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Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details	
Product Status	Active
Programmable Type	In System Programmable
Delay Time tpd(1) Max	5 ns
Voltage Supply - Internal	2.375V ~ 2.625V
Number of Logic Elements/Blocks	16
Number of Macrocells	256
Number of Gates	5000
Number of I/O	164
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=epm7256bqc208-5

The Altera development system automatically optimizes product-term allocation according to the logic requirements of the design.

For registered functions, each macrocell flipflop can be individually programmed to implement D, T, JK, or SR operation with programmable clock control. The flipflop can be bypassed for combinatorial operation. During design entry, the designer specifies the desired flipflop type; the MAX+PLUS II software then selects the most efficient flipflop operation for each registered function to optimize resource utilization.

Each programmable register can be clocked in three different modes:

- Global clock signal. This mode achieves the fastest clock-to-output performance.
- Global clock signal enabled by an active-high clock enable. A clock enable is generated by a product term. This mode provides an enable on each flipflop while still achieving the fast clock-to-output performance of the global clock.
- Array clock implemented with a product term. In this mode, the flipflop can be clocked by signals from buried macrocells or I/O pins.

Two global clock signals are available in MAX 7000B devices. As shown in Figure 1, these global clock signals can be the true or the complement of either of the global clock pins, `GCLK1` or `GCLK2`.

Each register also supports asynchronous preset and clear functions. As shown in Figure 2, the product-term select matrix allocates product terms to control these operations. Although the product-term-driven preset and clear from the register are active high, active-low control can be obtained by inverting the signal within the logic array. In addition, each register clear function can be individually driven by the active-low dedicated global clear pin (`GCLRn`). Upon power-up, each register in a MAX 7000B device may be set to either a high or low state. This power-up state is specified at design entry.

All MAX 7000B I/O pins have a fast input path to a macrocell register. This dedicated path allows a signal to bypass the PIA and combinatorial logic and be clocked to an input D flipflop with an extremely fast input setup time. The input path from the I/O pin to the register has a programmable delay element that can be selected to either guarantee zero hold time or to get the fastest possible set-up time (as fast as 1.0 ns).

Expander Product Terms

Although most logic functions can be implemented with the five product terms available in each macrocell, more complex logic functions require additional product terms. Another macrocell can be used to supply the required logic resources. However, the MAX 7000B architecture also offers both shareable and parallel expander product terms ("expanders") that provide additional product terms directly to any macrocell in the same LAB. These expanders help ensure that logic is synthesized with the fewest possible logic resources to obtain the fastest possible speed.

Shareable Expanders

Each LAB has 16 shareable expanders that can be viewed as a pool of uncommitted single product terms (one from each macrocell) with inverted outputs that feed back into the logic array. Each shareable expander can be used and shared by any or all macrocells in the LAB to build complex logic functions. A small delay (t_{SEXP}) is incurred when shareable expanders are used. Figure 3 shows how shareable expanders can feed multiple macrocells.

Figure 3. MAX 7000B Shareable Expanders

Shareable expanders can be shared by any or all macrocells in an LAB.

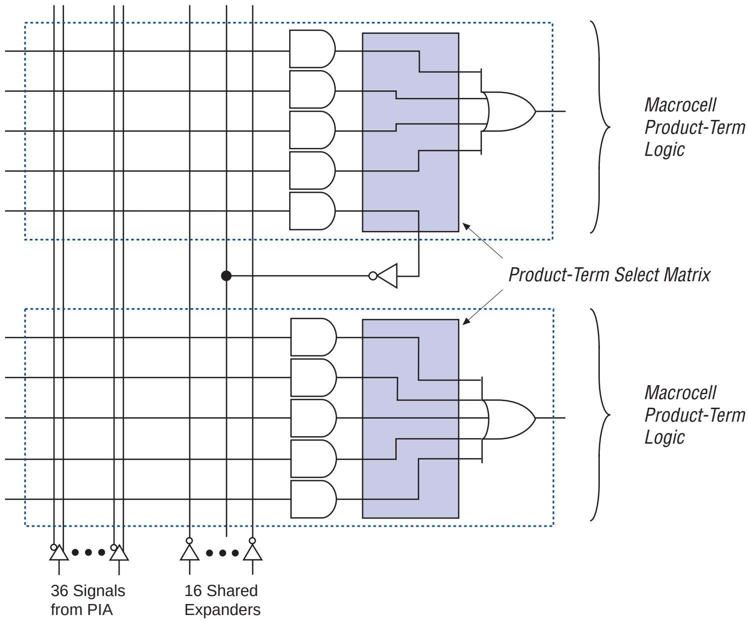
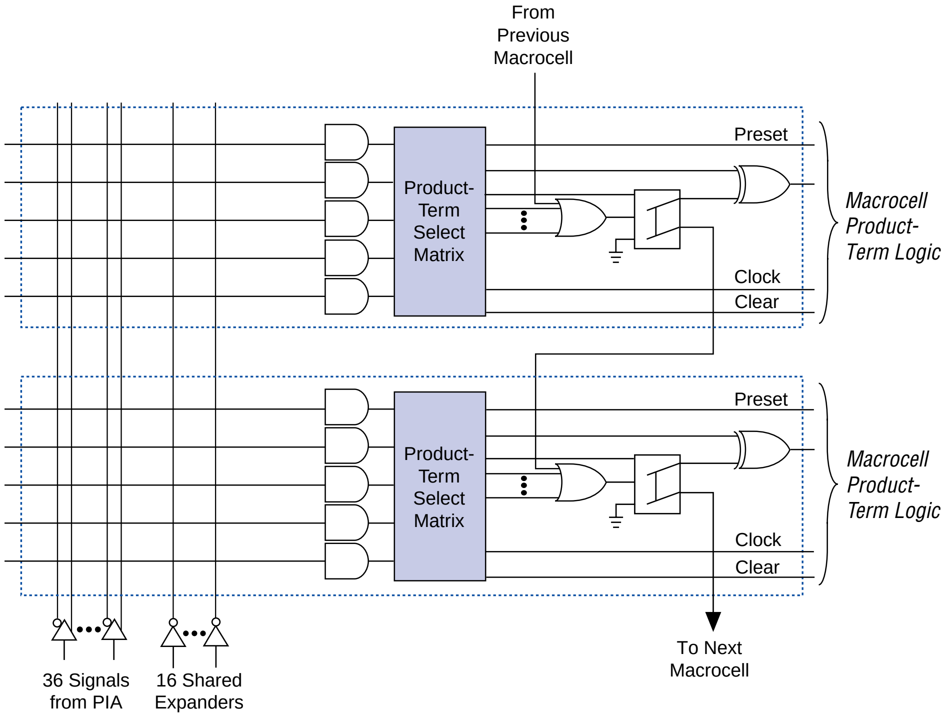


