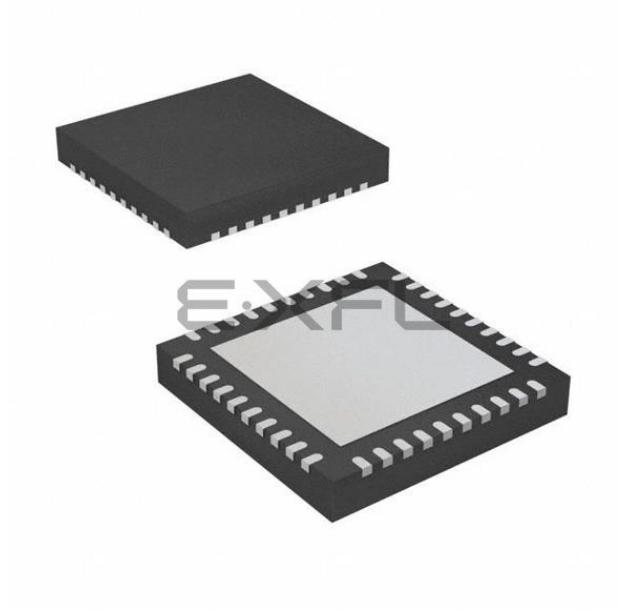
Infineon Technologies - XMC1302Q040X0128ABXUMA1 Datasheet



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"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	ARM® Cortex®-M0
Core Size	32-Bit Single-Core
Speed	32MHz
Connectivity	I²C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, I²S, POR, PWM, WDT
Number of I/O	27
Program Memory Size	128KB (128K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	16K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 5.5V
Data Converters	A/D 16x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	40-VFQFN Exposed Pad
Supplier Device Package	PG-VQFN-40-13
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/xmc1302q040x0128abxuma1

XMC1300 AB-Step

Microcontroller Series
for Industrial Applications

XMC1000 Family

ARM® Cortex®-M0
32-bit processor core

Data Sheet

V1.9 2017-03

Microcontrollers

- Configurable pad hysteresis

On-Chip Debug Support

- Support for debug features: 4 breakpoints, 2 watchpoints
- Various interfaces: ARM serial wire debug (SWD), single pin debug (SPD)

1.1 Ordering Information

The ordering code for an Infineon microcontroller provides an exact reference to a specific product. The code “XMC1<DDD>-<Z><PPP><T><FFFF>” identifies:

- <DDD> the derivatives function set
- <Z> the package variant
  - T: TSSOP
  - Q: VQFN
- <PPP> package pin count
- <T> the temperature range:
  - F: -40°C to 85°C
  - X: -40°C to 105°C
- <FFFF> the Flash memory size.

For ordering codes for the XMC1300 please contact your sales representative or local distributor.

This document describes several derivatives of the XMC1300 series, some descriptions may not apply to a specific product. Please see **Table 1**.

For simplicity the term **XMC1300** is used for all derivatives throughout this document.

1.2 Device Types

These device types are available and can be ordered through Infineon’s direct and/or distribution channels.

Table 1 Synopsis of XMC1300 Device Types

Derivative	Package	Flash Kbytes	SRAM Kbytes
XMC1301-T016F0008	PG-TSSOP-16-8	8	16
XMC1301-T016F0016	PG-TSSOP-16-8	16	16
XMC1301-T016F0032	PG-TSSOP-16-8	32	16
XMC1301-T016X0008	PG-TSSOP-16-8	8	16
XMC1301-T016X0016	PG-TSSOP-16-8	16	16
XMC1302-T016X0008	PG-TSSOP-16-8	8	16

Table 4 XMC1300 Chip Identification Number (cont'd)

Derivative	Value	Marking
XMC1302-Q040X0128	00013043 01FF00FF 00001FF7 0000900F 00000C00 00001000 00021000 $201ED083_H$	AB
XMC1302-Q040X0200	00013043 01FF00FF 00001FF7 0000900F 00000C00 00001000 00033000 $201ED083_H$	AB

2.2.1 Package Pin Summary

The following general building block is used to describe each pin:

Table 5 Package Pin Mapping Description

Function	Package A	Package B	...	Pad Type
Px.y	N	N		Pad Class

The table is sorted by the “Function” column, starting with the regular Port pins (Px.y), followed by the supply pins.

