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Details

Product Status	Active
Core Processor	8051
Core Size	8-Bit
Speed	25MHz
Connectivity	I ² C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	13
Program Memory Size	4KB (4K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	2.2V ~ 3.6V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	16-SOIC (0.154", 3.90mm Width)
Supplier Device Package	-
Purchase URL	https://www.e-xfl.com/product-detail/silicon-labs/c8051f864-c-isr

Table 1.10. Comparators

Parameter	Symbol	Test Condition	Min	Typ	Max	Unit
Positive Hysteresis Mode 3 (CPnMD = 11)	HYS _{CP+}	CPnHYP = 00	—	1.5	—	mV
		CPnHYP = 01	—	4	—	mV
		CPnHYP = 10	—	8	—	mV
		CPnHYP = 11	—	16	—	mV
Negative Hysteresis Mode 3 (CPnMD = 11)	HYS _{CP-}	CPnHYN = 00	—	-1.5	—	mV
		CPnHYN = 01	—	-4	—	mV
		CPnHYN = 10	—	-8	—	mV
		CPnHYN = 11	—	-16	—	mV
Input Range (CP+ or CP-)	V _{IN}		-0.25	—	V _{DD} +0.25	V
Input Pin Capacitance	C _{CP}		—	7.5	—	pF
Common-Mode Rejection Ratio	CMRR _{CP}		—	70	—	dB
Power Supply Rejection Ratio	PSRR _{CP}		—	72	—	dB
Input Offset Voltage	V _{OFF}	T _A = 25 °C	-10	0	10	mV
Input Offset Tempco	TC _{OFF}		—	3.5	—	μV/°C

