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Intel - EP1SGX25FF1020C5 Datasheet



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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	2566
Number of Logic Elements/Cells	25660
Total RAM Bits	1944576
Number of I/O	607
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	1020-BBGA
Supplier Device Package	1020-FBGA (33x33)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1sgx25ff1020c5

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Figure 1–1. Stratix GX I/O Blocks Note (1)

Notes to Figure 1–1:

- (1) Figure 1–1 is a top view of the Stratix GX silicon die.
- (2) Banks 9 through 12 are enhanced PLL external clock output banks.
- (3) If the high-speed differential I/O pins are not used for high-speed differential signaling, they can support all of the I/O standards except HSTL class I and II, GTL, SSTL-18 Class II, PCI, PCI-X, and AGP 1×/2×.
- (4) For guidelines for placing single-ended I/O pads next to differential I/O pads, see the Selectable I/O Standards in Stratix & Stratix GX Devices chapter of the Stratix GX Device Handbook, Volume 2.
- (5) These I/O banks in Stratix GX devices also support the LVDS, LVPECL, and 3.3-V PCML I/O standards on reference clocks and receiver input pins (AC coupled).

FPGA Functional Description

Stratix GX devices contain a two-dimensional row- and column-based architecture to implement custom logic. A series of column and row interconnects of varying length and speed provide signal interconnects between logic array blocks (LABs), memory block structures, and DSP blocks.





Note to Figure 2–2:(1) There are four transceiver channels in a transceiver block.

Receiver State Machine

The receiver state machine operates in GIGE and XAUI modes. In GIGE mode, the receiver state machine replaces invalid code groups with 9 ' h1FE. In XAUI mode, the receiver state machine translates the XAUI PCS code group to the XAUI XGMII code group. Table 2–8 shows the code conversion. The conversion adheres to the IEEE 802.3ae specification.

Table 2–8. Code Conversion							
XGMII RXC	XGMII RXD	PCS code-group	Description				
0	00 through FF	Dxx.y	Normal Data				
1	07	K28.0 or K28.3 or K28.5	Idle in I				
1	07	K28.5	Idle in T				
1	9C	K28.4	Sequence				
1	FB	K27.7	Start				
1	FD	K29.7	Terminate				
1	FE	K30.7	Error				
1	FE	Invalid code group	Invalid XGMII character				
1	See IEEE 802.3 reserved code groups	See IEEE 802.3 reserved code groups	Reserved code groups				

Byte Deserializer

The byte deserializer takes a single width word (8 or 10 bits) from the transceiver logic and converts it into double-width words (16 or 20 bits) to the phase compensation FIFO buffer. The byte deserializer is bypassed when single width mode (8 or 10 bits) is used at the PLD interface.

Phase Compensation FIFO Buffer

The receiver phase compensation FIFO buffer resides in the transceiver block at the programmable logic device (PLD) boundary. This buffer compensates for the phase difference between the recovered clock within the transceiver and the recovered clock after it has transferred to the PLD core. The phase compensation FIFO buffer is four words deep and cannot be bypassed.

Table 2–10. Possible Clocking Connections for Transceivers (Part 2 of 2)						
	Destination					
Source	Transmitter PLL	Receiver PLL	GCLK	RCLK	FCLK	IQ Lines
IQ lines	✓ (2)	✓ (2)				

Notes to Table 2–10:

(1) REFCLKB from transceiver block 0 and transceiver block 4 does not drive the inter-transceiver lines or the GCLK lines.

(2) Inter-transceiver line 0 and inter-transceiver line 1 drive the transmitter PLL, while inter-transceiver line 2 drives the receiver PLLs.

Other Transceiver Features

Other important features of the Stratix GX transceivers are the power down and reset capabilities, the external voltage reference and bias circuitry, and hot swapping.

Individual Power-Down & Reset for the Transmitter & Receiver

Stratix GX transceivers offer a power saving advantage with their ability to shut off functions that are not needed. The device can individually reset the receiver and transmitter blocks and the PLLs. The Stratix GX device can either globally power down and reset the transmitter and receiver channels or do each channel separately. Table 2–11 shows the connectivity between the reset signals and the Stratix GX logical blocks.

multiplication value. The $\times 1$ and $\times 2$ operation is also possible by bypassing the SERDES. The SERDES DPA cannot support $\times 1$, $\times 2$, or $\times 4$ natively.

