



Welcome to **E-XFL.COM** 

**Understanding Embedded - CPLDs (Complex Programmable Logic Devices)** 

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

#### **Applications of Embedded - CPLDs**

Details	
Product Status	Obsolete
Programmable Type	In System Programmable
Delay Time tpd(1) Max	5 ns
Voltage Supply - Internal	2.375V ~ 2.625V
Number of Logic Elements/Blocks	16
Number of Macrocells	256
Number of Gates	5000
Number of I/O	120
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm7256btc144-5n

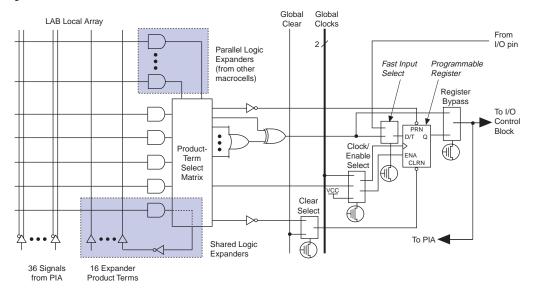
Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

#### **Macrocells**

The MAX 7000B macrocell can be individually configured for either sequential or combinatorial logic operation. The macrocell consists of three functional blocks: the logic array, the product-term select matrix, and the programmable register. Figure 2 shows the MAX 7000B macrocell.

Figure 2. MAX 7000B Macrocell



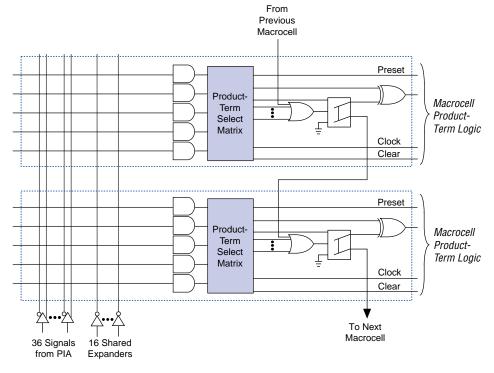
Combinatorial logic is implemented in the logic array, which provides five product terms per macrocell. The product-term select matrix allocates these product terms for use as either primary logic inputs (to the OR and XOR gates) to implement combinatorial functions, or as secondary inputs to the macrocell's register preset, clock, and clock enable control functions.

Two kinds of expander product terms ("expanders") are available to supplement macrocell logic resources:

- Shareable expanders, which are inverted product terms that are fed back into the logic array
- Parallel expanders, which are product terms borrowed from adjacent macrocells

Figure 4. MAX 7000B Parallel Expanders

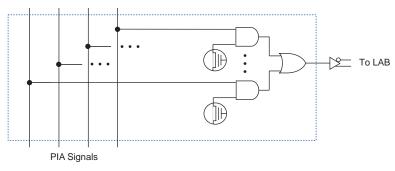
Unused product terms in a macrocell can be allocated to a neighboring macrocell.



#### **Programmable Interconnect Array**

Logic is routed between LABs on the PIA. This global bus is a programmable path that connects any signal source to any destination on the device. All MAX 7000B dedicated inputs, I/O pins, and macrocell outputs feed the PIA, which makes the signals available throughout the entire device. Only the signals required by each LAB are actually routed from the PIA into the LAB. Figure 5 shows how the PIA signals are routed into the LAB. An EEPROM cell controls one input to a two-input AND gate, which selects a PIA signal to drive into the LAB.

Figure 5. MAX 7000B PIA Routing



While the routing delays of channel-based routing schemes in masked or field-programmable gate arrays (FPGAs) are cumulative, variable, and path-dependent, the MAX 7000B PIA has a predictable delay. The PIA makes a design's timing performance easy to predict.

#### I/O Control Blocks

The I/O control block allows each I/O pin to be individually configured for input, output, or bidirectional operation. All I/O pins have a tri-state buffer that is individually controlled by one of the global output enable signals or directly connected to ground or  $V_{CC}.$  Figure 6 shows the I/O control block for MAX 7000B devices. The I/O control block has six or ten global output enable signals that are driven by the true or complement of two output enable signals, a subset of the I/O pins, or a subset of the I/O macrocells.

