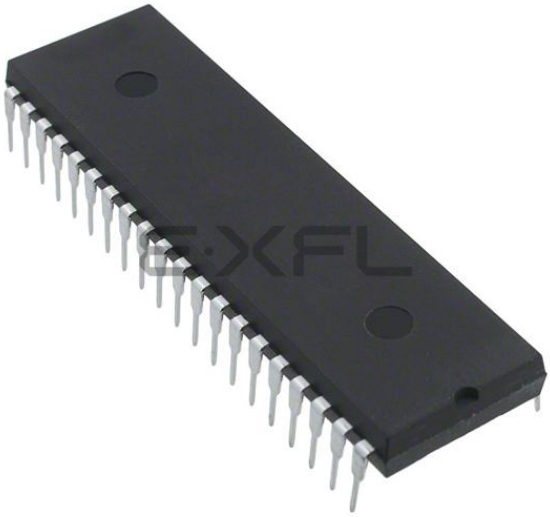

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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I²C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	36
Program Memory Size	14KB (8K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 35x10b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Through Hole
Package / Case	40-DIP (0.600", 15.24mm)
Supplier Device Package	40-PDIP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16lf18875-e-p

TABLE 1-3: PIC16F18875 PINOUT DESCRIPTION (CONTINUED)

Name	Function	Input Type	Output Type	Description
RC6/ANC6/CK(3)/IOCC6	RC6	TTL/ST	CMOS/OD	General purpose I/O.
	ANC6	AN	—	ADC Channel C6 input.
	CK(3)	TTL/ST	CMOS/OD	EUSART synchronous mode clock input/output.
	IOCC6	TTL/ST	—	Interrupt-on-change input.
RC7/ANC7/RX(1)/DT(3)/IOCC7	RC7	TTL/ST	CMOS/OD	General purpose I/O.
	ANC7	AN	—	ADC Channel C7 input.
	RX(1)	TTL/ST	—	EUSART Asynchronous mode receiver data input.
	DT(3)	TTL/ST	CMOS/OD	EUSART Synchronous mode data input/output.
	IOCC7	TTL/ST	—	Interrupt-on-change input.
RD0	RD0	TTL/ST	CMOS/OD	General purpose I/O.
	AND0	AN	—	ADC Channel D0 input.
RD1	RD1	TTL/ST	CMOS/OD	General purpose I/O.
	AND1	AN	—	ADC Channel D1 input.
RD2	RD2	TTL/ST	CMOS/OD	General purpose I/O.
	AND2	AN	—	ADC Channel D2 input.
RD3	RD3	TTL/ST	CMOS/OD	General purpose I/O.
	AND3	AN	—	ADC Channel D3 input.
RD4	RD4	TTL/ST	CMOS/OD	General purpose I/O.
	AND4	AN	—	ADC Channel D4 input.
RD5	RD5	TTL/ST	CMOS/OD	General purpose I/O.
	AND5	AN	—	ADC Channel D5 input.
RD6	RD6	TTL/ST	CMOS/OD	General purpose I/O.
	AND6	AN	—	ADC Channel D6 input.
RD7	RD7	TTL/ST	CMOS/OD	General purpose I/O.
	AND7	AN	—	ADC Channel D7 input.
RE0	RE0	TTL/ST	CMOS/OD	General purpose I/O.
	ANE0	AN	—	ADC Channel E0 input.
RE1	RE1	TTL/ST	CMOS/OD	General purpose I/O.
	ANE1	AN	—	ADC Channel E1 input.
RE2	RE2	TTL/ST	CMOS/OD	General purpose I/O.
	ANE2	AN	—	ADC Channel E2 input.
RE3/IOCE3/$\overline{MCLR}$/VPP	RE3	TTL/ST	—	General purpose input-only (when $\overline{MCLR}$ is disabled by config bit).
	IOCE3	TTL/ST	—	Interrupt-on-change input.
	$\overline{MCLR}$	ST	—	Master clear input with internal weak pull-up resistor.
	VPP	HV	—	ICSP™ high voltage programming mode entry input.
VDD	VDD	Power	—	Positive supply voltage input.

Legend: AN = Analog input or output CMOS = CMOS compatible input or output OD = Open-Drain
TTL = TTL compatible input ST = Schmitt Trigger input with CMOS levels I^2C = Schmitt Trigger input with I^2CHV= High Voltage XTAL= Crystal levels

Note 1: This is a PPS remappable input signal. The input function may be moved from the default location shown to one of several other PORTx pins. Refer to Table 13-1 for details on which PORT pins may be used for this signal.

2: All output signals shown in this row are PPS remappable. These signals may be mapped to output onto one of several PORTx pin options as described in Table 13-3.

3: This is a bidirectional signal. For normal module operation, the firmware should map this signal to the same pin in both the PPS input and PPS output registers.

4: These pins are configured for I^2C logic levels. The SCLx/SDAx signals may be assigned to any of the RB1/RB2/RC3/RC4 pins. PPS assignments to the other pins (e.g., RA5) will operate, but input logic levels will be standard TTL/ST, as selected by the INLVL register, instead of the I^2C specific or SMBus input buffer thresholds.

TABLE 3-13: SPECIAL FUNCTION REGISTER SUMMARY BANKS 0-31 (CONTINUED)

Address	Name	PIC16(L)F18855	PIC16(L)F18875	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets
Banks 15													
CPU CORE REGISTERS; see Table 3-2 for specifics													
78Ch	—	—		Unimplemented								—	—
78Dh	—	—		Unimplemented								—	—
78Eh	—	—		Unimplemented								—	—
78Fh	—	—		Unimplemented								—	—
790h	—	—		Unimplemented								—	—
791h	—	—		Unimplemented								—	—
792h	—	—		Unimplemented								—	—
793h	—	—		Unimplemented								—	—
794h	—	—		Unimplemented								—	—
795h	—	—		Unimplemented								—	—
796h	PMD0			SYSCMD	FVRMD	—	CRCMD	SCANMD	NVMMD	CLKRMD	IOCMD	00-0 0000	00-0 0000
797h	PMD1			NCOMD	TMR6MD	TMR5MD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	TMR0MD	0000 0000	0000 0000
798h	PMD2			—	DACMD	ADCMD	—	—	CMP2MD	CMP1MD	ZCDMD	-00- -000	-00- -000
799h	PMD3			—	PWM7MD	PWM6MD	CCP5MD	CCP4MD	CCP3MD	CCP2MD	CCP1MD	-000 0000	-000 0000
79Ah	PMD4			—	UART1MD	MSSP2MD	MSSP1MD	—	CWG3MD	CWG2MD	CWG1MD	-000 -000	-000 -000
79Bh	PMD5			SMT2MD	SMT1MD	—	CLC4MD	CLC3MD	CLC2MD	CLC1MD	DSMMD	00-0 0000	00-0 0000
79Ch	—	—		Unimplemented								—	—
79Dh	—	—		Unimplemented								—	—
79Eh	—	—		Unimplemented								—	—
79Fh	—	—		Unimplemented								—	—

Legend: x = unknown, u = unchanged, q =depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations unimplemented, read as '0'.

Note 1: Register present on PIC16F18855/75 devices only.

2: Unimplemented, read as '1'.

TABLE 6-1: NOSC/COSC BIT SETTINGS

NOSC<2:0>/ COSC<2:0>	Clock Source
111	EXTOSC(1)
110	HFINTOSC(2)
101	LFINTOSC
100	SOSC
011	Reserved (it operates like NOSC = 110)
010	EXTOSC with 4x PLL(1)
001	HFINTOSC with 2x PLL(1)
000	Reserved (it operates like NOSC = 110)

Note 1: EXTOSC configured by the FEXTOSC bits of Configuration Word 1 (Register 4-1).

