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### Applications of "<u>Embedded -</u> <u>Microcontrollers</u>"

#### Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	25MHz
Connectivity	I <sup>2</sup> C, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, HLVD, POR, PWM, WDT
Number of I/O	54
Program Memory Size	128KB (64K x 16)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	3.8K x 8
Voltage - Supply (Vcc/Vdd)	4.2V ~ 5.5V
Data Converters	A/D 12x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	64-TQFP
Supplier Device Package	64-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18f6722t-e-pt

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Register	Applicable Devices		Applicable Devices Bro		Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset, RESET Instruction, Stack Resets	Wake-up via WDT or Interrupt
PORTA <sup>(5)</sup>	6X27	6X22	8X27	8X22	xx0x 0000 <b>(5)</b>	uu0u 0000 <b>(5)</b>	uuuu uuuu <b>(5)</b>
SPBRGH1	6X27	6X22	8X27	8X22	0000 0000	0000 0000	uuuu uuuu
BAUDCON1	6X27	6X22	8X27	8X22	01-0 0-00	01-0 0-00	uu-u u-uu
SPBRGH2	6X27	6X22	8X27	8X22	0000 0000	0000 0000	uuuu uuuu
BAUDCON2	6X27	6X22	8X27	8X22	01-0 0-00	01-0 0-00	uu-u u-uu
ECCP1DEL	6X27	6X22	8X27	8X22	0000 0000	0000 0000	uuuu uuuu
TMR4	6X27	6X22	8X27	8X22	0000 0000	0000 0000	սսսս սսսս
PR4	6X27	6X22	8X27	8X22	1111 1111	սսսս սսսս	սսսս սսսս
T4CON	6X27	6X22	8X27	8X22	-000 0000	-000 0000	-uuu uuuu
CCPR4H	6X27	6X22	8X27	8X22	xxxx xxxx	սսսս սսսս	սսսս սսսս
CCPR4L	6X27	6X22	8X27	8X22	xxxx xxxx	uuuu uuuu	սսսս սսսս
CCP4CON	6X27	6X22	8X27	8X22	00 0000	00 0000	uu uuuu
CCPR5H	6X27	6X22	8X27	8X22	xxxx xxxx	սսսս սսսս	սսսս սսսս
CCPR5L	6X27	6X22	8X27	8X22	xxxx xxxx	uuuu uuuu	սսսս սսսս
CCP5CON	6X27	6X22	8X27	8X22	00 0000	00 0000	uu uuuu
SPBRG2	6X27	6X22	8X27	8X22	0000 0000	0000 0000	սսսս սսսս
RCREG2	6X27	6X22	8X27	8X22	0000 0000	0000 0000	uuuu uuuu
TXREG2	6X27	6X22	8X27	8X22	0000 0000	0000 0000	սսսս սսսս
TXSTA2	6X27	6X22	8X27	8X22	0000 0010	0000 0010	սսսս սսսս
RCSTA2	6X27	6X22	8X27	8X22	x000 0000	0000 000x	uuuu uuuu
ECCP3AS	6X27	6X22	8X27	8X22	0000 0000	0000 0000	uuuu uuuu
ECCP3DEL	6X27	6X22	8X27	8X22	0000 0000	0000 0000	uuuu uuuu
ECCP2AS	6X27	6X22	8X27	8X22	0000 0000	0000 0000	uuuu uuuu
ECCP2DEL	6X27	6X22	8X27	8X22	0000 0000	0000 0000	uuuu uuuu
SSP2BUF	6X27	6X22	8X27	8X22	xxxx xxxx	սսսս սսսս	սսսս սսսս
SSP2ADD	6X27	6X22	8X27	8X22	0000 0000	0000 0000	uuuu uuuu
SSP2STAT	6X27	6X22	8X27	8X22	0000 0000	0000 0000	սսսս սսսս
SSP2CON1	6X27	6X22	8X27	8X22	0000 0000	0000 0000	นนนน นนนน
SSP2CON2	6X27	6X22	8X27	8X22	0000 0000	0000 0000	սսսս սսսս

### TABLE 4-4: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)

**Legend:** u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition. Shaded cells indicate conditions do not apply for the designated device.

Note 1: One or more bits in the INTCONx or PIRx registers will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the PC is loaded with the interrupt vector (0008h or 0018h).

**3:** When the wake-up is due to an interrupt and the GIEL or GIEH bit is set, the TOSU, TOSH and TOSL are updated with the current value of the PC. The STKPTR is modified to point to the next location in the hardware stack.

4: See Table 4-3 for Reset value for specific condition.

5: Bits 6 and 7 of PORTA, LATA and TRISA are enabled, depending on the oscillator mode selected. When not enabled as PORTA pins, they are disabled and read '0'.

### 5.4.3.1 FSR Registers and the INDF Operand

At the core of Indirect Addressing are three sets of registers: FSR0, FSR1 and FSR2. Each represents a pair of 8-bit registers, FSRnH and FSRnL. The four upper bits of the FSRnH register are not used so each FSR pair holds a 12-bit value. This represents a value that can address the entire range of the data memory in a linear fashion. The FSR register pairs, then, serve as pointers to data memory locations.

Indirect Addressing is accomplished with a set of Indirect File Operands, INDF0 through INDF2. These can be thought of as "virtual" registers: they are mapped in the SFR space but are not physically implemented. Reading or writing to a particular INDF register actually accesses its corresponding FSR register pair. A read from INDF1, for example, reads the data at the address indicated by FSR1H:FSR1L. Instructions that use the INDF registers as operands actually use the contents of their corresponding FSR as a pointer to the instruction's target. The INDF operand is just a convenient way of using the pointer.