1.2.2. ADC Supply Current

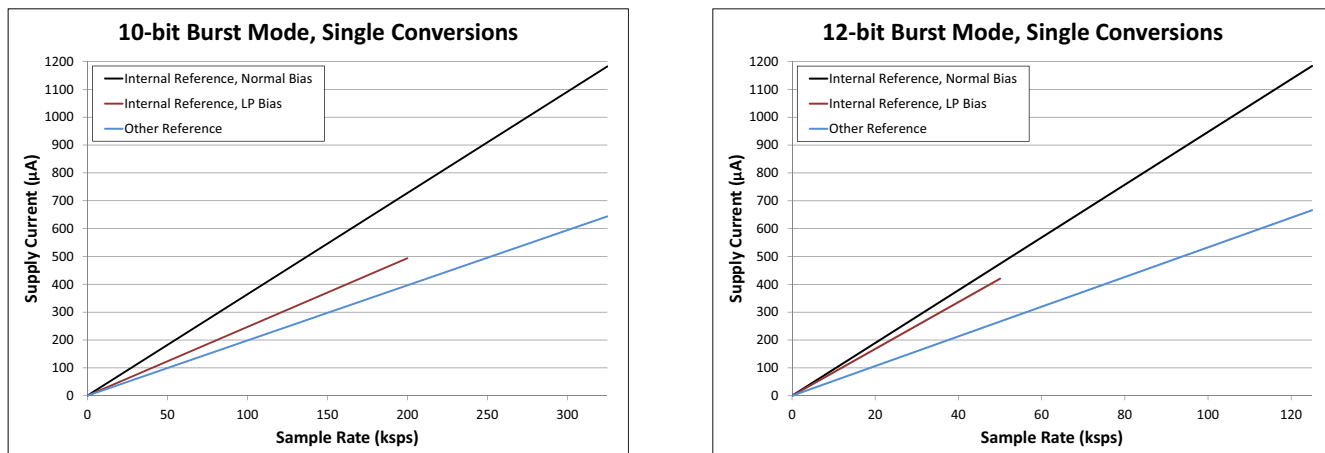


Figure 1.3. Typical ADC and Internal Reference Power Consumption in Burst Mode

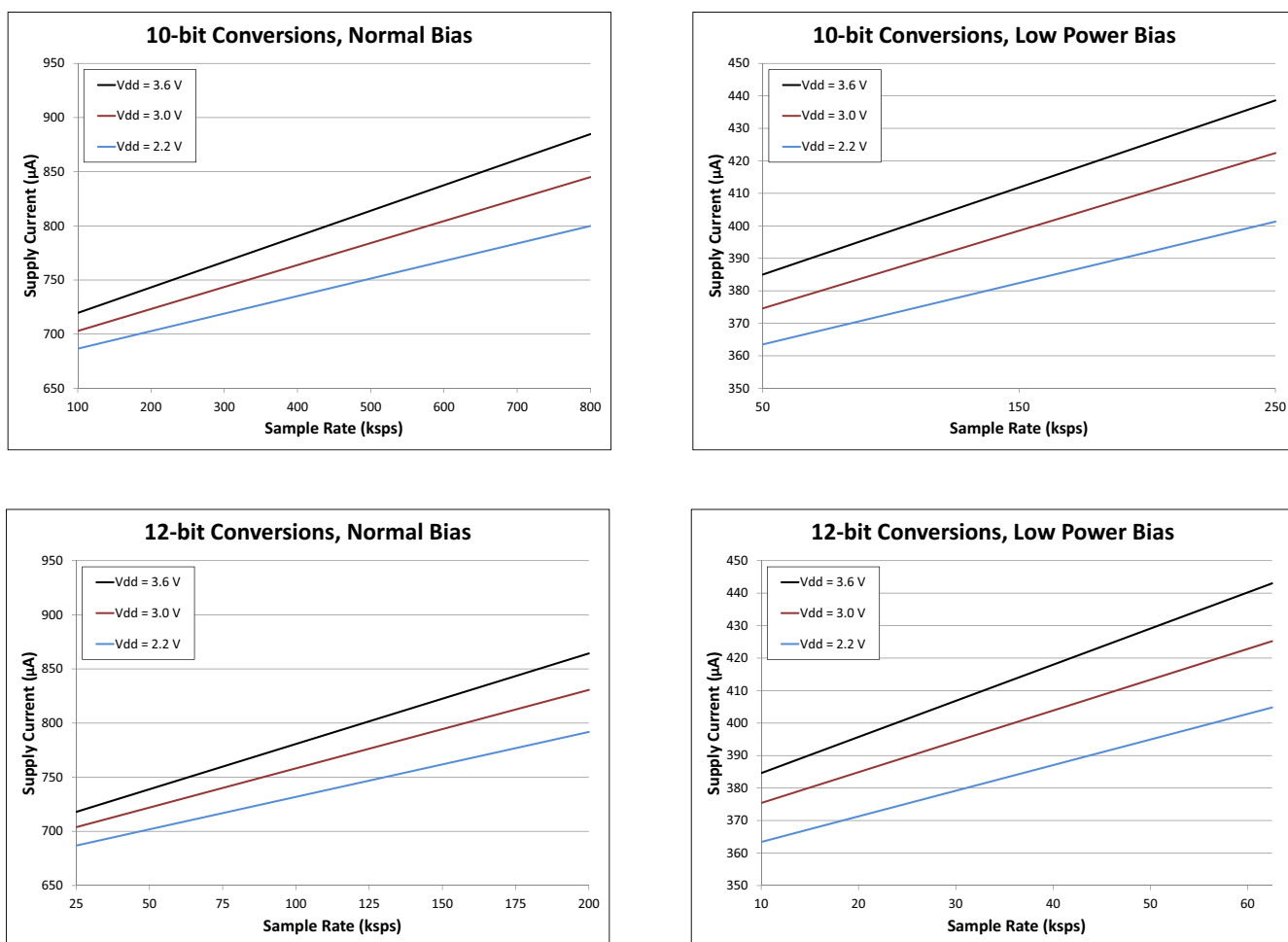


Figure 1.4. Typical ADC Power Consumption in Normal (Always-On) Mode

6. QFN-20 Package Specifications

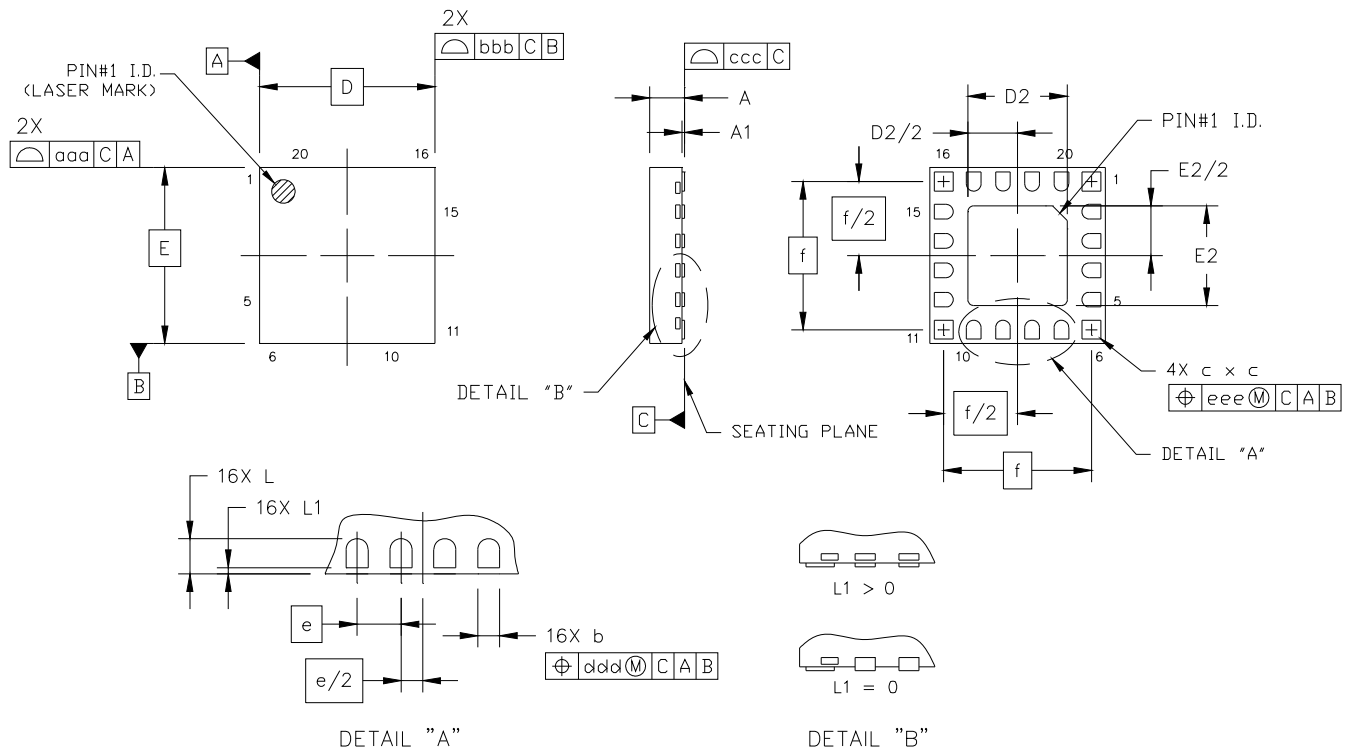


Figure 6.1. QFN-20 Package Drawing

Table 6.1. QFN-20 Package Dimensions

Symbol	Millimeters		
	Min	Nom	Max
A	0.70	0.75	0.80
A1	0.00	0.02	0.05
b	0.20	0.25	0.30
c	0.25	0.30	0.35
D	3.00 BSC		
D2	1.6	1.70	1.8
e	0.50 BSC		
E	3.00 BSC		
E2	1.6	1.70	1.8
f	2.53 BSC		
L	0.3	0.40	0.5
L1	0.00	—	0.10
aaa	—	—	0.05
bbb	—	—	0.05
ccc	—	—	0.08
ddd	—	—	0.10
eee	—	—	0.10

