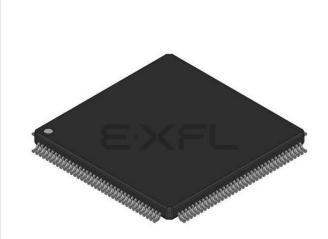
E·XFL

Altera - EPM7128BTC144-4 Datasheet



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Understanding <u>Embedded - CPLDs (Complex</u> <u>Programmable Logic Devices)</u>

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixedfunction ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details

Details	
Product Status	Active
Programmable Type	EE PLD
Delay Time tpd(1) Max	4 ns
Voltage Supply - Internal	2.375V ~ 2.625V
Number of Logic Elements/Blocks	8
Number of Macrocells	128
Number of Gates	2500
Number of I/O	100
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=epm7128btc144-4

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

Macrocells

The MAX 7000B macrocell can be individually configured for either sequential or combinatorial logic operation. The macrocell consists of three functional blocks: the logic array, the product-term select matrix, and the programmable register. Figure 2 shows the MAX 7000B macrocell.

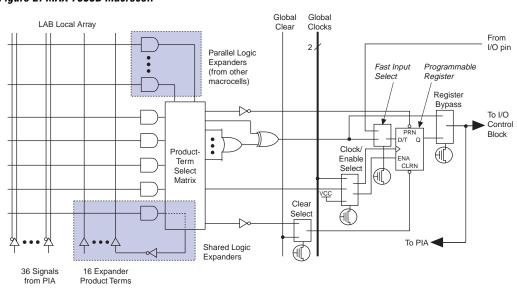


Figure 2. MAX 7000B Macrocell

Combinatorial logic is implemented in the logic array, which provides five product terms per macrocell. The product-term select matrix allocates these product terms for use as either primary logic inputs (to the OR and XOR gates) to implement combinatorial functions, or as secondary inputs to the macrocell's register preset, clock, and clock enable control functions.

Two kinds of expander product terms ("expanders") are available to supplement macrocell logic resources:

- Shareable expanders, which are inverted product terms that are fed back into the logic array
- Parallel expanders, which are product terms borrowed from adjacent macrocells

The Altera development system automatically optimizes product-term allocation according to the logic requirements of the design.

For registered functions, each macrocell flipflop can be individually programmed to implement D, T, JK, or SR operation with programmable clock control. The flipflop can be bypassed for combinatorial operation. During design entry, the designer specifies the desired flipflop type; the MAX+PLUS II software then selects the most efficient flipflop operation for each registered function to optimize resource utilization.

Each programmable register can be clocked in three different modes:

- Global clock signal. This mode achieves the fastest clock-to-output performance.
- Global clock signal enabled by an active-high clock enable. A clock enable is generated by a product term. This mode provides an enable on each flipflop while still achieving the fast clock-to-output performance of the global clock.
- Array clock implemented with a product term. In this mode, the flipflop can be clocked by signals from buried macrocells or I/O pins.

Two global clock signals are available in MAX 7000B devices. As shown in Figure 1, these global clock signals can be the true or the complement of either of the global clock pins, GCLK1 or GCLK2.

Each register also supports asynchronous preset and clear functions. As shown in Figure 2, the product-term select matrix allocates product terms to control these operations. Although the product-term-driven preset and clear from the register are active high, active-low control can be obtained by inverting the signal within the logic array. In addition, each register clear function can be individually driven by the active-low dedicated global clear pin (GCLRn). Upon power-up, each register in a MAX 7000B device may be set to either a high or low state. This power-up state is specified at design entry.

All MAX 7000B I/O pins have a fast input path to a macrocell register. This dedicated path allows a signal to bypass the PIA and combinatorial logic and be clocked to an input D flipflop with an extremely fast input setup time. The input path from the I/O pin to the register has a programmable delay element that can be selected to either guarantee zero hold time or to get the fastest possible set-up time (as fast as 1.0 ns).