2: HFINTOSC settings are configured with the HFFRQ bits of the OSCFRQ register (Register 6-6).

TABLE 6-2: NDIV/CDIV BIT SETTINGS

NDIV<3:0>/ CDIV<3:0>	Clock divider
1111-1010	Reserved
1001	512
1000	256
0111	128
0110	64
0101	32
0100	16
0011	8
0010	4
0001	2
0000	1

REGISTER 6-3: OSCCON3: OSCILLATOR CONTROL REGISTER 3

R/W/HC-0/0	R/W-0/0	U-0	R-0/0	R-0/0	U-0	U-0	U-0
CSWHOLD	SOSCPWR	—	ORDY	NOSCR	—	—	—
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	HC = Bit is cleared by hardware

bit 7 **CSWHOLD:** Clock Switch Hold bit
- 1 = Clock switch will hold (with interrupt) when the oscillator selected by NOSC is ready
- 0 = Clock switch may proceed when the oscillator selected by NOSC is ready; if this bit is clear at the time that NOSCR becomes ‘1’, the switch will occur

bit 6 **SOSCPWR:** Secondary Oscillator Power Mode Select bit
- 1 = Secondary oscillator operating in High-power mode
- 0 = Secondary oscillator operating in Low-power mode

bit 5 **Unimplemented:** Read as ‘0’.

bit 4 **ORDY:** Oscillator Ready bit (read-only)
- 1 = OSCCON1 = OSCCON2; the current system clock is the clock specified by NOSC
- 0 = A clock switch is in progress

bit 3 **NOSCR:** New Oscillator is Ready bit (read-only)
- 1 = A clock switch is in progress and the oscillator selected by NOSC indicates a “ready” condition
- 0 = A clock switch is not in progress, or the NOSC-selected oscillator is not yet ready

bit 2-0 **Unimplemented:** Read as ‘0’

9.1 Independent Clock Source

The WDT can derive its time base from either the 31 kHz LFINTOSC or 31.25 kHz MFINTOSC/16 internal oscillators, depending on the value of either the WDTCCS<2:0> Configuration bits or the WDTCS<2:0> bits of WDTCON1. Time intervals in this chapter are based on a minimum nominal interval of 1 ms. See **Section 37.0 "Electrical Specifications"** for LFINTOSC and MFINTOSC tolerances.

9.2 WDT Operating Modes

The Watchdog Timer module has four operating modes controlled by the WDTE<1:0> bits in Configuration Words. See Table 9-1.

9.2.1 WDT IS ALWAYS ON

When the WDTE bits of Configuration Words are set to '11', the WDT is always on.

WDT protection is active during Sleep.

9.2.2 WDT IS OFF IN SLEEP

When the WDTE bits of Configuration Words are set to '10', the WDT is on, except in Sleep.

WDT protection is not active during Sleep.

9.2.3 WDT CONTROLLED BY SOFTWARE

When the WDTE bits of Configuration Words are set to '01', the WDT is controlled by the SEN bit of the WDTCON0 register.

WDT protection is unchanged by Sleep. See Table 9-1 for more details.

TABLE 9-1: WDT OPERATING MODES

WDTE<1:0>	SEN	Device Mode	WDT Mode
11	X	X	Active
10	X	Awake	Active
		Sleep	Disabled
01	1	X	Active
	0	X	Disabled
00	X	X	Disabled

9.3 Time-Out Period

The WDTPS bits of the WDTCON0 register set the time-out period from 1 ms to 256 seconds (nominal). After a Reset, the default time-out period is two seconds.

9.4 Watchdog Window

The Watchdog Timer has an optional Windowed mode that is controlled by the WDTCWS<2:0> Configuration bits and WINDOW<2:0> bits of the WDTCON1 register. In the Windowed mode, the CLRWDT instruction must occur within the allowed window of the WDT period. Any CLRWDT instruction that occurs outside of this window will trigger a window violation and will cause a WDT Reset, similar to a WDT time out. See Figure 9-2 for an example.

The window size is controlled by the WDTCWS<2:0> Configuration bits, or the WINDOW<2:0> bits of WDTCON1, if WDTCWS<2:0> = 111.

In the event of a window violation, a Reset will be generated and the WDTWV bit of the PCON register will be cleared. This bit is set by a POR or can be set in firmware.

9.5 Clearing the WDT

The WDT is cleared when any of the following conditions occur:

- Any Reset
- Valid CLRWDT instruction is executed
- Device enters Sleep
- Device wakes up from Sleep
- WDT is disabled
- Oscillator Start-up Timer (OST) is running
- Any write to the WDTCON0 or WDTCON1 registers

9.5.1 CLRWDT CONSIDERATIONS (WINDOWED MODE)

When in Windowed mode, the WDT must be armed before a CLRWDT instruction will clear the timer. This is performed by reading the WDTCON0 register. Executing a CLRWDT instruction without performing such an arming action will trigger a window violation.

See Table 9-2 for more information.

9.6 Operation During Sleep

When the device enters Sleep, the WDT is cleared. If the WDT is enabled during Sleep, the WDT resumes counting. When the device exits Sleep, the WDT is cleared again.

The WDT remains clear until the OST, if enabled, completes. See **Section 6.0 "Oscillator Module (with Fail-Safe Clock Monitor)"** for more information on the OST.

When a WDT time-out occurs while the device is in Sleep, no Reset is generated. Instead, the device wakes up and resumes operation. The TO and PD bits in the STATUS register are changed to indicate the event. The RWDT bit in the PCON register can also be used. See **Section 3.0 "Memory Organization"** for more information.

In general, if INTM = 0, the scanner will take precedence over the interrupt, resulting in decreased interrupt processing speed and/or increased interrupt response latency. If INTM = 1, the interrupt will take precedence and have a better speed, delaying the memory scan.

11.10.6 WDT INTERACTION

Operation of the WDT is not affected by scanner activity. Hence, it is possible that long scans, particularly in Burst mode, may exceed the WDT time-out period and result in an undesired device Reset. This should be considered when performing memory scans with an application that also utilizes WDT.

11.10.7 IN-CIRCUIT DEBUG (ICD) INTERACTION

The scanner freezes when an ICD halt occurs, and remains frozen until user-mode operation resumes. The debugger may inspect the SCANCON0 and SCANLADR registers to determine the state of the scan.

The ICD interaction with each operating mode is summarized in Table 11-3.

TABLE 11-3: ICD AND SCANNER INTERACTIONS

ICD Halt	Scanner Operating Mode		
	Peek	**Concurrent Triggered**	**Burst**
External Halt	If Scanner would peek an instruction that is not executed (because of ICD entry), the peek will occur after ICD exit, when the instruction executes.	If external halt is asserted during a scan cycle, the instruction (delayed by scan) may or may not execute before ICD entry, depending on external halt timing.	If external halt is asserted during the BSF(SCANCON.GO), ICD entry occurs, and the burst is delayed until ICD exit. Otherwise, the current NVM-access cycle will complete, and then the scanner will be interrupted for ICD entry.
		If external halt is asserted during the cycle immediately prior to the scan cycle, both scan and instruction execution happen after the ICD exits.	If external halt is asserted during the burst, the burst is suspended and will resume with ICD exit.
PC Breakpoint		Scan cycle occurs before ICD entry and instruction execution happens after the ICD exits.	If PCPB (or single step) is on BSF(SCANCON.GO), the ICD is entered before execution; execution of the burst will occur at ICD exit, and the burst will run to completion. Note that the burst can be interrupted by an external halt.
Data Breakpoint		The instruction with the dataBP executes and ICD entry occurs immediately after. If scan is requested during that cycle, the scan cycle is postponed until the ICD exits.	
Single Step		If a scan cycle is ready after the debug instruction is executed, the scan will read PFM and then the ICD is re-entered.	
SWBP and ICDINST		If scan would stall a SWBP, the scan cycle occurs and the ICD is entered.	If SWBP replaces BSF(SCANCON.GO), the ICD will be entered; instruction execution will occur at ICD exit (from ICDINSTR register), and the burst will run to completion.

