3	-	0.28*(V_{DD}-2 V)+0.8 V	V
	IO FT(1) input low level voltage		–0.3	-	0.32*(V_{DD}-2 V)+0.75 V	V
V_{IH}	Standard IO input high level voltage	-	0.41*(V_{DD}-2 V)+1.3 V	-	V_{DD}+0.3	V
	IO FT(1) input high level voltage	V_{DD} > 2 V	0.42*(V_{DD}-2 V)+1 V	-	5.5	V
		V_{DD} ≤2 V			5.2	
V_{hys}	Standard IO Schmitt trigger voltage hysteresis(2)	-	200	-	-	mV
	IO FT Schmitt trigger voltage hysteresis(2)		5% V_{DD}(3)	-	-	mV
I_{lkg}	Input leakage current (4)	V_{SS} ≤V_{IN} ≤V_{DD} Standard I/Os	-	-	±1	µA
		V_{IN}= 5 V, I/O FT	-	-	3	
R_{PU}	Weak pull-up equivalent resistor(5)	V_{IN} = V_{SS}	30	40	50	kΩ
R_{PD}	Weak pull-down equivalent resistor(5)	V_{IN} = V_{DD}	30	40	50	kΩ
C_{IO}	I/O pin capacitance	-	-	5	-	pF

1. FT = Five-volt tolerant. In order to sustain a voltage higher than V_{DD}+0.3 the internal pull-up/pull-down resistors must be disabled.
2. Hysteresis voltage between Schmitt trigger switching levels. Guaranteed by characterization results.
3. With a minimum of 100 mV.
4. Leakage could be higher than max. if negative current is injected on adjacent pins.
5. Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This MOS/NMOS contribution to the series resistance is minimum (~10% order).

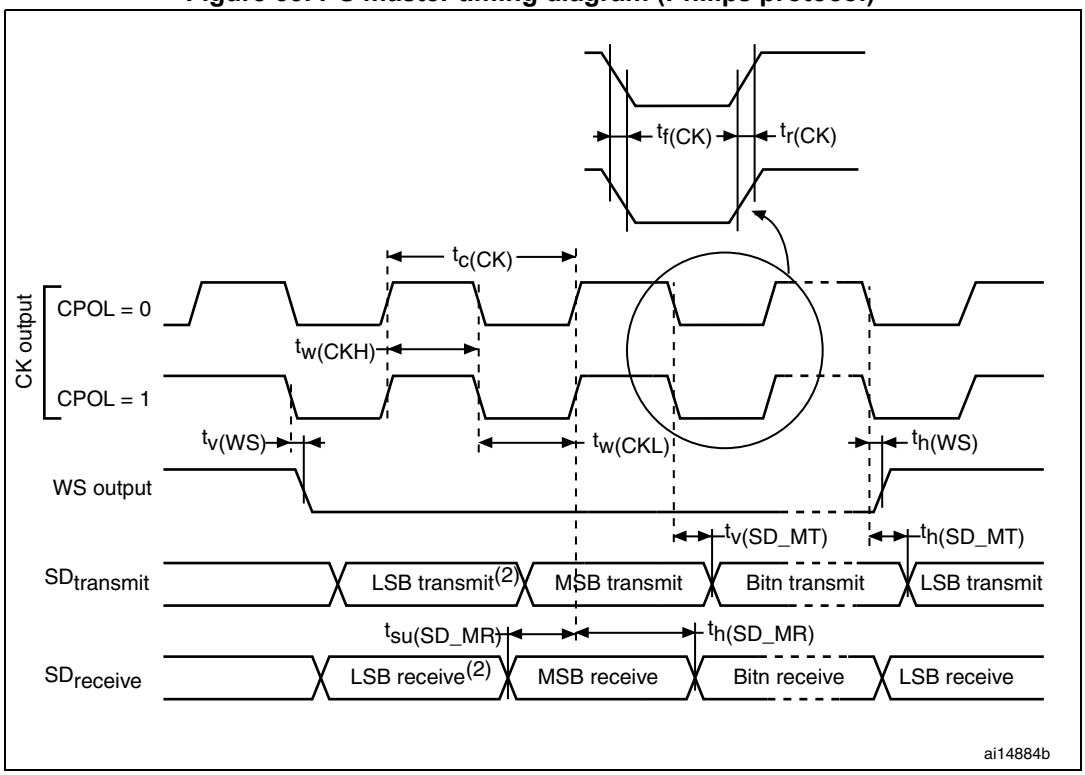
All I/Os are CMOS and TTL compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements is shown in *Figure 42* and *Figure 43* for standard I/Os, and in *Figure 44* and *Figure 45* for 5 V tolerant I/Os.