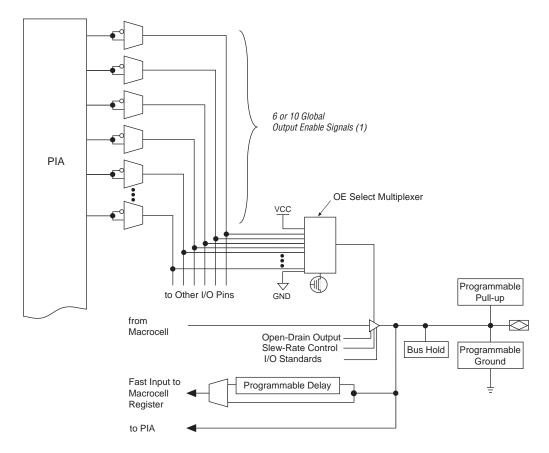
Parallel Expanders

Parallel expanders are unused product terms that can be allocated to a neighboring macrocell to implement fast, complex logic functions. Parallel expanders allow up to 20 product terms to directly feed the macrocell OR logic, with five product terms provided by the macrocell and 15 parallel expanders provided by neighboring macrocells in the LAB.

The Altera Compiler can automatically allocate up to three sets of up to five parallel expanders to the macrocells that require additional product terms. Each set of five parallel expanders incurs a small, incremental timing delay (t_{PEXP}). For example, if a macrocell requires 14 product terms, the Compiler uses the five dedicated product terms within the macrocell and allocates two sets of parallel expanders; the first set includes five product terms and the second set includes four product terms, increasing the total delay by $2 \times t_{PEXP}$.

Two groups of eight macrocells within each LAB (e.g., macrocells 1 through 8, and 9 through 16) form two chains to lend or borrow parallel expanders. A macrocell borrows parallel expanders from lower-numbered macrocells. For example, macrocell 8 can borrow parallel expanders from macrocell 7, from macrocells 7 and 6, or from macrocells 7, 6, and 5. Within each group of eight, the lowest-numbered macrocell can only lend parallel expanders and the highest-numbered macrocell can only borrow them. Figure 4 shows how parallel expanders can be borrowed from a neighboring macrocell.





Note:

(1) EPM7032B, EPM7064B, EPM7128B, and EPM7256B devices have six output enable signals. EPM7512B devices have ten output enable signals.

When the tri-state buffer control is connected to ground, the output is tri-stated (high impedance) and the I/O pin can be used as a dedicated input. When the tri-state buffer control is connected to $V_{CC'}$, the output is enabled.

The MAX 7000B architecture provides dual I/O feedback, in which macrocell and pin feedbacks are independent. When an I/O pin is configured as an input, the associated macrocell can be used for buried logic.

In-System Programmability (ISP)

MAX 7000B devices can be programmed in-system via an industrystandard 4-pin IEEE Std. 1149.1 (JTAG) interface. ISP offers quick, efficient iterations during design development and debugging cycles. The MAX 7000B architecture internally generates the high programming voltages required to program EEPROM cells, allowing in-system programming with only a single 2.5-V power supply. During in-system programming, the I/O pins are tri-stated and weakly pulled-up to eliminate board conflicts. The pull-up value is nominally 50 k³/₄.

MAX 7000B devices have an enhanced ISP algorithm for faster programming. These devices also offer an ISP_Done bit that provides safe operation when in-system programming is interrupted. This ISP_Done bit, which is the last bit programmed, prevents all I/O pins from driving until the bit is programmed.

ISP simplifies the manufacturing flow by allowing devices to be mounted on a PCB with standard pick-and-place equipment before they are programmed. MAX 7000B devices can be programmed by downloading the information via in-circuit testers, embedded processors, the Altera MasterBlaster communications cable, and the ByteBlasterMV parallel port download cable. Programming the devices after they are placed on the board eliminates lead damage on high-pin-count packages (e.g., QFP packages) due to device handling. MAX 7000B devices can be reprogrammed after a system has already shipped to the field. For example, product upgrades can be performed in the field via software or modem.