By combining the pulse and shift times for each of the programming stages, the program or verify time can be derived as a function of the TCK frequency, the number of devices, and specific target device(s). Because different ISP-capable devices have a different number of EEPROM cells, both the total fixed and total variable times are unique for a single device.

#### Programming a Single MAX 7000B Device

The time required to program a single MAX 7000B device in-system can be calculated from the following formula:

$$t_{PROG} = t_{PPULSE} + \frac{Cycle_{PTCK}}{f_{TCK}}$$

where:  $t_{PROG} = \text{Programming time}$   $t_{PPULSE} = \text{Sum of the fixed times to erase, program, and}$ 

verify the EEPROM cells

 $Cycle_{PTCK}$  = Number of TCK cycles to program a device

= TCK frequency

The ISP times for a stand-alone verification of a single MAX 7000B device can be calculated from the following formula:

$$t_{VER} = t_{VPULSE} + \frac{Cycle_{VTCK}}{f_{TCK}}$$

where:  $t_{VER}$  = Verify time  $t_{VPULSE}$  = Sum of the fixed times to verify the EEPROM cells  $Cycle_{VTCK}$  = Number of TCK cycles to verify a device

The programming times described in Tables 4 through 6 are associated with the worst-case method using the enhanced ISP algorithm.

Table 4. MAX 7000B t <sub>PUL</sub>	Table 4. MAX 7000B t <sub>PULSE</sub> & Cycle <sub>TCK</sub> Values											
Device	Progra	Programming Stand-Alone Verification										
	t <sub>PPULSE</sub> (s)	Cycle <sub>PTCK</sub>	t <sub>VPULSE</sub> (s)	Cycle <sub>VTCK</sub>								
EMP7032B	2.12	70,000	0.002	18,000								
EMP7064B	2.12	120,000	0.002	35,000								
EMP7128B	2.12	222,000	0.002	69,000								
EMP7256B	2.12	466,000	0.002	151,000								
EMP7512B	2.12	914,000	0.002	300,000								

Tables 5 and 6 show the in-system programming and stand alone verification times for several common test clock frequencies.

Table 5. MAX 7000B In-System Programming Times for Different Test Clock Frequencies											
Device		f <sub>TCK</sub>									
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz			
EMP7032B	2.13	2.13	2.15	2.19	2.26	2.47	2.82	3.52	S		
EMP7064B	2.13	2.14	2.18	2.24	2.36	2.72	3.32	4.52	S		
EMP7128B	2.14	2.16	2.23	2.34	2.56	3.23	4.34	6.56	S		
EMP7256B	2.17	2.21	2.35	2.58	3.05	4.45	6.78	11.44	S		
EMP7512B	2.21	2.30	2.58	3.03	3.95	6.69	11.26	20.40	S		

Table 1. MAX 7000B Stand-Alone Verification Times for Different Test Clock Frequencies											
Device		f <sub>TCK</sub>									
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz			
EMP7032B	0.00	0.01	0.01	0.02	0.04	0.09	0.18	0.36	S		
EMP7064B	0.01	0.01	0.02	0.04	0.07	0.18	0.35	0.70	S		
EMP7128B	0.01	0.02	0.04	0.07	0.14	0.35	0.69	1.38	S		
EMP7256B	0.02	0.03	0.08	0.15	0.30	0.76	1.51	3.02	S		
EMP7512B	0.03	0.06	0.15	0.30	0.60	1.50	3.00	6.00	S		

# Programming with External Hardware

MAX 7000B devices can be programmed on Windows-based PCs with an Altera Logic Programmer card, the Master Programming Unit (MPU), and the appropriate device adapter. The MPU performs continuity checking to ensure adequate electrical contact between the adapter and the device.



For more information, see the Altera Programming Hardware Data Sheet.

The Altera software can use text- or waveform-format test vectors created with the Altera Text Editor or Waveform Editor to test the programmed device. For added design verification, designers can perform functional testing to compare the functional device behavior with the results of simulation.

Data I/O, BP Microsystems, and other programming hardware manufacturers provide programming support for Altera devices. For more information, see *Programming Hardware Manufacturers*.

### IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

MAX 7000B devices include the JTAG boundary-scan test circuitry defined by IEEE Std. 1149.1. Table 6 describes the JTAG instructions supported by MAX 7000B devices. The pin-out tables starting on page 59 of this data sheet show the location of the JTAG control pins for each device. If the JTAG interface is not required, the JTAG pins are available as user I/O pins.

Table 6. MAX 7000B	JTAG Instructions
JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern output at the device pins.
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the boundary-scan test data to pass synchronously through a selected device to adjacent devices during normal operation.
CLAMP	Allows the values in the boundary-scan register to determine pin states while placing the 1-bit bypass register between the TDI and TDO pins.
IDCODE	Selects the IDCODE register and places it between the TDI and TDO pins, allowing the IDCODE to be serially shifted out of TDO.
USERCODE	Selects the 32-bit USERCODE register and places it between the TDI and TDO pins, allowing the USERCODE value to be shifted out of TDO.
ISP Instructions	These instructions are used when programming MAX 7000B devices via the JTAG ports with the MasterBlaster or ByteBlasterMV download cable, or using a Jam File (.jam), Jam Byte-Code File (.jbc), or Serial Vector Format File (.svf) via an embedded processor or test equipment.

The instruction register length of MAX 7000B devices is ten bits. The MAX 7000B USERCODE register length is 32 bits. Tables 7 and 8 show the boundary-scan register length and device IDCODE information for MAX 7000B devices.

Table 7. MAX 7000B Boundary-Scan Register Length								
Device Boundary-Scan Register Length								
EPM7032B	96							
EPM7064B	192							
EPM7128B	288							
EPM7256B	480							
EPM7512B	624							

Table 8. 32-1	Table 8. 32-Bit MAX 7000B Device IDCODENote (1)											
Device		IDCODE (32 Bits)										
	Version (4 Bits)	Part Number (16 Bits)	Manufacturer's Identity (11 Bits)	1 (1 Bit) (2)								
EPM7032B	0010	0111 0000 0011 0010	00001101110	1								
EPM7064B	0010	0111 0000 0110 0100	00001101110	1								
EPM7128B	0010	0111 0001 0010 1000	00001101110	1								
EPM7256B	0010	0111 0010 0101 0110	00001101110	1								
EPM7512B	0010	0111 0101 0001 0010	00001101110	1								

#### Notes:

- (1) The most significant bit (MSB) is on the left.
- (2) The least significant bit (LSB) for all JTAG IDCODEs is 1.



See Application Note 39 (IEEE 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices) for more information on JTAG boundary-scan testing.

Figure 8 shows the timing information for the JTAG signals.

MAX 7000B devices contain two I/O banks. Both banks support all standards. Each I/O bank has its own VCCIO pins. A single device can support 1.8-V, 2.5-V, and 3.3-V interfaces; each bank can support a different standard independently. Within a bank, any one of the terminated standards can be supported.

Figure 9 shows the arrangement of the MAX 7000B I/O banks.

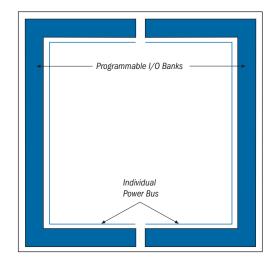


Figure 9. MAX 7000B I/O Banks for Various Advanced I/O Standards

Table 11 shows which macrocells have pins in each I/O bank.

Table 11. Macrocell Pins Contained in Each I/O Bank									
Device Bank 1 Bank 2									
EPM7032B	1-16	17-32							
EPM7064B	1-32	33-64							
EPM7128B	1-64	65-128							
EPM7256B	1-128, 177-181	129-176, 182-256							
EPM7512B	1-265	266-512							

Each MAX 7000B device has two VREF pins. Each can be set to a separate  $V_{REF}$  level. Any I/O pin that uses one of the voltage-referenced standards (GTL+, SSTL-2, or SSTL-3) may use either of the two VREF pins. If these pins are not required as VREF pins, they may be individually programmed to function as user I/O pins.