13.0 PERIPHERAL PIN SELECT (PPS) MODULE

The Peripheral Pin Select (PPS) module connects peripheral inputs and outputs to the device I/O pins. Only digital signals are included in the selections. All analog inputs and outputs remain fixed to their assigned pins. Input and output selections are independent as shown in the simplified block diagram Figure 13-1.

TABLE 13-1: PPS INPUT SIGNAL ROUTING OPTIONS

Input Signal Name	Input Register Name	Default Location at POR	Remappable to Pins of PORTx							
			PIC16F18855			PIC16F18875				
			PORTA	PORTB	PORTC	PORTA	PORTB	PORTC	PORTD	PORTE
INT	INTPPS	RB0	●	●		●	●			
T0CKI	T0CKIPPS	RA4	●	●		●	●			
T1CKI	T1CKIPPS	RC0	●		●	●		●		
T1G	T1GPPS	RB5		●	●		●	●		
T3CKI	T3CKIPPS	RC0		●	●		●	●		
T3G	T3GPPS	RC0	●		●	●		●		
T5CKI	T5CKIPPS	RC2	●		●	●		●		
T5G	T5GPPS	RB4		●	●		●		●	
T2IN	T2INPPS	RC3	●		●		●		●	
T4IN	T4INPPS	RC5		●	●		●	●		
T6IN	T6INPPS	RB7		●	●		●		●	
CCP1	CCP1PPS	RC2		●	●		●	●		
CCP2	CCP2PPS	RC1		●	●		●	●		
CCP3	CCP3PPS	RB5		●	●		●		●	
CCP4	CCP4PPS	RB0		●	●		●		●	
CCP5	CCP5PPS	RA4	●		●	●				●
SMTWIN1	SMTWIN1PPS	RC0		●	●		●			
SMTSIG1	SMTSIG1PPS	RC1		●	●		●			
SMTWIN2	SMTWIN2PPS	RB4		●	●		●		●	
SMTSIG2	SMTSIG2PPS	RB5		●	●		●		●	
CWG1IN	CWG1PPS	RB0		●	●		●		●	
CWG2IN	CWG2PPS	RB1		●	●		●		●	
CWG3IN	CWG3PPS	RB2		●	●		●		●	
MDCARL	MDCARLPPS	RA3	●		●	●			●	
MDCARH	MDCARHPPS	RA4	●		●	●			●	
MDMSRC	MDSRCPPS	RA5	●		●	●			●	
CLCIN0	CLCIN0PPS	RA0	●		●	●		●		
CLCIN1	CLCIN1PPS	RA1	●		●	●		●		
CLCIN2	CLCIN2PPS	RB6		●	●		●		●	

19.1.1 PWM CLOCK SELECTION

The PIC16(L)F18855/75 allows each individual CCP and PWM module to select the timer source that controls the module. Each module has an independent selection.

As there are up to three 8-bit timers with auto-reload (Timer2/4/6), PWM mode on the CCP and PWM modules can use any of these timers.

The CCPTMRS0 and CCPTMRS1 register are used to select which timer is used.

19.1.2 USING THE TMR2/4/6 WITH THE PWM MODULE

This device has a newer version of the TMR2 module that has many new modes, which allow for greater customization and control of the PWM signals than on older parts. Refer to Section 29.5, Operation Examples for examples of PWM signal generation using the different modes of Timer2. PWM operation requires that the timer used as the PWM time base has the FOSC/4 clock source selected.

19.1.3 PWM PERIOD

Referring to Figure 19-1, the PWM output has a period and a pulse width. The frequency of the PWM is the inverse of the period (1/period).

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following formula:

EQUATION 19-1: PWM PERIOD

$$PWM\ Period = [(PR2) + 1] \cdot 4 \cdot TOSC \cdot (TMR2\ Prescale\ Value)$$

Note 1: TOSC = 1/FOSC

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The PWMx pin is set (Exception: If the PWM duty cycle = 0%, the pin will not be set.)
- The PWM pulse width is latched from PWMxDC.

Note:	If the pulse width value is greater than the period the assigned PWM pin(s) will remain unchanged.

19.1.4 PWM DUTY CYCLE

The PWM duty cycle is specified by writing a 10-bit value to the PWMxDC register. The PWMxDCH contains the eight MSbs and the PWMxDCL<7:6> bits contain the two LSbs.

The PWMDC register is double-buffered and can be updated at any time. This double buffering is essential for glitch-free PWM operation. New values take effect when TMR2 = PR2. Note that PWMDC is left-justified.

The 8-bit timer TMR2 register is concatenated with either the 2-bit internal system clock (FOSC), or two bits of the prescaler, to create the 10-bit time base. The system clock is used if the Timer2 prescaler is set to 1:1.

Equation 19-2 is used to calculate the PWM pulse width.

Equation 19-3 is used to calculate the PWM duty cycle ratio.

EQUATION 19-2: PULSE WIDTH

$$Pulse\ Width = (PWMxDC) \cdot TOSC \cdot (TMR2\ Prescale\ Value)$$

EQUATION 19-3: DUTY CYCLE RATIO

$$Duty\ Cycle\ Ratio = \frac{(PWMxDC)}{4(PR2 + 1)}$$

19.1.5 PWM RESOLUTION

The resolution determines the number of available duty cycles for a given period. For example, a 10-bit resolution will result in 1024 discrete duty cycles, whereas an 8-bit resolution will result in 256 discrete duty cycles.

The maximum PWM resolution is ten bits when PR2 is 255. The resolution is a function of the PR2 register value as shown by Equation 19-4.