Notes:

1. All dimensions are shown in millimeters unless otherwise noted.
2. Dimensioning and tolerancing per ANSI Y14.5M-1994.

needed, it is recommended that AD12SM be set to 1 and ADTK to 0x3F, and that the ADC be placed in always-on mode (ADEN = 1). For sample rates under 180 ksps, or when accumulating multiple samples, AD12SM should normally be cleared to 0, and ADTK should be configured to provide the appropriate settling time for the subsequent conversions.

14.5. Power Considerations

The ADC has several power-saving features which can help the user optimize power consumption according to the needs of the application. The most efficient way to use the ADC for slower sample rates is by using burst mode. Burst mode dynamically controls power to the ADC and (if used) the internal voltage reference. By completely powering off these circuits when the ADC is not tracking or converting, the average supply current required for lower sampling rates is reduced significantly.

The ADC also provides low power options that allow reduction in operating current when operating at low SAR clock frequencies or with longer tracking times. The internal common-mode buffer can be configured for low power mode by setting the ADLPM bit in ADC0PWR to 1. Two other fields in the ADC0PWR register (ADBIAS and ADMXLP) may be used together to adjust the power consumed by the ADC and its multiplexer and reference buffers, respectively. In general, these options are used together, when operating with a SAR conversion clock frequency of 4 MHz.

Table 14.2. ADC0 Optimal Power Configuration (8- and 10-bit Mode)

Required Throughput	Reference Source	Mode Configuration	SAR Clock Speed	Other Register Field Settings
325-800 ksps	Any	Always-On (ADEN = 1 ADBMEN = 0)	12.25 MHz (ADSC = 1)	ADC0PWR = 0x40 ADC0TK = N/A ADRPT = 0
0-325 ksps	External	Burst Mode (ADEN = 0 ADBMEN = 1)	12.25 MHz (ADSC = 1)	ADC0PWR = 0x44 ADC0TK = 0x3A ADRPT = 0
250-325 ksps	Internal	Burst Mode (ADEN = 0 ADBMEN = 1)	12.25 MHz (ADSC = 1)	ADC0PWR = 0x44 ADC0TK = 0x3A ADRPT = 0
200-250 ksps	Internal	Always-On (ADEN = 1 ADBMEN = 0)	4.08 MHz (ADSC = 5)	ADC0PWR = 0xF0 ADC0TK = N/A ADRPT = 0
0-200 ksps	Internal	Burst Mode (ADEN = 0 ADBMEN = 1)	4.08 MHz (ADSC = 5)	ADC0PWR = 0xF4 ADC0TK = 0x34 ADRPT = 0

Notes:

1. For always-on configuration, ADSC settings assume SYSCLK is the internal 24.5 MHz high-frequency oscillator. Adjust ADSC as needed if using a different source for SYSCLK.
2. ADRPT reflects the minimum setting for this bit field. When using the ADC in Burst Mode, up to 64 samples may be auto-accumulated per conversion start by adjusting ADRPT.

