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Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details	
Product Status	Active
Programmable Type	EE PLD
Delay Time tpd(1) Max	7.5 ns
Voltage Supply - Internal	2.375V ~ 2.625V
Number of Logic Elements/Blocks	4
Number of Macrocells	64
Number of Gates	1250
Number of I/O	68
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	100-TQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=epm7064btc100-7

- Additional design entry and simulation support provided by EDIF 2 0 0 and 3 0 0 netlist files, library of parameterized modules (LPMs), Verilog HDL, VHDL, and other interfaces to popular EDA tools from manufacturers such as Cadence, Exemplar Logic, Mentor Graphics, OrCAD, Synopsys, Synplicity, and VeriBest
- Programming support with Altera's Master Programming Unit (MPU), MasterBlaster™ serial/universal serial bus (USB) communications cable, and ByteBlasterMV™ parallel port download cable, as well as programming hardware from third-party manufacturers and any Jam™ STAPL File (**.jam**), Jam Byte-Code File (**.jbc**), or Serial Vector Format File (**.svf**)-capable in-circuit tester

General Description

MAX 7000B devices are high-density, high-performance devices based on Altera's second-generation MAX architecture. Fabricated with advanced CMOS technology, the EEPROM-based MAX 7000B devices operate with a 2.5-V supply voltage and provide 600 to 10,000 usable gates, ISP, pin-to-pin delays as fast as 3.5 ns, and counter speeds up to 303.0 MHz. See Table 2.

Table 2. MAX 7000B Speed Grades *Note (1)*

Device	Speed Grade				
	-3	-4	-5	-7	-10
EPM7032B	✓		✓	✓	
EPM7064B	✓		✓	✓	
EPM7128B		✓		✓	✓
EPM7256B			✓	✓	✓
EPM7512B			✓	✓	✓

Notes:
(1) Contact Altera Marketing for up-to-date information on available device speed grades.

The MAX 7000B architecture supports 100% TTL emulation and high-density integration of SSI, MSI, and LSI logic functions. It easily integrates multiple devices ranging from PALs, GALs, and 22V10s to MACH and pLSI devices. MAX 7000B devices are available in a wide range of packages, including PLCC, BGA, FineLine BGA, 0.8-mm Ultra FineLine BGA, PQFP, TQFP, and TQFP packages. See Table 3.

The Altera development system automatically optimizes product-term allocation according to the logic requirements of the design.

For registered functions, each macrocell flipflop can be individually programmed to implement D, T, JK, or SR operation with programmable clock control. The flipflop can be bypassed for combinatorial operation. During design entry, the designer specifies the desired flipflop type; the MAX+PLUS II software then selects the most efficient flipflop operation for each registered function to optimize resource utilization.

Each programmable register can be clocked in three different modes:

- Global clock signal. This mode achieves the fastest clock-to-output performance.
- Global clock signal enabled by an active-high clock enable. A clock enable is generated by a product term. This mode provides an enable on each flipflop while still achieving the fast clock-to-output performance of the global clock.
- Array clock implemented with a product term. In this mode, the flipflop can be clocked by signals from buried macrocells or I/O pins.

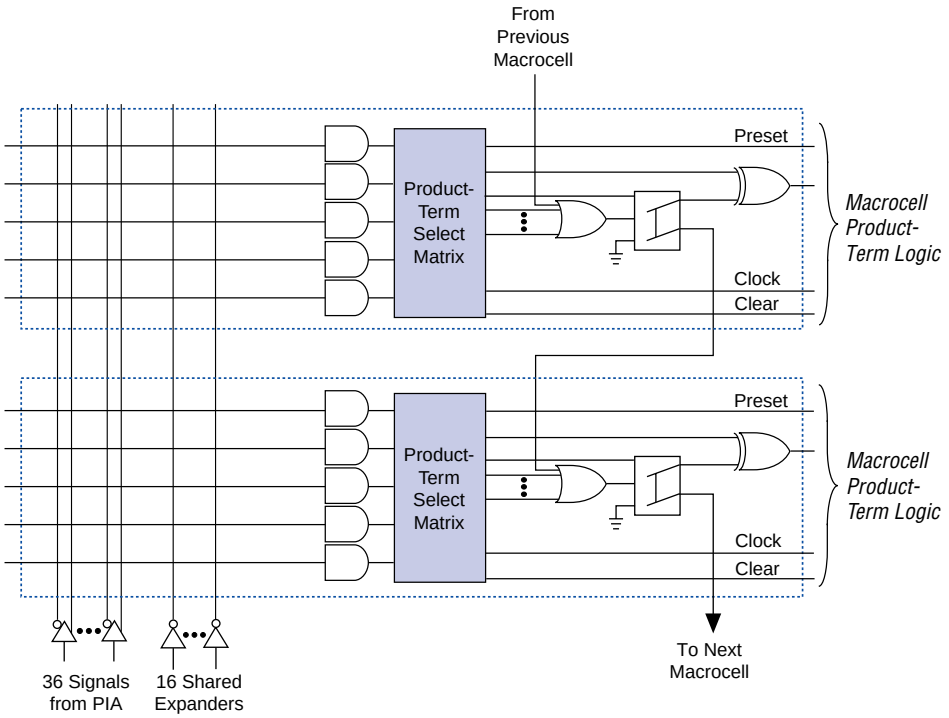
Two global clock signals are available in MAX 7000B devices. As shown in Figure 1, these global clock signals can be the true or the complement of either of the global clock pins, `GCLK1` or `GCLK2`.