The following columns, titled with the supported package variants, lists the package pin number to which the respective function is mapped in that package.

The “Pad Type” indicates the employed pad type:

- STD_INOUT(standard bi-directional pads)
- STD_INOUT/AN (standard bi-directional pads with analog input)
- High Current (high current bi-directional pads)
- STD_IN/AN (standard input pads with analog input)
- Power (power supply)

Details about the pad properties are defined in the Electrical Parameters.

Table 6 Package Pin Mapping

Function	VQFN 40	TSSOP 38	TSSOP 28	VQFN 24	TSSOP 16	Pad Type	Notes
P0.0	23	17	13	15	7	STD_INOUT	
P0.1	24	18	-	-	-	STD_INOUT	
P0.2	25	19	-	-	-	STD_INOUT	
P0.3	26	20	-	-	-	STD_INOUT	
P0.4	27	21	14	-	-	STD_INOUT	
P0.5	28	22	15	16	8	STD_INOUT	
P0.6	29	23	16	17	9	STD_INOUT	

Table 6 Package Pin Mapping (cont'd)

Function	VQFN 40	TSSOP 38	TSSOP 28	VQFN 24	TSSOP 16	Pad Type	Notes
P2.1	2	36	26	2	-	STD_IN OUT/AN	
P2.2	3	37	27	3	-	STD_IN/ AN	
P2.3	4	38	-	-	-	STD_IN/ AN	
P2.4	5	1	-	-	-	STD_IN/ AN	
P2.5	6	2	28	-	-	STD_IN/ AN	
P2.6	7	3	1	4	16	STD_IN/ AN	
P2.7	8	4	2	5	1	STD_IN/ AN	
P2.8	9	5	3	5	1	STD_IN/ AN	
P2.9	10	6	4	6	2	STD_IN/ AN	
P2.10	11	7	5	7	3	STD_IN OUT/AN	
P2.11	12	8	6	8	4	STD_IN OUT/AN	
VSS	13	9	7	9	5	Power	Supply GND, ADC reference GND
VDD	14	10	8	10	6	Power	Supply VDD, ADC reference voltage/ ORC reference voltage
VDDP	15	10	8	10	6	Power	When VDD is supplied, VDDP has to be supplied with the same voltage.

2.2.3 Hardware Controlled I/O Function Description

The following general building block is used to describe the hardware I/O and pull control functions of each PORT pin:

Table 8 Hardware Controlled I/O Function Description

Function	Outputs	Inputs	Pull Control	
	HWO0	**HWI0**	**HW0_PD**	**HW0_PU**
P0.0	MODB.OUT	MODB.INA		
Pn.y			$\overline{\text{MODC.OUT}}$	MODC.OUT

By Pn_HWSEL, it is possible to select between different hardware “masters” (HWO0/HWI0, HWO1/HWI1). The selected peripheral can take control of the pin(s). Hardware control overrules settings in the respective port pin registers. Additional hardware signals HW0_PD/HW1_PD and HW0_PU/HW1_PU controlled by the peripherals can be used to control the pull devices of the pin.

Please refer to the **Hardware Controlled I/O Functions** table for the complete hardware I/O and pull control function mapping.

Table 9 Port I/O Functions (cont'd)