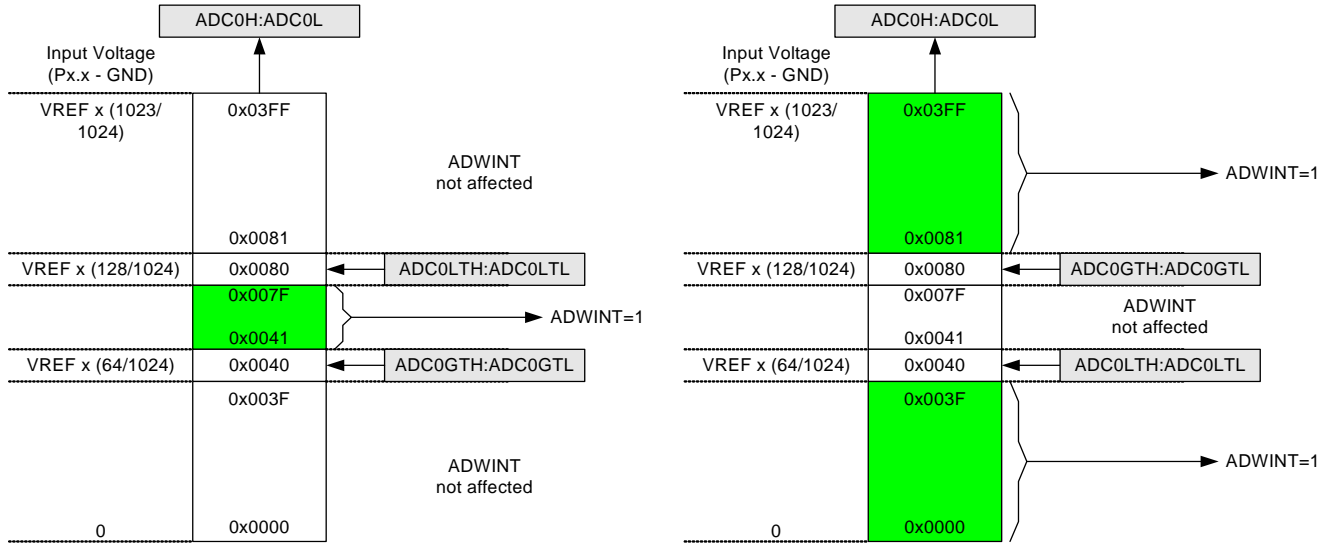


Figure 14.6. ADC Window Compare Example: Right-Justified Single-Ended Data

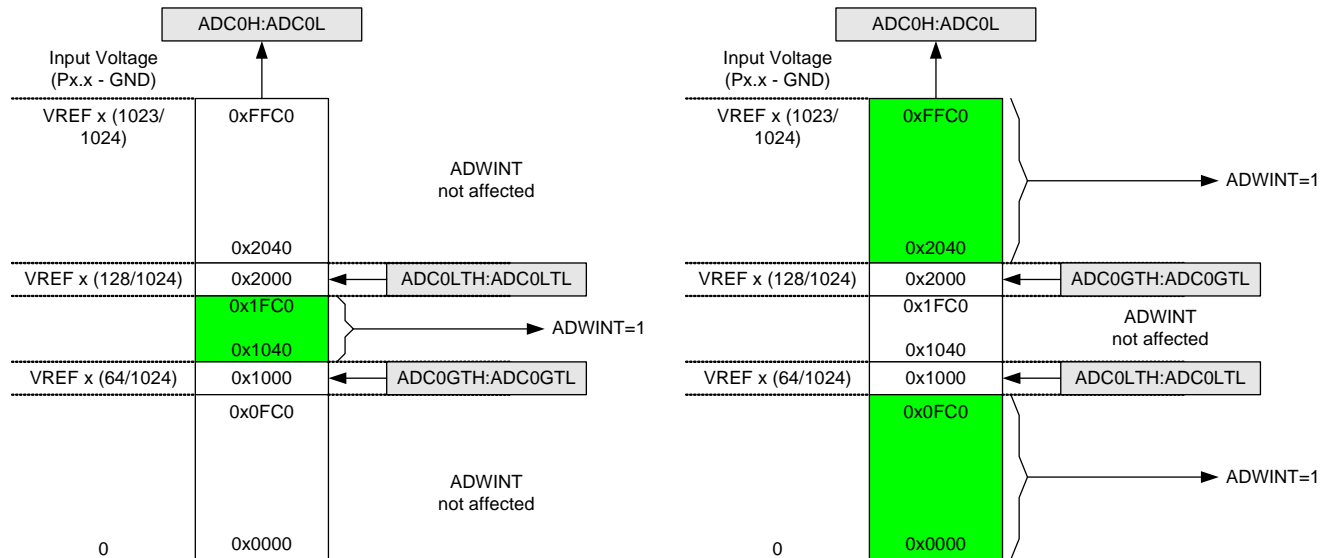


Figure 14.7. ADC Window Compare Example: Left-Justified Single-Ended Data

17.2. Functional Description

The comparator offers programmable response time and hysteresis, an analog input multiplexer, and two outputs that are optionally available at the port pins: a synchronous “latched” output (CPn), or an asynchronous “raw” output (CPnA). The asynchronous CPnA signal is available even when the system clock is not active. This allows the comparator to operate and generate an output with the device in STOP mode.

When disabled, the comparator output (if assigned to a port I/O pin via the crossbar) defaults to the logic low state, and the power supply to the comparator is turned off.

The comparator response time may be configured in software via the CPTnMD register. Selecting a longer response time reduces the comparator supply current.

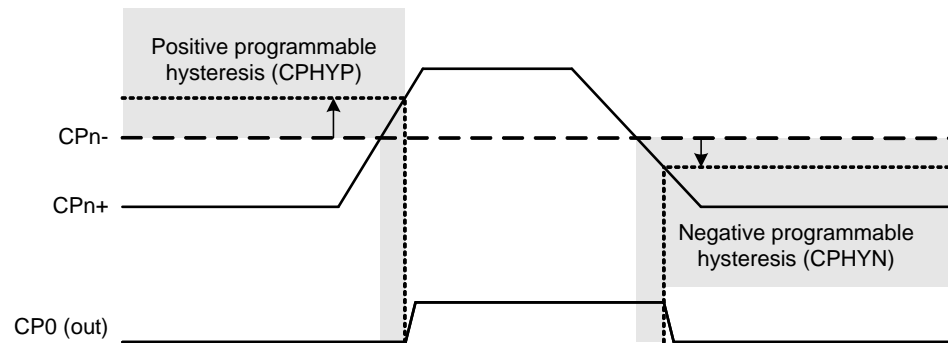


Figure 17.2. Comparator Hysteresis Plot

The comparator hysteresis is software-programmable via its Comparator Control register CPTnCN. The user can program both the amount of hysteresis voltage (referred to the input voltage) and the positive and negative-going symmetry of this hysteresis around the threshold voltage.

The comparator hysteresis is programmable using the CPHYN and CPHYP fields in the Comparator Control Register CPTnCN. The amount of negative hysteresis voltage is determined by the settings of the CPHYN bits. As shown in Figure 17.2, settings of 20, 10, or 5 mV (nominal) of negative hysteresis can be programmed, or negative hysteresis can be disabled. In a similar way, the amount of positive hysteresis is determined by the setting the CPHYP bits.

Comparator interrupts can be generated on both rising-edge and falling-edge output transitions. The CPFIF flag is set to logic 1 upon a comparator falling-edge occurrence, and the CPRIF flag is set to logic 1 upon the comparator rising-edge occurrence. Once set, these bits remain set until cleared by software. The comparator rising-edge interrupt mask is enabled by setting CPRIE to a logic 1. The comparator falling-edge interrupt mask is enabled by setting CPFIE to a logic 1.