Figure 4. MAX 7000B Parallel Expanders

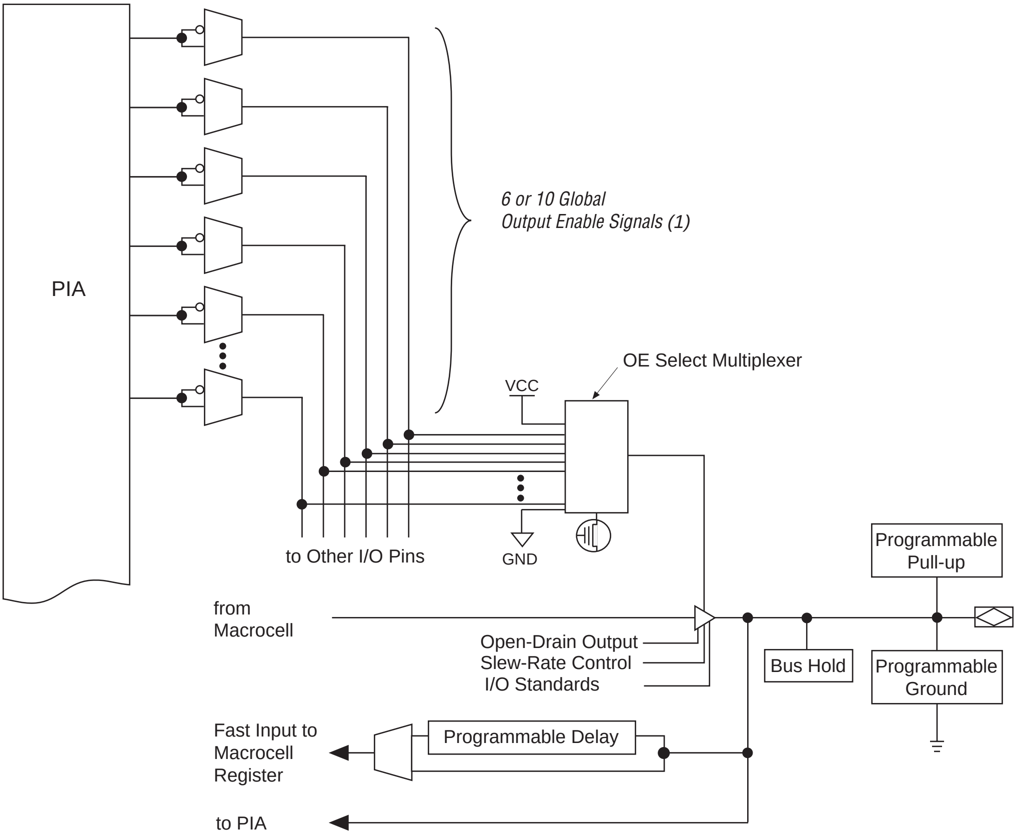
Unused product terms in a macrocell can be allocated to a neighboring macrocell.



Programmable Interconnect Array

Logic is routed between LABs on the PIA. This global bus is a programmable path that connects any signal source to any destination on the device. All MAX 7000B dedicated inputs, I/O pins, and macrocell outputs feed the PIA, which makes the signals available throughout the entire device. Only the signals required by each LAB are actually routed from the PIA into the LAB. Figure 5 shows how the PIA signals are routed into the LAB. An EEPROM cell controls one input to a two-input AND gate, which selects a PIA signal to drive into the LAB.

Figure 6. I/O Control Block of MAX 7000B Devices



Note:

(1) EPM7032B, EPM7064B, EPM7128B, and EPM7256B devices have six output enable signals. EPM7512B devices have ten output enable signals.

When the tri-state buffer control is connected to ground, the output is tri-stated (high impedance) and the I/O pin can be used as a dedicated input. When the tri-state buffer control is connected to V_{CC}, the output is enabled.