Each register also supports asynchronous preset and clear functions. As shown in Figure 2, the product-term select matrix allocates product terms to control these operations. Although the product-term-driven preset and clear from the register are active high, active-low control can be obtained by inverting the signal within the logic array. In addition, each register clear function can be individually driven by the active-low dedicated global clear pin (`GCLRn`). Upon power-up, each register in a MAX 7000B device may be set to either a high or low state. This power-up state is specified at design entry.

All MAX 7000B I/O pins have a fast input path to a macrocell register. This dedicated path allows a signal to bypass the PIA and combinatorial logic and be clocked to an input D flipflop with an extremely fast input setup time. The input path from the I/O pin to the register has a programmable delay element that can be selected to either guarantee zero hold time or to get the fastest possible set-up time (as fast as 1.0 ns).

Figure 4. MAX 7000B Parallel Expanders

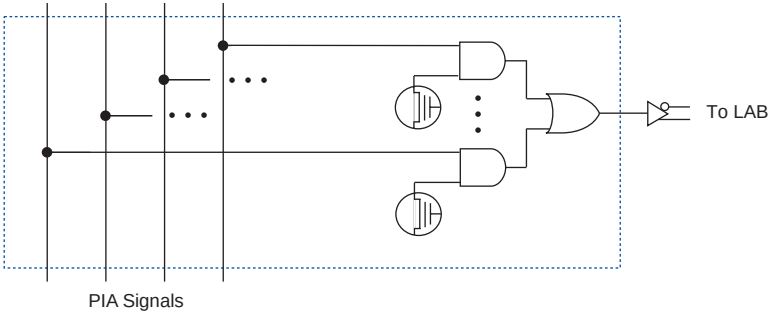
Unused product terms in a macrocell can be allocated to a neighboring macrocell.



Programmable Interconnect Array

Logic is routed between LABs on the PIA. This global bus is a programmable path that connects any signal source to any destination on the device. All MAX 7000B dedicated inputs, I/O pins, and macrocell outputs feed the PIA, which makes the signals available throughout the entire device. Only the signals required by each LAB are actually routed from the PIA into the LAB. Figure 5 shows how the PIA signals are routed into the LAB. An EEPROM cell controls one input to a two-input AND gate, which selects a PIA signal to drive into the LAB.

Figure 5. MAX 7000B PIA Routing



While the routing delays of channel-based routing schemes in masked or field-programmable gate arrays (FPGAs) are cumulative, variable, and path-dependent, the MAX 7000B PIA has a predictable delay. The PIA makes a design's timing performance easy to predict.

I/O Control Blocks

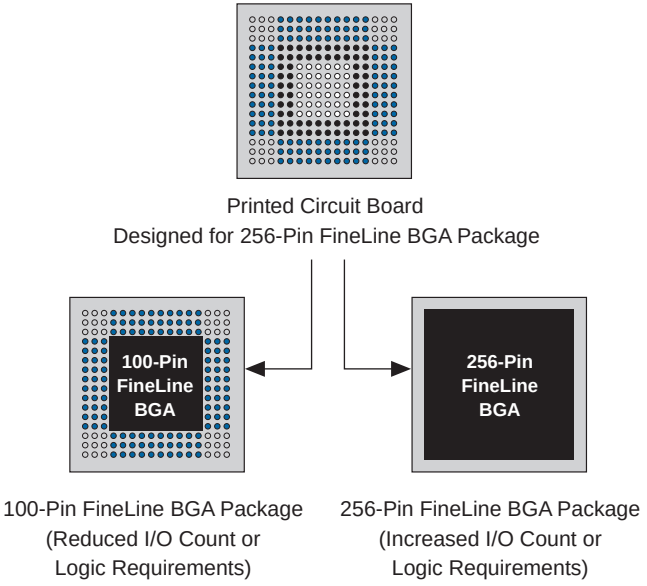
The I/O control block allows each I/O pin to be individually configured for input, output, or bidirectional operation. All I/O pins have a tri-state buffer that is individually controlled by one of the global output enable signals or directly connected to ground or V_{CC}. Figure 6 shows the I/O control block for MAX 7000B devices. The I/O control block has six or ten global output enable signals that are driven by the true or complement of two output enable signals, a subset of the I/O pins, or a subset of the I/O macrocells.