The output state of the comparator can be obtained at any time by reading the CPOUT bit. The comparator is enabled by setting the CPEN bit to logic 1, and is disabled by clearing this bit to logic 0.

Note that false rising edges and falling edges can be detected when the comparator is first powered on or if changes are made to the hysteresis or response time control bits. Therefore, it is recommended that the rising-edge and falling-edge flags be explicitly cleared to logic 0 a short time after the comparator is enabled or its mode bits have been changed, before enabling comparator interrupts.

Register 18.2. CRC0IN: CRC0 Data Input

Bit	7	6	5	4	3	2	1	0
Name	CRC0IN							
Type	RW							
Reset	0	0	0	0	0	0	0	0
SFR Address: 0xDD								

Table 18.3. CRC0IN Register Bit Descriptions

Bit	Name	Function
7:0	CRC0IN	CRC Data Input. Each write to CRCIN results in the written data being computed into the existing CRC result according to the CRC algorithm.

Register 18.3. CRC0DAT: CRC0 Data Output

Bit	7	6	5	4	3	2	1	0
Name	CRC0DAT							
Type	RW							
Reset	0	0	0	0	0	0	0	0
SFR Address: 0xDE								

Table 18.4. CRC0DAT Register Bit Descriptions

Bit	Name	Function
7:0	CRC0DAT	CRC Data Output. Each read or write performed on CRC0DAT targets the CRC result bits pointed to by the CRC0 Result Pointer (CRC0PNT bits in CRC0CN).
Note: CRC0DAT may not be valid for one cycle after setting the CRC0INIT bit in the CRC0CN register to 1. Any time CRC0INIT is written to 1 by firmware, at least one instruction should be performed before reading CRC0DAT.		

20.3.5. Frequency Output Mode

Frequency Output Mode produces a programmable-frequency square wave on the module's associated CEXn pin. The capture/compare module high byte holds the number of PCA clocks to count before the output is toggled. The frequency of the square wave is then defined by Equation 20.1.

$$F_{CEXn} = \frac{F_{PCA}}{2 \times PCA0CPHn}$$

Note: A value of 0x00 in the PCA0CPHn register is equal to 256 for this equation.

Equation 20.1. Square Wave Frequency Output

Where F_{PCA} is the frequency of the clock selected by the CPS2–0 bits in the PCA mode register, PCA0MD. The lower byte of the capture/compare module is compared to the PCA counter low byte; on a match, n is toggled and the offset held in the high byte is added to the matched value in PCA0CPLn. Frequency Output Mode is enabled by setting the ECOMn, TOGn, and PWMn bits in the PCA0CPMn register. Note that the MATn bit should normally be set to 0 in this mode. If the MATn bit is set to 1, the CCFn flag for the channel will be set when the 16-bit PCA0 counter and the 16-bit capture/compare register for the channel are equal.

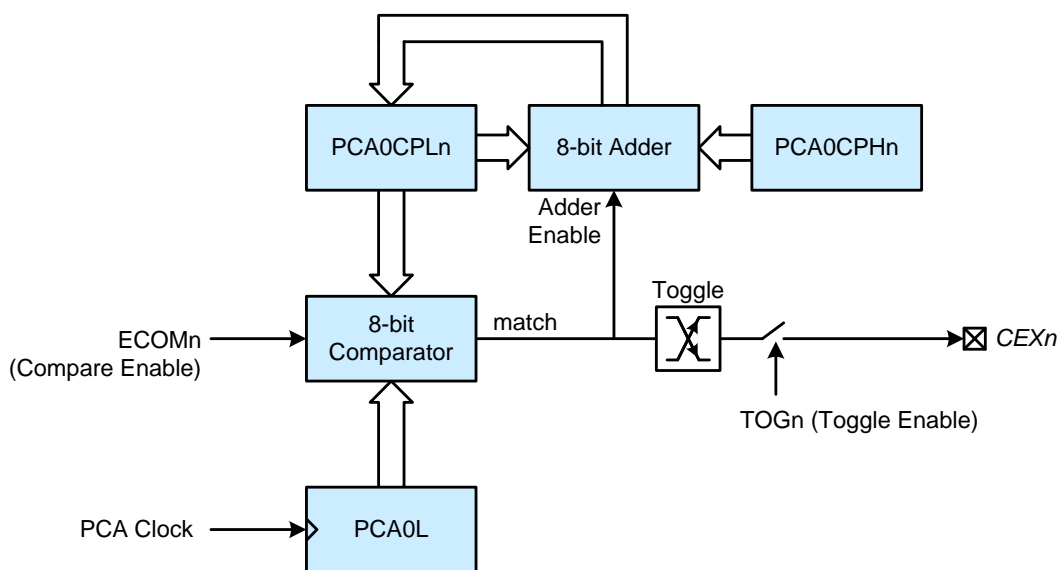


Figure 20.5. PCA Frequency Output Mode

Register 20.10. PCA0POL: PCA Output Polarity

Bit	7	6	5	4	3	2	1	0
Name	Reserved					CEX2POL	CEX1POL	CEX0POL
Type	R					RW	RW	RW
Reset	0	0	0	0	0	0	0	0

SFR Address: 0x96**Table 20.12. PCA0POL Register Bit Descriptions**

Bit	Name	Function
7:3	Reserved	Must write reset value.
2	CEX2POL	CEX2 Output Polarity. Selects the polarity of the CEX2 output channel. When this bit is modified, the change takes effect at the pin immediately. 0: Use default polarity. 1: Invert polarity.
1	CEX1POL	CEX1 Output Polarity. Selects the polarity of the CEX1 output channel. When this bit is modified, the change takes effect at the pin immediately. 0: Use default polarity. 1: Invert polarity.
0	CEX0POL	CEX0 Output Polarity. Selects the polarity of the CEX0 output channel. When this bit is modified, the change takes effect at the pin immediately. 0: Use default polarity. 1: Invert polarity.