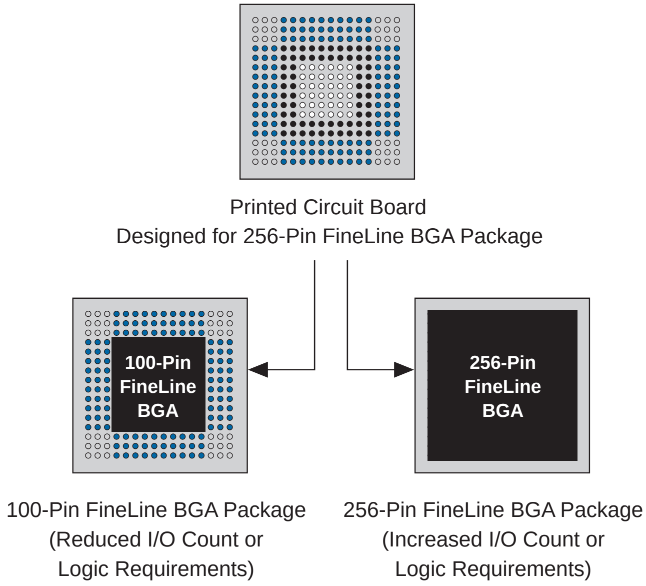
The MAX 7000B architecture provides dual I/O feedback, in which macrocell and pin feedbacks are independent. When an I/O pin is configured as an input, the associated macrocell can be used for buried logic.

SameFrame Pin-Outs

MAX 7000B devices support the SameFrame pin-out feature for FineLine BGA and 0.8-mm Ultra FineLine BGA packages. The SameFrame pin-out feature is the arrangement of balls on FineLine BGA and 0.8-mm Ultra FineLine BGA packages such that the lower-ball-count packages form a subset of the higher-ball-count packages. SameFrame pin-outs provide the flexibility to migrate not only from device to device within the same package, but also from one package to another. FineLine BGA packages are compatible with other FineLine BGA packages, and 0.8-mm Ultra FineLine BGA packages are compatible with other 0.8-mm Ultra FineLine BGA packages. A given printed circuit board (PCB) layout can support multiple device density/package combinations. For example, a single board layout can support a range of devices from an EPM7064B device in a 100-pin FineLine BGA package to an EPM7512B device in a 256-pin FineLine BGA package.

The Altera software provides support to design PCBs with SameFrame pin-out devices. Devices can be defined for present and future use. The Altera software generates pin-outs describing how to layout a board to take advantage of this migration (see Figure 7).

Figure 7. SameFrame Pin-Out Example



By combining the pulse and shift times for each of the programming stages, the program or verify time can be derived as a function of the TCK frequency, the number of devices, and specific target device(s). Because different ISP-capable devices have a different number of EEPROM cells, both the total fixed and total variable times are unique for a single device.

Programming a Single MAX 7000B Device

The time required to program a single MAX 7000B device in-system can be calculated from the following formula:

$$t_{PROG} = t_{PPULSE} + \frac{Cycle_{PTCK}}{f_{TCK}}$$

where: t_{PROG} = Programming time
t_{PPULSE} = Sum of the fixed times to erase, program, and verify the EEPROM cells
$Cycle_{PTCK}$ = Number of TCK cycles to program a device
f_{TCK} = TCK frequency

The ISP times for a stand-alone verification of a single MAX 7000B device can be calculated from the following formula:

$$t_{VER} = t_{VPULSE} + \frac{Cycle_{VTCK}}{f_{TCK}}$$

where: t_{VER} = Verify time
t_{VPULSE} = Sum of the fixed times to verify the EEPROM cells
$Cycle_{VTCK}$ = Number of TCK cycles to verify a device

The programming times described in Tables 4 through 6 are associated with the worst-case method using the enhanced ISP algorithm.

Table 4. MAX 7000B t_{PULSE} & $Cycle_{TCK}$ Values				
Device	**Programming**		**Stand-Alone Verification**	
	t_{PPULSE} (s)	$Cycle_{PTCK}$	t_{VPULSE} (s)	$Cycle_{VTCK}$
EMP7032B	2.12	70,000	0.002	18,000
EMP7064B	2.12	120,000	0.002	35,000
EMP7128B	2.12	222,000	0.002	69,000
EMP7256B	2.12	466,000	0.002	151,000
EMP7512B	2.12	914,000	0.002	300,000

Tables 5 and 6 show the in-system programming and stand alone verification times for several common test clock frequencies.

Table 5. MAX 7000B In-System Programming Times for Different Test Clock Frequencies									
Device	f_{TCK}								**Units**
	10 MHz	**5 MHz**	**2 MHz**	**1 MHz**	**500 kHz**	**200 kHz**	**100 kHz**	**50 kHz**	
EMP7032B	2.13	2.13	2.15	2.19	2.26	2.47	2.82	3.52	s
EMP7064B	2.13	2.14	2.18	2.24	2.36	2.72	3.32	4.52	s
EMP7128B	2.14	2.16	2.23	2.34	2.56	3.23	4.34	6.56	s
EMP7256B	2.17	2.21	2.35	2.58	3.05	4.45	6.78	11.44	s
EMP7512B	2.21	2.30	2.58	3.03	3.95	6.69	11.26	20.40	s

Table 1. MAX 7000B Stand-Alone Verification Times for Different Test Clock Frequencies									
Device	f_{TCK}								**Units**
	10 MHz	**5 MHz**	**2 MHz**	**1 MHz**	**500 kHz**	**200 kHz**	**100 kHz**	**50 kHz**	
EMP7032B	0.00	0.01	0.01	0.02	0.04	0.09	0.18	0.36	s
EMP7064B	0.01	0.01	0.02	0.04	0.07	0.18	0.35	0.70	s
EMP7128B	0.01	0.02	0.04	0.07	0.14	0.35	0.69	1.38	s
EMP7256B	0.02	0.03	0.08	0.15	0.30	0.76	1.51	3.02	s
EMP7512B	0.03	0.06	0.15	0.30	0.60	1.50	3.00	6.00	s

