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# Intel - EPM7064BTC100-3N Datasheet



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# Understanding <u>Embedded - CPLDs (Complex</u> <u>Programmable Logic Devices)</u>

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixedfunction ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

## Applications of Embedded - CPLDs

# Details

Product Status	Obsolete
Programmable Type	In System Programmable
Delay Time tpd(1) Max	3.5 ns
Voltage Supply - Internal	2.375V ~ 2.625V
Number of Logic Elements/Blocks	4
Number of Macrocells	64
Number of Gates	1250
Number of I/O	68
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	100-TQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm7064btc100-3n

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MAX 7000B devices provide programmable speed/power optimization. Speed-critical portions of a design can run at high speed/full power, while the remaining portions run at reduced speed/low power. This speed/power optimization feature enables the designer to configure one or more macrocells to operate up to 50% lower power while adding only a nominal timing delay. MAX 7000B devices also provide an option that reduces the slew rate of the output buffers, minimizing noise transients when non-speed-critical signals are switching. The output drivers of all MAX 7000B devices can be set for 3.3 V, 2.5 V, or 1.8 V and all input pins are 3.3-V, 2.5-V, and 1.8-V tolerant, allowing MAX 7000B devices to be used in mixed-voltage systems.

MAX 7000B devices are supported by Altera development systems, which are integrated packages that offer schematic, text—including VHDL, Verilog HDL, and the Altera Hardware Description Language (AHDL)— and waveform design entry, compilation and logic synthesis, simulation and timing analysis, and device programming. Altera software provides EDIF 2 0 0 and 3 0 0, LPM, VHDL, Verilog HDL, and other interfaces for additional design entry and simulation support from other industry-standard PC- and UNIX-workstation-based EDA tools. Altera software runs on Windows-based PCs, as well as Sun SPARCstation, and HP 9000 Series 700/800 workstations.

For more information on development tools, see the *MAX+PLUS II Programmable Logic Development System & Software Data Sheet* and the *Quartus Programmable Logic Development System & Software Data Sheet*.

# Functional Description

The MAX 7000B architecture includes the following elements:

- LABs
- Macrocells
- Expander product terms (shareable and parallel)
- PIA
- I/O control blocks

The MAX 7000B architecture includes four dedicated inputs that can be used as general-purpose inputs or as high-speed, global control signals (clock, clear, and two output enable signals) for each macrocell and I/O pin. Figure 1 shows the architecture of MAX 7000B devices.

# Macrocells

The MAX 7000B macrocell can be individually configured for either sequential or combinatorial logic operation. The macrocell consists of three functional blocks: the logic array, the product-term select matrix, and the programmable register. Figure 2 shows the MAX 7000B macrocell.



Figure 2. MAX 7000B Macrocell

Combinatorial logic is implemented in the logic array, which provides five product terms per macrocell. The product-term select matrix allocates these product terms for use as either primary logic inputs (to the OR and XOR gates) to implement combinatorial functions, or as secondary inputs to the macrocell's register preset, clock, and clock enable control functions.

Two kinds of expander product terms ("expanders") are available to supplement macrocell logic resources:

- Shareable expanders, which are inverted product terms that are fed back into the logic array
- Parallel expanders, which are product terms borrowed from adjacent macrocells

The Altera development system automatically optimizes product-term allocation according to the logic requirements of the design.

For registered functions, each macrocell flipflop can be individually programmed to implement D, T, JK, or SR operation with programmable clock control. The flipflop can be bypassed for combinatorial operation. During design entry, the designer specifies the desired flipflop type; the MAX+PLUS II software then selects the most efficient flipflop operation for each registered function to optimize resource utilization.

Each programmable register can be clocked in three different modes:

- Global clock signal. This mode achieves the fastest clock-to-output performance.
- Global clock signal enabled by an active-high clock enable. A clock enable is generated by a product term. This mode provides an enable on each flipflop while still achieving the fast clock-to-output performance of the global clock.
- Array clock implemented with a product term. In this mode, the flipflop can be clocked by signals from buried macrocells or I/O pins.