On the receiver side, the high-frequency clock generated by the PLL shifts the serial data through a shift register (also called deserializer). The parallel data is clocked out to the logic array synchronized with the lowfrequency clock. On the transmitter side, the parallel data from the logic array is first clocked into a parallel-in, serial-out shift register synchronized with the low-frequency clock and then transmitted out by the output buffers.

There are two dedicated fast PLLs each in EP1SGX10 to EP1SGX25 devices, and four in EP1SGX40 devices. These PLLs are used for the SERDES operations as well as general-purpose use.

Stratix GX Differential I/O Receiver Operation (Non-DPA Mode)

You can configure any of the Stratix GX source synchronous differential input channels as a receiver channel (see Figure 3–1). The differential receiver deserializes the incoming high-speed data. The input shift register continuously clocks the incoming data on the negative transition of the high-frequency clock generated by the PLL clock (×*W*).

The data in the serial shift register is shifted into a parallel register by the RXLOADEN signal generated by the fast PLL counter circuitry on the third falling edge of the high-frequency clock. However, you can select which falling edge of the high frequency clock loads the data into the parallel register, using the data-realignment circuit.

In normal mode, the enable signal RXLOADEN loads the parallel data into the next parallel register on the second rising edge of the low-frequency clock. You can also load data to the parallel register through the TXLOADEN signal when using the data-realignment circuit.

Figure 3–1 shows the block diagram of a single SERDES receiver channel. Figure 3–2 shows the timing relationship between the data and clocks in Stratix GX devices in ×10 mode. *W* is the low-frequency multiplier and *J* is the data parallelization division factor.



Figure 3–1. Stratix GX High-Speed Interface Deserialized in ×10 Mode

Notes to Figure 3–1:

 W = 1, 2, 4, 7, 8, or 10. *J* = 4, 7, 8, or 10 for non-DPA (*J* = 8 or 10 for DPA). W does not have to equal *J*. When *J* = 1 or 2, the deserializer is bypassed. When *J* = 2, the device uses DDRIO registers.
(2) This figure does not show additional circuitry for clock or data manipulation.

Figure 3–2. Receiver Timing Diagram



Stratix GX Differential I/O Transmitter Operation

You can configure any of the Stratix GX differential output channels as a transmitter channel. The differential transmitter serializes outbound parallel data.

its own fan-out LUT. This provides another mechanism for improved fitting. The LE can also drive out registered and unregistered versions of the LUT output.

LUT Chain & Register Chain

In addition to the three general routing outputs, the LEs within an LAB have LUT chain and register chain outputs. LUT chain connections allow LUTs within the same LAB to cascade together for wide input functions. Register chain outputs allow registers within the same LAB to cascade together. The register chain output allows an LAB to use LUTs for a single combinatorial function and the registers to be used for an unrelated shift register implementation. These resources speed up connections between LABs while saving local interconnect resources. See "MultiTrack Interconnect" on page 4–11 for more information on LUT chain and register chain connections.

addnsub Signal

The LE's dynamic adder/subtractor feature saves logic resources by using one set of LEs to implement both an adder and a subtractor. This feature is controlled by the LAB-wide control signal addnsub. The addnsub signal sets the LAB to perform either A + B or A – B. The LUT computes addition, and subtraction is computed by adding the two's complement of the intended subtractor. The LAB-wide signal converts to two's complement by inverting the B bits within the LAB and setting carry-in = 1 to add one to the least significant bit (LSB). The LSB of an adder/subtractor must be placed in the first LE of the LAB, where the LAB-wide addnsub signal automatically sets the carry-in to 1. The Quartus II Compiler automatically places and uses the adder/subtractor feature when using adder/subtractor parameterized functions.