Figure 8. MAX 7000B JTAG Waveforms

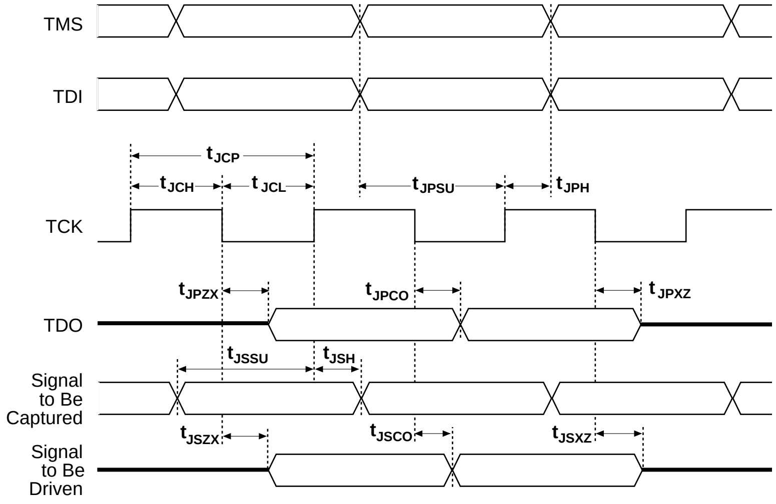


Table 9 shows the JTAG timing parameters and values for MAX 7000B devices.

Table 9. JTAG Timing Parameters & Values for MAX 7000B Devices *Note (1)*

Symbol	Parameter	Min	Max	Unit
t_{JCP}	TCK clock period	100		ns
t_{JCH}	TCK clock high time	50		ns
t_{JCL}	TCK clock low time	50		ns
t_{JPSU}	JTAG port setup time	20		ns
t_{JPH}	JTAG port hold time	45		ns
t_{JPCO}	JTAG port clock to output		25	ns
t_{JPZX}	JTAG port high impedance to valid output		25	ns
t_{JPXZ}	JTAG port valid output to high impedance		25	ns
t_{JSSU}	Capture register setup time	20		ns
t_{JSH}	Capture register hold time	45		ns
t_{JSCO}	Update register clock to output		25	ns
t_{JSZX}	Update register high impedance to valid output		25	ns
t_{JSXZ}	Update register valid output to high impedance		25	ns

Note:
(1) Timing parameters in this table apply to all V_{CCIO} levels.

Programmable Pull-Up Resistor

Each MAX 7000B device I/O pin provides an optional programmable pull-up resistor during user mode. When this feature is enabled for an I/O pin, the pull-up resistor (typically 50 k¾) weakly holds the output to V_{CCIO} level.

Bus Hold

Each MAX 7000B device I/O pin provides an optional bus-hold feature. When this feature is enabled for an I/O pin, the bus-hold circuitry weakly holds the signal at its last driven state. By holding the last driven state of the pin until the next input signals is present, the bus-hold feature can eliminate the need to add external pull-up or pull-down resistors to hold a signal level when the bus is tri-stated. The bus-hold circuitry also pulls undriven pins away from the input threshold voltage where noise can cause unintended high-frequency switching. This feature can be selected individually for each I/O pin. The bus-hold output will drive no higher than V_{CCIO} to prevent overdriving signals. The propagation delays through the input and output buffers in MAX 7000B devices are not affected by whether the bus-hold feature is enabled or disabled.

The bus-hold circuitry weakly pulls the signal level to the last driven state through a resistor with a nominal resistance (R_{BH}) of approximately 8.5 k¾. Table 12 gives specific sustaining current that will be driven through this resistor and overdrive current that will identify the next driven input level. This information is provided for each VCCIO voltage level.

Table 12. Bus Hold Parameters

Parameter	Conditions	VCCIO Level						Units
		1.8 V		2.5 V		3.3 V		
		Min	Max	Min	Max	Min	Max	
Low sustaining current	$V_{IN} > V_{IL}$ (max)	30		50		70		µA
High sustaining current	$V_{IN} < V_{IH}$ (min)	–30		–50		–70		µA
Low overdrive current	$0 V < V_{IN} < V_{CCIO}$		200		300		500	µA
High overdrive current	$0 V < V_{IN} < V_{CCIO}$		–295		–435		–680	µA

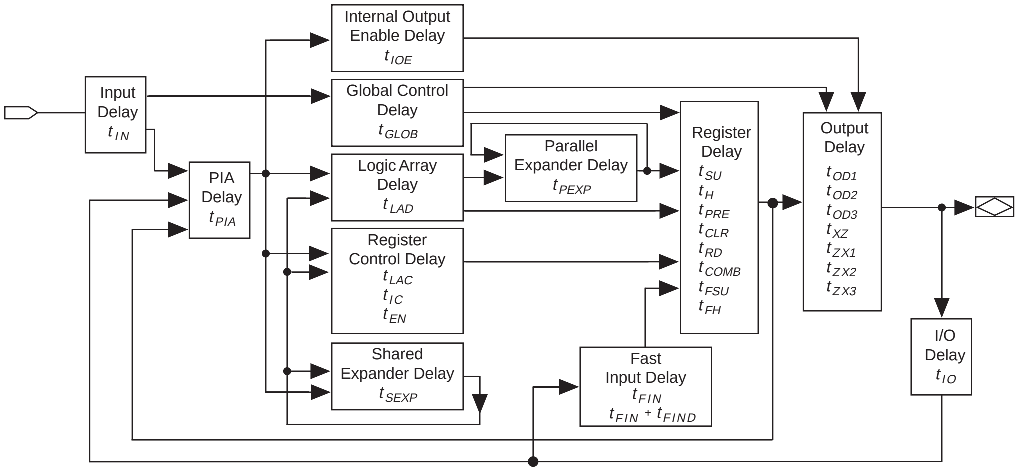
The bus-hold circuitry is active only during user operation. At power-up, the bus-hold circuit initializes its initial hold value as V_{CC} approaches the recommended operation conditions. When transitioning from ISP to User Mode with bus hold enabled, the bus-hold circuit captures the value present on the pin at the end of programming.