Function	Outputs							Inputs									
	ALT1	ALT2	ALT3	ALT4	ALT5	ALT6	ALT7	Input	Input	Input	Input	Input	Input	Input	Input	Input	Input
P2.6								ACMP1.INN	VADC0.G0CH0		ERU0.2A1	USIC0_CH0.DX3E	USIC0_CH0.DX4E	USIC0_CH1.DX5D	ORC4.AIN		
P2.7								ACMP1.INP	VADC0.G1CH1		ERU0.3A1	USIC0_CH0.DX5C	USIC0_CH1.DX3D	USIC0_CH1.DX4D	ORC5.AIN		
P2.8								ACMP0.INN	VADC0.G0CH1	VADC0.G1CH0	ERU0.3B1	USIC0_CH0.DX3D	USIC0_CH0.DX4D	USIC0_CH1.DX5C	ORC6.AIN		
P2.9								ACMP0.INP	VADC0.G0CH2	VADC0.G1CH4	ERU0.3B0	USIC0_CH0.DX5A	USIC0_CH1.DX3B	USIC0_CH1.DX4B	ORC7.AIN		
P2.10	ERU0.PDOUT1	CCU40.OUT2	ERU0.GOUT1		CCU80.OUT30	ACMP0.OUT	USIC0_CH1.DOUT0		VADC0.G0CH3	VADC0.G1CH2	ERU0.2B0	USIC0_CH0.DX3C	USIC0_CH0.DX4C	USIC0_CH1.DX0F			
P2.11	ERU0.PDOUT0	CCU40.OUT3	ERU0.GOUT0		CCU80.OUT31	USIC0_CH1.SCLKOUT	USIC0_CH1.DOUT0	ACMP.REF	VADC0.G0CH4	VADC0.G1CH3	ERU0.2B1	USIC0_CH1.DX0E	USIC0_CH1.DX1E				

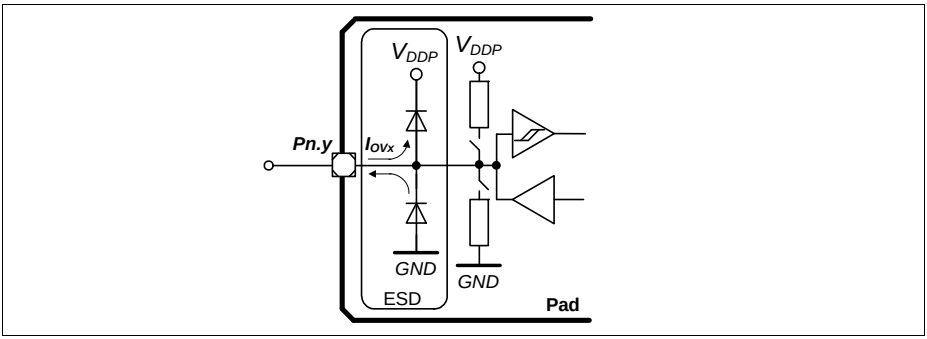

Figure 10 Input Overload Current via ESD structures

Table 13 and **Table 14** list input voltages that can be reached under overload conditions. Note that the absolute maximum input voltages as defined in the **Absolute Maximum Ratings** must not be exceeded during overload.

Table 13 PN-Junction Characterisitics for positive Overload

Pad Type	I_{OV} = 5 mA
Standard, High-current, AN/DIG_IN	$V_{IN} = V_{DDP} + 0.5$ V $V_{AIN} = V_{DDP} + 0.5$ V $V_{AREF} = V_{DDP} + 0.5$ V
P2.[1,2,6:9,11]	$V_{INP2} = V_{DDP} + 0.3$ V

Table 14 PN-Junction Characterisitics for negative Overload

Pad Type	I_{OV} = 5 mA
Standard, High-current, AN/DIG_IN	$V_{IN} = V_{SS} - 0.5$ V $V_{AIN} = V_{SS} - 0.5$ V $V_{AREF} = V_{SS} - 0.5$ V
P2.[1,2,6:9,11]	$V_{INP2} = V_{SS} - 0.3$ V

Table 16 Input/Output Characteristics (Operating Conditions apply) (cont'd)