#### **Programmable Pull-Up Resistor**

Each MAX 7000B device I/O pin provides an optional programmable pull-up resistor during user mode. When this feature is enabled for an I/O pin, the pull-up resistor (typically 50 k¾) weakly holds the output to  $V_{\rm CCIO}$  level.

#### **Bus Hold**

Each MAX 7000B device I/O pin provides an optional bus-hold feature. When this feature is enabled for an I/O pin, the bus-hold circuitry weakly holds the signal at its last driven state. By holding the last driven state of the pin until the next input signals is present, the bus-hold feature can eliminate the need to add external pull-up or pull-down resistors to hold a signal level when the bus is tri-stated. The bus-hold circuitry also pulls undriven pins away from the input threshold voltage where noise can cause unintended high-frequency switching. This feature can be selected individually for each I/O pin. The bus-hold output will drive no higher than  $V_{\rm CCIO}$  to prevent overdriving signals. The propagation delays through the input and output buffers in MAX 7000B devices are not affected by whether the bus-hold feature is enabled or disabled.

The bus-hold circuitry weakly pulls the signal level to the last driven state through a resistor with a nominal resistance ( $R_{BH}$ ) of approximately 8.5 k¾. Table 12 gives specific sustaining current that will be driven through this resistor and overdrive current that will identify the next driven input level. This information is provided for each VCCIO voltage level.

Table 12. Bus Hold Parameters										
Parameter	Conditions	Conditions VCCIO Level								
		1.8 V 2.5 V		3.3	3 V					
		Min	Max	Min	Max	Min	Max			
Low sustaining current	$V_{IN} > V_{IL} (max)$	30		50		70		μΑ		
High sustaining current	V <sub>IN</sub> < V <sub>IH</sub> (min)	-30		-50		-70		μΑ		
Low overdrive current	0 V < V <sub>IN</sub> < V <sub>CCIO</sub>		200		300		500	μΑ		
High overdrive current	$0 \text{ V} < \text{V}_{\text{IN}} < \text{V}_{\text{CCIO}}$		-295		-435		-680	μΑ		

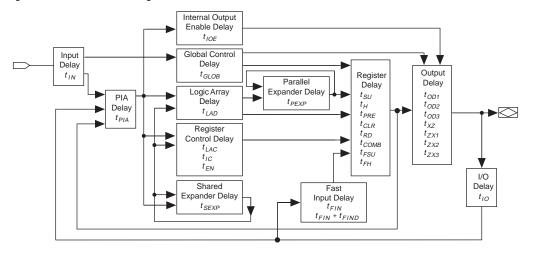
The bus-hold circuitry is active only during user operation. At power-up, the bus-hold circuit initializes its initial hold value as  $V_{CC}$  approaches the recommended operation conditions. When transitioning from ISP to User Mode with bus hold enabled, the bus-hold circuit captures the value present on the pin at the end of programming.