SameFrame Pin-Outs

MAX 7000B devices support the SameFrame pin-out feature for FineLine BGA and 0.8-mm Ultra FineLine BGA packages. The SameFrame pin-out feature is the arrangement of balls on FineLine BGA and 0.8-mm Ultra FineLine BGA packages such that the lower-ball-count packages form a subset of the higher-ball-count packages. SameFrame pin-outs provide the flexibility to migrate not only from device to device within the same package, but also from one package to another. FineLine BGA packages are compatible with other FineLine BGA packages, and 0.8-mm Ultra FineLine BGA packages are compatible with other 0.8-mm Ultra FineLine BGA packages. A given printed circuit board (PCB) layout can support multiple device density/package combinations. For example, a single board layout can support a range of devices from an EPM7064B device in a 100-pin FineLine BGA package to an EPM7512B device in a 256-pin FineLine BGA package.

The Altera software provides support to design PCBs with SameFrame pin-out devices. Devices can be defined for present and future use. The Altera software generates pin-outs describing how to layout a board to take advantage of this migration (see Figure 7).

Figure 7. SameFrame Pin-Out Example



By combining the pulse and shift times for each of the programming stages, the program or verify time can be derived as a function of the TCK frequency, the number of devices, and specific target device(s). Because different ISP-capable devices have a different number of EEPROM cells, both the total fixed and total variable times are unique for a single device.

Programming a Single MAX 7000B Device

The time required to program a single MAX 7000B device in-system can be calculated from the following formula:

$$t_{PROG} = t_{PPULSE} + \frac{Cycle_{PTCK}}{f_{TCK}}$$

where: t_{PROG} = Programming time
t_{PPULSE} = Sum of the fixed times to erase, program, and verify the EEPROM cells
$Cycle_{PTCK}$ = Number of TCK cycles to program a device
f_{TCK} = TCK frequency

The ISP times for a stand-alone verification of a single MAX 7000B device can be calculated from the following formula:

$$t_{VER} = t_{VPULSE} + \frac{Cycle_{VTCK}}{f_{TCK}}$$

where: t_{VER} = Verify time
t_{VPULSE} = Sum of the fixed times to verify the EEPROM cells
$Cycle_{VTCK}$ = Number of TCK cycles to verify a device

The programming times described in Tables 4 through 6 are associated with the worst-case method using the enhanced ISP algorithm.

Table 4. MAX 7000B t_{PULSE} & $Cycle_{TCK}$ Values

Device	Programming		Stand-Alone Verification	
	t_{PPULSE} (s)	$Cycle_{PTCK}$	t_{VPULSE} (s)	$Cycle_{VTCK}$
EMP7032B	2.12	70,000	0.002	18,000
EMP7064B	2.12	120,000	0.002	35,000
EMP7128B	2.12	222,000	0.002	69,000
EMP7256B	2.12	466,000	0.002	151,000
EMP7512B	2.12	914,000	0.002	300,000

Tables 5 and 6 show the in-system programming and stand alone verification times for several common test clock frequencies.

Table 5. MAX 7000B In-System Programming Times for Different Test Clock Frequencies

Device	f_{TCK}								Units
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EMP7032B	2.13	2.13	2.15	2.19	2.26	2.47	2.82	3.52	s
EMP7064B	2.13	2.14	2.18	2.24	2.36	2.72	3.32	4.52	s
EMP7128B	2.14	2.16	2.23	2.34	2.56	3.23	4.34	6.56	s
EMP7256B	2.17	2.21	2.35	2.58	3.05	4.45	6.78	11.44	s
EMP7512B	2.21	2.30	2.58	3.03	3.95	6.69	11.26	20.40	s

Table 1. MAX 7000B Stand-Alone Verification Times for Different Test Clock Frequencies

Device	f_{TCK}								Units
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EMP7032B	0.00	0.01	0.01	0.02	0.04	0.09	0.18	0.36	s
EMP7064B	0.01	0.01	0.02	0.04	0.07	0.18	0.35	0.70	s
EMP7128B	0.01	0.02	0.04	0.07	0.14	0.35	0.69	1.38	s
EMP7256B	0.02	0.03	0.08	0.15	0.30	0.76	1.51	3.02	s
EMP7512B	0.03	0.06	0.15	0.30	0.60	1.50	3.00	6.00	s

MAX 7000B devices contain two I/O banks. Both banks support all standards. Each I/O bank has its own VCCIO pins. A single device can support 1.8-V, 2.5-V, and 3.3-V interfaces; each bank can support a different standard independently. Within a bank, any one of the terminated standards can be supported.