Parameter	Symbol		Limit Values		Unit	Test Conditions
			Min.	Max.		
Input Hysteresis[8)]	HYS	CC	0.08 × V_{DDP}	–	V	CMOS Mode (5 V), Standard Hysteresis
			0.03 × V_{DDP}	–	V	CMOS Mode (3.3 V), Standard Hysteresis
			0.02 × V_{DDP}	–	V	CMOS Mode (2.2 V), Standard Hysteresis
			0.5 × V_{DDP}	0.75 × V_{DDP}	V	CMOS Mode(5 V), Large Hysteresis
			0.4 × V_{DDP}	0.75 × V_{DDP}	V	CMOS Mode(3.3 V), Large Hysteresis
			0.2 × V_{DDP}	0.65 × V_{DDP}	V	CMOS Mode(2.2 V), Large Hysteresis
Pin capacitance (digital inputs/outputs)	C_{IO}	CC	–	10	pF	
Pull-up resistor on port pins	R_{PUP}	CC	20	50	kohm	$V_{IN} = V_{SSP}$
Pull-down resistor on port pins	R_{PDP}	CC	20	50	kohm	$V_{IN} = V_{DDP}$
Input leakage current[9)]	I_{OZP}	CC	-1	1	µA	$0 < V_{IN} < V_{DDP}$, $T_A \leq 105$ °C
Voltage on any pin during V_{DDP} power off	V_{PO}	SR	–	0.3	V	[10)]
Maximum current per pin (excluding P1, V_{DDP} and V_{SS})	I_{MP}	SR	-10	11	mA	–
Maximum current per high currrent pins	I_{MP1A}	SR	-10	50	mA	–
Maximum current into V_{DDP} (TSSOP16, VQFN24)	I_{MVDD1}	SR	–	130	mA	[18)]
Maximum current into V_{DDP} (TSSOP38, VQFN40)	I_{MVDD2}	SR	–	260	mA	[18)]

3.2.2 Analog to Digital Converters (ADC)

Table 17 shows the Analog to Digital Converter (ADC) characteristics.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Table 17 ADC Characteristics (Operating Conditions apply)[1)]

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Supply voltage range (internal reference)	V_{DD_int} SR	2.0	–	3.0	V	SHSCFG.AREF = 11_B CALCTR.CALGNSTC = $0C_H$
		3.0	–	5.5	V	SHSCFG.AREF = 10_B
Supply voltage range (external reference)	V_{DD_ext} SR	3.0	–	5.5	V	SHSCFG.AREF = 00_B
Analog input voltage range	V_{AIN} SR	V_{SSP} - 0.05	–	V_{DDP} + 0.05	V	
Auxiliary analog reference ground	V_{REFGND} SR	V_{SSP} - 0.05	–	1.0	V	G0CH0
		V_{SSP} - 0.05	–	0.2	V	G1CH0
Internal reference voltage (full scale value)	V_{REFINT} CC	5			V	
Switched capacitance of an analog input	C_{AINS} CC	–	1.2	2	pF	GNCTRxz.GAINy = 00_B (unity gain)
		–	1.2	2	pF	GNCTRxz.GAINy = 01_B (gain g1)
		–	4.5	6	pF	GNCTRxz.GAINy = 10_B (gain g2)
		–	4.5	6	pF	GNCTRxz.GAINy = 11_B (gain g3)
Total capacitance of an analog input	C_{AINT} CC	–	–	10	pF	
Total capacitance of the reference input	C_{AREFT} CC	–	–	10	pF	

Table 17 ADC Characteristics (Operating Conditions apply)[1)] (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Gain settings	G_{IN} CC	1			–	GNCTRxz.GAINy = 00_B (unity gain)
		3			–	GNCTRxz.GAINy = 01_B (gain g1)
		6			–	GNCTRxz.GAINy = 10_B (gain g2)
		12			–	GNCTRxz.GAINy = 11_B (gain g3)
Sample Time	t_{sample} CC	3	–	–	1 / f_{ADC}	V_{DD} = 5.0 V
		3	–	–	1 / f_{ADC}	V_{DD} = 3.3 V
		30	–	–	1 / f_{ADC}	V_{DD} = 2.0 V
Sigma delta loop hold time	t_{SD_hold} CC	20	–	–	µs	Residual charge stored in an active sigma delta loop remains available
Conversion time in fast compare mode	t_{CF} CC	9			1 / f_{ADC}	2)
Conversion time in 12-bit mode	t_{C12} CC	20			1 / f_{ADC}	2)
Maximum sample rate in 12-bit mode 3)	f_{C12} CC	–	–	f_{ADC} / 42.5	–	1 sample pending
		–	–	f_{ADC} / 62.5	–	2 samples pending
Conversion time in 10-bit mode	t_{C10} CC	18			1 / f_{ADC}	2)
Maximum sample rate in 10-bit mode 3)	f_{C10} CC	–	–	f_{ADC} / 40.5	–	1 sample pending
		–	–	f_{ADC} / 58.5	–	2 samples pending
Conversion time in 8-bit mode	t_{C8} CC	16			1 / f_{ADC}	2)

