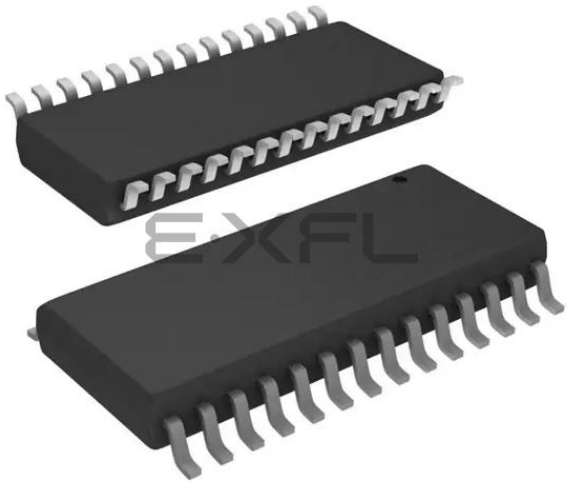

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	40MHz
Connectivity	I²C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LVD, Power Control PWM, QEI, POR, PWM, WDT
Number of I/O	24
Program Memory Size	16KB (8K x 16)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	768 x 8
Voltage - Supply (Vcc/Vdd)	4.2V ~ 5.5V
Data Converters	A/D 5x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-SOIC (0.295", 7.50mm Width)
Supplier Device Package	28-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18f2431-i-so

1.2 Other Special Features

- **Memory Endurance:** The enhanced Flash cells for both program memory and data EEPROM are rated to last for many thousands of erase/write cycles – up to 100,000 for program memory and 1,000,000 for EEPROM. Data retention without refresh is conservatively estimated to be greater than 100 years.
- **Self-Programmability:** These devices can write to their own program memory spaces under internal software control. By using a bootloader routine located in the protected boot block at the top of program memory, it becomes possible to create an application that can update itself in the field.
- **Power Control PWM Module:** In PWM mode, this module provides 1, 2 or 4 modulated outputs for controlling half-bridge and full-bridge drivers. Other features include auto-shutdown on Fault detection and auto-restart to reactivate outputs once the condition has cleared.
- **Enhanced Addressable USART:** This serial communication module is capable of standard RS-232 operation and provides support for the LIN/J2602 bus protocol. Other enhancements include automatic baud rate detection and a 16-bit Baud Rate Generator for improved resolution. When the microcontroller is using the internal oscillator block, the EUSART provides stable operation for applications that talk to the outside world without using an external crystal (or its accompanying power requirement).
- **Extended Watchdog Timer (WDT):** This enhanced version incorporates a 16-bit prescaler, allowing an extended time-out range that is stable across operating voltage and temperature. See **Section 26.0 “Electrical Characteristics”** for time-out periods.
- **High-Speed 10-Bit A/D Converter:** This module incorporates programmable acquisition time, allowing for a channel to be selected and a conversion to be initiated without waiting for a sampling period and thus, reducing code overhead.
- **Motion Feedback Module** (MFM): This module features a Quadrature Encoder Interface (QEI) and an Input Capture (IC) module. The QEI accepts two phase inputs (QEA, QEB) and one index input (INDX) from an incremental encoder. The QEI supports high and low precision position tracking, direction status and change of direction interrupt and velocity measurement. The input capture features 3 channels of independent input capture with Timer5 as the time base, a Special Event Trigger to other modules and an adjustable noise filter on each IC input.
- **Extended Watchdog Timer (WDT):** This enhanced version incorporates a 16-bit prescaler, allowing a time-out range from 4 ms to over 2 minutes, that is stable across operating voltage and temperature.

5.6 Reset State of Registers

Most registers are unaffected by a Reset. Their status is unknown on POR and unchanged by all other Resets. The other registers are forced to a “Reset state” depending on the type of Reset that occurred.

Most registers are not affected by a WDT wake-up, since this is viewed as the resumption of normal operation.

Status bits from the RCON register ($\overline{RI}$, $\overline{TO}$, $\overline{PD}$, $\overline{POR}$ and $\overline{BOR}$) are set or cleared differently in different Reset situations, as indicated in Table 5-2. These bits are used in software to determine the nature of the Reset.

Table 5-3 describes the Reset states for all of the Special Function Registers. These are categorized by Power-on and Brown-out Resets, Master Clear and WDT Resets, and WDT wake-ups.

FIGURE 5-3: TIME-OUT SEQUENCE ON POWER-UP ($\overline{MCLR}$ TIED TO V_{DD}, V_{DD} RISE < T_{PWRT})

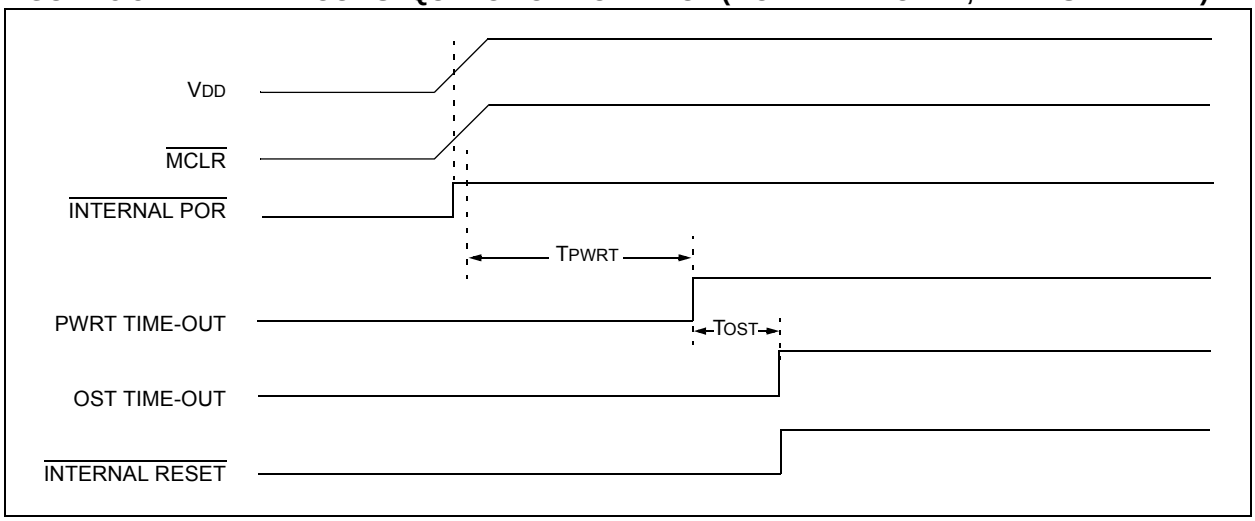
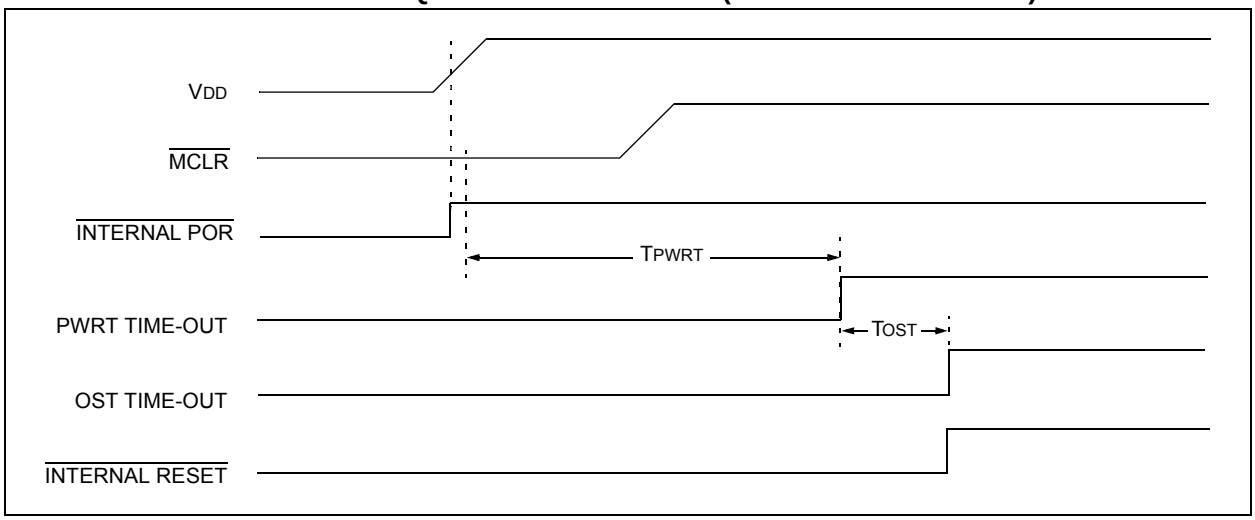