In-system programming can be accomplished with either an adaptive or constant algorithm. An adaptive algorithm reads information from the unit and adapts subsequent programming steps to achieve the fastest possible programming time for that unit. A constant algorithm uses a pre-defined (non-adaptive) programming sequence that does not take advantage of adaptive algorithm programming time improvements. Some in-circuit testers cannot program using an adaptive algorithm. Therefore, a constant algorithm must be used. MAX 7000B devices can be programmed with either an adaptive or constant (non-adaptive) algorithm.

The Jam Standard Test and Programming Language (STAPL), JEDEC standard JESD-71, can be used to program MAX 7000B devices with in-circuit testers, PCs, or embedded processors.

For more information on using the Jam language, see *Application Note 88* (Using the Jam Language for ISP & ICR via an Embedded Processor) and Application Note 122 (Using STAPL for ISP & ICR via an Embedded Processor).

The ISP circuitry in MAX 7000B devices is compliant with the IEEE Std. 1532 specification. The IEEE Std. 1532 is a standard developed to allow concurrent ISP between multiple PLD vendors.

Programming Sequence

During in-system programming, instructions, addresses, and data are shifted into the MAX 7000B device through the TDI input pin. Data is shifted out through the TDO output pin and compared against the expected data.

Programming a pattern into the device requires the following six ISP stages. A stand-alone verification of a programmed pattern involves only stages 1, 2, 5, and 6.

- 1. *Enter ISP*. The enter ISP stage ensures that the I/O pins transition smoothly from user mode to ISP mode. The enter ISP stage requires 1 ms.
- 2. *Check ID*. Before any program or verify process, the silicon ID is checked. The time required to read this silicon ID is relatively small compared to the overall programming time.
- 3. *Bulk Erase.* Erasing the device in-system involves shifting in the instructions to erase the device and applying one erase pulse of 100 ms.
- 4. *Program*. Programming the device in-system involves shifting in the address and data and then applying the programming pulse to program the EEPROM cells. This process is repeated for each EEPROM address.
- 5. *Verify.* Verifying an Altera device in-system involves shifting in addresses, applying the read pulse to verify the EEPROM cells, and shifting out the data for comparison. This process is repeated for each EEPROM address.
- 6. *Exit ISP*. An exit ISP stage ensures that the I/O pins transition smoothly from ISP mode to user mode. The exit ISP stage requires 1 ms.

Programming Times

The time required to implement each of the six programming stages can be broken into the following two elements:

- A pulse time to erase, program, or read the EEPROM cells.
- A shifting time based on the test clock (TCK) frequency and the number of TCK cycles to shift instructions, address, and data into the device.

The programming times described in Tables 4 through 6 are associated with the worst-case method using the enhanced ISP algorithm.

Table 4. MAX 7000B t _{PUL}	ble 4. MAX 7000B t _{PULSE} & Cycle _{TCK} Values							
Device	Programming		Stand-Alone Verif					
	t _{PPULSE} (s)	Cycle _{PTCK}	t _{VPULSE} (s)	Cycle _{VTCK}				
EMP7032B	2.12	70,000	0.002	18,000				
EMP7064B	2.12	120,000	0.002	35,000				
EMP7128B	2.12	222,000	0.002	69,000				
EMP7256B	2.12	466,000	0.002	151,000				
EMP7512B	2.12	914,000	0.002	300,000				

Tables 5 and 6 show the in-system programming and stand alone verification times for several common test clock frequencies.