Registers XBR0, XBR1 and XBR2 are used to assign the digital I/O resources to the physical I/O port pins. Note that when the SMBus is selected, the crossbar assigns both pins associated with the SMBus (SDA and SCL); when UART0 is selected, the crossbar assigns both pins associated with UART0 (TX and RX). UART0 pin assignments are fixed for bootloading purposes: UART0 TX is always assigned to P0.4; UART0 RX is always assigned to P0.5. Standard port I/Os appear contiguously after the prioritized functions have been assigned.

Figure 21.3 shows an example of the resulting pin assignments of the device with UART0 and SPI0 enabled and the EXTCLK (P0.3) pin skipped (P0SKIP = 0x08). UART0 is the highest priority and it will be assigned first. The UART0 pins can only appear on P0.4 and P0.5, so that is where it is assigned. The next-highest enabled peripheral is SPI0. P0.0, P0.1 and P0.2 are free, so SPI0 takes these three pins. The fourth pin, NSS, is routed to P0.6 because P0.3 is skipped and P0.4 and P0.5 are already occupied by the UART. The other pins on the device are available for use as general-purpose digital I/O or analog functions.

Port	P0								P1								P2	
Pin Number	0	1	2	3	4	5	6	7	0	1	2	3	4	5	6	7	0	1
SOIC-16 Package													N/A	N/A	N/A			
QFN-20 Package	VREF			EXTCLK			CNVSTR										C2D	N/A
QSOP-24 Package																		
UART0-TX																		
UART0-RX																		
SPI0-SCK																		
SPI0-MISO																		
SPI0-MOSI																		
SPI0-NSS*																		
SMB0-SDA																		
SMB0-SCL																		
CMP0-CP0																		
CMP0-CP0A																		
CMP1-CP1																		
CMP1-CP1A																		
SYSCLK																		
PCA0-CEX0																		
PCA0-CEX1																		
PCA0-CEX2																		
PCA0-ECI																		
Timer0-T0																		
Timer1-T1																		
Timer2-T2																		
Pin Skip Settings	0	0	0	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0
	P0SKIP								P1SKIP									

The crossbar peripherals are assigned in priority order from top to bottom.

■ These boxes represent Port pins which can potentially be assigned to a peripheral.

□ Special Function Signals are not assigned by the crossbar. When these signals are enabled, the Crossbar should be manually configured to skip the corresponding port pins.

□ Pins can be "skipped" by setting the corresponding bit in PnSKIP to 1.

* NSS is only pinned out when the SPI is in 4-wire mode.

Figure 21.3. Crossbar Priority Decoder Example

Register 21.4. PRTDRV: Port Drive Strength

Bit	7	6	5	4	3	2	1	0
Name	Reserved					P2DRV	P1DRV	P0DRV
Type	R					RW	RW	RW
Reset	0	0	0	0	0	1	1	1

SFR Address: 0xF6

Table 21.7. PRTDRV Register Bit Descriptions

Bit	Name	Function
7:3	Reserved	Must write reset value.
2	P2DRV	Port 2 Drive Strength. 0: All pins on P2 use low drive strength. 1: All pins on P2 use high drive strength.
1	P1DRV	Port 1 Drive Strength. 0: All pins on P1 use low drive strength. 1: All pins on P1 use high drive strength.
0	P0DRV	Port 0 Drive Strength. 0: All pins on P0 use low drive strength. 1: All pins on P0 use high drive strength.

22. Reset Sources and Supply Monitor

Reset circuitry allows the controller to be easily placed in a predefined default condition. Upon entering this reset state, the following events occur:

- CIP-51 halts program execution
- Special Function Registers (SFRs) are initialized to their defined reset values
- External port pins are placed in a known state
- Interrupts and timers are disabled.

All SFRs are reset to the predefined values noted in the SFR detailed descriptions. The contents of internal data memory are unaffected during a reset; any previously stored data is preserved. However, since the stack pointer SFR is reset, the stack is effectively lost, even though the data on the stack is not altered.

The Port I/O latches are reset to 0xFF (all logic ones) in open-drain, low-drive mode. Weak pullups are enabled during and after the reset. For V_{DD} Monitor and power-on resets, the \overline{RST} pin is driven low until the device exits the reset state. Note that during a power-on event, there may be a short delay before the POR circuitry fires and the \overline{RST} pin is driven low. During that time, the \overline{RST} pin will be weakly pulled to the V_{DD} supply pin.

On exit from the reset state, the program counter (PC) is reset, the Watchdog Timer is enabled and the system clock defaults to the internal oscillator. Program execution begins at location 0x0000.