Figure 9 shows the arrangement of the MAX 7000B I/O banks.

Figure 9. MAX 7000B I/O Banks for Various Advanced I/O Standards

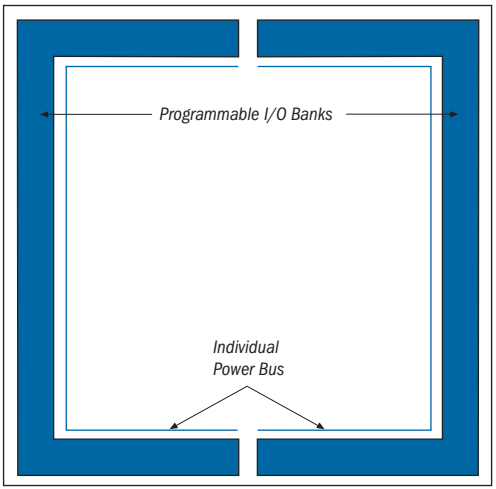


Table 11 shows which macrocells have pins in each I/O bank.

Table 11. Macrocell Pins Contained in Each I/O Bank

Device	Bank 1	Bank 2
EPM7032B	1-16	17-32
EPM7064B	1-32	33-64
EPM7128B	1-64	65-128
EPM7256B	1-128, 177-181	129-176, 182-256
EPM7512B	1-265	266-512

Each MAX 7000B device has two VREF pins. Each can be set to a separate V_{REF} level. Any I/O pin that uses one of the voltage-referenced standards (GTL+, SSTL-2, or SSTL-3) may use either of the two VREF pins. If these pins are not required as VREF pins, they may be individually programmed to function as user I/O pins.

Programmable Pull-Up Resistor

Each MAX 7000B device I/O pin provides an optional programmable pull-up resistor during user mode. When this feature is enabled for an I/O pin, the pull-up resistor (typically 50 k¾) weakly holds the output to V_{CCIO} level.

Bus Hold

Each MAX 7000B device I/O pin provides an optional bus-hold feature. When this feature is enabled for an I/O pin, the bus-hold circuitry weakly holds the signal at its last driven state. By holding the last driven state of the pin until the next input signals is present, the bus-hold feature can eliminate the need to add external pull-up or pull-down resistors to hold a signal level when the bus is tri-stated. The bus-hold circuitry also pulls undriven pins away from the input threshold voltage where noise can cause unintended high-frequency switching. This feature can be selected individually for each I/O pin. The bus-hold output will drive no higher than V_{CCIO} to prevent overdriving signals. The propagation delays through the input and output buffers in MAX 7000B devices are not affected by whether the bus-hold feature is enabled or disabled.

The bus-hold circuitry weakly pulls the signal level to the last driven state through a resistor with a nominal resistance (R_{BH}) of approximately 8.5 k¾. Table 12 gives specific sustaining current that will be driven through this resistor and overdrive current that will identify the next driven input level. This information is provided for each VCCIO voltage level.

Table 12. Bus Hold Parameters

Parameter	Conditions	VCCIO Level						Units
		1.8 V		2.5 V		3.3 V		
		Min	Max	Min	Max	Min	Max	
Low sustaining current	$V_{IN} > V_{IL}$ (max)	30		50		70		µA
High sustaining current	$V_{IN} < V_{IH}$ (min)	–30		–50		–70		µA
Low overdrive current	$0 V < V_{IN} < V_{CCIO}$		200		300		500	µA
High overdrive current	$0 V < V_{IN} < V_{CCIO}$		–295		–435		–680	µA

The bus-hold circuitry is active only during user operation. At power-up, the bus-hold circuit initializes its initial hold value as V_{CC} approaches the recommended operation conditions. When transitioning from ISP to User Mode with bus hold enabled, the bus-hold circuit captures the value present on the pin at the end of programming.

Power Sequencing & Hot-Socketing

Because MAX 7000B devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The V_{CCIO} and V_{CCINT} power planes can be powered in any order.

Signals can be driven into MAX 7000B devices before and during power-up (and power-down) without damaging the device. Additionally, MAX 7000B devices do not drive out during power-up. Once operating conditions are reached, MAX 7000B devices operate as specified by the user.