Table 1	6. MAX 7000B Device DC Opera	ating Conditions Note (4)			
Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>IH</sub>	High-level input voltage for 3.3-V TTL/CMOS		2.0	3.9	V
	High-level input voltage for 2.5-V TTL/CMOS		1.7	3.9	V
	High-level input voltage for 1.8-V TTL/CMOS		0.65 × V <sub>CCIO</sub>	3.9	V
V <sub>IL</sub>	Low-level input voltage for 3.3-V TTL/CMOS and PCI compliance		-0.5	0.8	V
	Low-level input voltage for 2.5-V TTL/CMOS		-0.5	0.7	V
	Low-level input voltage for 1.8-V TTL/CMOS		-0.5	0.35 × V <sub>CCIO</sub>	
V <sub>OH</sub>	3.3-V high-level TTL output voltage	$I_{OH} = -8 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (5)$	2.4		V
	3.3-V high-level CMOS output voltage	$I_{OH} = -0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (5)$	V <sub>CCIO</sub> - 0.2		V
	2.5-V high-level output voltage	$I_{OH} = -100 \mu A DC, V_{CCIO} = 2.30 V (5)$	2.1		V
		$I_{OH} = -1 \text{ mA DC}, V_{CCIO} = 2.30 \text{ V } (5)$	2.0		V
		$I_{OH} = -2 \text{ mA DC}, V_{CCIO} = 2.30 \text{ V } (5)$	1.7		V
	1.8-V high-level output voltage	$I_{OH} = -2 \text{ mA DC}, V_{CCIO} = 1.65 \text{ V } (5)$	1.2		V
$V_{OL}$	3.3-V low-level TTL output voltage	I <sub>OL</sub> = 8 mA DC, V <sub>CCIO</sub> = 3.00 V (6)		0.4	V
	3.3-V low-level CMOS output voltage	$I_{OL} = 0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (6)$		0.2	V
	2.5-V low-level output voltage	$I_{OL}$ = 100 $\mu$ A DC, $V_{CCIO}$ = 2.30 $V$ (6)		0.2	V
		$I_{OL}$ = 1 mA DC, $V_{CCIO}$ = 2.30 V (6)		0.4	V
		$I_{OL}$ = 2 mA DC, $V_{CCIO}$ = 2.30 V (6)		0.7	V
	1.8-V low-level output voltage	I <sub>OL</sub> = 2 mA DC, V <sub>CCIO</sub> = 1.7 V (6)		0.4	V
I <sub>I</sub>	Input leakage current	$V_1 = -0.5 \text{ to } 3.9 \text{ V } (7)$	-10	10	μΑ
I <sub>OZ</sub>	Tri-state output off-state current	$V_1 = -0.5 \text{ to } 3.9 \text{ V } (7)$	-10	10	μΑ
R <sub>ISP</sub>	Value of I/O pin pull-up resistor during in-system programming or during power up	V <sub>CCIO</sub> = 1.7 to 3.6 V (8)	20	74	k¾

### **Timing Model**

MAX 7000B device timing can be analyzed with the Altera software, with a variety of popular industry-standard EDA simulators and timing analyzers, or with the timing model shown in Figure 13. MAX 7000B devices have predictable internal delays that enable the designer to determine the worst-case timing of any design. The Altera software provides timing simulation, point-to-point delay prediction, and detailed timing analysis for device-wide performance evaluation.

Figure 13. MAX 7000B Timing Model



The timing characteristics of any signal path can be derived from the timing model and parameters of a particular device. External timing parameters, which represent pin-to-pin timing delays, can be calculated as the sum of internal parameters. Figure 14 shows the timing relationship between internal and external delay parameters.



See Application Note 94 (Understanding MAX 7000 Timing) for more information.

Table 20. EPM7032B	Table 20. EPM7032B Selectable I/O Standard Timing Adder Delays Notes (1)											
I/O Standard	Parameter		Speed Grade									
		-3.5		-3.5 -5.0		5.0 -7.5		.5				
		Min	Max	Min	Max	Min	Max					
PCI	Input to PIA		0.0		0.0		0.0	ns				
	Input to global clock and clear		0.0		0.0		0.0	ns				
	Input to fast input register		0.0		0.0		0.0	ns				
	All outputs		0.0		0.0		0.0	ns				

#### Notes to tables:

- (1) These values are specified under the Recommended Operating Conditions in Table 15 on page 29. See Figure 14 for more information on switching waveforms.
- (2) These values are specified for a PIA fan-out of all LABs.
- (3) Measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
- (4) The  $t_{LPA}$  parameter must be added to the  $t_{LAD}$ ,  $t_{LAC}$ ,  $t_{IC}$ ,  $t_{ACL}$ ,  $t_{CPPW}$ ,  $t_{EN}$ , and  $t_{SEXP}$  parameters for macrocells running in low-power mode.