Two global clock signals are available in MAX 7000B devices. As shown in Figure 1, these global clock signals can be the true or the complement of either of the global clock pins, GCLK1 or GCLK2.

Each register also supports asynchronous preset and clear functions. As shown in Figure 2, the product-term select matrix allocates product terms to control these operations. Although the product-term-driven preset and clear from the register are active high, active-low control can be obtained by inverting the signal within the logic array. In addition, each register clear function can be individually driven by the active-low dedicated global clear pin (GCLRn). Upon power-up, each register in a MAX 7000B device may be set to either a high or low state. This power-up state is specified at design entry.

All MAX 7000B I/O pins have a fast input path to a macrocell register. This dedicated path allows a signal to bypass the PIA and combinatorial logic and be clocked to an input D flipflop with an extremely fast input setup time. The input path from the I/O pin to the register has a programmable delay element that can be selected to either guarantee zero hold time or to get the fastest possible set-up time (as fast as 1.0 ns).

## Parallel Expanders

Parallel expanders are unused product terms that can be allocated to a neighboring macrocell to implement fast, complex logic functions. Parallel expanders allow up to 20 product terms to directly feed the macrocell OR logic, with five product terms provided by the macrocell and 15 parallel expanders provided by neighboring macrocells in the LAB.

The Altera Compiler can automatically allocate up to three sets of up to five parallel expanders to the macrocells that require additional product terms. Each set of five parallel expanders incurs a small, incremental timing delay ( $t_{PEXP}$ ). For example, if a macrocell requires 14 product terms, the Compiler uses the five dedicated product terms within the macrocell and allocates two sets of parallel expanders; the first set includes five product terms and the second set includes four product terms, increasing the total delay by  $2 \times t_{PEXP}$ .

Two groups of eight macrocells within each LAB (e.g., macrocells 1 through 8, and 9 through 16) form two chains to lend or borrow parallel expanders. A macrocell borrows parallel expanders from lower-numbered macrocells. For example, macrocell 8 can borrow parallel expanders from macrocell 7, from macrocells 7 and 6, or from macrocells 7, 6, and 5. Within each group of eight, the lowest-numbered macrocell can only lend parallel expanders and the highest-numbered macrocell can only borrow them. Figure 4 shows how parallel expanders can be borrowed from a neighboring macrocell.





#### Note:

(1) EPM7032B, EPM7064B, EPM7128B, and EPM7256B devices have six output enable signals. EPM7512B devices have ten output enable signals.

When the tri-state buffer control is connected to ground, the output is tri-stated (high impedance) and the I/O pin can be used as a dedicated input. When the tri-state buffer control is connected to  $V_{CC'}$ , the output is enabled.

The MAX 7000B architecture provides dual I/O feedback, in which macrocell and pin feedbacks are independent. When an I/O pin is configured as an input, the associated macrocell can be used for buried logic.

# **Programming Sequence**

During in-system programming, instructions, addresses, and data are shifted into the MAX 7000B device through the TDI input pin. Data is shifted out through the TDO output pin and compared against the expected data.

Programming a pattern into the device requires the following six ISP stages. A stand-alone verification of a programmed pattern involves only stages 1, 2, 5, and 6.

- 1. *Enter ISP*. The enter ISP stage ensures that the I/O pins transition smoothly from user mode to ISP mode. The enter ISP stage requires 1 ms.
- 2. *Check ID*. Before any program or verify process, the silicon ID is checked. The time required to read this silicon ID is relatively small compared to the overall programming time.
- 3. *Bulk Erase.* Erasing the device in-system involves shifting in the instructions to erase the device and applying one erase pulse of 100 ms.
- 4. *Program*. Programming the device in-system involves shifting in the address and data and then applying the programming pulse to program the EEPROM cells. This process is repeated for each EEPROM address.
- 5. *Verify.* Verifying an Altera device in-system involves shifting in addresses, applying the read pulse to verify the EEPROM cells, and shifting out the data for comparison. This process is repeated for each EEPROM address.
- 6. *Exit ISP*. An exit ISP stage ensures that the I/O pins transition smoothly from ISP mode to user mode. The exit ISP stage requires 1 ms.