EQUATION 19-4: PWM RESOLUTION

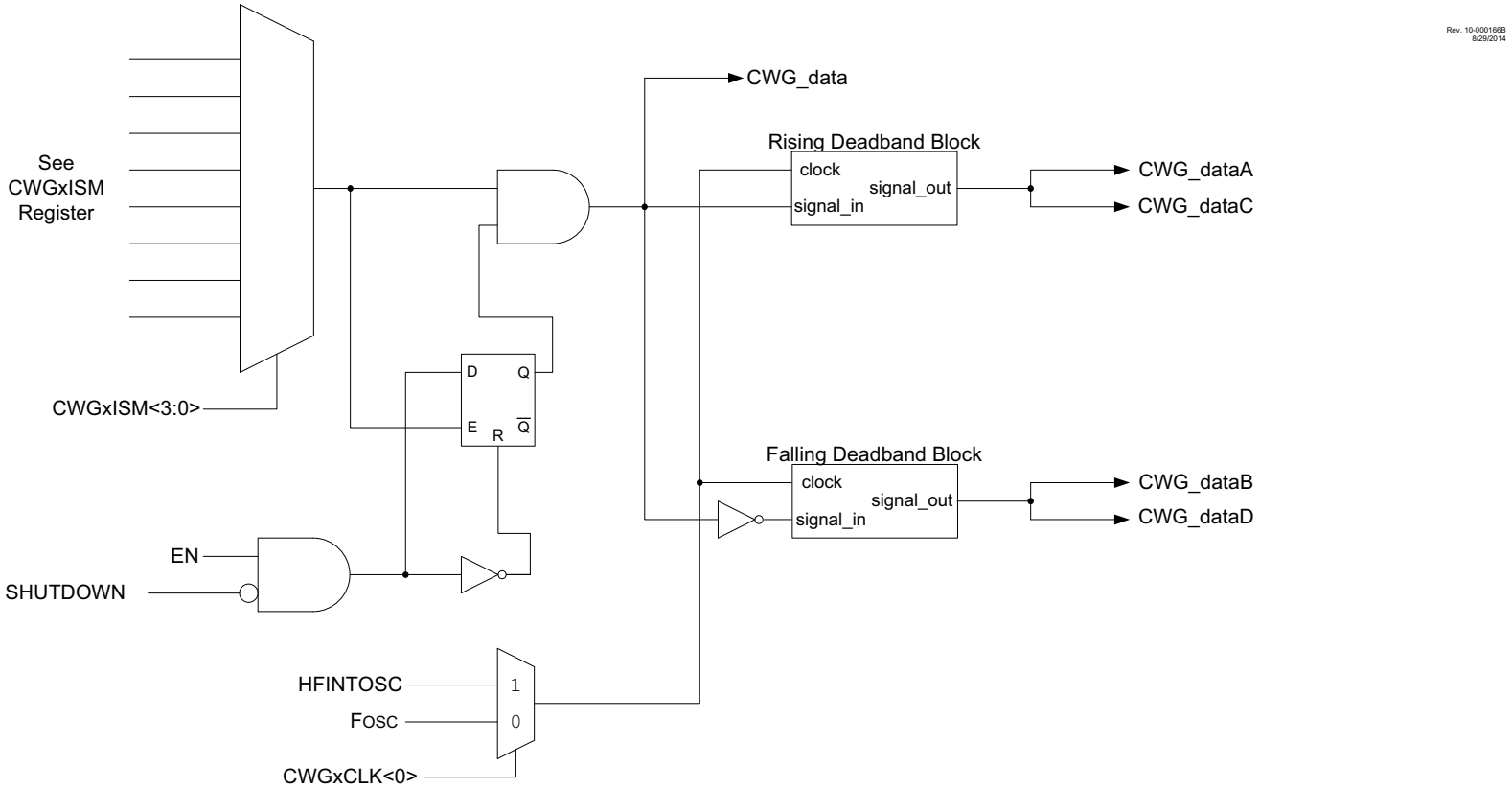
$$Resolution = \frac{\log[4(PR2 + 1)]}{\log(2)}\ bits$$

Note:	If the pulse width value is greater than the period the assigned PWM pin(s) will remain unchanged.

19.1.6 OPERATION IN SLEEP MODE

In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the PWMx pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

FIGURE 20-1: SIMPLIFIED CWG BLOCK DIAGRAM (HALF-BRIDGE MODE)



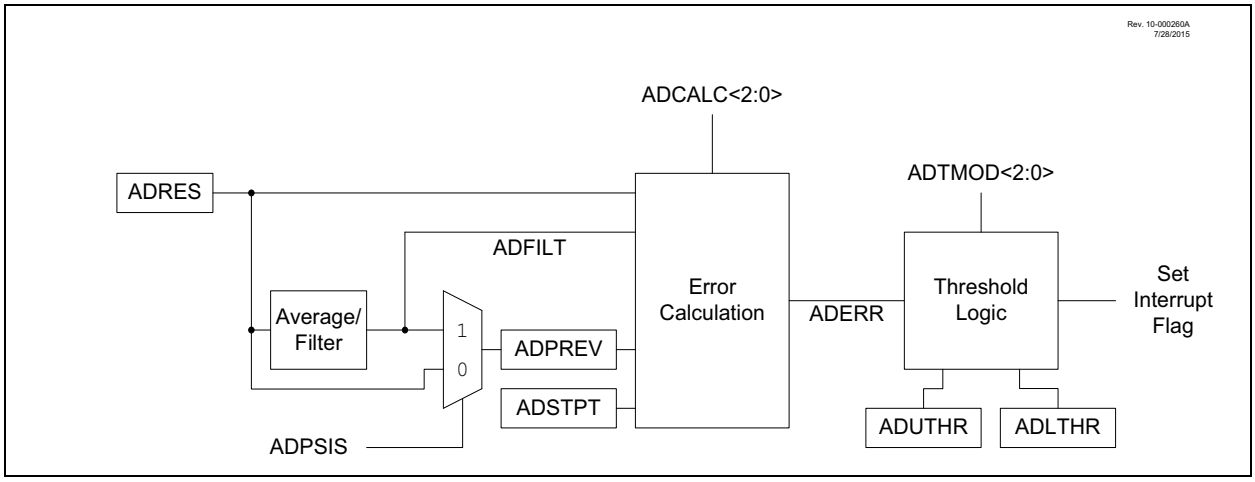
23.4.5 ADDITIONAL SAMPLE AND HOLD CAPACITANCE

Additional capacitance can be added in parallel with the internal sample and hold capacitor (CHOLD) by means of the ADCAP register. This register selects a digitally programmable capacitance which is added to the ADC conversion bus, increasing the effective internal capacitance of the sample and hold capacitor in the ADC module. This is used to improve the match between internal and external capacitance for a better sensing performance. The additional capacitance does not affect analog performance of the ADC because it is not connected during conversion. See Figure 23-10.

23.5 Computation Operation

The ADC module hardware is equipped with post conversion computation features. These features provide data post-processing functions that can be operated on the ADC conversion result, including digital filtering/averaging and threshold comparison functions.

FIGURE 23-10: COMPUTATIONAL FEATURES SIMPLIFIED BLOCK DIAGRAM



The operation of the ADC computational features is controlled by the ADMD <2:0> bits in the ADCON2 register.

The module can be operated in one of five modes:

• **Basic**: This is a legacy mode. In this mode, ADC conversion occurs on single (ADDSEN=0) or double (ADDSEN=1) samples. ADIF is set after each conversion completes.

• **Accumulate**: With each trigger, the ADC conversion result is added to accumulator and ADCNT increments. ADIF is set after each conversion. ADTIF is set according to the Calculation mode.

• **Average**: With each trigger, the ADC conversion result is added to the accumulator. When the ADRPT number of samples have been accumulated, a threshold test is performed. Upon the next trigger, the counter is reset to '1' and the accumulator is replaced with the first ADC conversion cleared. For the subsequent threshold tests, additional ADRPT samples are required to be accumulated.

• **Burst Average**: At the trigger, the accumulator and counter are cleared. The ADC conversion results are then collected repetitively until ADRPT samples are accumulated and finally the threshold is tested.

• **Low-Pass Filter (LPF)**: With each trigger, the ADC conversion result is sent through a filter. When ADRPT samples have occurred, a threshold test is performed. Every trigger after that the ADC conversion result is sent through the filter and another threshold test is performed.

The five modes are summarized in Table 23-3 below.

