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**Understanding Embedded - CPLDs (Complex Programmable Logic Devices)** 

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

### **Applications of Embedded - CPLDs**

Details	
Product Status	Obsolete
Programmable Type	In System Programmable
Delay Time tpd(1) Max	5 ns
Voltage Supply - Internal	2.375V ~ 2.625V
Number of Logic Elements/Blocks	16
Number of Macrocells	256
Number of Gates	5000
Number of I/O	120
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm7256btc144-5

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Table 3. MAX	7000B	Maximu	m User i	I/O Pins	Note (	(1)					
Device	44-Pin PLCC	44-Pin TQFP	48-Pin TQFP (2)	49-Pin 0.8-mm Ultra FineLine BGA (3)	100- Pin TQFP	100-Pin FineLine BGA (4)	144- Pin TQFP	169-Pin 0.8-mm Ultra FineLine BGA (3)	208- Pin PQFP	256- Pin BGA	256-Pin FineLine BGA (4)
EPM7032B	36	36	36	36							
EPM7064B	36	36	40	41	68	68					
EPM7128B				41	84	84	100	100			100
EPM7256B					84		120	141	164		164
EPM7512B							120	141	176	212	212

#### Notes:

- When the IEEE Std. 1149.1 (JTAG) interface is used for in-system programming or boundary-scan testing, four I/O pins become JTAG pins.
- (2) Contact Altera for up-to-date information on available device package options.
- (3) All 0.8-mm Ultra FineLine BGA packages are footprint-compatible via the SameFrame<sup>TM</sup> pin-out feature. Therefore, designers can design a board to support a variety of devices, providing a flexible migration path across densities and pin counts. Device migration is fully supported by Altera development tools. See "SameFrame Pin-Outs" on page 14 for more details.
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MAX 7000B devices use CMOS EEPROM cells to implement logic functions. The user-configurable MAX 7000B architecture accommodates a variety of independent combinatorial and sequential logic functions. The devices can be reprogrammed for quick and efficient iterations during design development and debug cycles, and can be programmed and erased up to 100 times.

MAX 7000B devices contain 32 to 512 macrocells that are combined into groups of 16 macrocells, called logic array blocks (LABs). Each macrocell has a programmable-AND/fixed-OR array and a configurable register with independently programmable clock, clock enable, clear, and preset functions. To build complex logic functions, each macrocell can be supplemented with both shareable expander product terms and high-speed parallel expander product terms to provide up to 32 product terms per macrocell.

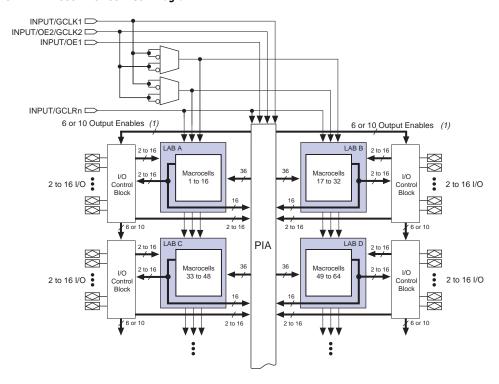


Figure 1. MAX 7000B Device Block Diagram

#### Note:

(1) EPM7032B, EPM7064B, EPM7128B, and EPM7256B devices have six output enables. EPM7512B devices have ten output enables.

# **Logic Array Blocks**

The MAX 7000B device architecture is based on the linking of high-performance LABs. LABs consist of 16 macrocell arrays, as shown in Figure 1. Multiple LABs are linked together via the PIA, a global bus that is fed by all dedicated input pins, I/O pins, and macrocells.

Each LAB is fed by the following signals:

- 36 signals from the PIA that are used for general logic inputs
- Global controls that are used for secondary register functions
- Direct input paths from I/O pins to the registers that are used for fast setup times

The Altera development system automatically optimizes product-term allocation according to the logic requirements of the design.

For registered functions, each macrocell flipflop can be individually programmed to implement D, T, JK, or SR operation with programmable clock control. The flipflop can be bypassed for combinatorial operation. During design entry, the designer specifies the desired flipflop type; the MAX+PLUS II software then selects the most efficient flipflop operation for each registered function to optimize resource utilization.