FIGURE 5-4: TIME-OUT SEQUENCE ON POWER-UP ($\overline{MCLR}$ NOT TIED TO V_{DD}): CASE 1



8.5.1 FLASH PROGRAM MEMORY WRITE SEQUENCE

The sequence of events for programming an internal program memory location should be:

1. Read 64 bytes into RAM.
2. Update data values in RAM as necessary.
3. Load Table Pointer with address being erased.
4. Do the row erase procedure (see **Section 8.4.1 "Flash Program Memory Erase Sequence"**).
5. Load Table Pointer with the address of the first byte being written.
6. Write the first 8 bytes into the holding registers with auto-increment.
7. Set the EECON1 register for the write operation by doing the following:
   - Set the EEPGD bit to point to program memory
   - Clear the CFGS bit to access program memory
   - Set the WREN bit to enable byte writes
8. Disable interrupts.
9. Write 55h to EECON2.
10. Write 0AAh to EECON2.
11. Set the WR bit. This will begin the write cycle.
12. The CPU will stall for the duration of the write (about 2 ms using internal timer).
13. Execute a NOP.
14. Re-enable interrupts.
15. Repeat Steps 6-14 seven times to write 64 bytes.
16. Verify the memory (table read).

This procedure will require about 18 ms to update one row of 64 bytes of memory. An example of the required code is given in Example 8-3.

REGISTER 10-2: INTCON2: INTERRUPT CONTROL REGISTER 2

R/W-1	R/W-1	R/W-1	R/W-1	U-0	R/W-1	U-0	R/W-1
$\overline{\text{RBPU}}$	INTEDG0	INTEDG1	INTEDG2	—	TMR0IP	—	RBIP
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	‘1’ = Bit is set	‘0’ = Bit is cleared	x = Bit is unknown

bit 7 **$\overline{\text{RBPU}}$:** PORTB Pull-up Enable bit
1 = All PORTB pull-ups are disabled
0 = PORTB pull-ups are enabled by individual port latch values

bit 6 **INTEDG0:** External Interrupt 0 Edge Select bit
1 = Interrupt on rising edge
0 = Interrupt on falling edge

bit 5 **INTEDG1:** External Interrupt 1 Edge Select bit
1 = Interrupt on rising edge
0 = Interrupt on falling edge

bit 4 **INTEDG2:** External Interrupt 2 Edge Select bit
1 = Interrupt on rising edge
0 = Interrupt on falling edge

bit 3 **Unimplemented:** Read as ‘0’

bit 2 **TMR0IP:** TMR0 Overflow Interrupt Priority bit
1 = High priority
0 = Low priority

bit 1 **Unimplemented:** Read as ‘0’

bit 0 **RBIP:** RB Port Change Interrupt Priority bit
1 = High priority
0 = Low priority

Note: Interrupt flag bits are set when an interrupt condition occurs, regardless of the state of its corresponding enable bit or the global enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt. This feature allows for software polling.

14.2 Timer2 Interrupt

Timer2 can also generate an optional device interrupt. The Timer2 output signal (TMR2 to PR2 match) provides the input for the 4-bit output counter/postscaler. This counter generates the TMR2 match interrupt flag which is latched in TMR2IF (PIR1<1>).

The interrupt is enabled by setting the TMR2 Match Interrupt Enable bit, TMR2IE (PIE1<1>). A range of 16 postscale options (from 1:1 through 1:16 inclusive) can be selected with the postscaler control bits, T2OUTPS<3:0> (T2CON<6:3>).

14.3 Output of TMR2

The unscaled output of TMR2 is available primarily to the CCP modules, where it is used as a time base for operations in PWM mode. Timer2 can be optionally used as the shift clock source for the SSP module operating in SPI mode.

For additional information, see **Section 19.0 "Synchronous Serial Port (SSP) Module"**.

FIGURE 14-1: TIMER2 BLOCK DIAGRAM

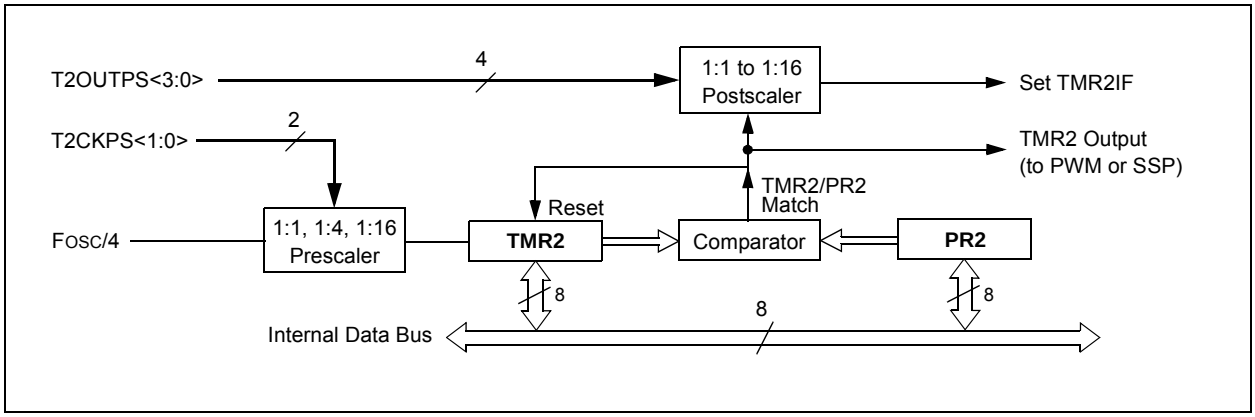


TABLE 14-1: REGISTERS ASSOCIATED WITH TIMER2 AS A TIMER/COUNTER

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page:
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	54
PIR1	—	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	57
PIE1	—	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	57
IPR1	—	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	57
TMR2	Timer2 Register								55
T2CON	—	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	55
PR2	Timer2 Period Register								55

Legend: — = unimplemented, read as '0'. Shaded cells are not used by the Timer2 module.