Table 17 ADC Characteristics (Operating Conditions apply)[1] (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Maximum sample rate in 8-bit mode [3]	f_{C8} CC	–	–	f_{ADC} / 38.5	–	1 sample pending
		–	–	f_{ADC} / 54.5	–	2 samples pending
RMS noise [4]	EN_{RMS} CC	–	1.5	–	LSB 12	DC input, V_{DD} = 5.0 V, $V_{AIN = 2.5 V}$, 25°C
DNL error	EA_{DNL} CC	–	±2.0	–	LSB 12	
INL error	EA_{INL} CC	–	±4.0	–	LSB 12	
Gain error with external reference	EA_{GAIN} CC	–	±0.5	–	%	SHSCFG.AREF = 00_B (calibrated)
Gain error with internal reference [5]	EA_{GAIN} CC	–	±3.6	–	%	SHSCFG.AREF = $1X_B$ (calibrated), -40°C - 105°C
		–	±2.0	–	%	SHSCFG.AREF = $1X_B$ (calibrated), 0°C - 85°C
Offset error	EA_{OFF} CC	–	±8.0	–	mV	Calibrated, V_{DD} = 5.0 V

1) The parameters are defined for ADC clock frequency f_{SH} = 32MHz, SHSCFG.DIVS = 0000_B. Usage of any other frequencies may affect the ADC performance.

2) No pending samples assumed, excluding sampling time and calibration.

3) Includes synchronization and calibration (average of gain and offset calibration).

4) This parameter can also be defined as an SNR value: SNR[dB] = 20 × log(A_{MAXeff} / N_{RMS}).
With $A_{MAXeff} = 2^N / 2$, SNR[dB] = 20 × log (2048 / N_{RMS}) [N = 12].
N_{RMS} = 1.5 LSB12, therefore, equals SNR = 20 × log (2048 / 1.5) = 62.7 dB.

5) Includes error from the reference voltage.

3.3.2 Power-Up and Supply Monitoring Characteristics

Table 24 provides the characteristics of the power-up and supply monitoring in XMC1300.

The guard band between the lowest valid operating voltage and the brownout reset threshold provides a margin for noise immunity and hysteresis. The electrical parameters may be violated while V_{DDP} is outside its operating range.

The brownout detection triggers a reset within the defined range. The prewarning detection can be used to trigger an early warning and issue corrective and/or fail-safe actions in case of a critical supply voltage drop.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Table 24 Power-Up and Supply Monitoring Parameters (Operating Conditions apply)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
V_{DDP} ramp-up time	t_{RAMPUP} SR	V_{DDP}/ $S_{VDDPrise}$	–	10^7	µs	
V_{DDP} slew rate	S_{VDDPOP} SR	0	–	0.1	V/µs	Slope during normal operation
	S_{VDDP10} SR	0	–	10	V/µs	Slope during fast transient within +/- 10% of V_{DDP}
	$S_{VDDPrise}$ SR	0	–	10	V/µs	Slope during power-on or restart after brownout event
	$S_{VDDPfall}$ [1] SR	0	–	0.25	V/µs	Slope during supply falling out of the +/-10% limits[2]
V_{DDP} prewarning voltage	V_{DDPPW} CC	2.1	2.25	2.4	V	ANAVDEL.VDEL_SELECT = 00_B
		2.85	3	3.15	V	ANAVDEL.VDEL_SELECT = 01_B
		4.2	4.4	4.6	V	ANAVDEL.VDEL_SELECT = 10_B

3.3.5 SPD Timing Requirements

The optimum SPD decision time between 0_B and 1_B is 0.75 µs. With this value the system has maximum robustness against frequency deviations of the sampling clock on tool and on device side. However it is not always possible to exactly match this value with the given constraints for the sample clock. For instance for a oversampling rate of 4, the sample clock will be 8 MHz and in this case the closest possible effective decision time is 5.5 clock cycles (0.69 µs).