Table 5. MAX 7000B In-System Programming Times for Different Test Clock Frequencies									
Device		f _{TCK}							
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EMP7032B	2.13	2.13	2.15	2.19	2.26	2.47	2.82	3.52	s
EMP7064B	2.13	2.14	2.18	2.24	2.36	2.72	3.32	4.52	S
EMP7128B	2.14	2.16	2.23	2.34	2.56	3.23	4.34	6.56	S
EMP7256B	2.17	2.21	2.35	2.58	3.05	4.45	6.78	11.44	S
EMP7512B	2.21	2.30	2.58	3.03	3.95	6.69	11.26	20.40	S

Table 1. MAX 7000B Stand-Alone Verification Times for Different Test Clock Frequencies									
Device		f _{тск}							
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EMP7032B	0.00	0.01	0.01	0.02	0.04	0.09	0.18	0.36	S
EMP7064B	0.01	0.01	0.02	0.04	0.07	0.18	0.35	0.70	s
EMP7128B	0.01	0.02	0.04	0.07	0.14	0.35	0.69	1.38	s
EMP7256B	0.02	0.03	0.08	0.15	0.30	0.76	1.51	3.02	S
EMP7512B	0.03	0.06	0.15	0.30	0.60	1.50	3.00	6.00	S

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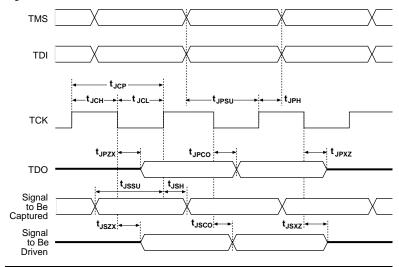


Figure 8. MAX 7000B JTAG Waveforms

Table 9 shows the JTAG timing parameters and values for MAX 7000B devices.

Table 9. Note (1)	JTAG Timing Parameters & Values for MAX 70	00B Dev	ices	
Symbol	Parameter	Min	Max	Unit
t _{JCP}	TCK clock period	100		ns
t _{JCH}	TCK clock high time	50		ns
t _{JCL}	TCK clock low time	50		ns
t _{JPSU}	JTAG port setup time	20		ns
t _{JPH}	JTAG port hold time	45		ns
t _{JPCO}	JTAG port clock to output		25	ns
t _{JPZX}	JTAG port high impedance to valid output		25	ns
t _{JPXZ}	JTAG port valid output to high impedance		25	ns
t _{JSSU}	Capture register setup time	20		ns
t _{JSH}	Capture register hold time	45		ns
t _{JSCO}	Update register clock to output		25	ns
t _{JSZX}	Update register high impedance to valid output		25	ns
t _{JSXZ}	Update register valid output to high impedance		25	ns

Note:

(1) Timing parameters in this table apply to all V_{CCIO} levels.

MAX 7000B devices contain two I/O banks. Both banks support all standards. Each I/O bank has its own VCCIO pins. A single device can support 1.8-V, 2.5-V, and 3.3-V interfaces; each bank can support a different standard independently. Within a bank, any one of the terminated standards can be supported.

Figure 9 shows the arrangement of the MAX 7000B I/O banks.

Programmable I/O Banks

Figure 9. MAX 7000B I/O Banks for Various Advanced I/O Standards

Table 11 shows which macrocells have pins in each I/O bank.

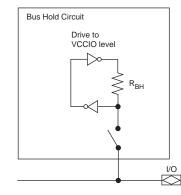
Table 11. Macrocell Pins Co	ntained in Each I/O Bank	
Device	Bank 1	Bank 2
EPM7032B	1-16	17-32
EPM7064B	1-32	33-64
EPM7128B	1-64	65-128
EPM7256B	1-128, 177-181	129-176, 182-256
EPM7512B	1-265	266-512

Each MAX 7000B device has two VREF pins. Each can be set to a separate V_{REF} level. Any I/O pin that uses one of the voltage-referenced standards (GTL+, SSTL-2, or SSTL-3) may use either of the two VREF pins. If these pins are not required as VREF pins, they may be individually programmed to function as user I/O pins.

Two inverters implement the bus-hold circuitry in a loop that weakly drives back to the I/O pin in user mode.

Figure 10 shows a block diagram of the bus-hold circuit.