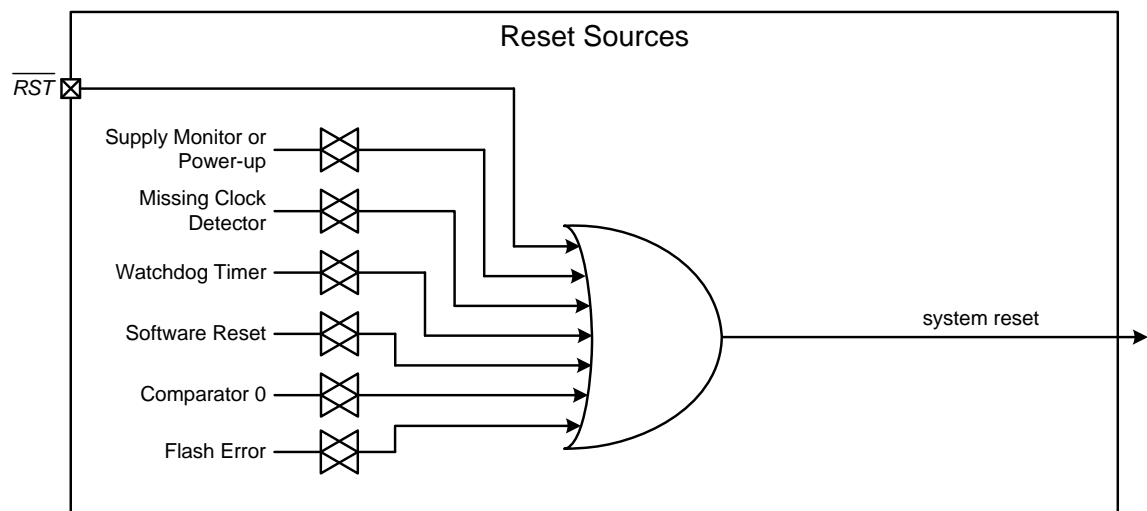


Figure 22.1. Reset Sources

Table 24.4. Hardware Address Recognition Examples (EHACK = 1)

Hardware Slave Address SLV	Slave Address Mask SLVM	GC bit	Slave Addresses Recognized by Hardware
0x34	0x7E	1	0x34, 0x35, 0x00 (General Call)
0x70	0x73	0	0x70, 0x74, 0x78, 0x7C

24.4.6. Data Register

The SMBus Data register SMB0DAT holds a byte of serial data to be transmitted or one that has just been received. Software may safely read or write to the data register when the SI flag is set. Software should not attempt to access the SMB0DAT register when the SMBus is enabled and the SI flag is cleared to logic 0, as the interface may be in the process of shifting a byte of data into or out of the register.

Data in SMB0DAT is always shifted out MSB first. After a byte has been received, the first bit of received data is located at the MSB of SMB0DAT. While data is being shifted out, data on the bus is simultaneously being shifted in. SMB0DAT always contains the last data byte present on the bus. In the event of lost arbitration, the transition from master transmitter to slave receiver is made with the correct data or address in SMB0DAT.

24.5.2. Read Sequence (Master)

During a read sequence, an SMBus master reads data from a slave device. The master in this transfer will be a transmitter during the address byte, and a receiver during all data bytes. The SMBus interface generates the START condition and transmits the first byte containing the address of the target slave and the data direction bit. In this case the data direction bit (R/W) will be logic 1 (READ). Serial data is then received from the slave on SDA while the SMBus outputs the serial clock. The slave transmits one or more bytes of serial data.

If hardware ACK generation is disabled, the ACKRQ is set to 1 and an interrupt is generated after each received byte. Software must write the ACK bit at that time to ACK or NACK the received byte.

With hardware ACK generation enabled, the SMBus hardware will automatically generate the ACK/NACK, and then post the interrupt. **It is important to note that the appropriate ACK or NACK value should be set up by the software prior to receiving the byte when hardware ACK generation is enabled.**

Writing a 1 to the ACK bit generates an ACK; writing a 0 generates a NACK. Software should write a 0 to the ACK bit for the last data transfer, to transmit a NACK. The interface exits Master Receiver Mode after the STO bit is set and a STOP is generated. The interface will switch to Master Transmitter Mode if SMB0DAT is written while an active Master Receiver. Figure 24.6 shows a typical master read sequence. Two received data bytes are shown, though any number of bytes may be received. Notice that the 'data byte transferred' interrupts occur at different places in the sequence, depending on whether hardware ACK generation is enabled. The interrupt occurs **before** the ACK with hardware ACK generation disabled, and **after** the ACK when hardware ACK generation is enabled.

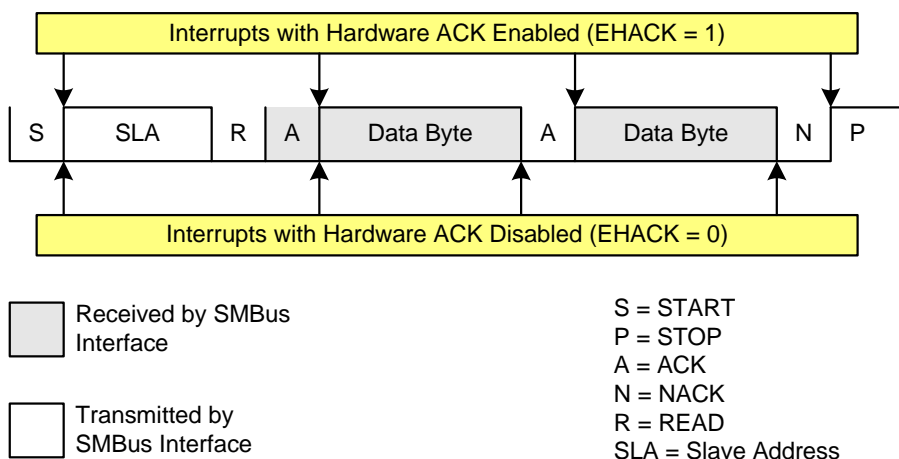


Figure 24.6. Typical Master Read Sequence

Register 24.5. SMB0ADM: SMBus0 Slave Address Mask

Bit	7	6	5	4	3	2	1	0
Name	SLVM							EHACK
Type	RW							RW
Reset	1	1	1	1	1	1	1	0

SFR Address: 0xD6**Table 24.11. SMB0ADM Register Bit Descriptions**

Bit	Name	Function
7:1	SLVM	SMBus0 Slave Address Mask. Defines which bits of register SMB0ADR are compared with an incoming address byte, and which bits are ignored. Any bit set to 1 in SLVM enables comparisons with the corresponding bit in SLV. Bits set to 0 are ignored (can be either 0 or 1 in the incoming address).
0	EHACK	Hardware Acknowledge Enable. Enables hardware acknowledgement of slave address and received data bytes. 0: Firmware must manually acknowledge all incoming address and data bytes. 1: Automatic slave address recognition and hardware acknowledge is enabled.