Each programmable register can be clocked in three different modes:

- Global clock signal. This mode achieves the fastest clock-to-output performance.
- Global clock signal enabled by an active-high clock enable. A clock enable is generated by a product term. This mode provides an enable on each flipflop while still achieving the fast clock-to-output performance of the global clock.
- Array clock implemented with a product term. In this mode, the flipflop can be clocked by signals from buried macrocells or I/O pins.

Two global clock signals are available in MAX 7000B devices. As shown in Figure 1, these global clock signals can be the true or the complement of either of the global clock pins, GCLK1 or GCLK2.

Each register also supports asynchronous preset and clear functions. As shown in Figure 2, the product-term select matrix allocates product terms to control these operations. Although the product-term-driven preset and clear from the register are active high, active-low control can be obtained by inverting the signal within the logic array. In addition, each register clear function can be individually driven by the active-low dedicated global clear pin (GCLRn). Upon power-up, each register in a MAX 7000B device may be set to either a high or low state. This power-up state is specified at design entry.

All MAX 7000B I/O pins have a fast input path to a macrocell register. This dedicated path allows a signal to bypass the PIA and combinatorial logic and be clocked to an input D flipflop with an extremely fast input setup time. The input path from the I/O pin to the register has a programmable delay element that can be selected to either guarantee zero hold time or to get the fastest possible set-up time (as fast as 1.0 ns).

## Parallel Expanders

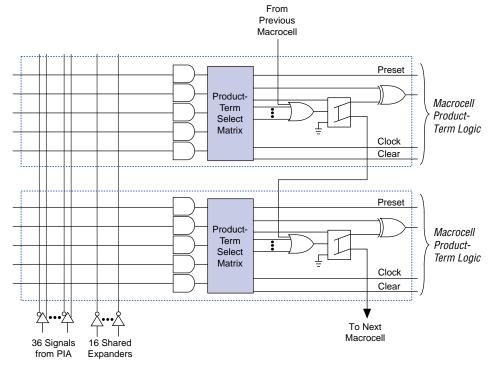
Parallel expanders are unused product terms that can be allocated to a neighboring macrocell to implement fast, complex logic functions. Parallel expanders allow up to 20 product terms to directly feed the macrocell OR logic, with five product terms provided by the macrocell and 15 parallel expanders provided by neighboring macrocells in the LAB.

The Altera Compiler can automatically allocate up to three sets of up to five parallel expanders to the macrocells that require additional product terms. Each set of five parallel expanders incurs a small, incremental timing delay ( $t_{PEXP}$ ). For example, if a macrocell requires 14 product terms, the Compiler uses the five dedicated product terms within the macrocell and allocates two sets of parallel expanders; the first set includes five product terms and the second set includes four product terms, increasing the total delay by  $2 \times t_{PEXP}$ .

Two groups of eight macrocells within each LAB (e.g., macrocells 1 through 8, and 9 through 16) form two chains to lend or borrow parallel expanders. A macrocell borrows parallel expanders from lower-numbered macrocells. For example, macrocell 8 can borrow parallel expanders from macrocell 7, from macrocells 7 and 6, or from macrocells 7, 6, and 5. Within each group of eight, the lowest-numbered macrocell can only lend parallel expanders and the highest-numbered macrocell can only borrow them. Figure 4 shows how parallel expanders can be borrowed from a neighboring macrocell.

Figure 4. MAX 7000B Parallel Expanders

Unused product terms in a macrocell can be allocated to a neighboring macrocell.



## **Programmable Interconnect Array**

Logic is routed between LABs on the PIA. This global bus is a programmable path that connects any signal source to any destination on the device. All MAX 7000B dedicated inputs, I/O pins, and macrocell outputs feed the PIA, which makes the signals available throughout the entire device. Only the signals required by each LAB are actually routed from the PIA into the LAB. Figure 5 shows how the PIA signals are routed into the LAB. An EEPROM cell controls one input to a two-input AND gate, which selects a PIA signal to drive into the LAB.

The instruction register length of MAX 7000B devices is ten bits. The MAX 7000B USERCODE register length is 32 bits. Tables 7 and 8 show the boundary-scan register length and device IDCODE information for MAX 7000B devices.