MAX 7000B device I/O pins will not source or sink more than 300 µA of DC current during power-up. All pins can be driven up to 4.1 V during hot-socketing.

Design Security

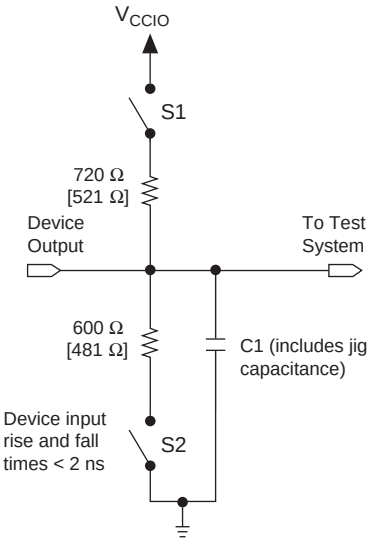
All MAX 7000B devices contain a programmable security bit that controls access to the data programmed into the device. When this bit is programmed, a design implemented in the device cannot be copied or retrieved. This feature provides a high level of design security, because programmed data within EEPROM cells is invisible. The security bit that controls this function, as well as all other programmed data, is reset only when the device is reprogrammed.

Generic Testing

MAX 7000B devices are fully functionally tested. Complete testing of each programmable EEPROM bit and all internal logic elements ensures 100% programming yield. AC test measurements are taken under conditions equivalent to those shown in Figure 11. Test patterns can be used and then erased during early stages of the production flow.

Figure 11. MAX 7000B AC Test Conditions

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers in brackets are for 2.5-V outputs. Numbers without brackets are for 3.3-V outputs. Switches S1 and S2 are open for all tests except output disable timing parameters.



Operating Conditions

Tables 14 through 17 provide information on absolute maximum ratings, recommended operating conditions, operating conditions, and capacitance for MAX 7000B devices.

Table 14. MAX 7000B Device Absolute Maximum Ratings *Note (1)*

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CCINT}	Supply voltage		–0.5	3.6	V
V_{CCIO}	Supply voltage		–0.5	3.6	V
V_I	DC input voltage	*(2)*	–2.0	4.6	V
I_{OUT}	DC output current, per pin		–33	50	mA
T_{STG}	Storage temperature	No bias	–65	150	° C
T_A	Ambient temperature	Under bias	–65	135	° C
T_J	Junction temperature	Under bias	–65	135	° C

Table 16. MAX 7000B Device DC Operating Conditions *Note (4)*

Symbol	Parameter	Conditions	Min	Max	Unit
V_{IH}	High-level input voltage for 3.3-V TTL/CMOS		2.0	3.9	V
	High-level input voltage for 2.5-V TTL/CMOS		1.7	3.9	V
	High-level input voltage for 1.8-V TTL/CMOS		0.65 × V_{CCIO}	3.9	V
V_{IL}	Low-level input voltage for 3.3-V TTL/CMOS and PCI compliance		–0.5	0.8	V
	Low-level input voltage for 2.5-V TTL/CMOS		–0.5	0.7	V
	Low-level input voltage for 1.8-V TTL/CMOS		–0.5	0.35 × V_{CCIO}	
V_{OH}	3.3-V high-level TTL output voltage	I_{OH} = –8 mA DC, V_{CCIO} = 3.00 V *(5)*	2.4		V
	3.3-V high-level CMOS output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 3.00 V *(5)*	V_{CCIO} – 0.2		V
	2.5-V high-level output voltage	I_{OH} = –100 µA DC, V_{CCIO} = 2.30 V *(5)*	2.1		V
		I_{OH} = –1 mA DC, V_{CCIO} = 2.30 V *(5)*	2.0		V
		I_{OH} = –2 mA DC, V_{CCIO} = 2.30 V *(5)*	1.7		V
	1.8-V high-level output voltage	I_{OH} = –2 mA DC, V_{CCIO} =1.65 V *(5)*	1.2		V
V_{OL}	3.3-V low-level TTL output voltage	I_{OL} = 8 mA DC, V_{CCIO} = 3.00 V *(6)*		0.4	V
	3.3-V low-level CMOS output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 3.00 V *(6)*		0.2	V
	2.5-V low-level output voltage	I_{OL} = 100 µA DC, V_{CCIO} = 2.30 V *(6)*		0.2	V
		I_{OL} = 1 mA DC, V_{CCIO} = 2.30 V *(6)*		0.4	V
		I_{OL} = 2 mA DC, V_{CCIO} = 2.30 V *(6)*		0.7	V
	1.8-V low-level output voltage	I_{OL} = 2 mA DC, V_{CCIO} = 1.7 V *(6)*		0.4	V
I_I	Input leakage current	V_I = –0.5 to 3.9 V *(7)*	–10	10	µA
I_{OZ}	Tri-state output off-state current	V_I = –0.5 to 3.9 V *(7)*	–10	10	µA
R_{ISP}	Value of I/O pin pull-up resistor during in-system programming or during power up	V_{CCIO} = 1.7 to 3.6 V *(8)*	20	74	k¾