TABLE 17-8: REGISTERS ASSOCIATED WITH THE MOTION FEEDBACK MODULE

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page:
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	54
IPR3	—	—	—	PTIP	IC3DRIP	IC2QEIP	IC1IP	TMR5IP	56
PIE3	—	—	—	PTIE	IC3DRIE	IC2QEIE	IC1IE	TMR5IE	56
PIR3	—	—	—	PTIF	IC3DRIF	IC2QEIF	IC1IF	TMR5IF	56
TMR5H	Timer5 Register High Byte								57
TMR5L	Timer5 Register Low Byte								57
PR5H	Timer5 Period Register High Byte								57
PR5L	Timer5 Period Register Low Byte								57
T5CON	T5SEN	$\overline{\text{RESEN}}$	T5MOD	T5PS1	T5PS0	$\overline{\text{T5SYNC}}$	TMR5CS	TMR5ON	57
CAP1BUFH/ VELRH	Capture 1 Register High Byte/Velocity Register High Byte(1)								58
CAP1BUFL/ VELRL	Capture 1 Register Low Byte/Velocity Register Low Byte(1)								58
CAP2BUFH/ POSCNTH	Capture 2 Register High Byte/QEI Position Counter Register High Byte(1)								58
CAP2BUFL/ POSCNTL	Capture 2 Register Low Byte/QEI Position Counter Register Low Byte(1)								58
CAP3BUFH/ MAXCNTH	Capture 3 Register High Byte/QEI Max. Count Limit Register High Byte(1)								58
CAP3BUFL/ MAXCNTL	Capture 3 Register Low Byte/QEI Max. Count Limit Register Low Byte(1)								58
CAP1CON	—	CAP1REN	—	—	CAP1M3	CAP1M2	CAP1M1	CAP1M0	59
CAP2CON	—	CAP2REN	—	—	CAP2M3	CAP2M2	CAP2M1	CAP2M0	59
CAP3CON	—	CAP3REN	—	—	CAP3M3	CAP3M2	CAP3M1	CAP3M0	59
DFLTCON	—	FLT4EN	FLT3EN	FLT2EN	FLT1EN	FLTCK2	FLTCK1	FLTCK0	59
QEICON	$\overline{\text{VELM}}$	QERR	UP/$\overline{\text{DOWN}}$	QEIM2	QEIM1	QEIM0	PDEC1	PDEC0	56

Legend: — = unimplemented. Shaded cells are not used by the Motion Feedback Module.

Note 1: Register name and function determined by which submodule is selected (IC/QEI, respectively). See **Section 17.1.10 "Other Operating Modes"** for more information.

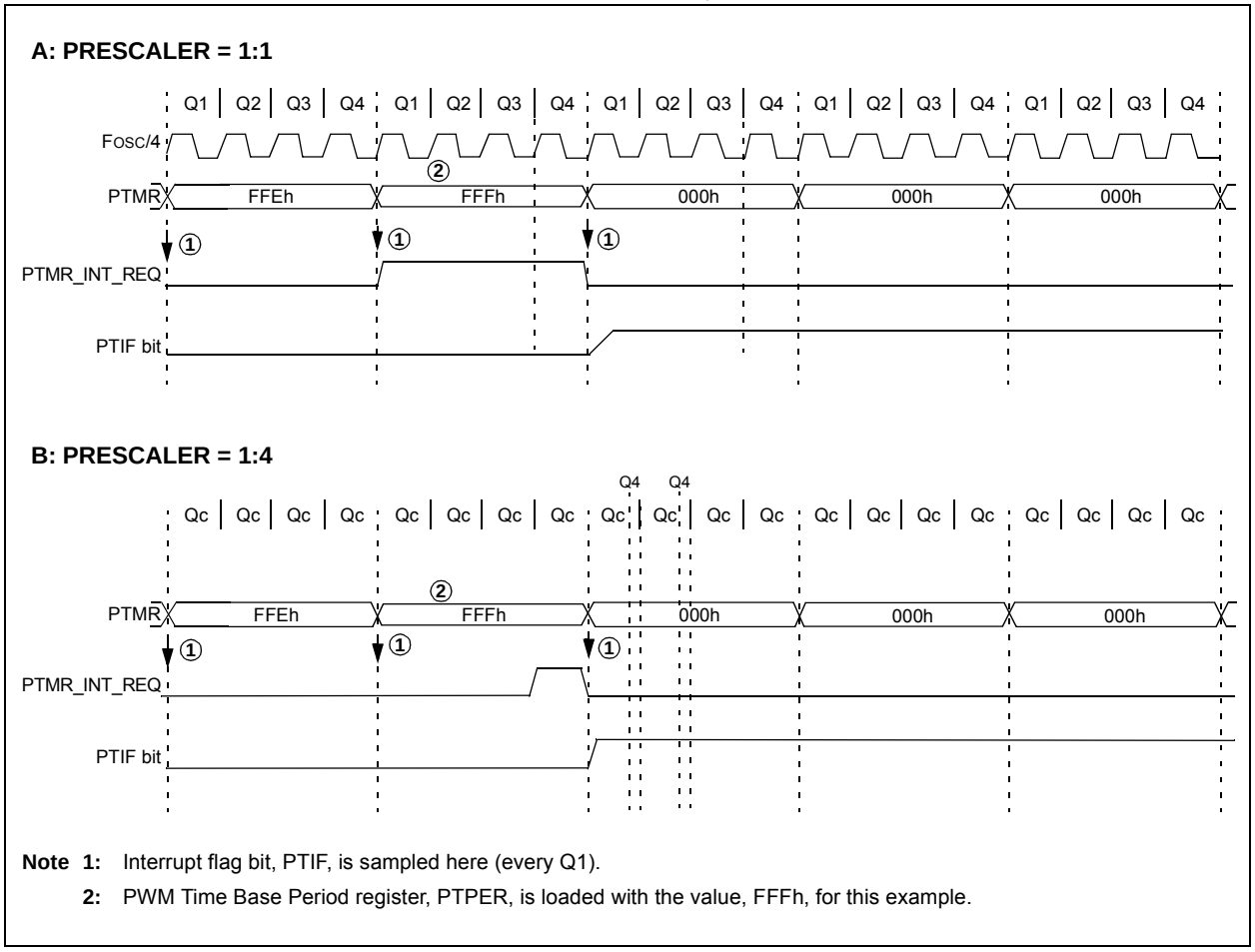
18.4.2 INTERRUPTS IN SINGLE-SHOT MODE

When the PWM time base is in the Single-Shot mode (PTMOD<1:0> = 01), an interrupt event is generated when a match with the PTPER register occurs. The PWM Time Base register (PTMR) is reset to zero on the following input clock edge and the PTEN bit is cleared. The postscaler selection bits have no effect in this Timer mode.

18.4.3 INTERRUPTS IN CONTINUOUS UP/DOWN COUNT MODE

In the Continuous Up/Down Count mode (PTMOD<1:0> = 10), an interrupt event is generated each time the value of the PTMR register becomes zero and the PWM time base begins to count upwards. The postscaler selection bits may be used in this mode of the timer to reduce the frequency of the interrupt events. Figure 18-7 shows the interrupts in Continuous Up/Down Count mode.