# **Programming Times**

The time required to implement each of the six programming stages can be broken into the following two elements:

- A pulse time to erase, program, or read the EEPROM cells.
- A shifting time based on the test clock (TCK) frequency and the number of TCK cycles to shift instructions, address, and data into the device.

The programming times described in Tables 4 through 6 are associated with the worst-case method using the enhanced ISP algorithm.

Table 4. MAX 7000B t <sub>PULSE</sub> & Cycle <sub>TCK</sub> Values										
Device	Progra	imming	Stand-Alone Verification							
	<i>t<sub>PPULSE</sub></i> (s)	Cycle <sub>PTCK</sub>	t <sub>VPULSE</sub> (s)	Cycle <sub>VTCK</sub>						
EMP7032B	2.12	70,000	0.002	18,000						
EMP7064B	2.12	120,000	0.002	35,000						
EMP7128B	2.12	222,000	0.002	69,000						
EMP7256B	2.12	466,000	0.002	151,000						
EMP7512B	2.12	914,000	0.002	300,000						

Tables 5 and 6 show the in-system programming and stand alone verification times for several common test clock frequencies.

Table 5. MAX 7000B In-System Programming Times for Different Test Clock Frequencies									
Device		t <sub>TCK</sub>							
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EMP7032B	2.13	2.13	2.15	2.19	2.26	2.47	2.82	3.52	S
EMP7064B	2.13	2.14	2.18	2.24	2.36	2.72	3.32	4.52	S
EMP7128B	2.14	2.16	2.23	2.34	2.56	3.23	4.34	6.56	S
EMP7256B	2.17	2.21	2.35	2.58	3.05	4.45	6.78	11.44	S
EMP7512B	2.21	2.30	2.58	3.03	3.95	6.69	11.26	20.40	S

Table 1. MAX 7000B Stand-Alone Verification Times for Different Test Clock Frequencies										
Device		f <sub>TCK</sub> U								
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz		
EMP7032B	0.00	0.01	0.01	0.02	0.04	0.09	0.18	0.36	S	
EMP7064B	0.01	0.01	0.02	0.04	0.07	0.18	0.35	0.70	S	
EMP7128B	0.01	0.02	0.04	0.07	0.14	0.35	0.69	1.38	s	
EMP7256B	0.02	0.03	0.08	0.15	0.30	0.76	1.51	3.02	S	
EMP7512B	0.03	0.06	0.15	0.30	0.60	1.50	3.00	6.00	S	

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Programming with External Hardware	MAX 7000B devices can be programmed on Windows-based PCs with an Altera Logic Programmer card, the Master Programming Unit (MPU), and the appropriate device adapter. The MPU performs continuity checking to ensure adequate electrical contact between the adapter and the device.
	For more information, see the <i>Altera Programming Hardware Data Sheet</i> .
	The Altera software can use text- or waveform-format test vectors created with the Altera Text Editor or Waveform Editor to test the programmed device. For added design verification, designers can perform functional testing to compare the functional device behavior with the results of simulation.
	Data I/O, BP Microsystems, and other programming hardware manufacturers provide programming support for Altera devices. For more information, see <i>Programming Hardware Manufacturers</i> .
IEEE Std. 1149.1 (JTAG) Boundary-Scan Support	MAX 7000B devices include the JTAG boundary-scan test circuitry defined by IEEE Std. 1149.1. Table 6 describes the JTAG instructions supported by MAX 7000B devices. The pin-out tables starting on page 59 of this data sheet show the location of the JTAG control pins for each device. If the JTAG interface is not required, the JTAG pins are available as user I/O pins.