Table 16. MAX 7000B Device DC Operating Conditions *Note (4)*					
Symbol	**Parameter**	**Conditions**	**Min**	**Max**	**Unit**
V_{IH}	High-level input voltage for 3.3-V TTL/CMOS		2.0	3.9	V
	High-level input voltage for 2.5-V TTL/CMOS		1.7	3.9	V
	High-level input voltage for 1.8-V TTL/CMOS		0.65 × V_{CCIO}	3.9	V
V_{IL}	Low-level input voltage for 3.3-V TTL/CMOS and PCI compliance		–0.5	0.8	V
	Low-level input voltage for 2.5-V TTL/CMOS		–0.5	0.7	V
	Low-level input voltage for 1.8-V TTL/CMOS		–0.5	0.35 × V_{CCIO}	
V_{OH}	3.3-V high-level TTL output voltage	I_{OH} = –8 mA DC, V_{CCIO} = 3.00 V *(5)*	2.4		V
	3.3-V high-level CMOS output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 3.00 V *(5)*	V_{CCIO} – 0.2		V
	2.5-V high-level output voltage	I_{OH} = –100 µA DC, V_{CCIO} = 2.30 V *(5)*	2.1		V
		I_{OH} = –1 mA DC, V_{CCIO} = 2.30 V *(5)*	2.0		V
		I_{OH} = –2 mA DC, V_{CCIO} = 2.30 V *(5)*	1.7		V
	1.8-V high-level output voltage	I_{OH} = –2 mA DC, V_{CCIO} =1.65 V *(5)*	1.2		V
V_{OL}	3.3-V low-level TTL output voltage	I_{OL} = 8 mA DC, V_{CCIO} = 3.00 V *(6)*		0.4	V
	3.3-V low-level CMOS output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 3.00 V *(6)*		0.2	V
	2.5-V low-level output voltage	I_{OL} = 100 µA DC, V_{CCIO} = 2.30 V *(6)*		0.2	V
		I_{OL} = 1 mA DC, V_{CCIO} = 2.30 V *(6)*		0.4	V
		I_{OL} = 2 mA DC, V_{CCIO} = 2.30 V *(6)*		0.7	V
	1.8-V low-level output voltage	I_{OL} = 2 mA DC, V_{CCIO} = 1.7 V *(6)*		0.4	V
I_I	Input leakage current	V_I = –0.5 to 3.9 V *(7)*	–10	10	µA
I_{OZ}	Tri-state output off-state current	V_I = –0.5 to 3.9 V *(7)*	–10	10	µA
R_{ISP}	Value of I/O pin pull-up resistor during in-system programming or during power up	V_{CCIO} = 1.7 to 3.6 V *(8)*	20	74	k¾

Timing Model

MAX 7000B device timing can be analyzed with the Altera software, with a variety of popular industry-standard EDA simulators and timing analyzers, or with the timing model shown in Figure 13. MAX 7000B devices have predictable internal delays that enable the designer to determine the worst-case timing of any design. The Altera software provides timing simulation, point-to-point delay prediction, and detailed timing analysis for device-wide performance evaluation.

Figure 13. MAX 7000B Timing Model



The timing characteristics of any signal path can be derived from the timing model and parameters of a particular device. External timing parameters, which represent pin-to-pin timing delays, can be calculated as the sum of internal parameters. Figure 14 shows the timing relationship between internal and external delay parameters.

See *Application Note 94 (Understanding MAX 7000 Timing)* for more information.

Table 20. EPM7032B Selectable I/O Standard Timing Adder Delays *Notes (1)*								
I/O Standard	**Parameter**	**Speed Grade**						**Unit**
		-3.5		**-5.0**		**-7.5**		
		Min	**Max**	**Min**	**Max**	**Min**	**Max**	
3.3 V TTL/CMOS	Input to (PIA)		0.0		0.0		0.0	ns
	Input to global clock and clear		0.0		0.0		0.0	ns
	Input to fast input register		0.0		0.0		0.0	ns
	All outputs		0.0		0.0		0.0	ns
2.5 V TTL/CMOS	Input to PIA		0.3		0.4		0.6	ns
	Input to global clock and clear		0.3		0.4		0.6	ns
	Input to fast input register		0.2		0.3		0.4	ns
	All outputs		0.2		0.3		0.4	ns
1.8 V TTL/CMOS	Input to PIA		0.5		0.8		1.1	ns
	Input to global clock and clear		0.5		0.8		1.1	ns
	Input to fast input register		0.4		0.5		0.8	ns
	All outputs		1.2		1.8		2.6	ns
SSTL-2 Class I	Input to PIA		1.3		1.9		2.8	ns
	Input to global clock and clear		1.2		1.8		2.6	ns
	Input to fast input register		0.9		1.3		1.9	ns
	All outputs		0.0		0.0		0.0	ns
SSTL-2 Class II	Input to PIA		1.3		1.9		2.8	ns
	Input to global clock and clear		1.2		1.8		2.6	ns
	Input to fast input register		0.9		1.3		1.9	ns
	All outputs		–0.1		–0.1		–0.2	ns
SSTL-3 Class I	Input to PIA		1.2		1.8		2.6	ns
	Input to global clock and clear		0.9		1.3		1.9	ns
	Input to fast input register		0.8		1.1		1.7	ns
	All outputs		0.0		0.0		0.0	ns
SSTL-3 Class II	Input to PIA		1.2		1.8		2.6	ns
	Input to global clock and clear		0.9		1.3		1.9	ns
	Input to fast input register		0.8		1.1		1.7	ns
	All outputs		0.0		0.0		0.0	ns
GTL+	Input to PIA		1.6		2.3		3.4	ns
	Input to global clock and clear		1.6		2.3		3.4	ns
	Input to fast input register		1.5		2.1		3.2	ns
	All outputs		0.0		0.0		0.0	ns