REGISTER 23-2: ADCON1: ADC CONTROL REGISTER 1

R/W-0/0	R/W-0/0	R/W-0/0	U-0	U-0	U-0	U-0	R/W-0/0
ADPPOL	ADIPEN	ADGPOL	—	—	—	—	ADDSEN
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7 **ADDPOL:** Precharge Polarity bit
If ADPRE>0x00:

ADPPOL	Action During 1st Precharge Stage	
	External (selected analog I/O pin)	Internal (AD sampling capacitor)
1	Shorted to AV_{DD}	C_{HOLD} shorted to V_{SS}
0	Shorted to V_{SS}	C_{HOLD} shorted to AV_{DD}

Otherwise
The bit is ignored

bit 6 **ADIPEN:** A/D Inverted Precharge Enable bit
If ADDSEN = 1:
1 = The precharge and guard signals in the second conversion cycle are the opposite polarity of the first cycle
0 = Both Conversion cycles use the precharge and guards specified by ADPPOL and ADGPOL
Otherwise:
The bit is ignored

bit 5 **ADGPOL:** Guard Ring Polarity Selection bit
1 = ADC guard ring outputs start as digital high during precharge stage
0 = ADC guard ring outputs start as digital low during precharge stage

bit 4-1 **Unimplemented:** Read as '0'

bit 0 **ADDSEN:** Double-Sample Enable bit
1 = See Table 23-5.
0 = One conversion is performed for each trigger

TABLE 23-5: EXAMPLE OF REGISTER VALUES FOR ACCUMULATE AND AVERAGE MODES

Trigger ADCONT		Sample n	ADRES	ADPREV ADPSIS		ADACC
0	1			0	1	
T1	T1	1	S(n)	S(n-1)	ADFLTR(n-1)	ADACC(n-1)-S(n-1)
T2	—	2	S(n)	S(n-1)	ADFLTR(n-2)	ADACC(n-1)+S(n-1)
T3	T2	3	S(n)	S(n-1)	ADFLTR(n-1)	ADACC(n-1)-S(n-1)
T4	—	4	S(n)	S(n-1)	ADFLTR(n-2)	ADACC(n-1)+S(n-1)
T5	T3	5	S(n)	S(n-1)	ADFLTR(n-1)	ADACC(n-1)-S(n-1)
T6	—	6	S(n)	S(n-1)	ADFLTR(n-2)	ADACC(n-1)+S(n-1)

27.0 TIMER0 MODULE

The Timer0 module is an 8/16-bit timer/counter with the following features:

- 16-bit timer/counter
- 8-bit timer/counter with programmable period
- Synchronous or asynchronous operation
- Selectable clock sources
- Programmable prescaler (independent of Watchdog Timer)
- Programmable postscaler
- Operation during Sleep mode
- Interrupt on match or overflow
- Output on I/O pin (via PPS) or to other peripherals

27.1 Timer0 Operation

Timer0 can operate as either an 8-bit timer/counter or a 16-bit timer/counter. The mode is selected with the T016BIT bit of the T0CON register.

When used with an internal clock source, the module is a timer and increments on every instruction cycle. When used with an external clock source, the module can be used as either a timer or a counter and increments on every rising edge of the external source.

27.1.1 16-BIT MODE

In normal operation, TMR0 increments on the rising edge of the clock source. A 15-bit prescaler on the clock input gives several prescale options (see prescaler control bits, T0CKPS<3:0> in the T0CON1 register).

27.1.1.1 Timer0 Reads and Writes in 16-Bit Mode

TMR0H is not the actual high byte of Timer0 in 16-bit mode. It is actually a buffered version of the real high byte of Timer0, which is neither directly readable nor writable (see Figure 27-1). TMR0H is updated with the contents of the high byte of Timer0 during a read of TMR0L. This provides the ability to read all 16 bits of Timer0 without having to verify that the read of the high and low byte was valid, due to a rollover between successive reads of the high and low byte.

Similarly, a write to the high byte of Timer0 must also take place through the TMR0H Buffer register. The high byte is updated with the contents of TMR0H when a write occurs to TMR0L. This allows all 16 bits of Timer0 to be updated at once.

27.1.2 8-BIT MODE

In normal operation, TMR0 increments on the rising edge of the clock source. A 15-bit prescaler on the clock input gives several prescale options (see prescaler control bits, T0CKPS<3:0> in the T0CON1 register).

The value of TMR0L is compared to that of the Period buffer, a copy of TMR0H, on each clock cycle. When the two values match, the following events happen:

- TMR0_out goes high for one prescaled clock period
- TMR0L is reset
- The contents of TMR0H are copied to the period buffer

In 8-bit mode, the TMR0L and TMR0H registers are both directly readable and writable. The TMR0L register is cleared on any device Reset, while the TMR0H register initializes at FFh.

Both the prescaler and postscaler counters are cleared on the following events:

- A write to the TMR0L register
- A write to either the T0CON0 or T0CON1 registers
- Any device Reset – Power-on Reset (POR), MCLR Reset, Watchdog Timer Reset (WDTR) or
- Brown-out Reset (BOR)

27.1.3 COUNTER MODE

In Counter mode, the prescaler is normally disabled by setting the T0CKPS bits of the T0CON1 register to '0000'. Each rising edge of the clock input (or the output of the prescaler if the prescaler is used) increments the counter by '1'.

27.1.4 TIMER MODE

In Timer mode, the Timer0 module will increment every instruction cycle as long as there is a valid clock signal and the T0CKPS bits of the T0CON1 register (Register 27-2) are set to '0000'. When a prescaler is added, the timer will increment at the rate based on the prescaler value.

27.1.5 ASYNCHRONOUS MODE

When the T0ASYNC bit of the T0CON1 register is set (T0ASYNC = '1'), the counter increments with each rising edge of the input source (or output of the prescaler, if used). Asynchronous mode allows the counter to continue operation during Sleep mode provided that the clock also continues to operate during Sleep.

27.1.6 SYNCHRONOUS MODE

When the T0ASYNC bit of the T0CON1 register is clear (T0ASYNC = 0), the counter clock is synchronized to the system oscillator (FOSC/4). When operating in Synchronous mode, the counter clock frequency cannot exceed FOSC/4.

27.2 Clock Source Selection

The T0CS<2:0> bits of the T0CON1 register are used to select the clock source for Timer0. Register 27-2 displays the clock source selections.

REGISTER 29-4: TxRST: TIMER2/4/6 EXTERNAL RESET SIGNAL SELECTION REGISTER

U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	RSEL<4:0>				
bit 7							bit 0

Legend:		
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-5 **Unimplemented:** Read as '0'

bit 4-0 **RSEL<4:0>:** Timer2 External Reset Signal Source Selection bits

11111 = Reserved
•
•
•
10010 = Reserved
10001 = LC4_out
10000 = LC3_out
01111 = LC2_out
01110 = LC1_out
01101 = ZCD1_output
01100 = C2OUT_sync
01011 = C1OUT_sync
01010 = PWM7_out
01001 = PWM6_out
01000 = CCP5_out
00111 = CCP4_out
00110 = CCP3_out
00101 = CCP2_out
00100 = CCP1_out
00011 = TMR6_postscaled(3)
00010 = TMR4_postscaled(2)
00001 = TMR2_postscaled(1)
00000 = Pin selected by TxINPPS

Note 1: For Timer2, this bit is Reserved.
2: For Timer4, this bit is Reserved.
3: For Timer6, this bit is Reserved.

30.0 CAPTURE/COMPARE/PWM MODULES

The Capture/Compare/PWM module is a peripheral that allows the user to time and control different events, and to generate Pulse-Width Modulation (PWM) signals. In Capture mode, the peripheral allows the timing of the duration of an event. The Compare mode allows the user to trigger an external event when a predetermined amount of time has expired. The PWM mode can generate Pulse-Width Modulated signals of varying frequency and duty cycle.

The Capture/Compare/PWM modules available are shown in Table 30-1.

TABLE 30-1: AVAILABLE CCP MODULES

Device	CCP1	CCP2	CCP3	CCP4	CCP5
PIC16(L)F18855/75	•	•	•	•	•

The Capture and Compare functions are identical for all CCP modules.

Note 1: In devices with more than one CCP module, it is very important to pay close attention to the register names used. A number placed after the module acronym is used to distinguish between separate modules. For example, the CCP1CON and CCP2CON control the same operational aspects of two completely different CCP modules.