FIGURE 18-6: PWM TIME BASE INTERRUPT TIMING, SINGLE-SHOT MODE



Note 1: Interrupt flag bit, PTIF, is sampled here (every Q1).

2: PWM Time Base Period register, PTPER, is loaded with the value, FFFh, for this example.

18.6 PWM Duty Cycle

PWM duty cycle is defined by the PDCx (PDCxL and PDCxH) registers. There are a total of four PWM Duty Cycle registers for four pairs of PWM channels. The Duty Cycle registers have 14-bit resolution by combining six LSbs of PDCxH with the 8 bits of PDCxL. PDCx is a double-buffered register used to set the counting period for the PWM time base.

18.6.1 PWM DUTY CYCLE REGISTERS

There are four 14-bit Special Function Registers used to specify duty cycle values for the PWM module:

- PDC0 (PDC0L and PDC0H)
- PDC1 (PDC1L and PDC1H)
- PDC2 (PDC2L and PDC2H)
- PDC3 (PDC3L and PDC3H)

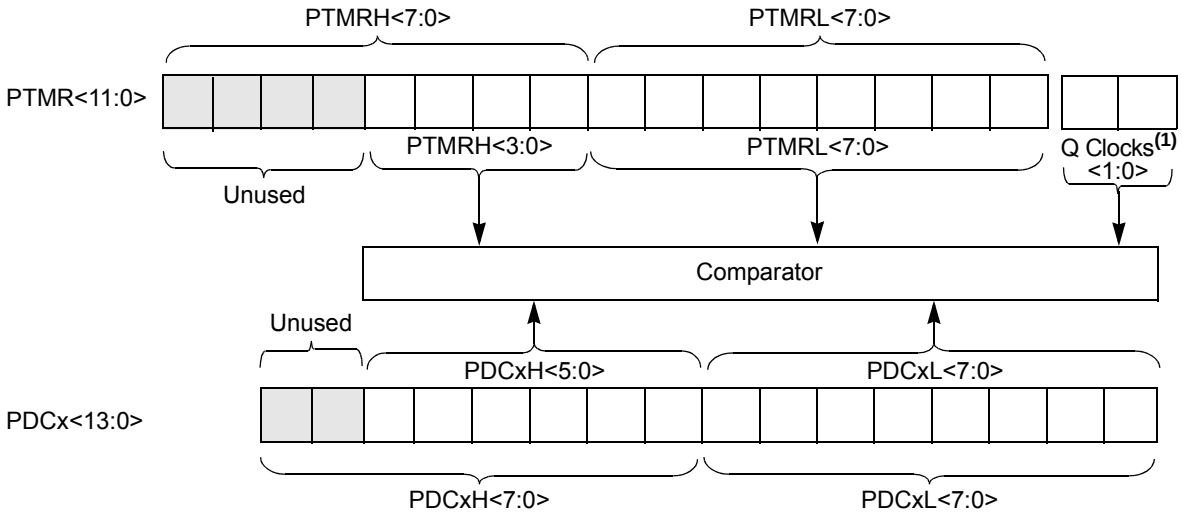
The value in each Duty Cycle register determines the amount of time that the PWM output is in the active state. The upper 12 bits of PDCx holds the actual duty cycle value from PTMRH/L<11:0>, while the lower 2 bits control which internal Q clock the duty cycle match will occur. This 2-bit value is decoded from the Q clocks as shown in Figure 18-11 (when the prescaler is 1:1 or PTCKPS<1:0> = 00).

In Edge-Aligned mode, the PWM period starts at Q1 and ends when the Duty Cycle register matches the PTMR register as follows. The duty cycle match is considered when the upper 12 bits of the PDCx are equal to the PTMR and the lower 2 bits are equal to Q1, Q2, Q3 or Q4, depending on the lower two bits of the PDCx (when the prescaler is 1:1 or PTCKPS<1:0> = 00).

Note:	When the prescaler is not 1:1 (PTCKPS<1:0> ≠ ~00), the duty cycle match occurs at the Q1 clock of the instruction cycle when the PTMR and PDCx match occurs.

Each compare unit has logic that allows override of the PWM signals. This logic also ensures that the PWM signals will complement each other (with dead-time insertion) in Complementary mode (see **Section 18.7 "Dead-Time Generators"**).

FIGURE 18-11: DUTY CYCLE COMPARISON



Note 1: This value is decoded from the Q clocks:
00 = duty cycle match occurs on Q1
01 = duty cycle match occurs on Q2
10 = duty cycle match occurs on Q3
11 = duty cycle match occurs on Q4

REGISTER 19-1: SSPSTAT: SYNCHRONOUS SERIAL PORT STATUS REGISTER

R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0
SMP	CKE	D/$\overline{A}$	P	S	R/$\overline{W}$	UA	BF
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **SMP:** Sample bit

SPI Master mode:
1 = Input data sampled at end of data output time
0 = Input data sampled at middle of data output time

SPI Slave mode:
SMP must be cleared when SPI is used in Slave mode.

bit 6 **CKE**: SPI Clock Edge Select bit (Figure 19-2, Figure 19-3 and Figure 19-4)

SPI mode, CKP = 0:
1 = Data transmitted on rising edge of SCK
0 = Data transmitted on falling edge of SCK

SPI mode, CKP = 1:
1 = Data transmitted on falling edge of SCK
0 = Data transmitted on rising edge of SCK

I^2C™ mode:
This bit must be maintained clear.

bit 5 **D/$\overline{A}$**: Data/$\overline{Address}$ bit (I^2C mode only)
1 = Indicates that the last byte received or transmitted was data
0 = Indicates that the last byte received or transmitted was address

bit 4 **P**: Stop bit (I^2C mode only)
This bit is cleared when the SSP module is disabled or when the Start bit is detected last; SSPEN is cleared.
1 = Indicates that a Stop bit has been detected last (this bit is '0' on Reset)
0 = Stop bit was not detected last

bit 3 **S**: Start bit (I^2C mode only)
This bit is cleared when the SSP module is disabled or when the Stop bit is detected last; SSPEN is cleared.
1 = Indicates that a Start bit has been detected last (this bit is '0' on Reset)
0 = Start bit was not detected last

bit 2 **R/$\overline{W}$**: Read/$\overline{Write}$ Information bit (I^2C mode only)
This bit holds the R/$\overline{W}$ bit information following the last address match. This bit is only valid from the address match to the next Start bit, Stop bit or $\overline{ACK}$ bit.
1 = Read
0 = Write

bit 1 **UA**: Update Address bit (10-Bit I^2C mode only)
1 = Indicates that the user needs to update the address in the SSPADD register
0 = Address does not need to be updated

bit 0 **BF**: Buffer Full Status bit

Receive (SPI and I^2C modes):
1 = Receive complete, SSPBUF is full
0 = Receive not complete, SSPBUF is empty

Transmit (I^2C mode only):
1 = Transmit in progress, SSPBUF is full
0 = Transmit complete, SSPBUF is empty