Table 20. EPM7032B Selectable I/O Standard Timing Adder Delays *Notes (1)*								
I/O Standard	**Parameter**	**Speed Grade**						**Unit**
		-3.5		**-5.0**		**-7.5**		
		Min	**Max**	**Min**	**Max**	**Min**	**Max**	
PCI	Input to PIA		0.0		0.0		0.0	ns
	Input to global clock and clear		0.0		0.0		0.0	ns
	Input to fast input register		0.0		0.0		0.0	ns
	All outputs		0.0		0.0		0.0	ns

Notes to tables:

(1) These values are specified under the Recommended Operating Conditions in Table 15 on page 29. See Figure 14 for more information on switching waveforms.

(2) These values are specified for a PIA fan-out of all LABs.

(3) Measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.

(4) The t_{LPA} parameter must be added to the t_{LAD}, t_{LAC}, t_{IC}, t_{ACL}, t_{CPPW}, t_{EN}, and t_{SEXP} parameters for macrocells running in low-power mode.

Table 25. EPM7128B Internal Timing Parameters *Note (1)*									
Symbol	Parameter	Conditions	Speed Grade						Unit
			-4		-7		-10		
			Min	Max	Min	Max	Min	Max	
t_{IN}	Input pad and buffer delay			0.3		0.6		0.8	ns
t_{IO}	I/O input pad and buffer delay			0.3		0.6		0.8	ns
t_{FIN}	Fast input delay			1.3		2.9		3.7	ns
t_{FIND}	Programmable delay adder for fast input			1.0		1.5		1.5	ns
t_{SEXP}	Shared expander delay			1.5		2.8		3.8	ns
t_{PEXP}	Parallel expander delay			0.4		0.8		1.0	ns
t_{LAD}	Logic array delay			1.6		2.9		3.8	ns
t_{LAC}	Logic control array delay			1.4		2.6		3.4	ns
t_{IOE}	Internal output enable delay			0.1		0.3		0.4	ns
t_{OD1}	Output buffer and pad delay slow slew rate = off V_{CCIO} = 3.3 V	C1 = 35 pF		0.9		1.7		2.2	ns
t_{OD3}	Output buffer and pad delay slow slew rate = on V_{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		5.9		6.7		7.2	ns
t_{ZX1}	Output buffer enable delay slow slew rate = off V_{CCIO} = 3.3 V	C1 = 35 pF		1.8		3.3		4.4	ns
t_{ZX3}	Output buffer enable delay slow slew rate = on V_{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		6.8		8.3		9.4	ns
t_{XZ}	Output buffer disable delay	C1 = 5 pF		1.8		3.3		4.4	ns
t_{SU}	Register setup time		1.0		1.9		2.6		ns
t_{H}	Register hold time		0.4		0.8		1.1		ns
t_{FSU}	Register setup time of fast input		0.8		0.9		0.9		ns
t_{FH}	Register hold time of fast input		1.2		1.6		1.6		ns
t_{RD}	Register delay			0.5		1.1		1.4	ns
t_{COMB}	Combinatorial delay			0.2		0.3		0.4	ns
t_{IC}	Array clock delay			1.4		2.8		3.6	ns
t_{EN}	Register enable time			1.4		2.6		3.4	ns
t_{GLOB}	Global control delay			1.1		2.3		3.1	ns
t_{PRE}	Register preset time			1.0		1.9		2.6	ns
t_{CLR}	Register clear time			1.0		1.9		2.6	ns
t_{PIA}	PIA delay	*(2)*		1.0		2.0		2.8	ns
t_{LPA}	Low-power adder	*(4)*		1.5		2.8		3.8	ns

Table 27. EPM7256B External Timing Parameters *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			-5		-7		-10		
			Min	Max	Min	Max	Min	Max	
t_{PD1}	Input to non-registered output	C1 = 35 pF *(2)*		5.0		7.5		10.0	ns
t_{PD2}	I/O input to non-registered output	C1 = 35 pF *(2)*		5.0		7.5		10.0	ns
t_{SU}	Global clock setup time	*(2)*	3.3		4.8		6.6		ns
t_{H}	Global clock hold time	*(2)*	0.0		0.0		0.0		ns
t_{FSU}	Global clock setup time of fast input		1.0		1.5		1.5		ns
t_{FH}	Global clock hold time for fast input		1.0		1.0		1.0		ns
t_{FZHSU}	Global clock setup time of fast input with zero hold time		2.5		3.0		3.0		ns
t_{FZHH}	Global clock hold time of fast input with zero hold time		0.0		0.0		0.0		ns
t_{CO1}	Global clock to output delay	C1 = 35 pF	1.0	3.3	1.0	5.1	1.0	6.7	ns
t_{CH}	Global clock high time		2.0		3.0		4.0		ns
t_{CL}	Global clock low time		2.0		3.0		4.0		ns
t_{ASU}	Array clock setup time	*(2)*	1.4		2.0		2.8		ns
t_{AH}	Array clock hold time	*(2)*	0.4		0.8		1.0		ns
t_{ACO1}	Array clock to output delay	C1 = 35 pF *(2)*	1.0	5.2	1.0	7.9	1.0	10.5	ns
t_{ACH}	Array clock high time		2.0		3.0		4.0		ns
t_{ACL}	Array clock low time		2.0		3.0		4.0		ns
t_{CPPW}	Minimum pulse width for clear and preset		2.0		3.0		4.0		ns
t_{CNT}	Minimum global clock period	*(2)*		5.3		7.9		10.6	ns
f_{CNT}	Maximum internal global clock frequency	*(2)*, *(3)*	188.7		126.6		94.3		MHz
t_{ACNT}	Minimum array clock period	*(2)*		5.3		7.9		10.6	ns
f_{ACNT}	Maximum internal array clock frequency	*(2)*, *(3)*	188.7		126.6		94.3		MHz