LE Operating Modes

The Stratix GX LE can operate in one of the following modes:

- Normal mode
- Dynamic arithmetic mode

Each mode uses LE resources differently. In each mode, eight available inputs to the LE—the four data inputs from the LAB local interconnect; carry-in0 and carry-in1 from the previous LE; the LAB carry-in from the previous carry-chain LAB; and the register chain connection—are directed to different destinations to implement the desired logic function. LAB-wide signals provide clock, asynchronous clear, asynchronous preset load, synchronous clear, synchronous load, and

Table 4–2. TriMatrix Memory Features (Part 2 of 2)						
Memory Feature	M512 RAM Block (32 × 18 Bits)	M4K RAM Block (128 × 36 Bits)	M-RAM Block (4K × 144 Bits)			
Configurations	512 ×1 256 × 2 128 × 4 64 × 8 64 × 9 32 × 16 32 × 18	$\begin{array}{l} 4{\rm K}\times1\\ 2{\rm K}\times2\\ 1{\rm K}\times4\\ 512\times8\\ 512\times9\\ 256\times16\\ 256\times18\\ 128\times32\\ 128\times36 \end{array}$	$64K \times 8$ $64K \times 9$ $32K \times 16$ $32K \times 18$ $16K \times 32$ $16K \times 36$ $8K \times 64$ $8K \times 72$ $4K \times 128$ $4K \times 144$			

Notes to Table 4–2:

- (1) See the DC & Switching Characteristics chapter of the Stratix GX Device Handbook, Volume 1 for maximum performance information.
- (2) The M-RAM block does not support memory initializations. However, the M-RAM block can emulate a ROM function using a dual-port RAM bock. The Stratix GX device must write to the dual-port memory once and then disable the write-enable ports afterwards.

Memory Modes

TriMatrix memory blocks include input registers that synchronize writes and output registers to pipeline designs and improve system performance. M4K and M-RAM memory blocks offer a true dual-port mode to support any combination of two-port operations: two reads, two writes, or one read and one write at two different clock frequencies. Figure 4–11 shows true dual-port memory.





In addition to true dual-port memory, the memory blocks support simple dual-port and single-port RAM. Simple dual-port memory supports a simultaneous read and write and can either read old data before the write

single block of RAM ideal for data packet storage. The different-sized blocks allow Stratix GX devices to efficiently support variable-sized memory in designs.

The Quartus II software automatically partitions the user-defined memory into the embedded memory blocks using the most efficient size combinations. You can also manually assign the memory to a specific block size or a mixture of block sizes.

M512 RAM Block

The M512 RAM block is a simple dual-port memory block and is useful for implementing small FIFO buffers, DSP, and clock domain transfer applications. Each block contains 576 RAM bits (including parity bits). M512 RAM blocks can be configured in the following modes:

- Simple dual-port RAM
- Single-port RAM
- FIFO
- ROM
- Shift register

When configured as RAM or ROM, you can use an initialization file to pre-load the memory contents.

The memory address depths and output widths can be configured as $512 \times 1, 256 \times 2, 128 \times 4, 64 \times 8$ (64×9 bits with parity), and 32×16 (32×18 bits with parity). Mixed-width configurations are also possible, allowing different read and write widths. Table 4–3 summarizes the possible M512 RAM block configurations.

Table 4–3. M512 RAM Block Configurations (Simple Dual-Port RAM)							
			V	Vrite Port			
neau run	512 × 1	256 × 2	128 × 4	64 × 8	32 × 16	64 × 9	32 × 18
512 × 1	\checkmark	\checkmark	\checkmark	\checkmark	~		
256 × 2	\checkmark	~	~	\checkmark	~		
128 × 4	\checkmark	~	~		~		
64 × 8	\checkmark	~		\checkmark			
32 × 16	~	~	~		~		
64 × 9						\checkmark	
32 × 18							\checkmark



Figure 4–15. M512 RAM Block LAB Row Interface

M4K RAM Blocks

The M4K RAM block includes support for true dual-port RAM. The M4K RAM block implements buffers for a wide variety of applications such as storing processor code, implementing lookup schemes, and implementing larger memory applications. Each block contains 4,608 RAM bits (including parity bits). M4K RAM blocks can be configured in the following modes:

- True dual-port RAM
- Simple dual-port RAM
- Single-port RAM
- FIFO
- ROM
- Shift register

When configured as RAM or ROM, you can use an initialization file to pre-load the memory contents.

the capabilities for dynamic and mixed sign multiplications are handled differently. The following list provides the largest functions that can fit into a single DSP block.