Symbol	Parameter	Conditions	Speed Grade						
			-	3	-	5	-7		]
			Min	Max	Min	Max	Min	Max	
t <sub>IN</sub>	Input pad and buffer delay			0.3		0.5		0.7	ns
$t_{IO}$	I/O input pad and buffer delay			0.3		0.5		0.7	ns
t <sub>FIN</sub>	Fast input delay			0.9		1.3		2.0	ns
t <sub>FIND</sub>	Programmable delay adder for fast input			1.0		1.5		1.5	ns
t <sub>SEXP</sub>	Shared expander delay			1.5		2.1		3.2	ns
t <sub>PEXP</sub>	Parallel expander delay			0.4		0.6		0.9	ns
$t_{LAD}$	Logic array delay			1.4		2.0		3.1	ns
$t_{LAC}$	Logic control array delay			1.2		1.7		2.6	ns
t <sub>IOE</sub>	Internal output enable delay			0.1		0.2		0.3	ns
t <sub>OD1</sub>	Output buffer and pad delay slow slew rate = off V <sub>CCIO</sub> = 3.3 V	C1 = 35 pF		0.9		1.2		1.8	ns
t <sub>OD3</sub>	Output buffer and pad delay slow slew rate = on V <sub>CCIO</sub> = 2.5 V or 3.3 V	C1 = 35 pF		5.9		6.2		6.8	ns
t <sub>ZX1</sub>	Output buffer enable delay slow slew rate = off V <sub>CCIO</sub> = 3.3 V	C1 = 35 pF		1.6		2.2		3.4	ns
t <sub>ZX3</sub>	Output buffer enable delay slow slew rate = on V <sub>CCIO</sub> = 2.5 V or 3.3 V	C1 = 35 pF		6.6		7.2		8.4	ns
$t_{XZ}$	Output buffer disable delay	C1 = 5 pF		1.6		2.2		3.4	ns
$t_{SU}$	Register setup time		0.7		1.1		1.6		ns
$t_H$	Register hold time		0.4		0.5		0.9		ns
t <sub>FSU</sub>	Register setup time of fast input		0.8		0.8		1.1		ns
$t_{FH}$	Register hold time of fast input		1.2		1.2		1.4		ns
$t_{RD}$	Register delay			0.5		0.6		0.9	ns
$t_{COMB}$	Combinatorial delay			0.2		0.3		0.5	ns
t <sub>IC</sub>	Array clock delay		İ	1.2		1.8		2.8	ns
$t_{EN}$	Register enable time		İ	1.2		1.7		2.6	ns
$t_{GLOB}$	Global control delay		İ	0.7		1.1		1.6	ns
$t_{PRE}$	Register preset time			1.0		1.3		1.9	ns
t <sub>CLR</sub>	Register clear time			1.0		1.3		1.9	ns
$t_{PIA}$	PIA delay	(2)		0.7		1.0		1.4	ns
$t_{LPA}$	Low-power adder	(4)		1.5		2.1		3.2	ns

Symbol	Parameter	Conditions		Speed Grade						
			-4		-7		-10			
			Min	Max	Min	Max	Min	Max		
t <sub>PD1</sub>	Input to non-registered output	C1 = 35 pF (2)		4.0		7.5		10.0	ns	
t <sub>PD2</sub>	I/O input to non-registered output	C1 = 35 pF (2)		4.0		7.5		10.0	ns	
t <sub>SU</sub>	Global clock setup time	(2)	2.5		4.5		6.1		ns	
t <sub>H</sub>	Global clock hold time	(2)	0.0		0.0		0.0		ns	
t <sub>FSU</sub>	Global clock setup time of fast input		1.0		1.5		1.5		ns	
t <sub>FH</sub>	Global clock hold time of fast input		1.0		1.0		1.0		ns	
t <sub>FZHSU</sub>	Global clock setup time of fast input with zero hold time		2.0		3.0		3.0		ns	
t <sub>FZHH</sub>	Global clock hold time of fast input with zero hold time		0.0		0.0		0.0		ns	
t <sub>CO1</sub>	Global clock to output delay	C1 = 35 pF	1.0	2.8	1.0	5.7	1.0	7.5	ns	
t <sub>CH</sub>	Global clock high time		1.5		3.0		4.0		ns	
t <sub>CL</sub>	Global clock low time		1.5		3.0		4.0		ns	
t <sub>ASU</sub>	Array clock setup time	(2)	1.2		2.0		2.8		ns	
t <sub>AH</sub>	Array clock hold time	(2)	0.2		0.7		0.9		ns	
t <sub>ACO1</sub>	Array clock to output delay	C1 = 35 pF (2)	1.0	4.1	1.0	8.2	1.0	10.8	ns	
t <sub>ACH</sub>	Array clock high time		1.5		3.0		4.0		ns	
t <sub>ACL</sub>	Array clock low time		1.5		3.0		4.0		ns	
t <sub>CPPW</sub>	Minimum pulse width for clear and preset		1.5		3.0		4.0		ns	
t <sub>CNT</sub>	Minimum global clock period	(2)		4.1		7.9		10.6	ns	
f <sub>CNT</sub>	Maximum internal global clock frequency	(2), (3)	243.9		126.6		94.3		MHz	
t <sub>ACNT</sub>	Minimum array clock period	(2)		4.1		7.9		10.6	ns	
f <sub>ACNT</sub>	Maximum internal array clock frequency	(2), (3)	243.9		126.6		94.3		MHz	