Table 7. MAX 7000B Boundary-Sca	n Register Length
Device	Boundary-Scan Register Length
EPM7032B	96
EPM7064B	192
EPM7128B	288
EPM7256B	480
EPM7512B	624

Table 8. 32-1	Bit MAX 700	OOB Device IDCODE No.	te (1)	
Device		IDCODE (32 I	Bits)	
	Version (4 Bits)	Part Number (16 Bits)	Manufacturer's Identity (11 Bits)	1 (1 Bit) (2)
EPM7032B	0010	0111 0000 0011 0010	00001101110	1
EPM7064B	0010	0111 0000 0110 0100	00001101110	1
EPM7128B	0010	0111 0001 0010 1000	00001101110	1
EPM7256B	0010	0111 0010 0101 0110	00001101110	1
EPM7512B	0010	0111 0101 0001 0010	00001101110	1

#### Notes:

- (1) The most significant bit (MSB) is on the left.
- (2) The least significant bit (LSB) for all JTAG IDCODEs is 1.



See Application Note 39 (IEEE 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices) for more information on JTAG boundary-scan testing.

Figure 8 shows the timing information for the JTAG signals.

# Programmable Speed/Power Control

MAX 7000B devices offer a power-saving mode that supports low-power operation across user-defined signal paths or the entire device. This feature allows total power dissipation to be reduced by 50% or more, because most logic applications require only a small fraction of all gates to operate at maximum frequency.

The designer can program each individual macrocell in a MAX 7000B device for either high-speed or low-power operation. As a result, speed-critical paths in the design can run at high speed, while the remaining paths can operate at reduced power. Macrocells that run at low power incur a nominal timing delay adder ( $t_{LPA}$ ) for the  $t_{LAD}$ ,  $t_{LAC}$ ,  $t_{IC}$ ,  $t_{ACL}$ ,  $t_{CPPW}$ ,  $t_{EN}$ , and  $t_{SEXP}$  parameters.

# Output Configuration

MAX 7000B device outputs can be programmed to meet a variety of system-level requirements.

### MultiVolt I/O Interface

The MAX 7000B device architecture supports the MultiVolt I/O interface feature, which allows MAX 7000B devices to connect to systems with differing supply voltages. MAX 7000B devices in all packages can be set for 3.3-V, 2.5-V, or 1.8-V pin operation. These devices have one set of  $\rm V_{CC}$  pins for internal operation and input buffers (VCCINT), and another set for I/O output drivers (VCCIO).

The VCCIO pins can be connected to either a 3.3-V, 2.5-V, or 1.8-V power supply, depending on the output requirements. When the VCCIO pins are connected to a 1.8-V power supply, the output levels are compatible with 1.8-V systems. When the VCCIO pins are connected to a 2.5-V power supply, the output levels are compatible with 2.5-V systems. When the VCCIO pins are connected to a 3.3-V power supply, the output high is at 3.3 V and is therefore compatible with 3.3-V or 5.0-V systems. Devices operating with  $V_{\rm CCIO}$  levels of 2.5 V or 1.8 V incur a nominal timing delay adder.

Table 10 describes the MAX 7000B MultiVolt I/O support.

Table 10. MAX 700	Table 10. MAX 7000B MultiVolt I/O Support									
V <sub>CCIO</sub> (V)		Input Si	ignal (V)			Output S	Signal (V)			
	1.8	2.5	3.3	5.0	1.8	2.5	3.3	5.0		
1.8	<b>✓</b>	<b>✓</b>	<b>✓</b>		<b>✓</b>					
2.5	<b>✓</b>	<b>✓</b>	<b>✓</b>			<b>✓</b>				
3.3	<b>✓</b>	<b>✓</b>	<b>✓</b>				<b>✓</b>	<b>✓</b>		

## **Open-Drain Output Option**

MAX 7000B devices provide an optional open-drain (equivalent to open-collector) output for each I/O pin. This open-drain output enables the device to provide system-level control signals (e.g., interrupt and write enable signals) that can be asserted by any of several devices. It can also provide an additional wired-OR plane.

## **Programmable Ground Pins**

Each unused I/O pin on MAX 7000B devices may be used as an additional ground pin. This programmable ground feature does not require the use of the associated macrocell; therefore, the buried macrocell is still available for user logic.

#### Slew-Rate Control

The output buffer for each MAX 7000B I/O pin has an adjustable output slew rate that can be configured for low-noise or high-speed performance. A faster slew rate provides high-speed transitions for high-performance systems. However, these fast transitions may introduce noise transients into the system. A slow slew rate reduces system noise, but adds a nominal delay of 4 to 5 ns. When the configuration cell is turned off, the slew rate is set for low-noise performance. Each I/O pin has an individual EEPROM bit that controls the slew rate, allowing designers to specify the slew rate on a pin-by-pin basis. The slew rate control affects both the rising and falling edges of the output signal.