Figure 10. Bus-Hold Circuit



PCI Compatibility

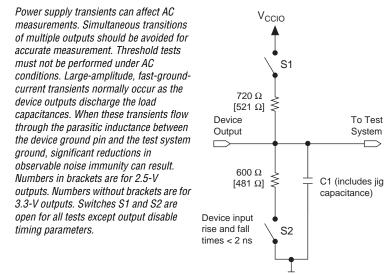
MAX 7000B devices are compatible with PCI applications as well as all 3.3-V electrical specifications in the *PCI Local Bus Specification Revision 2.2* except for the clamp diode. While having multiple clamp diodes on a signal trace may be redundant, designers can add an external clamp diode to meet the specification. Table 13 shows the MAX 7000B device speed grades that meet the PCI timing specifications.

Table 13. MAX 7000B Device Speed Grades that Meet PCI Timing Specifications						
Device	Specification					
	33-MHz PCI	66-MHz PCI				
EPM7032B	All speed grades	-3				
EPM7064B	All speed grades	-3				
EPM7128B	All speed grades	-4				
EPM7256B	All speed grades	-5 (1)				
EPM7512B	All speed grades	-5 (1)				

Note:

(1) The EPM7256B and EPM7512B devices in a -5 speed grade meet all PCI timing specifications for 66-MHz operation except the Input Setup Time to CLK—Bused Signal parameter. However, these devices are within 1 ns of that parameter. EPM7256B and EPM7512B devices meet all other 66-MHz PCI timing specifications.

Figure 11. MAX 7000B AC Test Conditions



Operating Conditions

Tables 14 through 17 provide information on absolute maximum ratings, recommended operating conditions, operating conditions, and capacitance for MAX 7000B devices.

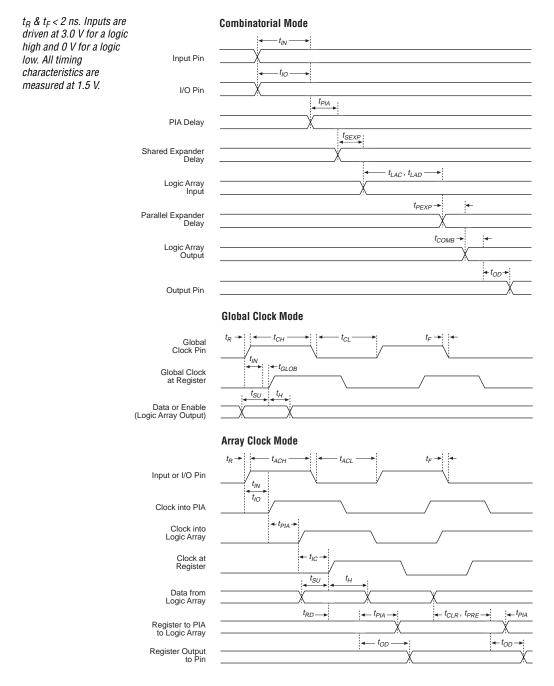
Table 1	Table 14. MAX 7000B Device Absolute Maximum Ratings Note (1)								
Symbol	Parameter	Conditions	Min	Max	Unit				
V _{CCINT}	Supply voltage		-0.5	3.6	V				
V _{CCIO}	Supply voltage		-0.5	3.6	V				
VI	DC input voltage	(2)	-2.0	4.6	V				
I _{OUT}	DC output current, per pin		-33	50	mA				
T _{STG}	Storage temperature	No bias	-65	150	°C				
T _A	Ambient temperature	Under bias	-65	135	°C				
TJ	Junction temperature	Under bias	-65	135	°C				

Table 1	able 17. MAX 7000B Device Capacitance Note (9)						
Symbol	Parameter	Parameter Conditions Min Max Unit					
C _{IN}	Input pin capacitance	V _{IN} = 0 V, f = 1.0 MHz		8	pF		
C _{I/O}	I/O pin capacitance	V _{OUT} = 0 V, f = 1.0 MHz		8	pF		

Notes to tables:

- (1) See the Operating Requirements for Altera Devices Data Sheet.
- (2) Minimum DC input voltage is -0.5 V. During transitions, the inputs may undershoot to -2.0 V or overshoot to 4.6 V for input currents less than 100 mA and periods shorter than 20 ns.
- (3) All pins, including dedicated inputs, I/O pins, and JTAG pins, may be driven before V_{CCINT} and V_{CCIO} are powered.
- (4) These values are specified under the Recommended Operating Conditions in Table 15 on page 29.
- (5) The parameter is measured with 50% of the outputs each sourcing the specified current. The I_{OH} parameter refers to high-level TTL or CMOS output current.
- (6) The parameter is measured with 50% of the outputs each sinking the specified current. The I_{OL} parameter refers to low-level TTL or CMOS output current.
- (7) This value is specified for normal device operation. During power-up, the maximum leakage current is ±300 μA.
- (8) This pull-up exists while devices are being programmed in-system and in unprogrammed devices during power-up. The pull-up resistor is from the pins to V_{CCIO}.
- (9) Capacitance is measured at 25° C and is sample-tested only. Two of the dedicated input pins (OE1 and GCLRN) have a maximum capacitance of 15 pF.
- (10) The POR time for all 7000B devices does not exceed 100 μs. The sufficient V_{CCINT} voltage level for POR is 2.375 V. The device is fully initialized within the POR time after V_{CCINT} reaches the sufficient POR voltage level.
- (11) These devices support in-system programming for -40° to 100° C. For in-system programming support between -40° and 0° C, contact Altera Applications.

Figure 14. MAX 7000B Switching Waveforms



Symbol	Parameter	Conditions	Speed Grade						Unit
			-3		-5		-7		
			Min	Max	Min	Max	Min	Max	
t _{IN}	Input pad and buffer delay			0.3		0.5		0.7	ns
t _{IO}	I/O input pad and buffer delay			0.3		0.5		0.7	ns
t _{FIN}	Fast input delay			0.9		1.3		2.0	ns
t _{FIND}	Programmable delay adder for fast input			1.0		1.5		1.5	ns
t _{SEXP}	Shared expander delay			1.5		2.1		3.2	ns
t _{PEXP}	Parallel expander delay			0.4		0.6		0.9	ns
t _{LAD}	Logic array delay			1.4		2.0		3.1	ns
t _{LAC}	Logic control array delay			1.2		1.7		2.6	ns
t _{IOE}	Internal output enable delay			0.1		0.2		0.3	ns
t _{OD1}	Output buffer and pad delay slow slew rate = off $V_{CCIO} = 3.3 \text{ V}$	C1 = 35 pF		0.9		1.2		1.8	ns
t _{OD3}	Output buffer and pad delay slow slew rate = on $V_{CCIO} = 2.5$ V or 3.3 V	C1 = 35 pF		5.9		6.2		6.8	ns
t _{ZX1}	Output buffer enable delay slow slew rate = off $V_{CCIO} = 3.3 V$	C1 = 35 pF		1.6		2.2		3.4	ns
t _{ZX3}	Output buffer enable delay slow slew rate = on $V_{CCIO} = 2.5 V \text{ or } 3.3 V$	C1 = 35 pF		6.6		7.2		8.4	ns
t _{XZ}	Output buffer disable delay	C1 = 5 pF		1.6		2.2		3.4	ns
t _{SU}	Register setup time		0.7		1.1		1.6		ns
t _H	Register hold time		0.4		0.5		0.9		ns
t _{FSU}	Register setup time of fast input		0.8		0.8		1.1		ns
t _{FH}	Register hold time of fast input		1.2		1.2		1.4		ns
t _{RD}	Register delay			0.5		0.6		0.9	ns
t _{COMB}	Combinatorial delay			0.2		0.3		0.5	ns
t _{IC}	Array clock delay			1.2		1.8		2.8	ns
t _{EN}	Register enable time			1.2		1.7		2.6	ns
t _{GLOB}	Global control delay			0.7		1.1		1.6	ns
t _{PRE}	Register preset time		1	1.0		1.3		1.9	ns
t _{CLR}	Register clear time			1.0		1.3		1.9	ns
t _{PIA}	PIA delay	(2)	1	0.7		1.0		1.4	ns
t _{LPA}	Low-power adder	(4)	1	1.5	1	2.1		3.2	ns