Table 6. MAX 7000B JTAG Instructions								
JTAG Instruction	Description							
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern output at the device pins.							
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.							
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the boundary-scan test data to pass synchronously through a selected device to adjacent devices during normal operation.							
CLAMP	Allows the values in the boundary-scan register to determine pin states while placing the 1-bit bypass register between the TDI and TDO pins.							
IDCODE	Selects the IDCODE register and places it between the TDI and TDO pins, allowing the IDCODE to be serially shifted out of TDO.							
USERCODE	Selects the 32-bit USERCODE register and places it between the TDI and TDO pins, allowing the USERCODE value to be shifted out of TDO.							
ISP Instructions	These instructions are used when programming MAX 7000B devices via the JTAG ports with the MasterBlaster or ByteBlasterMV download cable, or using a Jam File (.jam), Jam Byte-Code File (.jbc), or Serial Vector Format File (.svf) via an embedded processor or test equipment.							



Figure 8. MAX 7000B JTAG Waveforms

Table 9 shows the JTAG timing parameters and values for MAX 7000B devices.

<b>Table 9</b> Note (1)	. JTAG Timing Parameters & Values for MAX 70	00B Devi	ices	
Symbol	Parameter	Min	Max	Unit
t <sub>JCP</sub>	TCK clock period	100		ns
t <sub>JCH</sub>	TCK clock high time	50		ns
t <sub>JCL</sub>	TCK clock low time	50		ns
t <sub>JPSU</sub>	JTAG port setup time	20		ns
t <sub>JPH</sub>	JTAG port hold time	45		ns
t <sub>JPCO</sub>	JTAG port clock to output		25	ns
t <sub>JPZX</sub>	JTAG port high impedance to valid output		25	ns
t <sub>JPXZ</sub>	JTAG port valid output to high impedance		25	ns
t <sub>JSSU</sub>	Capture register setup time	20		ns
t <sub>JSH</sub>	Capture register hold time	45		ns
t <sub>JSCO</sub>	Update register clock to output		25	ns
t <sub>JSZX</sub>	Update register high impedance to valid output		25	ns
t <sub>JSXZ</sub>	Update register valid output to high impedance		25	ns

Note:

(1) Timing parameters in this table apply to all  $V_{CCIO}$  levels.

MAX 7000B devices contain two I/O banks. Both banks support all standards. Each I/O bank has its own VCCIO pins. A single device can support 1.8-V, 2.5-V, and 3.3-V interfaces; each bank can support a different standard independently. Within a bank, any one of the terminated standards can be supported.

Figure 9 shows the arrangement of the MAX 7000B I/O banks.

Programmable I/O Banks

Figure 9. MAX 7000B I/O Banks for Various Advanced I/O Standards

Table 11 shows which macrocells have pins in each I/O bank.

Table 11. Macrocell Pins Contained in Each I/O Bank								
Device	Bank 1	Bank 2						
EPM7032B	1-16	17-32						
EPM7064B	1-32	33-64						
EPM7128B	1-64	65-128						
EPM7256B	1-128, 177-181	129-176, 182-256						
EPM7512B	1-265	266-512						

Each MAX 7000B device has two VREF pins. Each can be set to a separate  $V_{REF}$  level. Any I/O pin that uses one of the voltage-referenced standards (GTL+, SSTL-2, or SSTL-3) may use either of the two VREF pins. If these pins are not required as VREF pins, they may be individually programmed to function as user I/O pins.

Two inverters implement the bus-hold circuitry in a loop that weakly drives back to the I/O pin in user mode.

Figure 10 shows a block diagram of the bus-hold circuit.

### Figure 10. Bus-Hold Circuit



# PCI Compatibility

MAX 7000B devices are compatible with PCI applications as well as all 3.3-V electrical specifications in the *PCI Local Bus Specification Revision 2.2* except for the clamp diode. While having multiple clamp diodes on a signal trace may be redundant, designers can add an external clamp diode to meet the specification. Table 13 shows the MAX 7000B device speed grades that meet the PCI timing specifications.

Table 13. MAX 7000B Device Speed Grades that Meet PCI Timing   Specifications								
Device	Specification							
	33-MHz PCI	66-MHz PCI						
EPM7032B	All speed grades	-3						
EPM7064B	All speed grades	-3						
EPM7128B	All speed grades	-4						
EPM7256B	All speed grades	-5 (1)						
EPM7512B	All speed grades	-5 (1)						

#### Note:

(1) The EPM7256B and EPM7512B devices in a -5 speed grade meet all PCI timing specifications for 66-MHz operation except the Input Setup Time to CLK—Bused Signal parameter. However, these devices are within 1 ns of that parameter. EPM7256B and EPM7512B devices meet all other 66-MHz PCI timing specifications.