# Advanced I/O Standard Support

The MAX 7000B I/O pins support the following I/O standards: LVTTL, LVCMOS, 1.8-V I/O, 2.5-V I/O, GTL+, SSTL-3 Class I and II, and SSTL-2 Class I and II.

## **Programmable Pull-Up Resistor**

Each MAX 7000B device I/O pin provides an optional programmable pull-up resistor during user mode. When this feature is enabled for an I/O pin, the pull-up resistor (typically 50 k¾) weakly holds the output to  $V_{\rm CCIO}$  level.

## **Bus Hold**

Each MAX 7000B device I/O pin provides an optional bus-hold feature. When this feature is enabled for an I/O pin, the bus-hold circuitry weakly holds the signal at its last driven state. By holding the last driven state of the pin until the next input signals is present, the bus-hold feature can eliminate the need to add external pull-up or pull-down resistors to hold a signal level when the bus is tri-stated. The bus-hold circuitry also pulls undriven pins away from the input threshold voltage where noise can cause unintended high-frequency switching. This feature can be selected individually for each I/O pin. The bus-hold output will drive no higher than  $V_{\rm CCIO}$  to prevent overdriving signals. The propagation delays through the input and output buffers in MAX 7000B devices are not affected by whether the bus-hold feature is enabled or disabled.

The bus-hold circuitry weakly pulls the signal level to the last driven state through a resistor with a nominal resistance ( $R_{BH}$ ) of approximately 8.5 k¾. Table 12 gives specific sustaining current that will be driven through this resistor and overdrive current that will identify the next driven input level. This information is provided for each VCCIO voltage level.

Table 12. Bus Hold Par	ameters							
Parameter	Conditions			VCCIO	Level			Units
		1.	8 V	2.	5 V	3.3	3 V	
		Min	Max	Min	Max	Min	Max	
Low sustaining current	$V_{IN} > V_{IL} (max)$	30		50		70		μΑ
High sustaining current	V <sub>IN</sub> < V <sub>IH</sub> (min)	-30		-50		-70		μΑ
Low overdrive current	0 V < V <sub>IN</sub> < V <sub>CCIO</sub>		200		300		500	μΑ
High overdrive current	$0 \text{ V} < \text{V}_{\text{IN}} < \text{V}_{\text{CCIO}}$		-295		-435		-680	μΑ

The bus-hold circuitry is active only during user operation. At power-up, the bus-hold circuit initializes its initial hold value as  $V_{CC}$  approaches the recommended operation conditions. When transitioning from ISP to User Mode with bus hold enabled, the bus-hold circuit captures the value present on the pin at the end of programming.

# Power Sequencing & Hot-Socketing

Because MAX 7000B devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The  $V_{\rm CCIO}$  and  $V_{\rm CCINT}$  power planes can be powered in any order.

Signals can be driven into MAX 7000B devices before and during power-up (and power-down) without damaging the device. Additionally, MAX 7000B devices do not drive out during power-up. Once operating conditions are reached, MAX 7000B devices operate as specified by the user.

MAX 7000B device I/O pins will not source or sink more than 300  $\mu$ A of DC current during power-up. All pins can be driven up to 4.1 V during hot-socketing.

# **Design Security**

All MAX 7000B devices contain a programmable security bit that controls access to the data programmed into the device. When this bit is programmed, a design implemented in the device cannot be copied or retrieved. This feature provides a high level of design security, because programmed data within EEPROM cells is invisible. The security bit that controls this function, as well as all other programmed data, is reset only when the device is reprogrammed.

# **Generic Testing**

MAX 7000B devices are fully functionally tested. Complete testing of each programmable EEPROM bit and all internal logic elements ensures 100% programming yield. AC test measurements are taken under conditions equivalent to those shown in Figure 11. Test patterns can be used and then erased during early stages of the production flow.