Symbol	Parameter	Conditions	Speed Grade						Unit
			-	-4		7	-10		1
			Min	Max	Min	Max	Min	Max	
t _{PD1}	Input to non-registered output	C1 = 35 pF (2)		4.0		7.5		10.0	ns
t _{PD2}	I/O input to non-registered output	C1 = 35 pF (2)		4.0		7.5		10.0	ns
t _{SU}	Global clock setup time	(2)	2.5		4.5		6.1		ns
t _H	Global clock hold time	(2)	0.0		0.0		0.0		ns
t _{FSU}	Global clock setup time of fast input		1.0		1.5		1.5		ns
t _{FH}	Global clock hold time of fast input		1.0		1.0		1.0		ns
t _{FZHSU}	Global clock setup time of fast input with zero hold time		2.0		3.0		3.0		ns
t _{FZHH}	Global clock hold time of fast input with zero hold time		0.0		0.0		0.0		ns
t _{CO1}	Global clock to output delay	C1 = 35 pF	1.0	2.8	1.0	5.7	1.0	7.5	ns
t _{CH}	Global clock high time		1.5		3.0		4.0		ns
t _{CL}	Global clock low time		1.5		3.0		4.0		ns
t _{ASU}	Array clock setup time	(2)	1.2		2.0		2.8		ns
t _{AH}	Array clock hold time	(2)	0.2		0.7		0.9		ns
t _{ACO1}	Array clock to output delay	C1 = 35 pF (2)	1.0	4.1	1.0	8.2	1.0	10.8	ns
t _{ACH}	Array clock high time		1.5		3.0		4.0		ns
t _{ACL}	Array clock low time		1.5		3.0		4.0		ns
t _{CPPW}	Minimum pulse width for clear and preset		1.5		3.0		4.0		ns
t _{CNT}	Minimum global clock period	(2)		4.1		7.9		10.6	ns
f _{CNT}	Maximum internal global clock frequency	(2), (3)	243.9		126.6		94.3		MHz
t _{acnt}	Minimum array clock period	(2)		4.1		7.9		10.6	ns
f _{acnt}	Maximum internal array clock frequency	(2), (3)	243.9		126.6		94.3		MHz

Symbol	Parameter	Conditions			Speed	Grade		U	
			-	-5		-7		-10	
			Min	Max	Min	Max	Min	Max	
t _{PD1}	Input to non-registered output	C1 = 35 pF (2)		5.0		7.5		10.0	ns
t _{PD2}	I/O input to non-registered output	C1 = 35 pF (2)		5.0		7.5		10.0	ns
t _{SU}	Global clock setup time	(2)	3.3		4.8		6.6		ns
t _H	Global clock hold time	(2)	0.0		0.0		0.0		ns
t _{FSU}	Global clock setup time of fast input		1.0		1.5		1.5		ns
t _{FH}	Global clock hold time for fast input		1.0		1.0		1.0		ns
t _{FZHSU}	Global clock setup time of fast input with zero hold time		2.5		3.0		3.0		ns
t _{FZHH}	Global clock hold time of fast input with zero hold time		0.0		0.0		0.0		ns
t _{CO1}	Global clock to output delay	C1 = 35 pF	1.0	3.3	1.0	5.1	1.0	6.7	ns
t _{CH}	Global clock high time		2.0		3.0		4.0		ns
t _{CL}	Global clock low time		2.0		3.0		4.0		ns
t _{ASU}	Array clock setup time	(2)	1.4		2.0		2.8		ns
t _{AH}	Array clock hold time	(2)	0.4		0.8		1.0		ns
t _{ACO1}	Array clock to output delay	C1 = 35 pF (2)	1.0	5.2	1.0	7.9	1.0	10.5	ns
t _{ACH}	Array clock high time		2.0		3.0		4.0		ns
t _{ACL}	Array clock low time		2.0		3.0		4.0		ns
t _{CPPW}	Minimum pulse width for clear and preset		2.0		3.0		4.0		ns
t _{CNT}	Minimum global clock period	(2)		5.3		7.9		10.6	ns
f _{CNT}	Maximum internal global clock frequency	(2), (3)	188.7		126.6		94.3		MHz
t _{acnt}	Minimum array clock period	(2)		5.3		7.9		10.6	ns
facnt	Maximum internal array clock frequency	(2), (3)	188.7		126.6		94.3		MHz