Table 26. EPM7128B Selectable I/O Standard Timing Adder Delays (Part 1 of 2) Note (1)								
I/O Standard	Parameter	Speed Grade						
		-4		-7		-10		1
		Min	Max	Min	Max	Min	Max	
3.3 V TTL/CMOS	Input to PIA		0.0		0.0		0.0	ns
	Input to global clock and clear		0.0		0.0		0.0	ns
	Input to fast input register		0.0		0.0		0.0	ns
	All outputs		0.0		0.0		0.0	ns
2.5 V TTL/CMOS	Input to PIA		0.3		0.6		0.8	ns
	Input to global clock and clear		0.3		0.6		0.8	ns
	Input to fast input register		0.2		0.4		0.5	ns
	All outputs		0.2		0.4		0.5	ns
1.8 V TTL/CMOS	Input to PIA		0.5		0.9		1.3	ns
	Input to global clock and clear		0.5		0.9		1.3	ns
	Input to fast input register		0.4		0.8		1.0	ns
	All outputs		1.2		2.3		3.0	ns
SSTL-2 Class I	Input to PIA		1.4		2.6		3.5	ns
	Input to global clock and clear		1.2		2.3		3.0	ns
	Input to fast input register		1.0		1.9		2.5	ns
	All outputs		0.0		0.0		0.0	ns
SSTL-2 Class II	Input to PIA		1.4		2.6		3.5	ns
	Input to global clock and clear		1.2		2.3		3.0	ns
	Input to fast input register		1.0		1.9		2.5	ns
	All outputs		-0.1		-0.2		-0.3	ns
SSTL-3 Class I	Input to PIA		1.3		2.4		3.3	ns
	Input to global clock and clear		1.0		1.9		2.5	ns
	Input to fast input register		0.9		1.7		2.3	ns
	All outputs		0.0		0.0		0.0	ns
SSTL-3 Class II	Input to PIA		1.3		2.4		3.3	ns
	Input to global clock and clear		1.0		1.9		2.5	ns
	Input to fast input register		0.9		1.7		2.3	ns
	All outputs		0.0		0.0		0.0	ns
GTL+	Input to PIA		1.7		3.2		4.3	ns
	Input to global clock and clear		1.7		3.2		4.3	ns
	Input to fast input register		1.6		3.0		4.0	ns
	All outputs		0.0		0.0		0.0	ns

Table 26. EPM7128B Selectable I/O Standard Timing Adder Delays (Part 2 of 2) Note (1)									
I/O Standard	Parameter	Speed Grade						Unit	
		-4		-7		-10			
		Min	Max	Min	Max	Min	Max		
PCI	Input to PIA		0.0		0.0		0.0	ns	
	Input to global clock and clear		0.0		0.0		0.0	ns	
	Input to fast input register		0.0		0.0		0.0	ns	
	All outputs		0.0		0.0		0.0	ns	

#### Notes to tables:

- (1) These values are specified under the Recommended Operating Conditions in Table 15 on page 29. See Figure 14 for more information on switching waveforms.
- (2) These values are specified for a PIA fan-out of one LAB (16 macrocells). For each additional LAB fan-out in these devices, add an additional 0.1 ns to the PIA timing value.
- (3) Measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
- (4) The  $t_{LPA}$  parameter must be added to the  $t_{LAD}$ ,  $t_{LAC}$ ,  $t_{IC}$ ,  $t_{ACL}$ ,  $t_{CPPW}$ ,  $t_{EN}$ , and  $t_{SEXP}$  parameters for macrocells running in low-power mode.