Table 32. EPM7512B Selectable I/O Standard Timing Adder Delays (Part 2 of 2) Note (1)

I/O Standard	Parameter	Speed Grade						Unit
		-5		-7		-10		
		Min	Max	Min	Max	Min	Max	
PCI	Input to PIA		0.0		0.0		0.0	ns
	Input to global clock and clear		0.0		0.0		0.0	ns
	Input to fast input register		0.0		0.0		0.0	ns
	All outputs		0.0		0.0		0.0	ns

Notes to tables:

(1) These values are specified under the Recommended Operating Conditions in Table 15 on page 29. See Figure 14 for more information on switching waveforms.

(2) These values are specified for a PIA fan-out of one LAB (16 macrocells). For each additional LAB fan-out in these devices, add an additional 0.12 ns to the PIA timing value.

(3) Measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.

(4) The t_{LPA} parameter must be added to the t_{LAD}, t_{LAC}, t_{IC}, t_{ACL}, t_{CPPW}, t_{EN}, and t_{SEXP} parameters for macrocells running in low-power mode.

Power Consumption

Supply power (P) versus frequency ($\mathbf{f_{MAX}}$, in MHz) for MAX 7000B devices is calculated with the following equation:

$$P = P_{INT} + P_{IO} = I_{CCINT} \times V_{CC} + P_{IO}$$

The P_{IO} value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

The I_{CCINT} value depends on the switching frequency and the application logic. The I_{CCINT} value is calculated with the following equation:

I_{CCINT} =

$$(A \times MC_{TON}) + [B \times (MC_{DEV} - MC_{TON})] + (C \times MC_{USED} \times \mathbf{f_{MAX}} \times \mathbf{tog_{LC}})$$

The parameters in this equation are:

MC_{TON}	=	Number of macrocells with the Turbo Bit™ option turned on, as reported in the MAX+PLUS II Report File (**.rpt**)
MC_{DEV}	=	Number of macrocells in the device
MC_{USED}	=	Total number of macrocells in the design, as reported in the Report File
$\mathbf{f_{MAX}}$	=	Highest clock frequency to the device
$\mathbf{tog_{LC}}$	=	Average percentage of logic cells toggling at each clock (typically 12.5%)
A, B, C	=	Constants, shown in Table 33

Table 33. MAX 7000B I_{CC} Equation Constants

Device	A	B	C
EPM7032B	0.91	0.54	0.010
EPM7064B	0.91	0.54	0.012
EPM7128B	0.91	0.54	0.016
EPM7256B	0.91	0.54	0.017
EPM7512B	0.91	0.54	0.019

This calculation provides an I_{CC} estimate based on typical conditions using a pattern of a 16-bit, loadable, enabled, up/down counter in each LAB with no output load. Actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

Figure 21. 48-Pin VTQFP Package Pin-Out Diagram

Package outlines not drawn to scale.

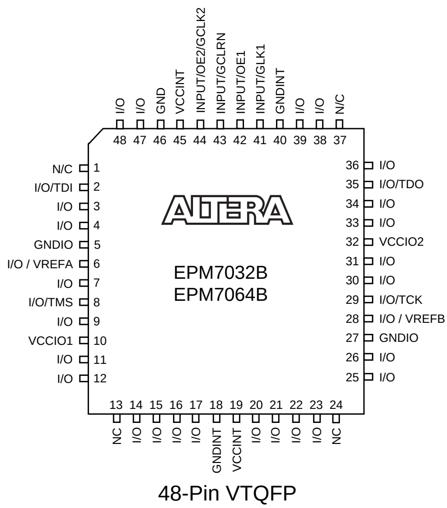


Figure 22. 49-Pin Ultra FineLine BGA Package Pin-Out Diagram

Package outline not drawn to scale.

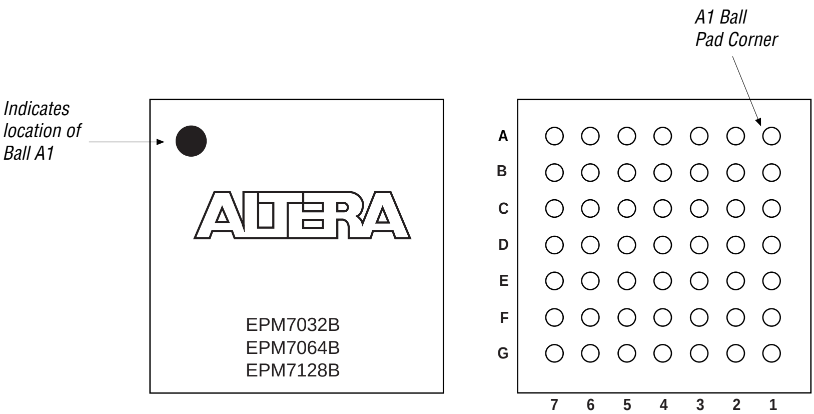


Figure 23. 100-Pin TQFP Package Pin-Out Diagram

Package outline not drawn to scale.