Table 18. EPM7032B External Timing Parameters Notes (1)									
Symbol	Parameter	Conditions		Speed Grade					
			-3	.5	-5	.0	-7	.5	
			Min	Max	Min	Max	Min	Max	
t <sub>PD1</sub>	Input to non-registered output	C1 = 35 pF (2)		3.5		5.0		7.5	ns
t <sub>PD2</sub>	I/O input to non-registered output	C1 = 35 pF (2)		3.5		5.0		7.5	ns
t <sub>SU</sub>	Global clock setup time	(2)	2.1		3.0		4.5		ns
t <sub>H</sub>	Global clock hold time	(2)	0.0		0.0		0.0		ns
t <sub>FSU</sub>	Global clock setup time of fast input		1.0		1.0		1.5		ns
t <sub>FH</sub>	Global clock hold time of fast input		1.0		1.0		1.0		ns
t <sub>fzhsu</sub>	Global clock setup time of fast input with zero hold time		2.0		2.5		3.0		ns
t <sub>FZHH</sub>	Global clock hold time of fast input with zero hold time		0.0		0.0		0.0		ns
t <sub>CO1</sub>	Global clock to output delay	C1 = 35 pF	1.0	2.4	1.0	3.4	1.0	5.0	ns
t <sub>CH</sub>	Global clock high time		1.5		2.0		3.0		ns
t <sub>CL</sub>	Global clock low time		1.5		2.0		3.0		ns
t <sub>ASU</sub>	Array clock setup time	(2)	0.9		1.3		1.9		ns
t <sub>AH</sub>	Array clock hold time	(2)	0.2		0.3		0.6		ns
t <sub>ACO1</sub>	Array clock to output delay	C1 = 35 pF (2)	1.0	3.6	1.0	5.1	1.0	7.6	ns
t <sub>ACH</sub>	Array clock high time		1.5		2.0		3.0		ns
t <sub>ACL</sub>	Array clock low time		1.5		2.0		3.0		ns
t <sub>CPPW</sub>	Minimum pulse width for clear and preset		1.5		2.0		3.0		ns
<sup>t</sup> CNT	Minimum global clock period	(2)		3.3		4.7		7.0	ns
fcnt	Maximum internal global clock frequency	(2), (3)	303.0		212.8		142.9		MHz
t <sub>acnt</sub>	Minimum array clock period	(2)		3.3		4.7		7.0	ns
f <sub>ACNT</sub>	Maximum internal array clock frequency	(2), (3)	303.0		212.8		142.9		MHz