REGISTER 20-2: RCSTA: RECEIVE STATUS AND CONTROL REGISTER

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x
SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7 **SPEN:** Serial Port Enable bit
1 = Serial port enabled
0 = Serial port disabled

bit 6 **RX9:** 9-Bit Receive Enable bit
1 = Selects 9-bit reception
0 = Selects 8-bit reception

bit 5 **SREN:** Single Receive Enable bit
Asynchronous mode:
Don't care.
Synchronous mode – Master:
1 = Enables single receive
0 = Disables single receive
This bit is cleared after reception is complete.
Synchronous mode – Slave:
Don't care.

bit 4 **CREN:** Continuous Receive Enable bit
Asynchronous mode:
1 = Enables receiver
0 = Disables receiver
Synchronous mode:
1 = Enables continuous receive until enable bit, CREN, is cleared (CREN overrides SREN)
0 = Disables continuous receive

bit 3 **ADDEN:** Address Detect Enable bit
Asynchronous mode 9-Bit (RX9 = 1):
1 = Enables address detection, enables interrupt and loads the receive buffer when RSR<8> is set
0 = Disables address detection, all bytes are received and ninth bit can be used as parity bit
Asynchronous mode 8-Bit (RX9 = 0):
Don't care.

bit 2 **FERR:** Framing Error bit
1 = Framing error (can be cleared by reading RCREGx register and receiving next valid byte)
0 = No framing error

bit 1 **OERR:** Overrun Error bit
1 = Overrun error (can be cleared by clearing bit, CREN)
0 = No overrun error

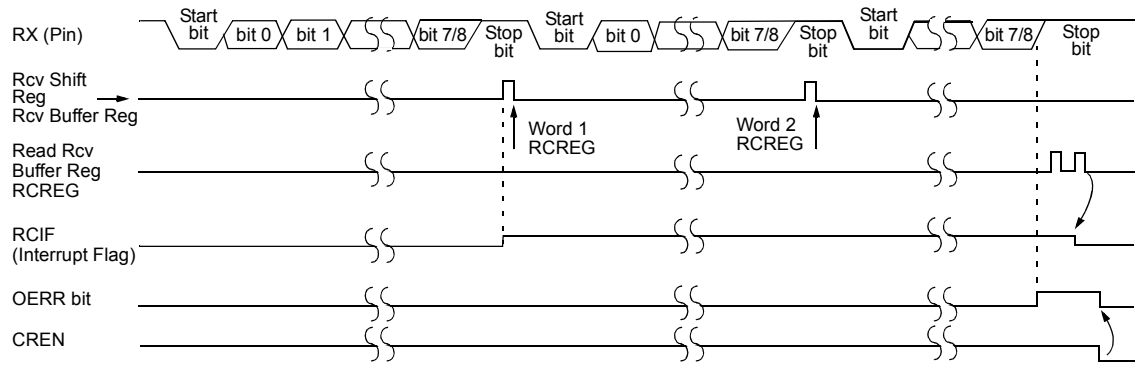
bit 0 **RX9D:** 9th Bit of Received Data
This can be address/data bit or a parity bit and must be calculated by user firmware.

To set up an Asynchronous Transmission:

1. Initialize the SPBRG register for the appropriate baud rate. If a high-speed baud rate is desired, set bit, BRGH (see **Section 20.2 "EUSART Baud Rate Generator (BRG)"**).
2. Enable the asynchronous serial port by clearing bit, SYNC, and setting bit, SPEN.
3. If interrupts are desired, set enable bit, TXIE.
4. If 9-bit transmission is desired, set transmit bit, TX9. Can be used as address/data bit.
5. Enable the transmission by setting bit, TXEN, which will also set bit, TXIF.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
7. Load data to the TXREG register (starts transmission).

If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

FIGURE 20-6: ASYNCHRONOUS RECEPTION



Note: This timing diagram shows three words appearing on the RX input. The RCREG (Receive Buffer) is read after the third word, causing the OERR (Overrun) bit to be set.

TABLE 20-6: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page:
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	54
PIR1	—	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	57
PIE1	—	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	57
IPR1	—	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	57
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	56
RCREG	EUSART Receive Register								56
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	56
BAUDCON	—	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	56
SPBRGH	EUSART Baud Rate Generator Register High Byte								56
SPBRG	EUSART Baud Rate Generator Register Low Byte								56

Legend: — = unimplemented, read as '0'. Shaded cells are not used for asynchronous reception.

20.3.4 AUTO-WAKE-UP ON SYNC BREAK CHARACTER

During Sleep mode, all clocks to the EUSART are suspended. Because of this, the Baud Rate Generator is inactive and a proper byte reception cannot be performed. The auto-wake-up feature allows the controller to wake-up due to activity on the RX/DT line, while the EUSART is operating in Asynchronous mode.

The auto-wake-up feature is enabled by setting the WUE bit (BAUDCON<1>). Once set, the typical receive sequence on RX/DT is disabled and the EUSART remains in an Idle state, monitoring for a wake-up event independent of the CPU mode. A wake-up event consists of a high-to-low transition on the RX/DT line. (This coincides with the start of a Sync Break or a Wake-up Signal character for the LIN/J2602 protocol.)

Following a wake-up event, the module generates an RCIF interrupt. The interrupt is generated synchronously to the Q clocks in normal operating modes (Figure 20-7), and asynchronously if the device is in Sleep mode (Figure 20-8). The interrupt condition is cleared by reading the RCREG register.

The WUE bit is automatically cleared once a low-to-high transition is observed on the RX line following the wake-up event. At this point, the EUSART module is in Idle mode and returns to normal operation. This signals to the user that the Sync Break event is over.

20.3.4.1 Special Considerations Using Auto-Wake-up

Since Auto-Wake-up functions by sensing rising edge transitions on RX/DT, information with any state changes before the Stop bit may signal a false end-of-character and cause data or framing errors. To work properly, therefore, the initial characters in the transmission must be all '0's. This can be 00h (8 bits) for standard RS-232 devices, or 000h (12 bits) for LIN/J2602 bus.

Oscillator start-up time must also be considered, especially in applications using oscillators with longer start-up intervals (i.e., LP, XT or HS/PLL mode). The Sync Break (or Wake-up Signal) character must be of sufficient length, and be followed by a sufficient interval, to allow enough time for the selected oscillator to start and provide proper initialization of the EUSART.

20.3.4.2 Special Considerations Using the WUE Bit

The timing of WUE and RCIF events may cause some confusion when it comes to determining the validity of received data. As noted, setting the WUE bit places the EUSART in an Idle mode. The wake-up event causes a receive interrupt by setting the RCIF bit. The WUE bit is cleared after this when a rising edge is seen on RX/DT. The interrupt condition is then cleared by reading the RCREG register. Ordinarily, the data in RCREG will be dummy data and should be discarded.

The fact that the WUE bit has been cleared (or is still set), and the RCIF flag is set, should not be used as an indicator of the integrity of the data in RCREG. Users should consider implementing a parallel method in firmware to verify received data integrity.

To assure that no actual data is lost, check the RCIDL bit to verify that a receive operation is not in process. If a receive operation is not occurring, the WUE bit may then be set just prior to entering the Sleep mode.