2: Throughout this section, generic references to a CCP module in any of its operating modes may be interpreted as being equally applicable to CCPx module. Register names, module signals, I/O pins, and bit names may use the generic designator ‘x’ to indicate the use of a numeral to distinguish a particular module, when required.

33.3 EUSART Baud Rate Generator (BRG)

The Baud Rate Generator (BRG) is an 8-bit or 16-bit timer that is dedicated to the support of both the asynchronous and synchronous EUSART operation. By default, the BRG operates in 8-bit mode. Setting the BRG16 bit of the BAUD1CON register selects 16-bit mode.

The SPBRGH, SPBRGL register pair determines the period of the free running baud rate timer. In Asynchronous mode the multiplier of the baud rate period is determined by both the BRGH bit of the TX1STA register and the BRG16 bit of the BAUD1CON register. In Synchronous mode, the BRGH bit is ignored.

Table 33-1 contains the formulas for determining the baud rate. Example 33-1 provides a sample calculation for determining the baud rate and baud rate error.

Typical baud rates and error values for various Asynchronous modes have been computed for your convenience and are shown in Table 33-3. It may be advantageous to use the high baud rate (BRGH = 1), or the 16-bit BRG (BRG16 = 1) to reduce the baud rate error. The 16-bit BRG mode is used to achieve slow baud rates for fast oscillator frequencies.

Writing a new value to the SPBRGH, SPBRGL register pair causes the BRG timer to be reset (or cleared). This ensures that the BRG does not wait for a timer overflow before outputting the new baud rate.

If the system clock is changed during an active receive operation, a receive error or data loss may result. To avoid this problem, check the status of the RCIDL bit to make sure that the receive operation is idle before changing the system clock.

EXAMPLE 33-1: CALCULATING BAUD RATE ERROR

For a device with FOSC of 16 MHz, desired baud rate of 9600, Asynchronous mode, 8-bit BRG:

$$Desired\ Baud\ Rate = \frac{FOSC}{64([SPBRGH:SPBRGL] + 1)}$$

Solving for SPBRGH:SPBRGL:

$$X = \frac{\frac{FOSC}{Desired\ Baud\ Rate}}{64} - 1$$

$$= \frac{\frac{16000000}{9600}}{64} - 1$$

$$= [25.042] = 25$$

$$Calculated\ Baud\ Rate = \frac{16000000}{64(25 + 1)}$$

$$= 9615$$

$$Error = \frac{Calc.\ Baud\ Rate - Desired\ Baud\ Rate}{Desired\ Baud\ Rate}$$

$$= \frac{(9615 - 9600)}{9600} = 0.16\%$$

FIGURE 33-7: AUTO-WAKE-UP BIT (WUE) TIMING DURING NORMAL OPERATION

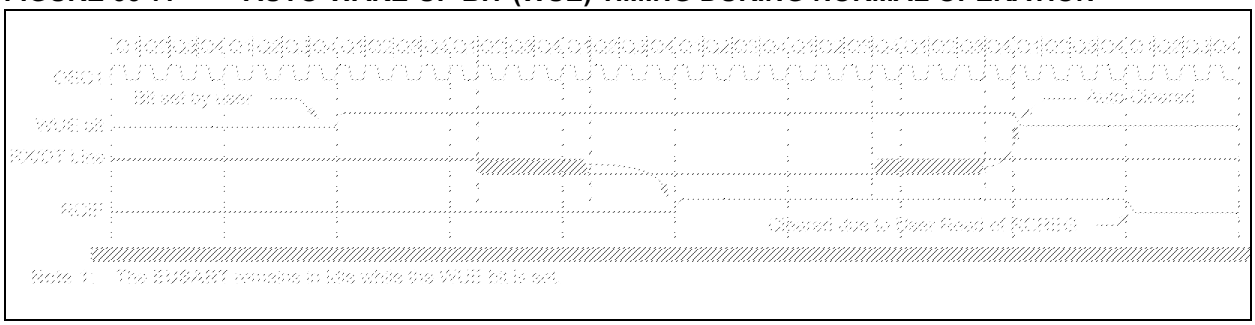
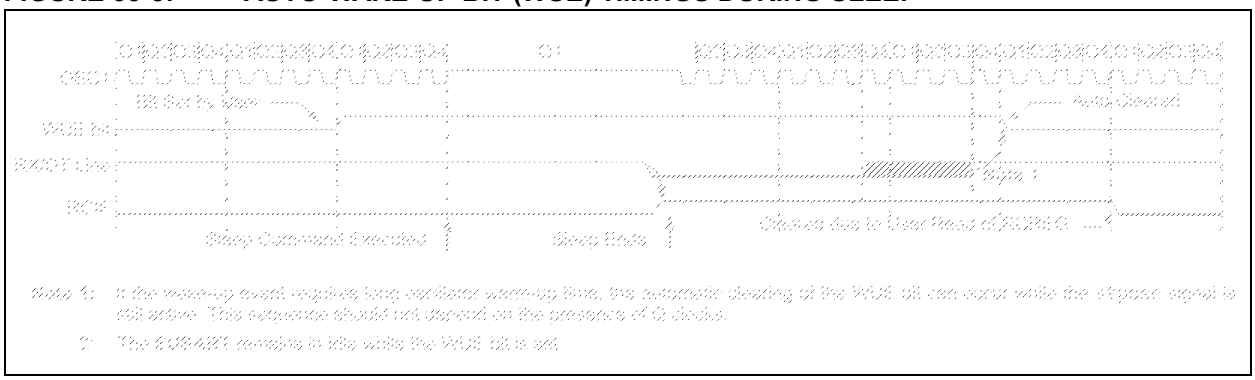


FIGURE 33-8: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING SLEEP



33.3.4 BREAK CHARACTER SEQUENCE

The EUSART module has the capability of sending the special Break character sequences that are required by the LIN bus standard. A Break character consists of a Start bit, followed by 12 ‘0’ bits and a Stop bit.

To send a Break character, set the SENDB and TXEN bits of the TX1STA register. The Break character transmission is then initiated by a write to the TXREG. The value of data written to TXREG will be ignored and all ‘0’s will be transmitted.

The SENDB bit is automatically reset by hardware after the corresponding Stop bit is sent. This allows the user to preload the transmit FIFO with the next transmit byte following the Break character (typically, the Sync character in the LIN specification).

The TRMT bit of the TX1STA register indicates when the transmit operation is active or idle, just as it does during normal transmission. See Figure 33-9 for the timing of the Break character sequence.

33.3.4.1 Break and Sync Transmit Sequence

The following sequence will start a message frame header made up of a Break, followed by an auto-baud Sync byte. This sequence is typical of a LIN bus master.

1. Configure the EUSART for the desired mode.
2. Set the TXEN and SENDB bits to enable the Break sequence.
3. Load the TXREG with a dummy character to initiate transmission (the value is ignored).
4. Write ‘55h’ to TXREG to load the Sync character into the transmit FIFO buffer.
5. After the Break has been sent, the SENDB bit is reset by hardware and the Sync character is then transmitted.

When the TXREG becomes empty, as indicated by the TXIF, the next data byte can be written to TXREG.