Table 20. EPM7032B Selectable I/O Standard Timing Adder Delays *Notes (1)*								
I/O Standard	**Parameter**	**Speed Grade**						**Unit**
		-3.5		**-5.0**		**-7.5**		
		Min	**Max**	**Min**	**Max**	**Min**	**Max**	
PCI	Input to PIA		0.0		0.0		0.0	ns
	Input to global clock and clear		0.0		0.0		0.0	ns
	Input to fast input register		0.0		0.0		0.0	ns
	All outputs		0.0		0.0		0.0	ns

Notes to tables:

(1) These values are specified under the Recommended Operating Conditions in Table 15 on page 29. See Figure 14 for more information on switching waveforms.
(2) These values are specified for a PIA fan-out of all LABs.
(3) Measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
(4) The t_{LPA} parameter must be added to the t_{LAD}, t_{LAC}, t_{IC}, t_{ACL}, t_{CPPW}, t_{EN}, and t_{SEXP} parameters for macrocells running in low-power mode.

Table 23. EPM7064B Selectable I/O Standard Timing Adder Delays (Part 2 of 2) *Note (1)*

I/O Standard	Parameter	Speed Grade						Unit
		-3		-5		-7		
		Min	Max	Min	Max	Min	Max	
PCI	Input to PIA		0.0		0.0		0.0	ns
	Input to global clock and clear		0.0		0.0		0.0	ns
	Input to fast input register		0.0		0.0		0.0	ns
	All outputs		0.0		0.0		0.0	ns

Notes to tables:

(1) These values are specified under the Recommended Operating Conditions in Table 15 on page 29. See Figure 14 for more information on switching waveforms.
(2) These values are specified for a PIA fan-out of all LABs.
(3) Measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
(4) The t_{LPA} parameter must be added to the t_{LAD}, t_{LAC}, t_{IC}, t_{ACL}, t_{CPPW}, t_{EN}, and t_{SEXP} parameters for macrocells running in low-power mode.

Table 24. EPM7128B External Timing Parameters *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			-4		-7		-10		
			Min	Max	Min	Max	Min	Max	
t_{PD1}	Input to non-registered output	C1 = 35 pF *(2)*		4.0		7.5		10.0	ns
t_{PD2}	I/O input to non-registered output	C1 = 35 pF *(2)*		4.0		7.5		10.0	ns
t_{SU}	Global clock setup time	*(2)*	2.5		4.5		6.1		ns
t_{H}	Global clock hold time	*(2)*	0.0		0.0		0.0		ns
t_{FSU}	Global clock setup time of fast input		1.0		1.5		1.5		ns
t_{FH}	Global clock hold time of fast input		1.0		1.0		1.0		ns
t_{FZHSU}	Global clock setup time of fast input with zero hold time		2.0		3.0		3.0		ns
t_{FZHH}	Global clock hold time of fast input with zero hold time		0.0		0.0		0.0		ns
t_{CO1}	Global clock to output delay	C1 = 35 pF	1.0	2.8	1.0	5.7	1.0	7.5	ns
t_{CH}	Global clock high time		1.5		3.0		4.0		ns
t_{CL}	Global clock low time		1.5		3.0		4.0		ns
t_{ASU}	Array clock setup time	*(2)*	1.2		2.0		2.8		ns
t_{AH}	Array clock hold time	*(2)*	0.2		0.7		0.9		ns
t_{ACO1}	Array clock to output delay	C1 = 35 pF *(2)*	1.0	4.1	1.0	8.2	1.0	10.8	ns
t_{ACH}	Array clock high time		1.5		3.0		4.0		ns
t_{ACL}	Array clock low time		1.5		3.0		4.0		ns
t_{CPPW}	Minimum pulse width for clear and preset		1.5		3.0		4.0		ns
t_{CNT}	Minimum global clock period	*(2)*		4.1		7.9		10.6	ns
f_{CNT}	Maximum internal global clock frequency	*(2)*, *(3)*	243.9		126.6		94.3		MHz
t_{ACNT}	Minimum array clock period	*(2)*		4.1		7.9		10.6	ns
f_{ACNT}	Maximum internal array clock frequency	*(2)*, *(3)*	243.9		126.6		94.3		MHz