- 36 × 36-bit unsigned by unsigned multiplication
- 36 × 36-bit signed by signed multiplication
- 35 × 36-bit unsigned by signed multiplication
- 36 × 35-bit signed by unsigned multiplication
- 36 × 35-bit signed by dynamic sign multiplication
- 35 × 36-bit dynamic sign by signed multiplication
- 35 × 36-bit unsigned by dynamic sign multiplication
- 36 × 35-bit dynamic sign by unsigned multiplication
- 35 × 35-bit dynamic sign multiplication when the sign controls for each operand are different
- 36 × 36-bit dynamic sign multiplication when the same sign control is used for both operands
- This list only shows functions that can fit into a single DSP block. Multiple DSP blocks can support larger multiplication functions.

Figure 4–28 shows one of the columns with surrounding LAB rows.





Note to Figure 4–34:

(1) These signals are not registered or registered once to match the data path pipeline.

DSP blocks can also implement one 36×36 -bit multiplier in multiplier mode. DSP blocks use four 18×18 -bit multipliers combined with dedicated adder and internal shift circuitry to achieve 36-bit multiplication. The input shift register feature is not available for the 36×36 -bit multiplier. In 36×36 -bit mode, the device can use the register that is normally a multiplier-result-output register as a pipeline stage for the 36×36 -bit multiplier. Figure 4–35 shows the 36×36 -bit multiply mode. You can use the two-multipliers adder mode for complex multiplications, which are written as:

$$(a + jb) \times (c + jd) = [(a \times c) - (b \times d)] + j \times [(a \times d) + (b \times c)]$$

The two-multipliers adder mode allows a single DSP block to calculate the real part $[(a \times c) - (b \times d)]$ using one subtractor and the imaginary part $[(a \times d) + (b \times c)]$ using one adder, for data widths up to 18 bits. Two complex multiplications are possible for data widths up to 9 bits using four adder/subtractor/accumulator blocks. Figure 4–37 shows an 18-bit two-multipliers adder.



Figure 4–37. Two-Multipliers Adder Mode Implementing Complex Multiply

Four-Multipliers Adder Mode

In the four-multipliers adder mode, the DSP block adds the results of two first -stage adder/subtractor blocks. One sum of four 18×18 -bit multipliers or two different sums of two sets of four 9×9 -bit multipliers can be implemented in a single DSP block. The product width for each multiplier must be the same size. The four-multipliers adder mode is useful for FIR filter applications. Figure 4–38 shows the four multipliers adder mode.



Figure 4–49. Global & Regional Clock Connections From Side Pins & Fast PLL Outputs Note (1)

Note to Figure 4–49:

(1) PLLs 1,27, and 8 are fast PLLs. PLLs 7 and 8 do not drive global clocks.

Figure 4–50 shows the global and regional clocking from enhanced PLL outputs and top CLK pins.





Notes to Figure 4–57:

- In high-speed differential I/O support mode, this high-speed PLL clock feeds the SERDES. Stratix GX devices only support one rate of data transfer per fast PLL in high-speed differential I/O support mode.
- (2) This signal is a high-speed differential I/O support SERDES control signal.

Clock Multiplication & Division

The Stratix GX device's fast PLLs provide clock synthesis for PLL output ports using *m*/(post scaler) scaling factors. The input clock is multiplied by the *m* feedback factor. Each output port has a unique post scale counter to divide down the high-frequency VCO. There is one multiply divider, *m*, per fast PLL with a range of 1 to 32. There are two post scale L dividers for regional and/or LVDS interface clocks, and *g*0 counter for global clock output port; all range from 1 to 32.

In the case of a high-speed differential interface, you can set the output counter to 1 to allow the high-speed VCO frequency to drive the SERDES.

External Clock Outputs

Each fast PLL supports differential or single-ended outputs for source-synchronous transmitters or for general-purpose external clocks. There are no dedicated external clock output pins. Any I/O pin can be driven by the fast PLL global or regional outputs as an external output When using the IOE for DDR inputs, the two input registers clock double rate input data on alternating edges. An input latch is also used within the IOE for DDR input acquisition. The latch holds the data that is present during the clock high times. This allows both bits of data to be synchronous with the same clock edge (either rising or falling). Figure 4–64 shows an IOE configured for DDR input. Figure 4–65 shows the DDR input timing diagram.