# Tables 18 through 32 show MAX 7000B device timing parameters.

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Symbol	Parameter	Conditions			Speed	Grade			Unit
			-	3	-	5	-	7	1
			Min	Max	Min	Max	Min	Max	
t <sub>PD1</sub>	Input to non-registered output	C1 = 35 pF (2)		3.5		5.0		7.5	ns
t <sub>PD2</sub>	I/O input to non-registered output	C1 = 35 pF (2)		3.5		5.0		7.5	ns
t <sub>SU</sub>	Global clock setup time	(2)	2.1		3.0		4.5		ns
t <sub>H</sub>	Global clock hold time	(2)	0.0		0.0		0.0		ns
t <sub>FSU</sub>	Global clock setup time of fast input		1.0		1.0		1.5		ns
t <sub>FH</sub>	Global clock hold time of fast input		1.0		1.0		1.0		ns
t <sub>FZHSU</sub>	Global clock setup time of fast input with zero hold time		2.0		2.5		3.0		ns
t <sub>FZHH</sub>	Global clock hold time of fast input with zero hold time		0.0		0.0		0.0		ns
t <sub>CO1</sub>	Global clock to output delay	C1 = 35 pF	1.0	2.4	1.0	3.4	1.0	5.0	ns
t <sub>CH</sub>	Global clock high time		1.5		2.0		3.0		ns
t <sub>CL</sub>	Global clock low time		1.5		2.0		3.0		ns
t <sub>ASU</sub>	Array clock setup time	(2)	0.9		1.3		1.9		ns
t <sub>AH</sub>	Array clock hold time	(2)	0.2		0.3		0.6		ns
t <sub>ACO1</sub>	Array clock to output delay	C1 = 35 pF (2)	1.0	3.6	1.0	5.1	1.0	7.6	ns
t <sub>ACH</sub>	Array clock high time		1.5		2.0		3.0		ns
t <sub>ACL</sub>	Array clock low time		1.5		2.0		3.0		ns
t <sub>CPPW</sub>	Minimum pulse width for clear and preset		1.5		2.0		3.0		ns
t <sub>CNT</sub>	Minimum global clock period	(2)		3.3		4.7		7.0	ns
f <sub>CNT</sub>	Maximum internal global clock frequency	(2), (3)	303.0		212.8		142.9		MHz
t <sub>acnt</sub>	Minimum array clock period	(2)		3.3		4.7		7.0	ns
facnt	Maximum internal array clock frequency	(2), (3)	303.0		212.8		142.9		MHz

Table 23. EPM7064B Selectable I/O Standard Timing Adder Delays (Part 2 of 2) Note (1)									
I/O Standard	Parameter	Speed Grade						Unit	
		-3		-5		-7			
		Min	Max	Min	Max	Min	Max		
PCI	Input to PIA		0.0		0.0		0.0	ns	
	Input to global clock and clear		0.0		0.0		0.0	ns	
	Input to fast input register		0.0		0.0		0.0	ns	
	All outputs		0.0		0.0		0.0	ns	

#### Notes to tables:

(1) These values are specified under the Recommended Operating Conditions in Table 15 on page 29. See Figure 14 for more information on switching waveforms.

(2) These values are specified for a PIA fan-out of all LABs.

(3) Measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.

(4) The  $t_{LPA}$  parameter must be added to the  $t_{LAD}$ ,  $t_{LAC}$ ,  $t_{IC}$ ,  $t_{ACL}$ ,  $t_{CPPW}$ ,  $t_{EN}$ , and  $t_{SEXP}$  parameters for macrocells running in low-power mode.

I/O Standard	Parameter	Speed Grade						Unit
		-4		-7		-10		1
		Min	Max	Min	Max	Min	Max	
3.3 V TTL/CMOS	Input to PIA		0.0		0.0		0.0	ns
	Input to global clock and clear		0.0		0.0		0.0	ns
	Input to fast input register		0.0		0.0		0.0	ns
	All outputs		0.0		0.0		0.0	ns
2.5 V TTL/CMOS	Input to PIA		0.3		0.6		0.8	ns
	Input to global clock and clear		0.3		0.6		0.8	ns
	Input to fast input register		0.2		0.4		0.5	ns
	All outputs		0.2		0.4		0.5	ns
1.8 V TTL/CMOS	Input to PIA		0.5		0.9		1.3	ns
	Input to global clock and clear		0.5		0.9		1.3	ns
	Input to fast input register		0.4		0.8		1.0	ns
	All outputs		1.2		2.3		3.0	ns
SSTL-2 Class I	Input to PIA		1.4		2.6		3.5	ns
	Input to global clock and clear		1.2		2.3		3.0	ns
	Input to fast input register		1.0		1.9		2.5	ns
	All outputs		0.0		0.0		0.0	ns
SSTL-2 Class II	Input to PIA		1.4		2.6		3.5	ns
	Input to global clock and clear		1.2		2.3		3.0	ns
	Input to fast input register		1.0		1.9		2.5	ns
	All outputs		-0.1		-0.2		-0.3	ns
SSTL-3 Class I	Input to PIA		1.3		2.4		3.3	ns
	Input to global clock and clear		1.0		1.9		2.5	ns
	Input to fast input register		0.9		1.7		2.3	ns
	All outputs		0.0		0.0		0.0	ns
SSTL-3 Class II	Input to PIA		1.3		2.4		3.3	ns
	Input to global clock and clear		1.0		1.9		2.5	ns
	Input to fast input register		0.9		1.7		2.3	ns
	All outputs		0.0		0.0		0.0	ns
GTL+	Input to PIA		1.7		3.2		4.3	ns
	Input to global clock and clear		1.7		3.2		4.3	ns
	Input to fast input register		1.6		3.0		4.0	ns
	All outputs		0.0		0.0		0.0	ns