Table 28 Optimum Number of Sample Clocks for SPD

Sample Freq.	Sampling Factor	Sample Clocks 0_B	Sample Clocks 1_B	Effective Decision Time[1)]	Remark
8 MHz	4	1 to 5	6 to 12	0.69 µs	The other closest option (0.81 µs) for the effective decision time is less robust.

1) Nominal sample frequency period multiplied with 0.5 + (max. number of 0_B sample clocks)

For a balanced distribution of the timing robustness of SPD between tool and device, the timing requirements for the tool are:

- Frequency deviation of the sample clock is +/- 5%
- Effective decision time is between 0.69 µs and 0.75 µs (calculated with nominal sample frequency)

3.3.6 Peripheral Timings

Note: These parameters are not subject to production test, but verified by design and/or characterization.

3.3.6.1 Synchronous Serial Interface (USIC SSC) Timing

The following parameters are applicable for a USIC channel operated in SSC mode.

Note: Operating Conditions apply.

Table 29 USIC SSC Master Mode Timing

Parameter	Symbol		Values			Unit	Note / Test Condition
			Min.	Typ.	Max.		
SCLKOUT master clock period	t_{CLK}	CC	62.5	–	–	ns	
Slave select output SELO active to first SCLKOUT transmit edge	t_1	CC	80	–	–	ns	
Slave select output SELO inactive after last SCLKOUT receive edge	t_2	CC	0	–	–	ns	
Data output DOUT[3:0] valid time	t_3	CC	-10	–	10	ns	
Receive data input DX0/DX[5:3] setup time to SCLKOUT receive edge	t_4	SR	80	–	–	ns	
Data input DX0/DX[5:3] hold time from SCLKOUT receive edge	t_5	SR	0	–	–	ns	