Figure 52. I²S slave timing diagram (Philips protocol)(1)



1. Measurement points are done at CMOS levels: $0.3 \times V_{DD}$ and $0.7 \times V_{DD}$.
2. LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

Figure 53. I²S master timing diagram (Philips protocol)(1)



1. Guaranteed by characterization results.
2. LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

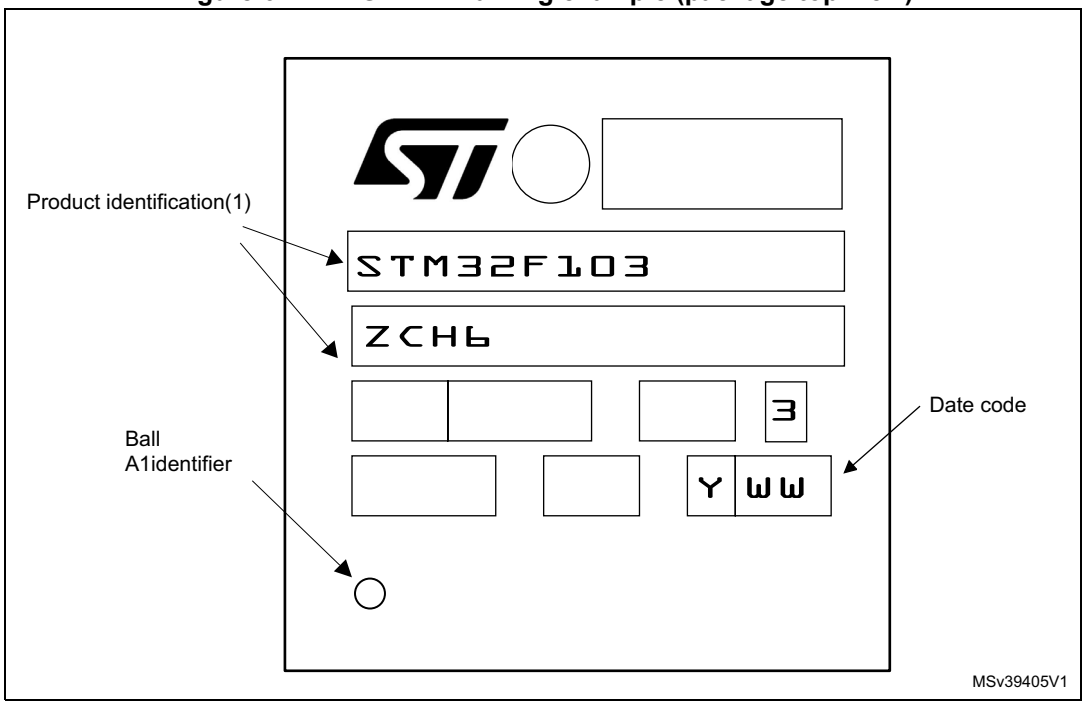
Table 66. LFBGA144 recommended PCB design rules (0.8 mm pitch BGA) (continued)

Dimension	Recommended values
Dsm	0.470 mm typ. (depends on the solder mask registration tolerance)
Stencil opening	0.400 mm
Stencil thickness	Between 0.100 mm to 0.125 mm
Pad trace width	0.120 mm
Ball Diameter	0.400 mm

Device marking for LFBGA144 package

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

Figure 64. LFBGA144 marking example (package top view)



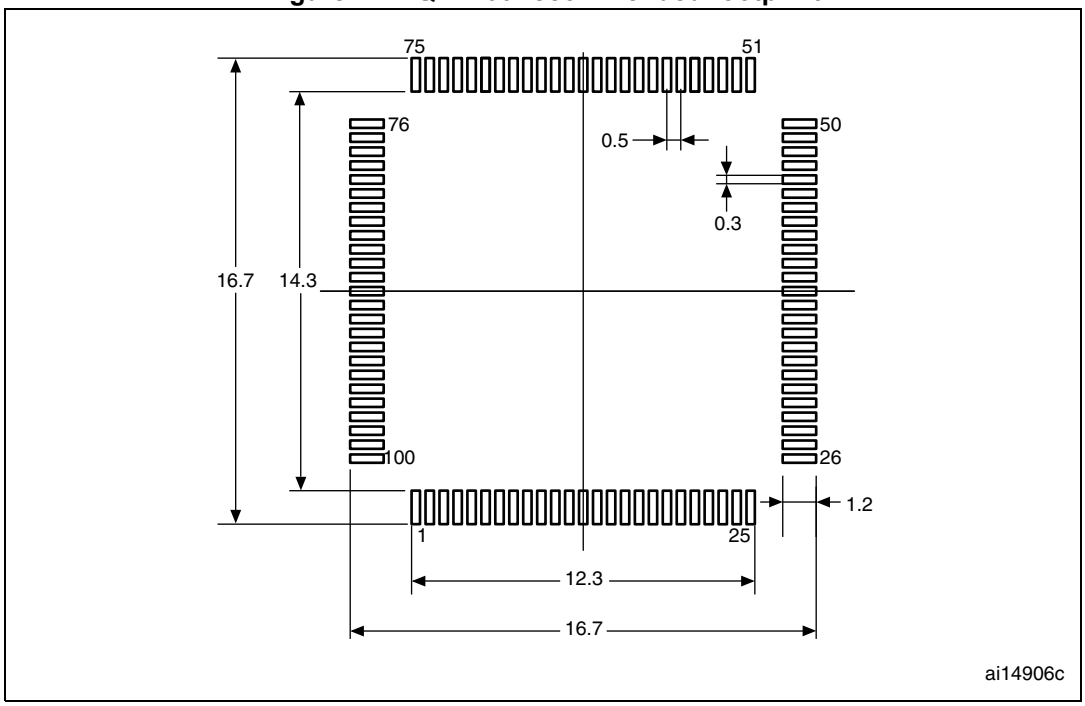
1. Parts marked as “ES”, “E” or accompanied by an Engineering Sample notification letter, are not yet qualified and therefore not yet ready to be used in production and any consequences deriving from such usage will not be at ST charge. In no event, ST will be liable for any customer usage of these engineering samples in production. ST Quality has to be contacted prior to any decision to use these Engineering samples to run qualification activity.