Table 28. EPM7256B Internal Timing Parameters Note (1)									
Symbol	Parameter	Conditions	Speed Grade						Unit
			-5		-7		-10		
			Min	Max	Min	Max	Min	Max	
t <sub>IN</sub>	Input pad and buffer delay			0.4		0.6		0.8	ns
t <sub>IO</sub>	I/O input pad and buffer delay			0.4		0.6		0.8	ns
t <sub>FIN</sub>	Fast input delay			1.5		2.5		3.1	ns
t <sub>FIND</sub>	Programmable delay adder for fast input			1.5		1.5		1.5	ns
t <sub>SEXP</sub>	Shared expander delay			1.5		2.3		3.0	ns
t <sub>PEXP</sub>	Parallel expander delay			0.4		0.6		0.8	ns
t <sub>LAD</sub>	Logic array delay			1.7		2.5		3.3	ns
t <sub>LAC</sub>	Logic control array delay			1.5		2.2		2.9	ns
t <sub>IOE</sub>	Internal output enable delay			0.1		0.2		0.3	ns
t <sub>OD1</sub>	Output buffer and pad delay slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		0.9		1.4		1.9	ns
t <sub>OD3</sub>	Output buffer and pad delay slow slew rate = on $V_{CCIO} = 2.5 V \text{ or } 3.3 V$	C1 = 35 pF		5.9		6.4		6.9	ns
t <sub>ZX1</sub>	Output buffer enable delay slow slew rate = off $V_{CCIO} = 3.3 V$	C1 = 35 pF		2.2		3.3		4.5	ns
t <sub>ZX3</sub>	Output buffer enable delay slow slew rate = on $V_{CCIO} = 2.5 V \text{ or } 3.3 V$	C1 = 35 pF		7.2		8.3		9.5	ns
t <sub>XZ</sub>	Output buffer disable delay	C1 = 5 pF		2.2		3.3		4.5	ns
t <sub>SU</sub>	Register setup time		1.2		1.8		2.5		ns
t <sub>H</sub>	Register hold time		0.6		1.0		1.3		ns
t <sub>FSU</sub>	Register setup time of fast input		0.8		1.1		1.1		ns
t <sub>FH</sub>	Register hold time of fast input		1.2		1.4		1.4		ns
t <sub>RD</sub>	Register delay			0.7		1.0		1.3	ns
t <sub>COMB</sub>	Combinatorial delay			0.3		0.4		0.5	ns
t <sub>IC</sub>	Array clock delay			1.5		2.3		3.0	ns
t <sub>EN</sub>	Register enable time			1.5		2.2		2.9	ns
t <sub>GLOB</sub>	Global control delay			1.3		2.1		2.7	ns
t <sub>PRE</sub>	Register preset time			1.0		1.6		2.1	ns
t <sub>CLR</sub>	Register clear time			1.0		1.6		2.1	ns
t <sub>PIA</sub>	PIA delay	(2)		1.7		2.6		3.3	ns
t <sub>LPA</sub>	Low-power adder	(4)		2.0		3.0		4.0	ns



Figure 17. I<sub>CC</sub> vs. Frequency for EPM7128B Devices







Package outline not drawn to scale.



Figure 26. 169-Pin Ultra FineLine BGA Pin-Out Diagram

Package outline not drawn to scale.



A1 Ball



Package outline not drawn to scale.



### Figure 28. 256-Pin BGA Package Pin-Out Diagram

Package outline not drawn to scale.