Table 32. EPM7512B Selectable I/O Standard Timing Adder Delays (Part 1 of 2) *Note (1)*

I/O Standard	Parameter	Speed Grade						Unit
		-5		-7		-10		
		Min	Max	Min	Max	Min	Max	
3.3 V TTL/CMOS	Input to PIA		0.0		0.0		0.0	ns
	Input to global clock and clear		0.0		0.0		0.0	ns
	Input to fast input register		0.0		0.0		0.0	ns
	All outputs		0.0		0.0		0.0	ns
2.5 V TTL/CMOS	Input to PIA		0.4		0.5		0.7	ns
	Input to global clock and clear		0.3		0.4		0.5	ns
	Input to fast input register		0.2		0.3		0.3	ns
	All outputs		0.2		0.3		0.3	ns
1.8 V TTL/CMOS	Input to PIA		0.7		1.0		1.3	ns
	Input to global clock and clear		0.6		0.8		1.0	ns
	Input to fast input register		0.5		0.6		0.8	ns
	All outputs		1.3		1.8		2.3	ns
SSTL-2 Class I	Input to PIA		1.5		2.0		2.7	ns
	Input to global clock and clear		1.4		1.9		2.5	ns
	Input to fast input register		1.1		1.5		2.0	ns
	All outputs		0.0		0.0		0.0	ns
SSTL-2 Class II	Input to PIA		1.5		2.0		2.7	ns
	Input to global clock and clear		1.4		1.9		2.5	ns
	Input to fast input register		1.1		1.5		2.0	ns
	All outputs		–0.1		–0.1		–0.2	ns
SSTL-3 Class I	Input to PIA		1.4		1.9		2.5	ns
	Input to global clock and clear		1.2		1.6		2.2	ns
	Input to fast input register		1.0		1.4		1.8	ns
	All outputs		0.0		0.0		0.0	ns
SSTL-3 Class II	Input to PIA		1.4		1.9		2.5	ns
	Input to global clock and clear		1.2		1.6		2.2	ns
	Input to fast input register		1.0		1.4		1.8	ns
	All outputs		0.0		0.0		0.0	ns
GTL+	Input to PIA		1.8		2.5		3.3	ns
	Input to global clock and clear		1.9		2.6		3.5	ns
	Input to fast input register		1.8		2.5		3.3	ns
	All outputs		0.0		0.0		0.0	ns

The I_{CCINT} value depends on the switching frequency and the application logic. The I_{CCINT} value is calculated with the following equation:

$$I_{CCINT} =$$

$$(A \times MC_{TON}) + [B \times (MC_{DEV} - MC_{TON})] + (C \times MC_{USED} \times \mathbf{f_{MAX}} \times \mathbf{tog_{LC}})$$

The parameters in this equation are:

MC_{TON} = Number of macrocells with the Turbo Bit™ option turned on, as reported in the MAX+PLUS II Report File (**.rpt**)
MC_{DEV} = Number of macrocells in the device
MC_{USED} = Total number of macrocells in the design, as reported in the Report File
$\mathbf{f_{MAX}}$ = Highest clock frequency to the device
$\mathbf{tog_{LC}}$ = Average percentage of logic cells toggling at each clock (typically 12.5%)
A, B, C = Constants, shown in Table 33

Table 33. MAX 7000B I_{CC} Equation Constants

Device	A	B	C
EPM7032B	0.91	0.54	0.010
EPM7064B	0.91	0.54	0.012
EPM7128B	0.91	0.54	0.016
EPM7256B	0.91	0.54	0.017
EPM7512B	0.91	0.54	0.019

This calculation provides an I_{CC} estimate based on typical conditions using a pattern of a 16-bit, loadable, enabled, up/down counter in each LAB with no output load. Actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