FIGURE 20-7: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING NORMAL OPERATION

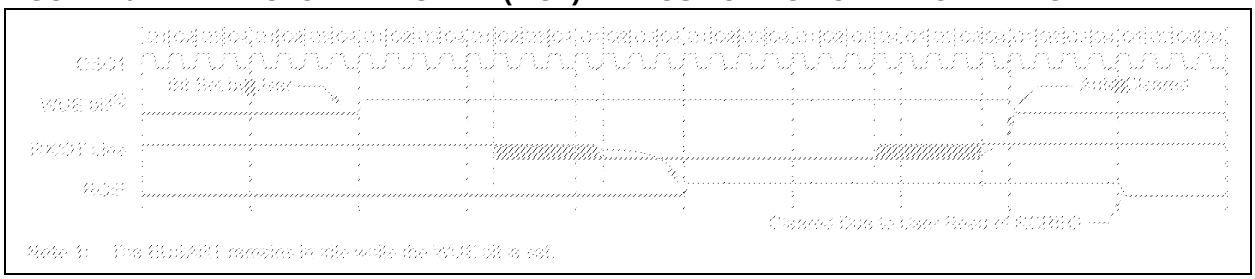
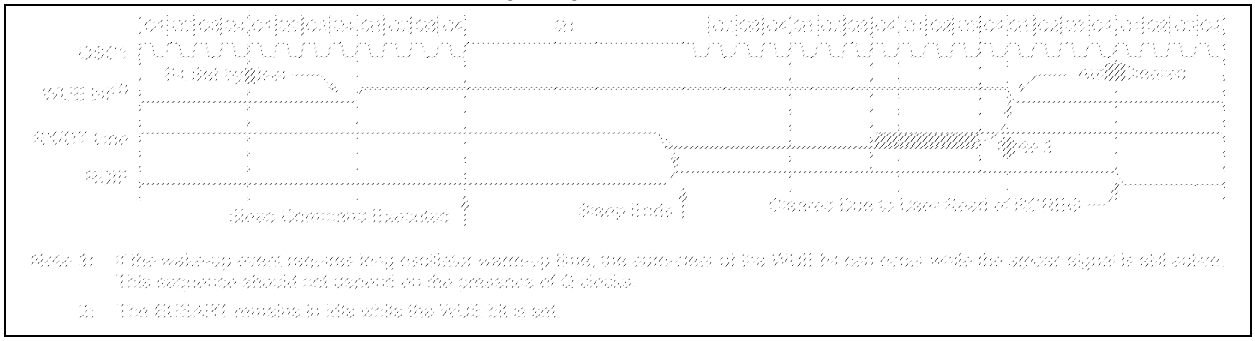


FIGURE 20-8: AUTO-WAKE-UP BIT (WUE) TIMINGS DURING SLEEP



20.4 EUSART Synchronous Master Mode

The Synchronous Master mode is entered by setting the CSRC bit (TXSTA<7>). In this mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit SYNC (TXSTA<4>). In addition, enable bit SPEN (RCSTA<7>) is set in order to configure the RC6/TX/CK/$\overline{SS}$ and RC7/RX/DT/SDO I/O pins to CK (clock) and DT (data) lines, respectively.

The Master mode indicates that the processor transmits the master clock on the CK line. Clock polarity is selected with the SCKP bit (BAUDCON<4>). Setting SCKP sets the Idle state on CK as high, while clearing the bit, sets the Idle state low. This option is provided to support Microwire devices with this module.

20.4.1 EUSART SYNCHRONOUS MASTER TRANSMISSION

The EUSART transmitter block diagram is shown in Figure 20-2. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The Shift register obtains its data from the Read/Write Transmit Buffer register, TXREG. The TXREG register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREG (if available).

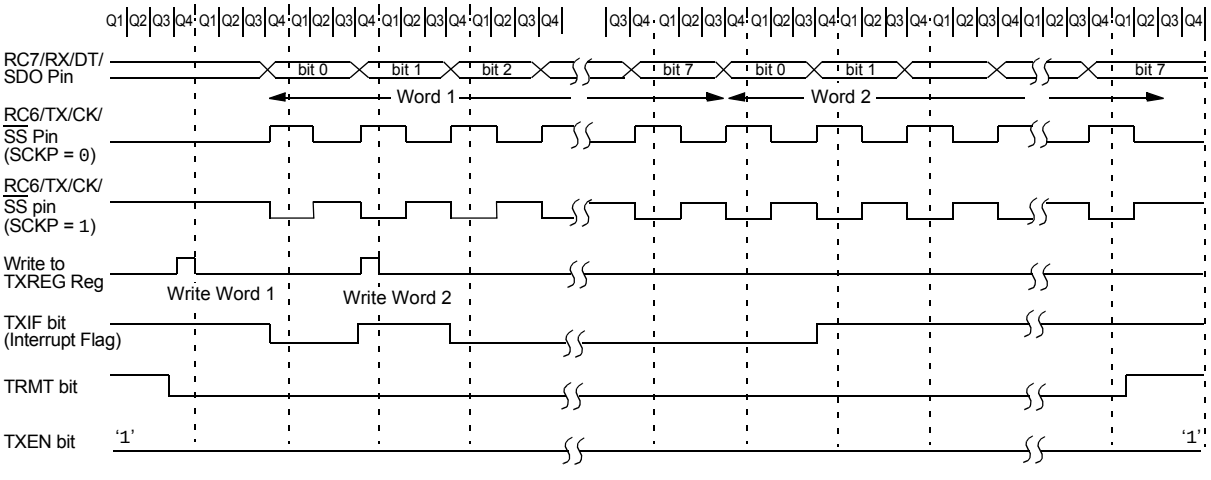
Once the TXREG register transfers the data to the TSR register (occurs in one TCYCLE), the TXREG is empty and interrupt bit, TXIF (PIR1<4>), is set. The interrupt can be enabled/disabled by setting/clearing enable bit, TXIE (PIE1<4>). Flag bit, TXIF, will be set, regardless of the state of enable bit, TXIE, and cannot be cleared in software. It will reset only when new data is loaded into the TXREG register.

While flag bit, TXIF, indicates the status of the TXREG register, another bit, TRMT (TXSTA<1>), shows the status of the TSR register. TRMT is a read-only bit which is set when the TSR is empty. No interrupt logic is tied to this bit, so the user must poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory, so it is not available to the user.

To set up a Synchronous Master Transmission:

1. Initialize the SPBRGH:SPBRG registers for the appropriate baud rate. Set or clear the BRGH and BRG16 bits, as required, to achieve the desired baud rate.
2. Enable the synchronous master serial port by setting bits, SYNC, SPEN and CSRC.
3. If interrupts are desired, set enable bit, TXIE.
4. If 9-bit transmission is desired, set bit, TX9.
5. Enable the transmission by setting bit, TXEN.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
7. Start transmission by loading data to the TXREG register.
8. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

FIGURE 20-10: SYNCHRONOUS TRANSMISSION



Note: Sync Master mode, SPBRG = 0, continuous transmission of two 8-bit words.

20.5 EUSART Synchronous Slave Mode

Synchronous Slave mode is entered by clearing bit, CSRC (TXSTA<7>). This mode differs from the Synchronous Master mode in that the shift clock is supplied externally at the RC6/TX/CK/SS pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in any low-power mode.

20.5.1 EUSART SYNCHRONOUS SLAVE TRANSMIT

The operation of the Synchronous Master and Slave modes are identical, except in the case of Sleep mode.