Table 1	6. MAX 7000B Device DC Opera	ating Conditions Note (4)			
Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>IH</sub>	High-level input voltage for 3.3-V TTL/CMOS		2.0	3.9	V
	High-level input voltage for 2.5-V TTL/CMOS		1.7	3.9	V
	High-level input voltage for 1.8-V TTL/CMOS		0.65 × V <sub>CCIO</sub>	3.9	V
V <sub>IL</sub>	Low-level input voltage for 3.3-V TTL/CMOS and PCI compliance		-0.5	0.8	V
	Low-level input voltage for 2.5-V TTL/CMOS		-0.5	0.7	V
	Low-level input voltage for 1.8-V TTL/CMOS		-0.5	0.35 × V <sub>CCIO</sub>	
V <sub>OH</sub>	3.3-V high-level TTL output voltage	$I_{OH} = -8 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (5)$	2.4		V
	3.3-V high-level CMOS output voltage	$I_{OH} = -0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (5)$	V <sub>CCIO</sub> - 0.2		V
	2.5-V high-level output voltage	$I_{OH} = -100 \mu A DC, V_{CCIO} = 2.30 V (5)$	2.1		V
		$I_{OH} = -1 \text{ mA DC}, V_{CCIO} = 2.30 \text{ V } (5)$	2.0		V
		$I_{OH} = -2 \text{ mA DC}, V_{CCIO} = 2.30 \text{ V } (5)$	1.7		V
	1.8-V high-level output voltage	$I_{OH} = -2 \text{ mA DC}, V_{CCIO} = 1.65 \text{ V } (5)$	1.2		V
$V_{OL}$	3.3-V low-level TTL output voltage	I <sub>OL</sub> = 8 mA DC, V <sub>CCIO</sub> = 3.00 V (6)		0.4	V
	3.3-V low-level CMOS output voltage	$I_{OL} = 0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (6)$		0.2	V
	2.5-V low-level output voltage	$I_{OL}$ = 100 $\mu$ A DC, $V_{CCIO}$ = 2.30 $V$ (6)		0.2	V
		$I_{OL}$ = 1 mA DC, $V_{CCIO}$ = 2.30 V (6)		0.4	V
		$I_{OL}$ = 2 mA DC, $V_{CCIO}$ = 2.30 V (6)		0.7	V
	1.8-V low-level output voltage	I <sub>OL</sub> = 2 mA DC, V <sub>CCIO</sub> = 1.7 V (6)		0.4	V
I <sub>I</sub>	Input leakage current	$V_1 = -0.5 \text{ to } 3.9 \text{ V } (7)$	-10	10	μΑ
I <sub>OZ</sub>	Tri-state output off-state current	$V_1 = -0.5 \text{ to } 3.9 \text{ V } (7)$	-10	10	μΑ
R <sub>ISP</sub>	Value of I/O pin pull-up resistor during in-system programming or during power up	V <sub>CCIO</sub> = 1.7 to 3.6 V (8)	20	74	k¾

Tables 18 through 32 show MAX 7000B device timing parameters.

Symbol	Parameter	Conditions			Speed	Grade			Unit
			-3	.5	-5	.0	-7	.5	
			Min	Max	Min	Max	Min	Max	
t <sub>PD1</sub>	Input to non-registered output	C1 = 35 pF (2)		3.5		5.0		7.5	ns
t <sub>PD2</sub>	I/O input to non-registered output	C1 = 35 pF (2)		3.5		5.0		7.5	ns
t <sub>SU</sub>	Global clock setup time	(2)	2.1		3.0		4.5		ns
t <sub>H</sub>	Global clock hold time	(2)	0.0		0.0		0.0		ns
t <sub>FSU</sub>	Global clock setup time of fast input		1.0		1.0		1.5		ns
t <sub>FH</sub>	Global clock hold time of fast input		1.0		1.0		1.0		ns
t <sub>FZHSU</sub>	Global clock setup time of fast input with zero hold time		2.0		2.5		3.0		ns
t <sub>FZHH</sub>	Global clock hold time of fast input with zero hold time		0.0		0.0		0.0		ns
t <sub>CO1</sub>	Global clock to output delay	C1 = 35 pF	1.0	2.4	1.0	3.4	1.0	5.0	ns
t <sub>CH</sub>	Global clock high time		1.5		2.0		3.0		ns
t <sub>CL</sub>	Global clock low time		1.5		2.0		3.0		ns
t <sub>ASU</sub>	Array clock setup time	(2)	0.9		1.3		1.9		ns
t <sub>AH</sub>	Array clock hold time	(2)	0.2		0.3		0.6		ns
t <sub>ACO1</sub>	Array clock to output delay	C1 = 35 pF (2)	1.0	3.6	1.0	5.1	1.0	7.6	ns
t <sub>ACH</sub>	Array clock high time		1.5		2.0		3.0		ns
t <sub>ACL</sub>	Array clock low time		1.5		2.0		3.0		ns
t <sub>CPPW</sub>	Minimum pulse width for clear and preset		1.5		2.0		3.0		ns
t <sub>CNT</sub>	Minimum global clock period	(2)		3.3		4.7		7.0	ns
f <sub>CNT</sub>	Maximum internal global clock frequency	(2), (3)	303.0		212.8		142.9		MHz
t <sub>ACNT</sub>	Minimum array clock period	(2)		3.3		4.7		7.0	ns
f <sub>ACNT</sub>	Maximum internal array clock frequency	(2), (3)	303.0		212.8		142.9		MHz