Device Pin-Outs

See the Altera web site (http://www.altera.com) or the *Altera Digital Library* for pin-out information.

Figures 20 through 29 show the package pin-out diagrams for MAX 7000B devices.



Package outlines not drawn to scale.

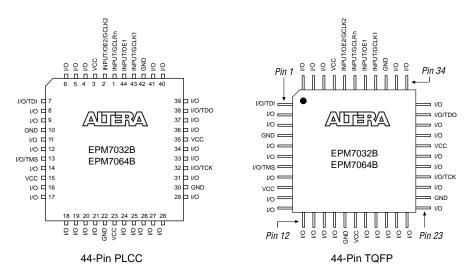


Figure 23. 100-Pin TQFP Package Pin-Out Diagram

Package outline not drawn to scale.

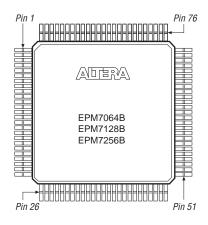
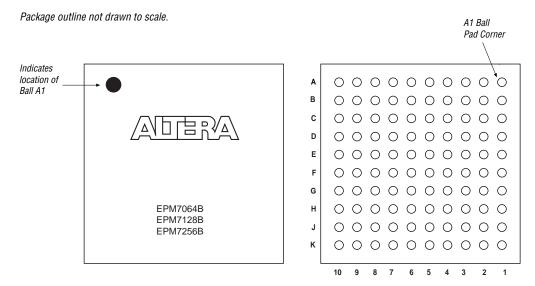


Figure 24. 100-Pin FineLine BGA Package Pin-Out Diagram





Package outline not drawn to scale.

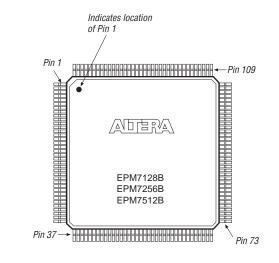
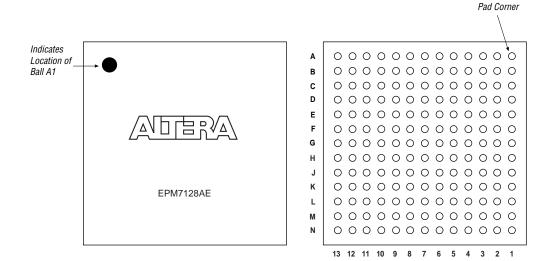


Figure 26. 169-Pin Ultra FineLine BGA Pin-Out Diagram

Package outline not drawn to scale.



A1 Ball

Figure 29. 256-Pin FineLine BGA Package Pin-Out Diagram

Package outline not drawn to scale.



Revision History

The information contained in the *MAX* 7000B Programmable Logic Device Family Data Sheet version 3.5 supersedes information published in previous versions.

Version 3.5

The following changes were made to the *MAX 7000B Programmable Logic Device Family Data Sheet* version 3.5:

■ Updated Figure 28.

Version 3.4

The following changes were made to the *MAX* 7000B Programmable Logic Device Family Data Sheet version 3.4:

Updated text in the "Power Sequencing & Hot-Socketing" section.