Figure 15. I_{CC} vs. Frequency for EPM7032B Devices

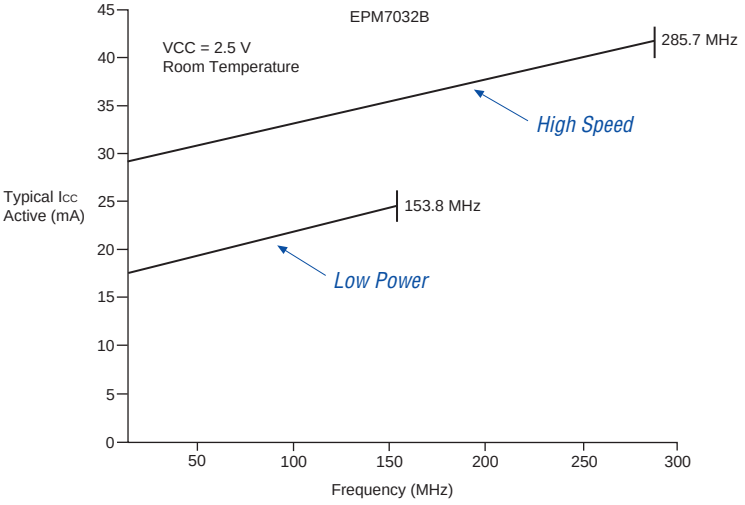


Figure 16. I_{CC} vs. Frequency for EPM7064B Devices

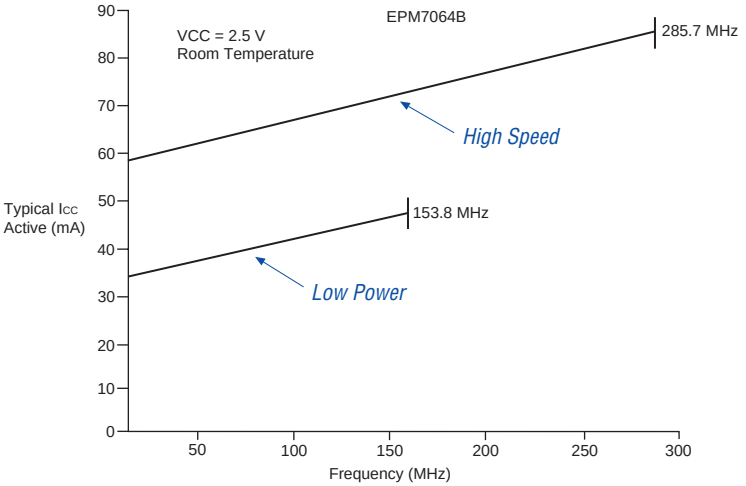


Figure 17. I_{CC} vs. Frequency for EPM7128B Devices

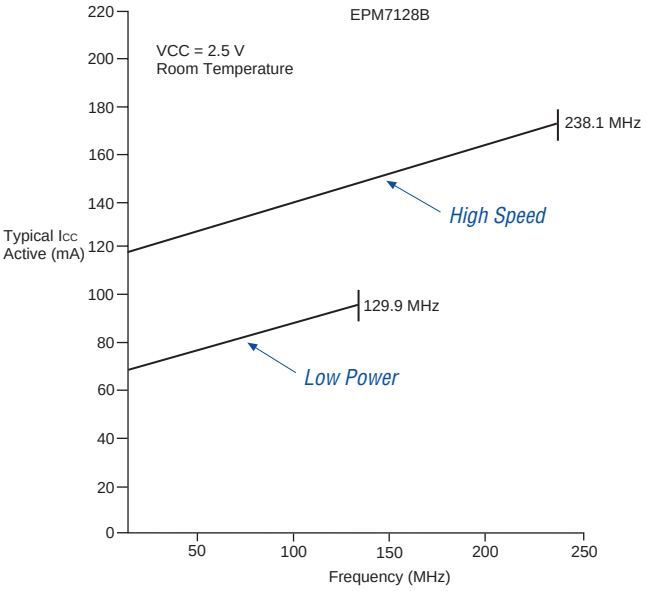


Figure 18. I_{CC} vs. Frequency for EPM7256B Devices

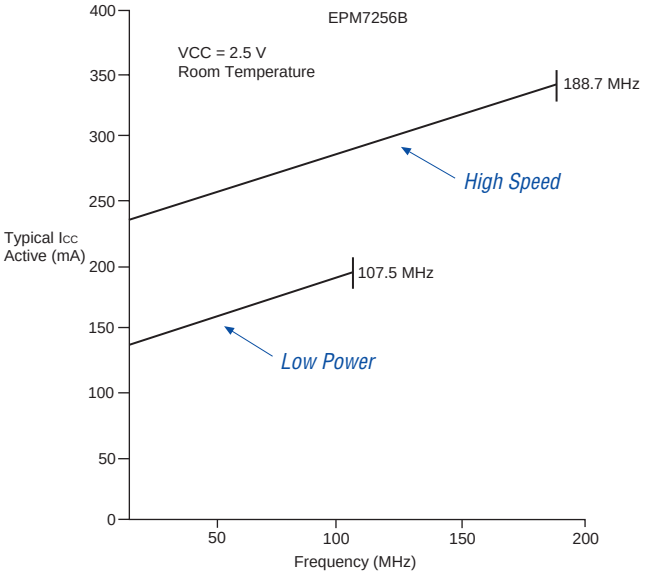


Figure 19. I_{CC} vs. Frequency for EPM7512B Devices