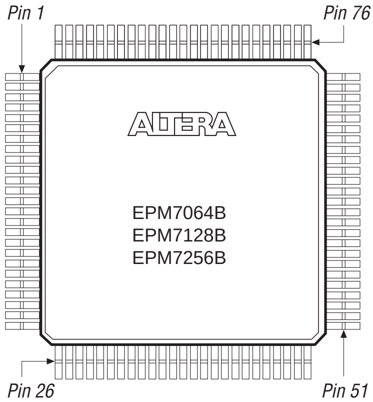


Figure 24. 100-Pin FineLine BGA Package Pin-Out Diagram

Package outline not drawn to scale.

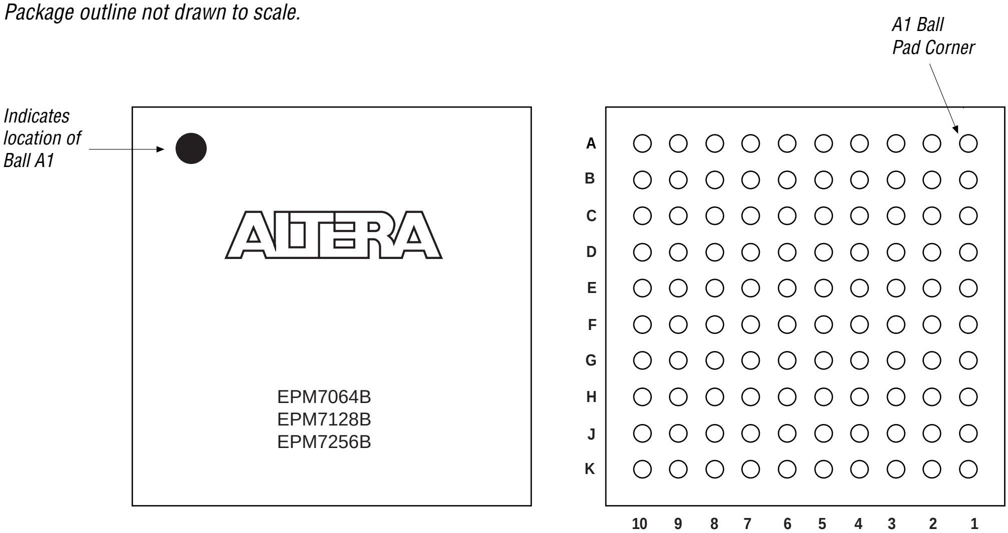
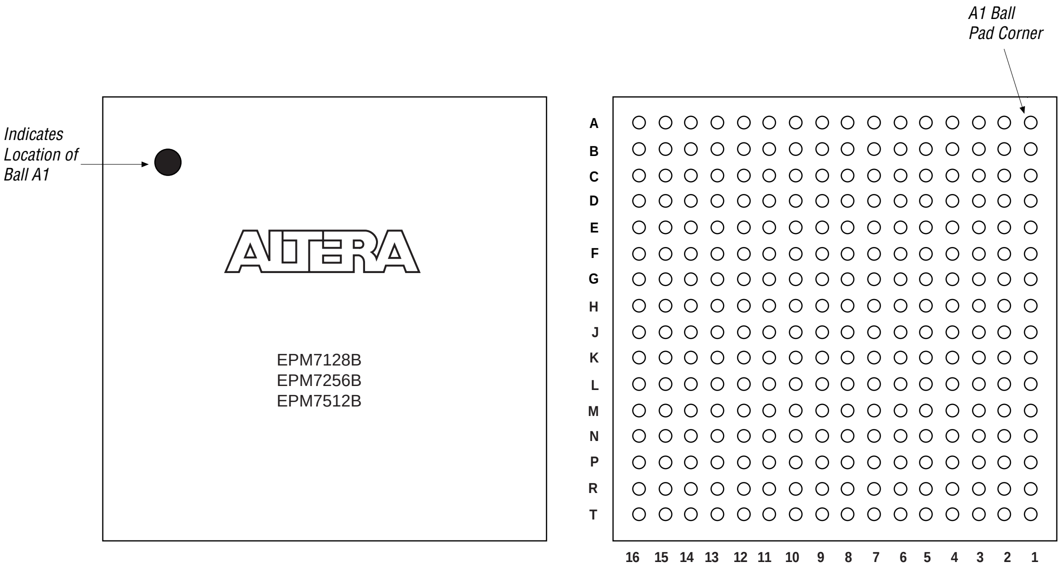


Figure 29. 256-Pin FineLine BGA Package Pin-Out Diagram

Package outline not drawn to scale.



Revision History

The information contained in the *MAX 7000B Programmable Logic Device Family Data Sheet* version 3.5 supersedes information published in previous versions.

Version 3.5

The following changes were made to the *MAX 7000B Programmable Logic Device Family Data Sheet* version 3.5:

- Updated Figure 28.

Version 3.4

The following changes were made to the *MAX 7000B Programmable Logic Device Family Data Sheet* version 3.4:

- Updated text in the "Power Sequencing & Hot-Socketing" section.