Symbol	Parameter	Conditions			Speed	Grade			Unit
			-	3	-	5	-	7	
			Min	Max	Min	Max	Min	Max	
t <sub>IN</sub>	Input pad and buffer delay			0.3		0.5		0.7	ns
$t_{IO}$	I/O input pad and buffer delay			0.3		0.5		0.7	ns
t <sub>FIN</sub>	Fast input delay			0.9		1.3		2.0	ns
t <sub>FIND</sub>	Programmable delay adder for fast input			1.0		1.5		1.5	ns
t <sub>SEXP</sub>	Shared expander delay			1.5		2.1		3.2	ns
t <sub>PEXP</sub>	Parallel expander delay			0.4		0.6		0.9	ns
$t_{LAD}$	Logic array delay			1.4		2.0		3.1	ns
$t_{LAC}$	Logic control array delay			1.2		1.7		2.6	ns
t <sub>IOE</sub>	Internal output enable delay			0.1		0.2		0.3	ns
t <sub>OD1</sub>	Output buffer and pad delay slow slew rate = off V <sub>CCIO</sub> = 3.3 V	C1 = 35 pF		0.9		1.2		1.8	ns
t <sub>OD3</sub>	Output buffer and pad delay slow slew rate = on V <sub>CCIO</sub> = 2.5 V or 3.3 V	C1 = 35 pF		5.9		6.2		6.8	ns
t <sub>ZX1</sub>	Output buffer enable delay slow slew rate = off V <sub>CCIO</sub> = 3.3 V	C1 = 35 pF		1.6		2.2		3.4	ns
t <sub>ZX3</sub>	Output buffer enable delay slow slew rate = on V <sub>CCIO</sub> = 2.5 V or 3.3 V	C1 = 35 pF		6.6		7.2		8.4	ns
$t_{XZ}$	Output buffer disable delay	C1 = 5 pF		1.6		2.2		3.4	ns
$t_{SU}$	Register setup time		0.7		1.1		1.6		ns
$t_H$	Register hold time		0.4		0.5		0.9		ns
t <sub>FSU</sub>	Register setup time of fast input		0.8		0.8		1.1		ns
$t_{FH}$	Register hold time of fast input		1.2		1.2		1.4		ns
$t_{RD}$	Register delay			0.5		0.6		0.9	ns
$t_{COMB}$	Combinatorial delay			0.2		0.3		0.5	ns
t <sub>IC</sub>	Array clock delay		İ	1.2		1.8		2.8	ns
$t_{EN}$	Register enable time		İ	1.2		1.7		2.6	ns
$t_{GLOB}$	Global control delay		İ	0.7		1.1		1.6	ns
$t_{PRE}$	Register preset time			1.0		1.3		1.9	ns
t <sub>CLR</sub>	Register clear time			1.0		1.3		1.9	ns
$t_{PIA}$	PIA delay	(2)		0.7		1.0		1.4	ns
$t_{LPA}$	Low-power adder	(4)		1.5		2.1		3.2	ns