Figure 4–64. Stratix GX IOE in DDR Input I/O Configuration Note (1)



Notes to Figure 4–64:

- (1) All input signals to the IOE can be inverted at the IOE.
- (2) This signal connection is only allowed on dedicated DQ function pins.
- (3) This signal is for dedicated DQS function pins only.

Table 6–9. LVCMOS Specifications								
Symbol	Parameter	Conditions	Minimum	Maximum	Units			
V _{OH}	High-level output voltage	$V_{CCIO} = 3.0,$ $I_{OH} = -0.1 \text{ mA}$	V _{CCIO} - 0.2		V			
V _{OL}	Low-level output voltage	$V_{CCIO} = 3.0,$ $I_{OL} = 0.1 \text{ mA}$		0.2	V			

Table 6–10. 2.5-V I/O Specifications Note (1)							
Symbol	Parameter	Conditions	Minimum	Maximum	Units		
V _{CCIO}	Output supply voltage		2.375	2.625	V		
V _{IH}	High-level input voltage		1.7	4.1	V		
V _{IL}	Low-level input voltage		-0.5	0.7	V		
V _{OH}	High-level output voltage	I _{OH} = -0.1 mA	2.1		V		
		I _{OH} = -1 mA	2.0		V		
		$I_{OH} = -2 \text{ to } -16 \text{ mA} (1)$	1.7		V		
V _{OL}	Low-level output voltage	I _{OL} = 0.1 mA		0.2	V		
		I _{OH} = 1 mA		0.4	V		
		I _{OH} = 2 to 16 mA <i>(1)</i>		0.7	V		

Table 6–11. 1.8-V I/O Specifications								
Symbol	Parameter	Conditions	Minimum	Maximum	Units			
V _{CCIO}	Output supply voltage		1.65	1.95	V			
V _{IH}	High-level input voltage		$0.65 \times V_{\text{CCIO}}$	2.25	V			
V _{IL}	Low-level input voltage		-0.3	$0.35 \times V_{\text{CCIO}}$	V			
V _{OH}	High-level output voltage	I _{OH} = -2 to -8 mA (1)	$V_{\text{CCIO}} - 0.45$		V			
V _{OL}	Low-level output voltage	I _{OL} = 2 to 8 mA (1)		0.45	V			

Table 6–12. 1.5-V I/O Specifications (Part 1 of 2)							
Symbol	Parameter	Conditions	Minimum	Maximum	Units		
V _{CCIO}	Output supply voltage		1.4	1.6	V		
V _{IH}	High-level input voltage		$0.65 \times V_{\text{CCIO}}$	$V_{CCIO} + 0.3$	V		
V _{IL}	Low-level input voltage		-0.3	$0.35 \times V_{\text{CCIO}}$	V		
V _{OH}	High-level output voltage	I _{OH} = -2 mA (1)	$0.75 \times V_{\text{CCIO}}$		V		

Table 6–47. M512 Block Internal Timing Microparameters							
	-5 Spee	ed Grade	-6 Speed Grade		-7 Speed Grade		
Symbol	Min	Max	Min	Max	Min	Max	Unit
t _{M512RC}		3,340		3,816		4,387	ps
t _{M512WC}		3,318		3,590		4,128	ps
t _{M512WERESU}	110		123		141		ps
t _{M512WERH}	34		38		43		ps
t _{M512DATASU}	110		123		141		ps
t _{M512DATAH}	34		38		43		ps
t _{M512WADDRASU}	110		123		141		ps
t _{M512WADDRH}	34		38		43		ps
t _{M512DATACO1}		424		472		541	ps
t _{M512DATACO2}		3,366		3,846		4,421	ps
t _{M512CLKHL}	150		167		192		ps
t _{M512CLR}	170		189		217		ps