Table 32. EPM7512B Selectable I/O Standard Timing Adder Delays (Part 2 of 2) Note (1)										
I/O Standard	Parameter	Speed Grade								
		-5		-7		-10				
		Min	Max	Min	Max	Min	Max			
PCI	Input to PIA		0.0		0.0		0.0	ns		
	Input to global clock and clear		0.0		0.0		0.0	ns		
	Input to fast input register		0.0		0.0		0.0	ns		
	All outputs		0.0		0.0		0.0	ns		

#### Notes to tables:

- These values are specified under the Recommended Operating Conditions in Table 15 on page 29. See Figure 14 for more information on switching waveforms.
- (2) These values are specified for a PIA fan-out of one LAB (16 macrocells). For each additional LAB fan-out in these devices, add an additional 0.12 ns to the PIA timing value.
- (3) Measured with a 16-bit loadable, enabled, up/down counter programmed into each LAB.
- (4) The  $t_{LPA}$  parameter must be added to the  $t_{LAD}$ ,  $t_{LAC}$ ,  $t_{IC}$ ,  $t_{ACL}$ ,  $t_{CPPW}$ ,  $t_{EN}$ , and  $t_{SEXP}$  parameters for macrocells running in low-power mode.

# Power Consumption

Supply power (P) versus frequency ( $f_{MAX}$ , in MHz) for MAX 7000B devices is calculated with the following equation:

$$P = P_{INT} + P_{IO} = I_{CCINT} \times V_{CC} + P_{IO}$$

The  $P_{\rm IO}$  value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

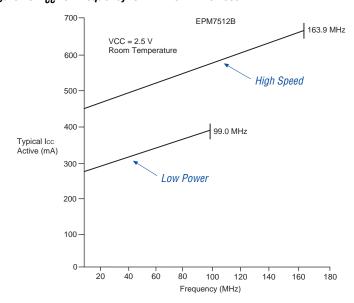


Figure 19.  $I_{CC}$  vs. Frequency for EPM7512B Devices

Figure 23. 100-Pin TQFP Package Pin-Out Diagram

Package outline not drawn to scale.

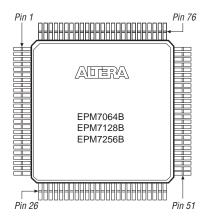


Figure 24. 100-Pin FineLine BGA Package Pin-Out Diagram

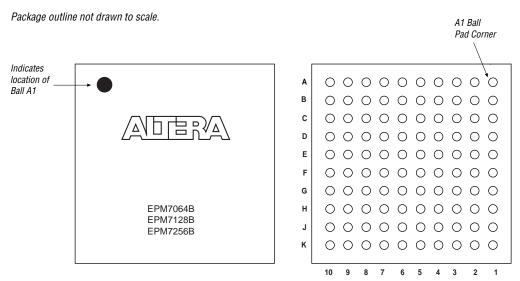
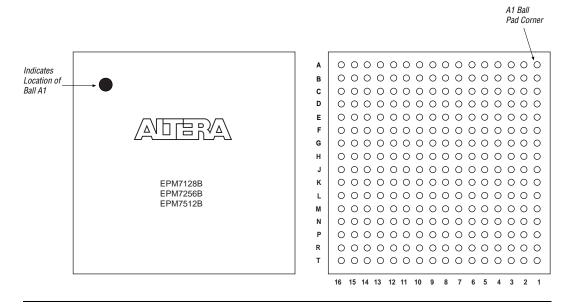


Figure 29. 256-Pin FineLine BGA Package Pin-Out Diagram

Package outline not drawn to scale.



## Revision History

The information contained in the MAX 7000B Programmable Logic Device Family Data Sheet version 3.5 supersedes information published in previous versions.

#### Version 3.5

The following changes were made to the *MAX 7000B Programmable Logic Device Family Data Sheet* version 3.5:

Updated Figure 28.

#### Version 3.4

The following changes were made to the MAX 7000B Programmable Logic Device Family Data Sheet version 3.4:

■ Updated text in the "Power Sequencing & Hot-Socketing" section.