I/O Standard	Parameter			Speed	Grade			Unit
		-	3	-	·5	-	7	
		Min	Max	Min	Max	Min	Max	
3.3 V TTL/CMOS	Input to PIA		0.0		0.0		0.0	ns
	Input to global clock and clear		0.0		0.0		0.0	ns
	Input to fast input register		0.0		0.0		0.0	ns
	All outputs		0.0		0.0		0.0	ns
2.5 V TTL/CMOS	Input to PIA		0.3		0.4		0.6	ns
	Input to global clock and clear		0.3		0.4		0.6	ns
	Input to fast input register		0.2		0.3		0.4	ns
	All outputs		0.2		0.3		0.4	ns
1.8 V TTL/CMOS	Input to PIA		0.5		0.7		1.1	ns
	Input to global clock and clear		0.5		0.7		1.1	ns
	Input to fast input register		0.4		0.6		0.9	ns
	All outputs		1.2		1.7		2.6	ns
SSTL-2 Class I	Input to PIA		1.3		1.9		2.8	ns
	Input to global clock and clear		1.2		1.7		2.6	ns
	Input to fast input register		0.9		1.3		1.9	ns
	All outputs		0.0		0.0		0.0	ns
SSTL-2 Class II	Input to PIA		1.3		1.9		2.8	ns
	Input to global clock and clear		1.2		1.7		2.6	ns
	Input to fast input register		0.9		1.3		1.9	ns
	All outputs		-0.1		-0.1		-0.2	ns
SSTL-3 Class I	Input to PIA		1.2		1.7		2.6	ns
	Input to global clock and clear		0.9		1.3		1.9	ns
	Input to fast input register		0.8		1.1		1.7	ns
	All outputs		0.0		0.0		0.0	ns
SSTL-3 Class II	Input to PIA		1.2		1.7		2.6	ns
	Input to global clock and clear		0.9		1.3		1.9	ns
	Input to fast input register		0.8		1.1		1.7	ns
	All outputs		0.0		0.0		0.0	ns
GTL+	Input to PIA		1.6		2.3		3.4	ns
	Input to global clock and clear		1.6		2.3		3.4	ns
	Input to fast input register		1.5		2.1		3.2	ns
	All outputs		0.0		0.0		0.0	ns

Symbol	Parameter	Conditions			Speed	Grade			Unit
			-	5	-	7		10	
			Min	Max	Min	Max	Min	Max	
t <sub>PD1</sub>	Input to non-registered output	C1 = 35 pF (2)		5.0		7.5		10.0	ns
t <sub>PD2</sub>	I/O input to non-registered output	C1 = 35 pF (2)		5.0		7.5		10.0	ns
t <sub>SU</sub>	Global clock setup time	(2)	3.3		4.8		6.6		ns
t <sub>H</sub>	Global clock hold time	(2)	0.0		0.0		0.0		ns
t <sub>FSU</sub>	Global clock setup time of fast input		1.0		1.5		1.5		ns
t <sub>FH</sub>	Global clock hold time for fast input		1.0		1.0		1.0		ns
t <sub>FZHSU</sub>	Global clock setup time of fast input with zero hold time		2.5		3.0		3.0		ns
t <sub>FZHH</sub>	Global clock hold time of fast input with zero hold time		0.0		0.0		0.0		ns
t <sub>CO1</sub>	Global clock to output delay	C1 = 35 pF	1.0	3.3	1.0	5.1	1.0	6.7	ns
t <sub>CH</sub>	Global clock high time		2.0		3.0		4.0		ns
t <sub>CL</sub>	Global clock low time		2.0		3.0		4.0		ns
t <sub>ASU</sub>	Array clock setup time	(2)	1.4		2.0		2.8		ns
t <sub>AH</sub>	Array clock hold time	(2)	0.4		0.8		1.0		ns
t <sub>ACO1</sub>	Array clock to output delay	C1 = 35 pF (2)	1.0	5.2	1.0	7.9	1.0	10.5	ns
t <sub>ACH</sub>	Array clock high time		2.0		3.0		4.0		ns
t <sub>ACL</sub>	Array clock low time		2.0		3.0		4.0		ns
t <sub>CPPW</sub>	Minimum pulse width for clear and preset		2.0		3.0		4.0		ns
t <sub>CNT</sub>	Minimum global clock period	(2)		5.3		7.9		10.6	ns
f <sub>CNT</sub>	Maximum internal global clock frequency	(2), (3)	188.7		126.6		94.3		MHz
t <sub>ACNT</sub>	Minimum array clock period	(2)		5.3		7.9		10.6	ns
f <sub>ACNT</sub>	Maximum internal array clock frequency	(2), (3)	188.7		126.6		94.3		MHz

Table 29. EPM7256B	Table 29. EPM7256B Selectable I/O Standard Timing Adder Delays (Part 2 of 2) Note (1)											
I/O Standard	Parameter	Speed Grade										
		-5 -7 -10		0								
		Min	Max	Min	Max	Min	Max					
PCI	Input to PIA		0.0		0.0		0.0	ns				
	Input to global clock and clear		0.0		0.0		0.0	ns				
	Input to fast input register		0.0		0.0		0.0	ns				
	All outputs		0.0		0.0		0.0	ns				