Table 6–48. M4K Block Internal Timing Microparameters (Part 1 of 2)								
Symbol	-5 Spee	d Grade	-6 Spee	-6 Speed Grade		-7 Speed Grade		
Symbol	Min	Max	Min	Max	Min	Max	Unit	
t _{M4KRC}		3,807		4,320		4,967	ps	
t _{M4KWC}		2,556		2,840		3,265	ps	
t _{M4KWERESU}	131		149		171		ps	
t _{M4KWERH}	34		38		43		ps	
t _{M4KDATASU}	131		149		171		ps	
t _{M4KDATAH}	34		38		43		ps	
t _{M4KWADDRASU}	131		149		171		ps	
t _{M4KWADDRH}	34		38		43		ps	
t _{M4KRADDRASU}	131		149		171		ps	
t _{M4KRADDRH}	34		38		43		ps	
t _{M4KDATABSU}	131		149		171		ps	
t _{M4KDATABH}	34		38		43		ps	
t _{M4KADDRBSU}	131		149		171		ps	
t _{M4KADDRBH}	34		38		43		ps	

Table 6–72. Stratix GX I/O Standard Column Pin Input Delay Adders (Part 2 of 2)							
I/O Standard	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	UIII
SSTL-2 class I		-70		-74		-86	ps
SSTL-2 class II		-70		-74		-86	ps
SSTL-18 class I		180		189		217	ps
SSTL-18 class II		180		189		217	ps
1.5-V HSTL class I		120		126		144	ps
1.5-V HSTL class II		120		126		144	ps
1.8-V HSTL class I		70		73		83	ps
1.8-V HSTL class II		70		73		83	ps

Table 6–73. Stratix GX I/O Standard Row Pin Input Delay Adders (Part 1 of 2)							
I/O Standard	-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		1114
	Min	Max	Min	Max	Min	Max	
LVCMOS		0		0		0	ps
3.3-V LVTTL		0		0		0	ps
2.5-V LVTTL		30		31		35	ps
1.8-V LVTTL		150		157		180	ps
1.5-V LVTTL		210		220		252	ps
GTL		0		0		0	ps
GTL+		220		231		265	ps
3.3-V PCI		0		0		0	ps
3.3-V PCI-X 1.0		0		0		0	ps
Compact PCI		0		0		0	ps
AGP 1×		0		0		0	ps
AGP 2×		0		0		0	ps
СТТ		80		84		96	ps
SSTL-3 class I		-30		-32		-37	ps
SSTL-3 class II		-30		-32		-37	ps
SSTL-2 class I		-70		-74		-86	ps
SSTL-2 class II		-70		-74		-86	ps
SSTL-18 class I		180		189		217	ps
SSTL-18 class II		0		0		0	ps
1.5-V HSTL class I		130		136		156	ps

Table 6–92. Fast PLL Specifications for -7 & -8 Speed Grades (Part 2 of 2)						
Symbol	Parameter	Min	Max	Unit		
t _{DUTY}	Duty cycle for DFFIO $1 \times CLKOUT$ pin (3)	45	55	%		
t _{JITTER}	Period jitter for DIFFIO clock out (3)		±80	ps		
	Period jitter for internal global or regional clock		±100 ps for >200 MHz outclk ±20 mUI for <200 MHz outclk	ps or mUI		
t _{LOCK}	Time required for PLL to acquire lock	10	100	μs		
m	Multiplication factors for <i>m</i> counter (4)	1	32	Integer		
<i>l</i> 0, <i>l</i> 1, <i>g</i> 0	Multiplication factors for IO, I1, and gO counter (4), (5)	1	32	Integer		
t _{ARESET}	Minimum pulse width on areset signal	10		ns		

Notes to Tables 6–91 & 6–92:

(1) See "Maximum Input & Output Clock Rates" on page 6–54.

- (2) When using the SERDES, high-speed differential I/O mode supports a maximum output frequency of 210 MHz to the global or regional clocks (that is, the maximum data rate 840 Mbps divided by the smallest SERDES J factor of 4).
- (3) This parameter is for high-speed differential I/O mode only.
- (4) These counters have a maximum of 32 if programmed for 50/50 duty cycle. Otherwise, they have a maximum of 16.
- (5) High-speed differential I/O mode supports W = 1 to 16 and J = 4, 7, 8, or 10.

DLL Jitter

Table 6–93 reports the jitter for the DLL in the DQS phase-shift reference circuit.

Table 6–93. DLL Jitter for DQS Phase Shift Reference Circuit					
Frequency (MHz)	DLL Jitter (ps)				
197 to 200	± 100				
160 to 196	± 300				
100 to 159	± 500				