Table 25.10. TMR2CN Register Bit Descriptions

Bit	Name	Function
0	T2XCLK	Timer 2 External Clock Select. This bit selects the external clock source for Timer 2. If Timer 2 is in 8-bit mode, this bit selects the external oscillator clock source for both timer bytes. However, the Timer 2 Clock Select bits (T2MH and T2ML in register CKCON) may still be used to select between the external clock and the system clock for either timer. 0: Timer 2 clock is the system clock divided by 12. 1: Timer 2 clock is the external clock divided by 8 (synchronized with SYSCCLK).

26. Universal Asynchronous Receiver/Transmitter (UART0)

UART0 is an asynchronous, full duplex serial port offering modes 1 and 3 of the standard 8051 UART. Enhanced baud rate support allows a wide range of clock sources to generate standard baud rates (details in Section “26.1. Enhanced Baud Rate Generation” on page 289). Received data buffering allows UART0 to start reception of a second incoming data byte before software has finished reading the previous data byte.

UART0 has two associated SFRs: Serial Control Register 0 (SCON0) and Serial Data Buffer 0 (SBUF0). The single SBUF0 location provides access to both transmit and receive registers. **Writes to SBUF0 always access the transmit register. Reads of SBUF0 always access the buffered receive register; it is not possible to read data from the transmit register.**

With UART0 interrupts enabled, an interrupt is generated each time a transmit is completed (TI is set in SCON0), or a data byte has been received (RI is set in SCON0). The UART0 interrupt flags are not cleared by hardware when the CPU vectors to the interrupt service routine. They must be cleared manually by software, allowing software to determine the cause of the UART0 interrupt (transmit complete or receive complete).

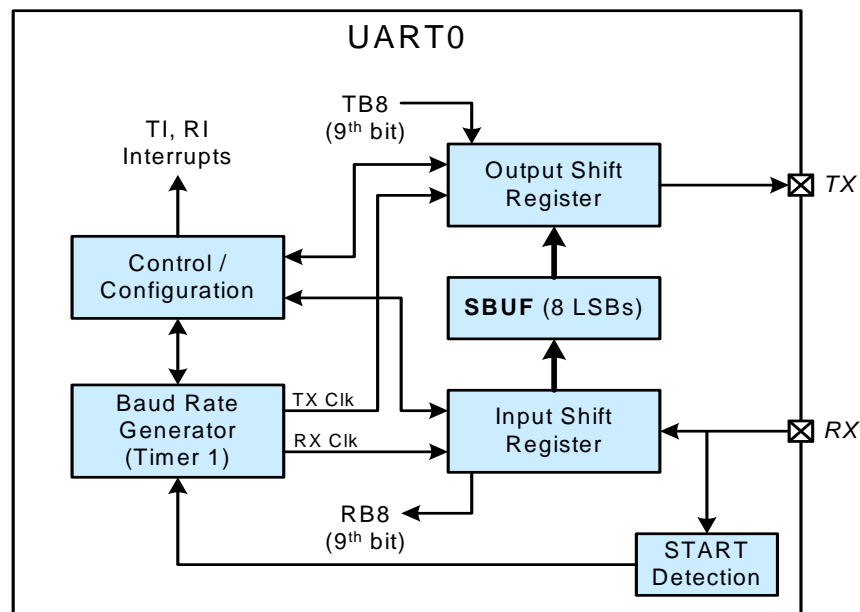


Figure 26.1. UART0 Block Diagram

26.1. Enhanced Baud Rate Generation

The UART0 baud rate is generated by Timer 1 in 8-bit auto-reload mode. The TX clock is generated by TL1; the RX clock is generated by a copy of TL1 (shown as RX Timer in Figure 26.2), which is not user-accessible. Both TX and RX Timer overflows are divided by two to generate the TX and RX baud rates. The RX Timer runs when Timer 1 is enabled, and uses the same reload value (TH1). However, an RX Timer reload is forced when a START condition is detected on the RX pin. This allows a receive to begin any time a START is detected, independent of the TX Timer state.

29.2. C2 Interface Registers

The following describes the C2 registers necessary to perform flash programming through the C2 interface. All C2 registers are accessed through the C2 interface, and are not available in the SFR map for firmware access.

Register 29.1. C2ADD: C2 Address

Bit	7	6	5	4	3	2	1	0
Name	C2ADD							
Type	RW							
Reset	0	0	0	0	0	0	0	0

This register is part of the C2 protocol.

Table 29.1. C2ADD Register Bit Descriptions

Bit	Name	Function
7:0	C2ADD	C2 Address. The C2ADD register is accessed via the C2 interface. The value written to C2ADD selects the target data register for C2 Data Read and Data Write commands. 0x00: C2DEVID 0x01: C2REVID 0x02: C2FPCTL 0xB4: C2FPDAT



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