#### Notes to tables:

- (1) These values are specified under the Recommended Operating Conditions in Table 15 on page 29. See Figure 14 for more information on switching waveforms.
- (2) These values are specified for a PIA fan-out of one LAB (16 macrocells). For each additional LAB fan-out in these devices, add an additional 0.1 ns to the PIA timing value.
- (3) Measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
- (4) The  $t_{LPA}$  parameter must be added to the  $t_{LAD}$ ,  $t_{LAC}$ ,  $t_{IC}$ ,  $t_{ACL}$ ,  $t_{CPPW}$ ,  $t_{EN}$ , and  $t_{SEXP}$  parameters for macrocells running in low-power mode.

Table 32. EPM7512B Selectable I/O Standard Timing Adder Delays (Part 2 of 2) Note (1)								
I/O Standard	Parameter	Speed Grade						Unit
		-5		-7		-10		
		Min	Max	Min	Max	Min	Max	
PCI	Input to PIA		0.0		0.0		0.0	ns
	Input to global clock and clear		0.0		0.0		0.0	ns
	Input to fast input register		0.0		0.0		0.0	ns
	All outputs		0.0		0.0		0.0	ns

#### Notes to tables:

- These values are specified under the Recommended Operating Conditions in Table 15 on page 29. See Figure 14 for more information on switching waveforms.
- (2) These values are specified for a PIA fan-out of one LAB (16 macrocells). For each additional LAB fan-out in these devices, add an additional 0.12 ns to the PIA timing value.
- (3) Measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
- (4) The  $t_{LPA}$  parameter must be added to the  $t_{LAD}$ ,  $t_{LAC}$ ,  $t_{IC}$ ,  $t_{ACL}$ ,  $t_{CPPW}$ ,  $t_{EN}$ , and  $t_{SEXP}$  parameters for macrocells running in low-power mode.

# Power Consumption

Supply power (P) versus frequency ( $f_{MAX}$ , in MHz) for MAX 7000B devices is calculated with the following equation:

$$P = P_{INT} + P_{IO} = I_{CCINT} \times V_{CC} + P_{IO}$$

The  $P_{\rm IO}$  value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

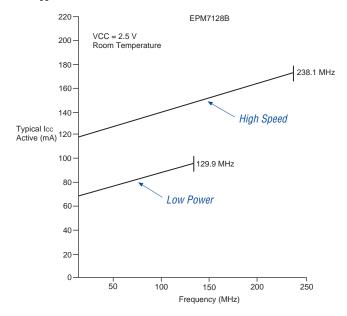
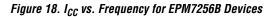


Figure 17.  $I_{CC}$  vs. Frequency for EPM7128B Devices



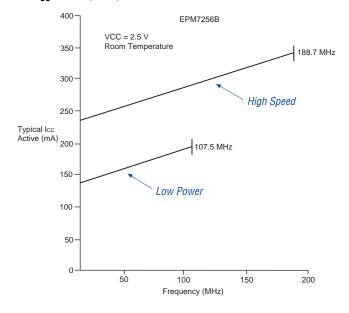


Figure 21. 48-Pin VTQFP Package Pin-Out Diagram

Package outlines not drawn to scale.

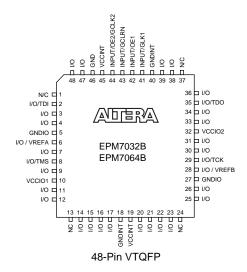


Figure 22. 49-Pin Ultra FineLine BGA Package Pin-Out Diagram

Package outline not drawn to scale.

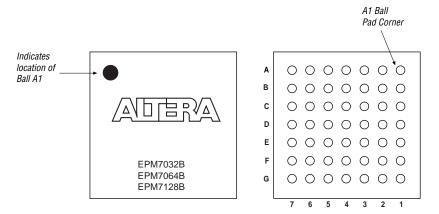


Figure 23. 100-Pin TQFP Package Pin-Out Diagram

Package outline not drawn to scale.

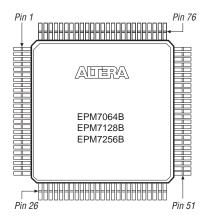


Figure 24. 100-Pin FineLine BGA Package Pin-Out Diagram