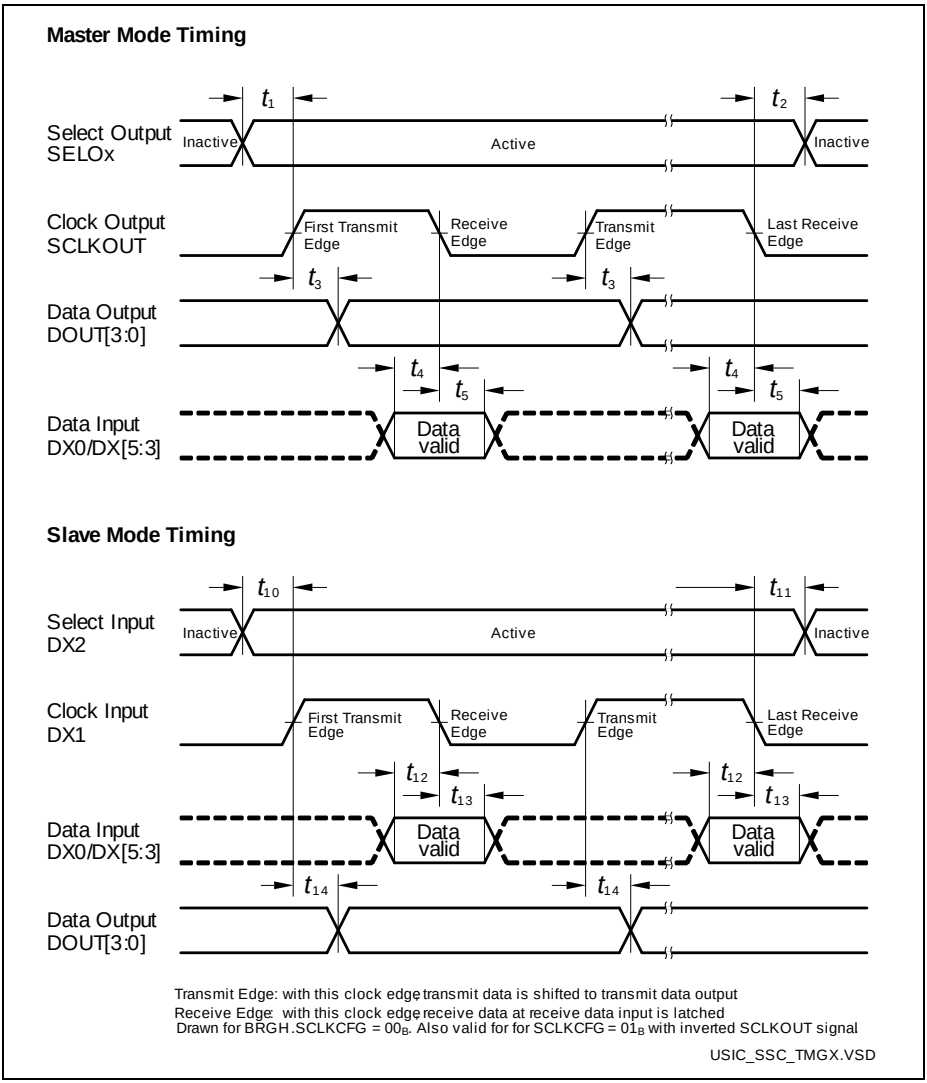

Figure 22 USIC - SSC Master/Slave Mode Timing

Note: This timing diagram shows a standard configuration, for which the slave select signal is low-active, and the serial clock signal is not shifted and not inverted.

4 Package and Reliability

The XMC1300 is a member of the XMC1000 Family of microcontrollers. It is also compatible to a certain extent with members of similar families or subfamilies.

Each package is optimized for the device it houses. Therefore, there may be slight differences between packages of the same pin-count but for different device types. In particular, the size of the exposed die pad may vary.

If different device types are considered or planned for an application, it must be ensured that the board layout fits all packages under consideration.

4.1 Package Parameters

Table 35 provides the thermal characteristics of the packages used in XMC1300.

Table 35 Thermal Characteristics of the Packages

Parameter	Symbol	Limit Values		Unit	Package Types
		Min.	Max.		
Exposed Die Pad Dimensions	Ex × Ey CC	-	2.7 × 2.7	mm	PG-VQFN-24-19
		-	3.7 × 3.7	mm	PG-VQFN-40-13
Thermal resistance Junction-Ambient	$R_{\Theta JA}$ CC	-	104.6	K/W	PG-TSSOP-16-8[1)]
		-	83.2	K/W	PG-TSSOP-28-16[1)]
		-	70.3	K/W	PG-TSSOP-38-9[1)]
		-	46.0	K/W	PG-VQFN-24-19[1)]
		-	38.4	K/W	PG-VQFN-40-13[1)]

1) Device mounted on a 4-layer JEDEC board (JESD 51-5); exposed pad soldered.

Note: For electrical reasons, it is required to connect the exposed pad to the board ground V_{SSP}, independent of EMC and thermal requirements.

4.1.1 Thermal Considerations

When operating the XMC1300 in a system, the total heat generated in the chip must be dissipated to the ambient environment to prevent overheating and the resulting thermal damage.

The maximum heat that can be dissipated depends on the package and its integration into the target board. The “Thermal resistance $R_{\Theta JA}$” quantifies these parameters. The power dissipation must be limited so that the average junction temperature does not exceed 115 °C.

The difference between junction temperature and ambient temperature is determined by $\Delta T = (P_{INT} + P_{IOSTAT} + P_{IODYN}) \times R_{\Theta JA}$

The internal power consumption is defined as
$P_{INT} = V_{DDP} \times I_{DDP}$ (switching current and leakage current).

The static external power consumption caused by the output drivers is defined as
$P_{IOSTAT} = \Sigma((V_{DDP}-V_{OH}) \times I_{OH}) + \Sigma(V_{OL} \times I_{OL})$

The dynamic external power consumption caused by the output drivers (P_{IODYN}) depends on the capacitive load connected to the respective pins and their switching frequencies.