TABLE 33-4: BAUD RATE FOR ASYNCHRONOUS MODES (CONTINUED)

BAUD RATE	SYNC = 0, BRGH = 0, BRG16 = 1											
	Fosc = 8.000 MHz			Fosc = 4.000 MHz			Fosc = 3.6864 MHz			Fosc = 1.000 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	299.9	-0.02	1666	300.1	0.04	832	300.0	0.00	767	300.5	0.16	207
1200	1199	-0.08	416	1202	0.16	207	1200	0.00	191	1202	0.16	51
2400	2404	0.16	207	2404	0.16	103	2400	0.00	95	2404	0.16	25
9600	9615	0.16	51	9615	0.16	25	9600	0.00	23	—	—	—
10417	10417	0.00	47	10417	0.00	23	10473	0.53	21	10417	0.00	5
19.2k	19.23k	0.16	25	19.23k	0.16	12	19.20k	0.00	11	—	—	—
57.6k	55556	-3.55	8	—	—	—	57.60k	0.00	3	—	—	—
115.2k	—	—	—	—	—	—	115.2k	0.00	1	—	—	—

BAUD RATE	SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1											
	Fosc = 32.000 MHz			Fosc = 20.000 MHz			Fosc = 18.432 MHz			Fosc = 11.0592 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	300.0	0.00	26666	300.0	0.00	16665	300.0	0.00	15359	300.0	0.00	9215
1200	1200	0.00	6666	1200	-0.01	4166	1200	0.00	3839	1200	0.00	2303
2400	2400	0.01	3332	2400	0.02	2082	2400	0.00	1919	2400	0.00	1151
9600	9604	0.04	832	9597	-0.03	520	9600	0.00	479	9600	0.00	287
10417	10417	0.00	767	10417	0.00	479	10425	0.08	441	10433	0.16	264
19.2k	19.18k	-0.08	416	19.23k	0.16	259	19.20k	0.00	239	19.20k	0.00	143
57.6k	57.55k	-0.08	138	57.47k	-0.22	86	57.60k	0.00	79	57.60k	0.00	47
115.2k	115.9k	0.64	68	116.3k	0.94	42	115.2k	0.00	39	115.2k	0.00	23

BAUD RATE	SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1											
	Fosc = 8.000 MHz			Fosc = 4.000 MHz			Fosc = 3.6864 MHz			Fosc = 1.000 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	300.0	0.00	6666	300.0	0.01	3332	300.0	0.00	3071	300.1	0.04	832
1200	1200	-0.02	1666	1200	0.04	832	1200	0.00	767	1202	0.16	207
2400	2401	0.04	832	2398	0.08	416	2400	0.00	383	2404	0.16	103
9600	9615	0.16	207	9615	0.16	103	9600	0.00	95	9615	0.16	25
10417	10417	0	191	10417	0.00	95	10473	0.53	87	10417	0.00	23
19.2k	19.23k	0.16	103	19.23k	0.16	51	19.20k	0.00	47	19.23k	0.16	12
57.6k	57.14k	-0.79	34	58.82k	2.12	16	57.60k	0.00	15	—	—	—
115.2k	117.6k	2.12	16	111.1k	-3.55	8	115.2k	0.00	7	—	—	—

TABLE 37-4: I/O PORTS

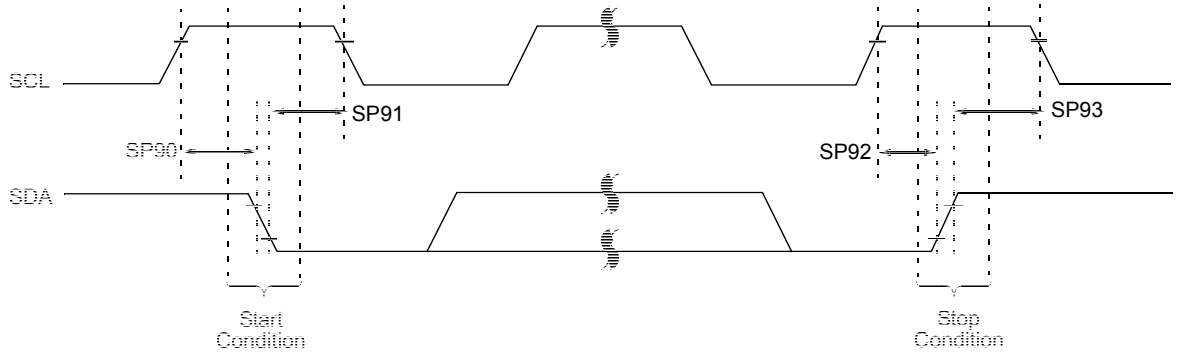
Standard Operating Conditions (unless otherwise stated)							
Param. No.	**Sym.**	**Characteristic**	**Min.**	**Typ†**	**Max.**	**Units**	**Conditions**
	V_{IL}	**Input Low Voltage**					
		I/O PORT:					
D300		with TTL buffer	—	—	0.8	V	$4.5V \le V_{DD} \le 5.5V$
D301			—	—	0.15 V_{DD}	V	$1.8V \le V_{DD} \le 4.5V$
D302		with Schmitt Trigger buffer	—	—	0.2 V_{DD}	V	$2.0V \le V_{DD} \le 5.5V$
D303		with I^2C levels	—	—	0.3 V_{DD}	V	
D304		with SMBus levels	—	—	0.8	V	$2.7V \le V_{DD} \le 5.5V$
D305		$\overline{MCLR}$	—	—	0.2 V_{DD}	V	
	V_{IH}	**Input High Voltage**					
		I/O PORT:					
D320		with TTL buffer	2.0	—	—	V	$4.5V \le V_{DD} \le 5.5V$
D321			0.25 V_{DD} + 0.8	—	—	V	$1.8V \le V_{DD} \le 4.5V$
D322		with Schmitt Trigger buffer	0.8 V_{DD}	—	—	V	$2.0V \le V_{DD} \le 5.5V$
D323		with I^2C levels	0.7 V_{DD}	—	—	V	
D324		with SMBus levels	2.1	—	—	V	$2.7V \le V_{DD} \le 5.5V$
D325		$\overline{MCLR}$	0.7 V_{DD}	—	—	V	
	I_{IL}	**Input Leakage Current(1)**					
D340		I/O Ports	—	± 5	± 125	nA	$V_{SS} \le V_{PIN} \le V_{DD}$, Pin at high-impedance, 85°C
D341			—	± 5	± 1000	nA	$V_{SS} \le V_{PIN} \le V_{DD}$, Pin at high-impedance, 125°C
D342		$\overline{MCLR}$(2)	—	± 50	± 200	nA	$V_{SS} \le V_{PIN} \le V_{DD}$, Pin at high-impedance, 85°C
	I_{PUR}	**Weak Pull-up Current**					
D350			25	120	200	µA	V_{DD} = 3.0V, V_{PIN} = V_{SS}
	V_{OL}	**Output Low Voltage**					
D360		I/O ports	—	—	0.6	V	I_{OL} = 10.0mA, V_{DD} = 3.0V
	V_{OH}	**Output High Voltage**					
D370		I/O ports	V_{DD} - 0.7	—	—	V	I_{OH} = 6.0 mA, V_{DD} = 3.0V
D380	C_{IO}	**All I/O pins**	—	5	50	pF	

† Data in "Typ" column is at 3.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Negative current is defined as current sourced by the pin.