If two words are written to the TXREG and then the SLEEP instruction is executed, the following will occur:

a) The first word will immediately transfer to the TSR register and transmit.
b) The second word will remain in TXREG register.
c) Flag bit, TXIF, will not be set.
d) When the first word has been shifted out of TSR, the TXREG register will transfer the second word to the TSR and flag bit, TXIF, will now be set.
e) If enable bit, TXIE, is set, the interrupt will wake the chip from Sleep. If the global interrupt is enabled, the program will branch to the interrupt vector.

To set up a Synchronous Slave Transmission:

1. Enable the synchronous slave serial port by setting bits, SYNC and SPEN, and clearing bit, CSRC.
2. Clear bits, CREN and SREN.
3. If interrupts are desired, set enable bit, TXIE.
4. If 9-bit transmission is desired, set bit, TX9.
5. Enable the transmission by setting enable bit, TXEN.
6. If 9-bit transmission is selected, the ninth bit should be loaded in bit, TX9D.
7. Start transmission by loading data to the TXREG register.
8. If using interrupts, ensure that the GIE and PEIE bits in the INTCON register (INTCON<7:6>) are set.

TABLE 20-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE TRANSMISSION

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on Page:
INTCON	GIE/GIEH	PEIE/GIEL	TMR0IE	INT0IE	RBIE	TMR0IF	INT0IF	RBIF	54
PIR1	—	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	57
PIE1	—	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	57
IPR1	—	ADIP	RCIP	TXIP	SSPIP	CCP1IP	TMR2IP	TMR1IP	57
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	56
TXREG	EUSART Transmit Register								56
TXSTA	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	56
BAUDCON	—	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	56
SPBRGH	EUSART Baud Rate Generator Register High Byte								56
SPBRG	EUSART Baud Rate Generator Register Low Byte								56

Legend: — = unimplemented, read as '0'. Shaded cells are not used for synchronous slave transmission.

REGISTER 23-4: CONFIG3L: CONFIGURATION REGISTER 3 LOW (BYTE ADDRESS 300004h)

U-0	U	R/P-1	R/P-1	R/P-1	R/P-1	U	U
—	—	$\overline{\text{T1OSCMX}}$	HPOL(1)	LPOL(1)	PWMPIN(3)	—	—
bit 7							bit 0

Legend:		
R = Readable bit	P = Programmable bit	U = Unimplemented bit, read as '0'
-n = Value when device is unprogrammed		U = Unchanged from programmed state

bit 7-6 **Unimplemented:** Read as '0'

bit 5 **T1OSCMX:** Timer1 Oscillator Mode bit
1 = Low-power Timer1 operation when microcontroller is in Sleep mode
0 = Standard (legacy) Timer1 oscillator operation

bit 4 **HPOL:** High Side Transistors Polarity bit (i.e., Odd PWM Output Polarity Control bit)(1)
1 = PWM1, 3, 5 and 7 are active-high (default)(2)
0 = PWM1, 3, 5 and 7 are active-low(2)

bit 3 **LPOL:** Low Side Transistors Polarity bit (i.e., Even PWM Output Polarity Control bit)(1)
1 = PWM0, 2, 4 and 6 are active-high (default)(2)
0 = PWM0, 2, 4 and 6 are active-low(2)

bit 2 **PWMPIN:** PWM Output Pins Reset State Control bit(3)
1 = PWM outputs are disabled upon Reset (default)
0 = PWM outputs drive active states upon Reset

bit 1-0 **Unimplemented:** Read as '0'

Note 1: Polarity control bits, HPOL and LPOL, define PWM signal output active and inactive states; PWM states generated by the Fault inputs or PWM manual override.

2: PWM6 and PWM7 output channels are only available on PIC18F4331/4431 devices.

3: When PWMPIN = 0, PWMEN<2:0> = 101 if the device has eight PWM output pins (40 and 44-pin devices) and PWMEN<2:0> = 100 if the device has six PWM output pins (28-pin devices). PWM output polarity is defined by HPOL and LPOL.

INCFSZ **Increment f, Skip if 0**

Syntax: [*label*] INCFSZ f [,d [,a]]

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

Operation: (f) + 1 → dest,
skip if result = 0

Status Affected: None

Encoding:

0011	11da	ffff	ffff

Description: The contents of register, 'f', are incremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register, 'f'.
If the result is '0', the next instruction, which is already fetched, is discarded, and a NOP is executed instead, making it a two-cycle instruction. If 'a' is '0', the Access Bank will be selected, overriding the BSR value. If 'a' = 1, then the bank will be selected as per the BSR value.

Words: 1

Cycles: 1(2)
Note: 3 cycles if skip and followed by a 2-word instruction.

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:
```
HERE    INCFSZ  CNT
NZERO   :
ZERO    :
```

Before Instruction
PC = Address (HERE)

After Instruction
CNT = CNT + 1
If CNT = 0;
PC = Address (ZERO)
If CNT ≠ 0;
PC = Address (NZERO)

INFSNZ **Increment f, Skip if Not 0**

Syntax: [*label*] INFSNZ f [,d [,a]]

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

Operation: (f) + 1 → dest,
skip if result ≠ 0

Status Affected: None

Encoding:

0100	10da	ffff	ffff

Description: The contents of register, 'f', are incremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register, 'f'.
If the result is not '0', the next instruction, which is already fetched, is discarded, and a NOP is executed instead, making it a two-cycle instruction. If 'a' is '0', the Access Bank will be selected, overriding the BSR value. If 'a' = 1, then the bank will be selected as per the BSR value.

Words: 1

Cycles: 1(2)
Note: 3 cycles if skip and followed by a 2-word instruction.

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:
```
HERE    INFSNZ  REG
ZERO
NZERO
```

Before Instruction
PC = Address (HERE)

After Instruction
REG = REG + 1
If REG ≠ 0;
PC = Address (NZERO)
If REG = 0;
PC = Address (ZERO)

RLNCF **Rotate Left f (No Carry)**

Syntax: [*label*] RLNCF f [,d [,a]]

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

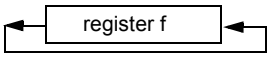
Operation: (f<n>) → dest<n + 1>,
(f<7>) → dest<0>

Status Affected: N, Z

Encoding:

0100	01da	ffff	ffff

Description: The contents of register, ‘f’, are rotated one bit to the left. If ‘d’ is ‘0’, the result is placed in W. If ‘d’ is ‘1’, the result is stored back in register, ‘f’. If ‘a’ is ‘0’, the Access Bank will be selected, overriding the BSR value. If ‘a’ is ‘1’, then the bank will be selected as per the BSR value.



Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register ‘f’	Process Data	Write to destination

Example: RLNCF REG

Before Instruction
REG = 1010 1011
After Instruction
REG = 0101 0111

RRCF **Rotate Right f through Carry**

Syntax: [*label*] RRCF f [,d [,a]]

Operands: $0 \le f \le 255$
$d \in [0,1]$
$a \in [0,1]$

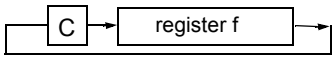
Operation: (f<n>) → dest<n – 1>,
(f<0>) → C,
(C) → dest<7>

Status Affected: C, N, Z

Encoding:

0011	00da	ffff	ffff

Description: The contents of register, ‘f’, are rotated one bit to the right through the Carry Flag. If ‘d’ is ‘0’, the result is placed in W. If ‘d’ is ‘1’, the result is placed back in register, ‘f’. If ‘a’ is ‘0’, the Access Bank will be selected, overriding the BSR value. If ‘a’ is ‘1’, then the bank will be selected as per the BSR value.



Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register ‘f’	Process Data	Write to destination

Example: RRCF REG, W

Before Instruction
REG = 1110 0110
C = 0
After Instruction
REG = 1110 0110
W = 0111 0011
C = 0

26.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings(†)

Ambient temperature under bias ..... -55°C to +125°C
Storage temperature ..... -65°C to +150°C
Voltage on any pin with respect to V_{SS} (except V_{DD}, $\overline{MCLR}$, and RA4) ..... -0.3V to (V_{DD} + 0.3V)
Voltage on V_{DD} with respect to V_{SS} ..... -0.3V to +7.5V
Voltage on $\overline{MCLR}$ with respect to V_{SS} **(Note 2)** ..... 0V to +13.25V
Voltage on RA4 with respect to Vss ..... 0V to +8.5V
Total power dissipation **(Note 1)** 1.0W
Maximum current out of V_{SS} pin ..... 300 mA
Maximum current into V_{DD} pin ..... 250 mA
Input clamp current, I_{IK} (V_I < 0 or V_I > V_{DD}) ..... ±20 mA
Output clamp current, I_{OK} (V_O < 0 or V_O > V_{DD}) ..... ±20 mA
Maximum output current sunk by any I/O pin ..... 25 mA
Maximum output current sourced by any I/O pin ..... 25 mA
Maximum current sunk by all ports ..... 200 mA
Maximum current sourced by all ports ..... 200 mA

Note 1: Power dissipation is calculated as follows:
$P_{dis} = V_{DD} \times \{I_{DD} - \sum I_{OH}\} + \sum \{(V_{DD} - V_{OH}) \times I_{OH}\} + \sum(V_{OL} \times I_{OL})$

2: Voltage spikes below V_{SS} at the $\overline{MCLR}/V_{PP}$ pin, inducing currents greater than 80 mA, may cause latch-up. Thus, a series resistor of 50-100Ω should be used when applying a "low" level to the $\overline{MCLR}/V_{PP}$ pin, rather than pulling this pin directly to V_{SS}.

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

26.2 DC Characteristics: Power-Down and Supply Current PIC18F2331/2431/4331/4431 (Industrial, Extended) PIC18LF2331/2431/4331/4431 (Industrial)

PIC18LF2331/2431/4331/4431 (Industrial)		**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial				
PIC18F2331/2431/4331/4431 (Industrial, Extended)		**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended				
Param No.	**Device**	**Typ**	**Max**	**Units**	**Conditions**	
	Power-Down Current (IPD)(1)					
	PIC18LF2X31/4X31	0.1	0.5	µA	-40°C	VDD = 2.0V (**Sleep** mode)
		0.1	0.5	µA	+25°C	
		0.2	1.9	µA	+85°C	
	PIC18LF2X31/4X31	0.1	0.5	µA	-40°C	VDD = 3.0V (**Sleep** mode)
		0.1	0.5	µA	+25°C	
		0.3	1.9	µA	+85°C	
	All devices	0.1	2.0	µA	-40°C	VDD = 5.0V (**Sleep** mode)
		0.1	2.0	µA	+25°C	
		0.4	6.5	µA	+85°C	
		5	33	µA	+125°C	

Legend: Shading of rows is to assist in readability of the table.

Note 1: The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or VSS, and all features that add delta current disabled (such as WDT, Timer1 Oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.
The test conditions for all IDD measurements in active operation mode are:
OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;
MCLR = VDD; WDT enabled/disabled as specified.

3: For RC oscillator configurations, current through REXT is not included. The current through the resistor can be estimated by the formula: Ir = VDD/2REXT (mA) with REXT in kΩ.

4: Standard, low-cost 32 kHz crystals have an operating temperature range of -10°C to +70°C. Extended temperature crystals are available at a much higher cost.

TABLE 26-2: LOW-VOLTAGE DETECT CHARACTERISTICS (CONTINUED)

PIC18LF2331/2431/4331/4431 (Industrial)				**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial				
PIC18F2331/2431/4331/4431 (Industrial, Extended)				**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial -40°C ≤ TA ≤ +125°C for extended				
Param No.	**Symbol**	**Characteristic**		**Min**	**Typ†**	**Max**	**Units**	**Conditions**
D420B	VLVD	LVD Voltage on VDD Transition High-to-Low		Industrial Low Voltage (-40°C to -10°C)				
		PIC18LF2X31/4X31	LVDL<3:0> = 0000	N/A	N/A	N/A	V	Reserved
			LVDL<3:0> = 0001	N/A	N/A	N/A	V	Reserved
			LVDL<3:0> = 0010	1.99	2.26	2.53	V	
			LVDL<3:0> = 0011	2.16	2.45	2.75	V	
			LVDL<3:0> = 0100	2.25	2.55	2.86	V	
			LVDL<3:0> = 0101	2.43	2.77	3.10	V	
			LVDL<3:0> = 0110	2.53	2.87	3.21	V	
			LVDL<3:0> = 0111	2.70	3.07	3.43	V	
			LVDL<3:0> = 1000	2.96	3.36	3.77	V	
			LVDL<3:0> = 1001	3.14	3.57	4.00	V	
			LVDL<3:0> = 1010	3.23	3.67	4.11	V	
			LVDL<3:0> = 1011	3.41	3.87	4.34	V	
			LVDL<3:0> = 1100	3.58	4.07	4.56	V	
			LVDL<3:0> = 1101	3.76	4.28	4.79	V	
			LVDL<3:0> = 1110	4.04	4.60	5.15	V	
D420C	VLVD	LVD Voltage on VDD Transition High-to-Low		Industrial (-10°C to +85°C)				
		PIC18F2X31/4X31	LVDL<3:0> = 1101	3.93	4.28	4.62	V	
			LVDL<3:0> = 1110	4.23	4.60	4.96	V	
D420D	VLVD	LVD Voltage on VDD Transition High-to-Low		Industrial (-40°C to -10°C)				
		PIC18F2X31/4X31	LVDL<3:0> = 1101	3.76	4.28	4.79	V	Reserved
			LVDL<3:0> = 1110	4.04	4.60	5.15	V	
D420E	VLVD	LVD Voltage on VDD Transition High-to-Low		Extended (-10°C to +85°C)				
		PIC18F2X31/4X31	LVDL<3:0> = 1101	3.94	4.28	4.62	V	
			LVDL<3:0> = 1110	4.23	4.60	4.96	V	
D420F	VLVD	LVD Voltage on VDD Transition High-to-Low		Extended (-40°C to -10°C, +85°C to +125°C)				
		PIC18F2X31/4X31	LVDL<3:0> = 1101	3.77	4.28	4.79	V	Reserved
			LVDL<3:0> = 1110	4.05	4.60	5.15	V	

Legend: Shading of rows is to assist in readability of the table.

† Production tested at TAMB = 25°C. Specifications over temperature limits ensured by characterization.