If the total power dissipation for a given system configuration exceeds the defined limit, countermeasures must be taken to ensure proper system operation:

- Reduce V_{DDP}, if possible in the system
- Reduce the system frequency
- Reduce the number of output pins
- Reduce the load on active output drivers

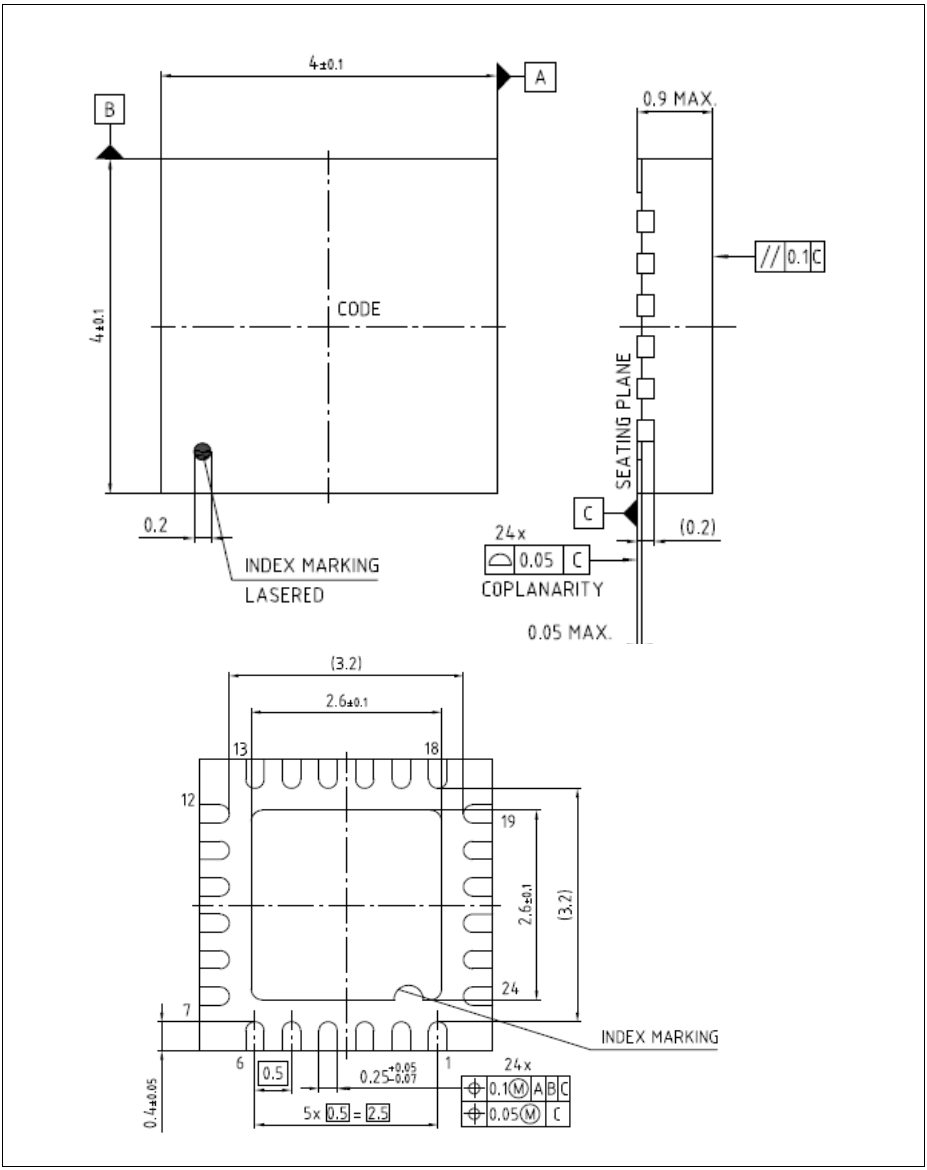

Figure 29 PG-VQFN-24-19

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