2: The leakage current on the $\overline{MCLR}$ pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

FIGURE 37-21: I²C BUS START/STOP BITS TIMING



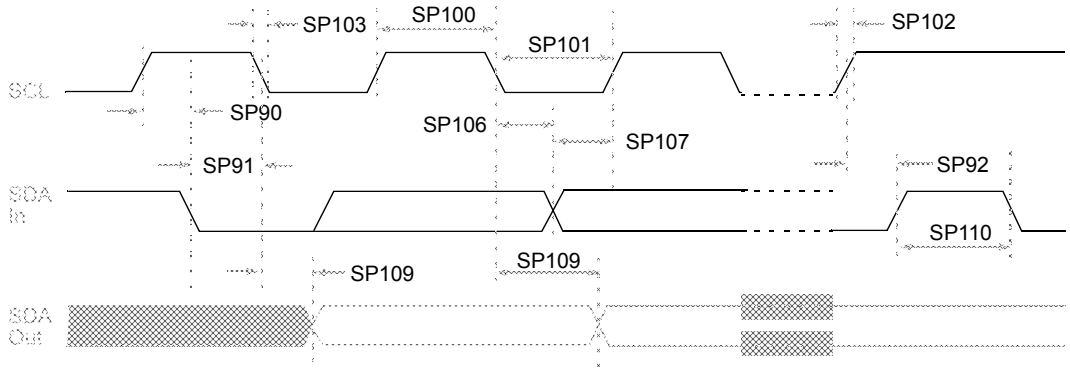
Note: Refer to Figure 37-4 for load conditions.

TABLE 37-24: I²C BUS START/STOP BITS REQUIREMENTS

Standard Operating Conditions (unless otherwise stated)								
Param. No.	**Symbol**	**Characteristic**		**Min.**	**Typ**	**Max.**	**Units**	**Conditions**
SP90*	TSU:STA	Start condition	100 kHz mode	4700	—	—	ns	Only relevant for Repeated Start condition
		Setup time	400 kHz mode	600	—	—		
SP91*	THD:STA	Start condition	100 kHz mode	4000	—	—	ns	After this period, the first clock pulse is generated
		Hold time	400 kHz mode	600	—	—		
SP92*	TSU:STO	Stop condition	100 kHz mode	4700	—	—	ns	
		Setup time	400 kHz mode	600	—	—		
SP93	THD:STO	Stop condition	100 kHz mode	4000	—	—	ns	
		Hold time	400 kHz mode	600	—	—		

* These parameters are characterized but not tested.

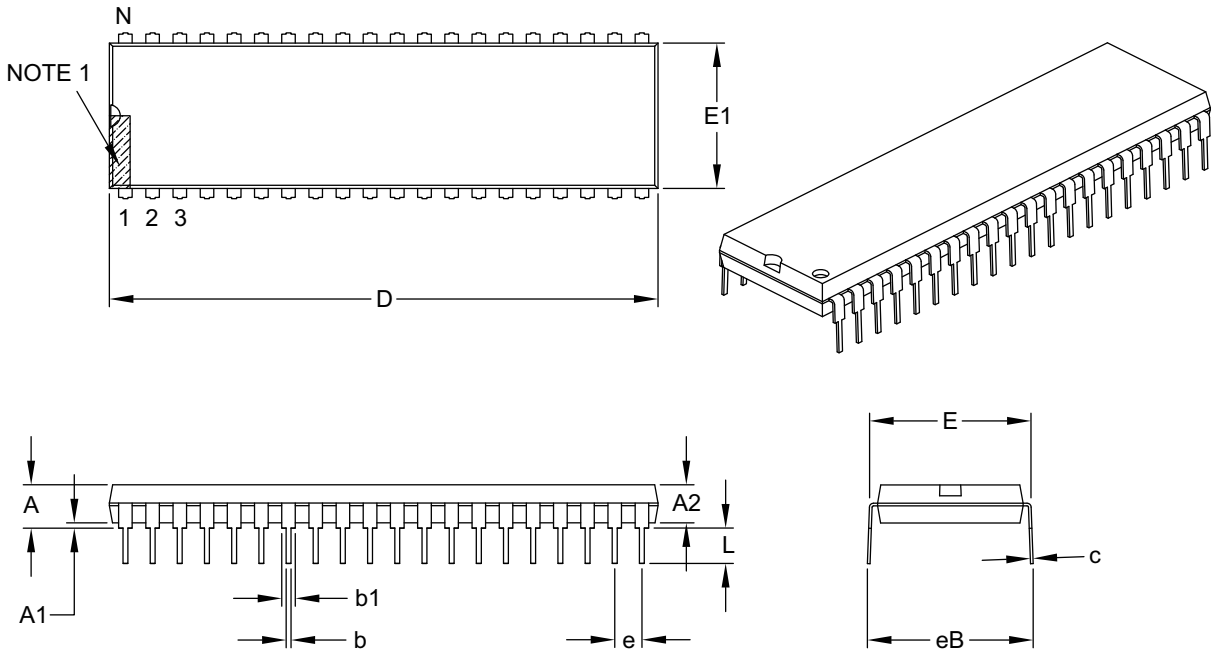
FIGURE 37-22: I²C BUS DATA TIMING



Note: Refer to Figure 37-4 for load conditions.

40-Lead Plastic Dual In-Line (P) – 600 mil Body [PDIP]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Units		INCHES		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N		40	
Pitch	e		.100 BSC	
Top to Seating Plane	A	–	–	.250
Molded Package Thickness	A2	.125	–	.195
Base to Seating Plane	A1	.015	–	–
Shoulder to Shoulder Width	E	.590	–	.625
Molded Package Width	E1	.485	–	.580
Overall Length	D	1.980	–	2.095
Tip to Seating Plane	L	.115	–	.200
Lead Thickness	c	.008	–	.015
Upper Lead Width	b1	.030	–	.070
Lower Lead Width	b	.014	–	.023
Overall Row Spacing §	eB	–	–	.700

Notes:

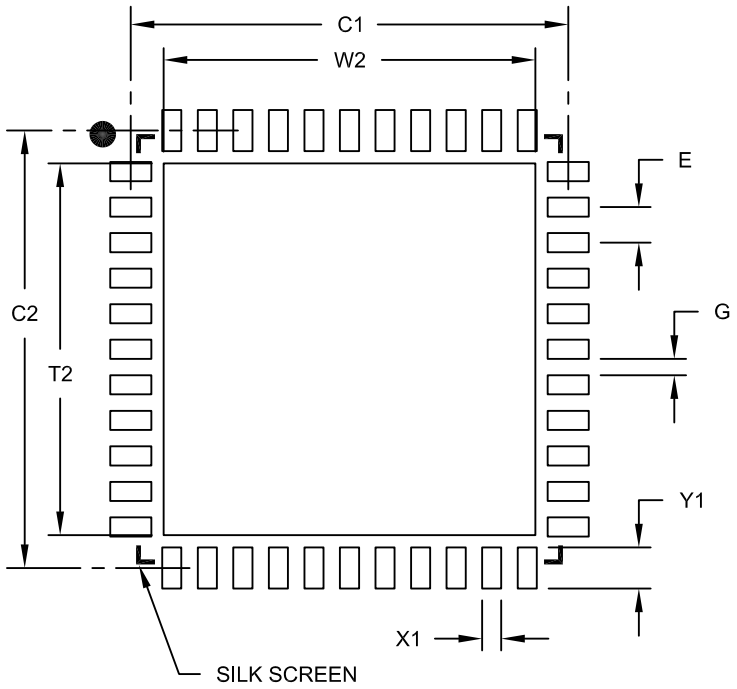
1. Pin 1 visual index feature may vary, but must be located within the hatched area.
2. § Significant Characteristic.
3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" per side.
4. Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-016B

44-Lead Plastic Quad Flat, No Lead Package (ML) - 8x8 mm Body [QFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Contact Pitch	E		0.65 BSC	
Optional Center Pad Width	W2			6.60
Optional Center Pad Length	T2			6.60
Contact Pad Spacing	C1		8.00	
Contact Pad Spacing	C2		8.00	
Contact Pad Width (X44)	X1			0.35
Contact Pad Length (X44)	Y1			0.85
Distance Between Pads	G	0.25		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2103B