Table 72. LQPF100 – 14 x 14 mm 100-pin low-profile quad flat package mechanical data (continued)

Symbol	millimeters			inches[1]		
	Min	Typ	Max	Min	Typ	Max
e	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.08	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

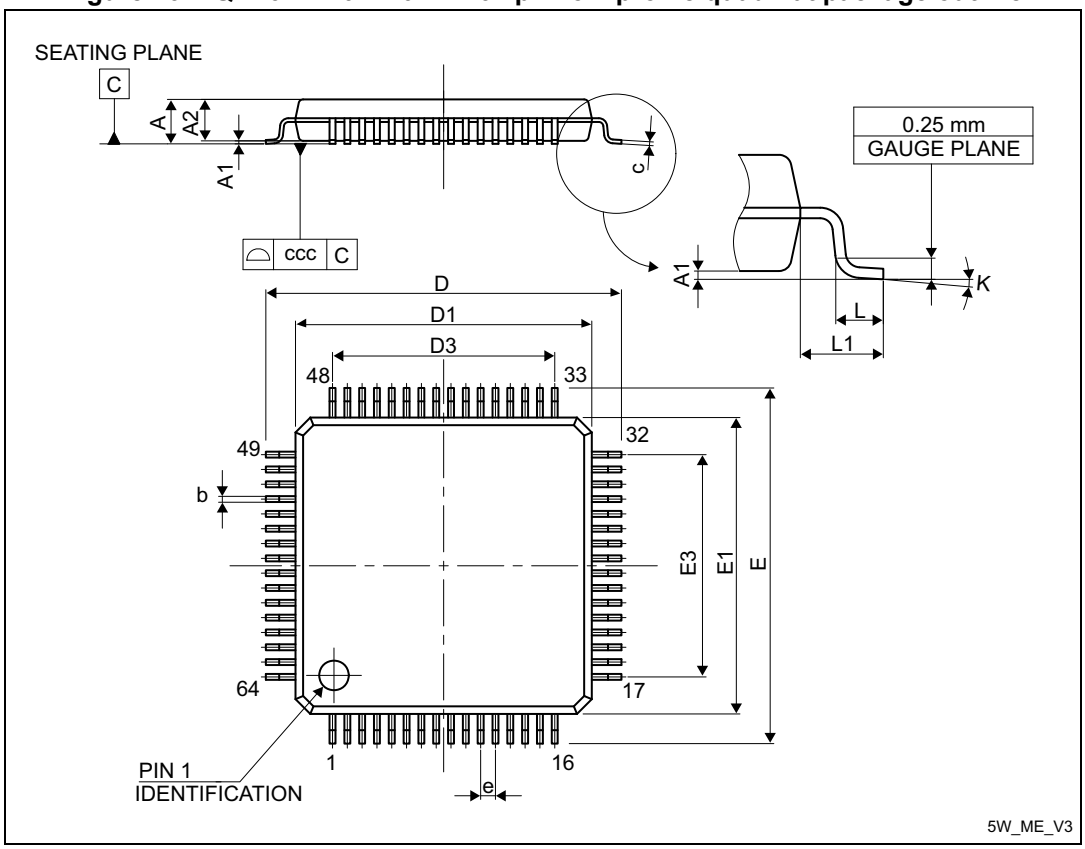
Figure 74. LQFP100 recommended footprint



1. Dimensions are in millimeters.

6.6 LQFP64 package information

Figure 76. LQFP64 – 10 x 10 mm 64 pin low-profile quad flat package outline



1. Drawing is not in scale.

Table 73. LQFP64 – 10 x 10 mm 64 pin low-profile quad flat package mechanical data

Symbol	millimeters			inches(1)		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	-	12.000	-	-	0.4724	-
D1	-	10.000	-	-	0.3937	-
D3	-	7.500	-	-	0.2953	-
E	-	12.000	-	-	0.4724	-
E1	-	10.000	-	-	0.3937	-