Because Indirect Addressing uses a full 12-bit address, data RAM banking is not necessary. Thus, the current contents of the BSR and the Access RAM bit have no effect on determining the target address.

### 5.4.3.2 FSR Registers and POSTINC, POSTDEC, PREINC and PLUSW

In addition to the INDF operand, each FSR register pair also has four additional indirect operands. Like INDF, these are "virtual" registers that cannot be indirectly read or written to. Accessing these registers actually accesses the associated FSR register pair, but also performs a specific action on its stored value. They are:

- POSTDEC: accesses the FSR value, then automatically decrements it by 1 afterwards
- POSTINC: accesses the FSR value, then automatically increments it by 1 afterwards
- PREINC: increments the FSR value by 1, then uses it in the operation
- PLUSW: adds the signed value of the W register (range of -127 to 128) to that of the FSR and uses the new value in the operation.

In this context, accessing an INDF register uses the value in the FSR registers without changing them. Similarly, accessing a PLUSW register gives the FSR value offset by the value in the W register; neither value is actually changed in the operation. Accessing the other virtual registers changes the value of the FSR registers.

Operations on the FSRs with POSTDEC, POSTINC and PREINC affect the entire register pair; that is, rollovers of the FSRnL register from FFh to 00h carry over to the FSRnH register. On the other hand, results of these operations do not change the value of any flags in the STATUS register (e.g., Z, N, OV, etc.).





### 6.0 FLASH PROGRAM MEMORY

The Flash program memory is readable, writable and erasable during normal operation over the entire VDD range.

A read from program memory is executed on one byte at a time. A write to program memory is executed on blocks of 64 bytes at a time. Program memory is erased in blocks of 64 bytes at a time. A bulk erase operation may not be issued from user code.

Writing or erasing program memory will cease instruction fetches until the operation is complete. The program memory cannot be accessed during the write or erase, therefore, code cannot execute. An internal programming timer terminates program memory writes and erases.

A value written to program memory does not need to be a valid instruction. Executing a program memory location that forms an invalid instruction results in a NOP.

### 6.1 Table Reads and Table Writes

In order to read and write program memory, there are two operations that allow the processor to move bytes between the program memory space and the data RAM:

- Table Read (TBLRD)
- Table Write (TBLWT)

The program memory space is 16 bits wide, while the data RAM space is 8 bits wide. Table reads and table writes move data between these two memory spaces through an 8-bit register (TABLAT).

Table read operations retrieve data from program memory and place it into the data RAM space. Figure 6-1 shows the operation of a table read with program memory and data RAM.

Table write operations store data from the data memory space into holding registers in program memory. The procedure to write the contents of the holding registers into program memory is detailed in **Section 6.5** "**Writing to Flash Program Memory**". Figure 6-2 shows the operation of a table write with program memory and data RAM.

Table operations work with byte entities. A table block containing data, rather than program instructions, is not required to be word aligned. Therefore, a table block can start and end at any byte address. If a table write is being used to write executable code into program memory, program instructions will need to be word aligned.

FIGURE 6-1: TABLE READ OPERATION



### EXAMPLE 6-3: WRITING TO FLASH PROGRAM MEMORY

	MOVLW MOVWF	D'64' COUNTER	;	number of bytes in erase block
	MOVLW	BUFFER_ADDR_HIGH	;	point to buffer
	MOVWF MOVLW	FSRUH BUFFER ADDR LOW		
	MOVWF	FSROL		
	MOVLW	CODE_ADDR_UPPER	;	Load TBLPTR with the base
	MOVWF MOVIT W	TBLPTRU	;	address of the memory block
	MOVHW	TBLPTRH		
	MOVLW	CODE_ADDR_LOW		
	MOVWF	TBLPTRL		
READ_BLOCK	TBLRD*+	-	;	read into TABLAT, and inc
	MOVF	TABLAT, W	;	get data
	MOVWF	POSTINC0	;	store data
	DECFSZ	COUNTER DEND BLOCK	;	done?
MODIFY WORD	BKA	READ_BLOCK	'	repeat
	MOVLWD	ATA_ADDR_HIGH	;	point to buffer
	MOVWF	FSROH		
	MOVLW	DATA_ADDR_LOW		
	MOVLW	NEW_DATA_LOW	;	update buffer word
	MOVWF	POSTINC0		-
	MOVLW	NEW_DATA_HIGH		
ERASE BLOCK	MOVWE	INDFU		
Lidibi_bioon	MOVLW	CODE_ADDR_UPPER	;	load TBLPTR with the base
	MOVWF	TBLPTRU	;	address of the memory block
	MOVLW	CODE_ADDR_HIGH		
	MOVWF	CODE ADDR LOW		
	MOVWF	TBLPTRL		
	BSF	EECON1, EEPGD	;	point to Flash program memory
	BCF	EECON1, CFGS FFCON1 WREN	;	access Flash program memory
	BSF	EECON1, FREE	;	enable Row Erase operation
	BCF	INTCON, GIE	;	disable interrupts
Domisional	MOVLW	55h		
Seguence	MOVWF	0AAh	'	write 55m
	MOVWF	EECON2	;	write OAAh
	BSF	EECON1, WR	;	start erase (CPU stall)
	BSF TRIPD*-	INTCON, GIE	;	re-enable interrupts
	MOVLW	BUFFER_ADDR_HIGH	;	point to buffer
	MOVWF	FSROH		
	MOVLW	BUFFER_ADDR_LOW		
WRITE BUFFER B	MOVWE	FSRUL		
	MOVLW	D'64'	;	number of bytes in holding register
	MOVWF	COUNTER		
WRITE_BYTE_TO_	HREGS	DOGUTNOO MODO		get low byte of buffer data
	MOVEF	TABLAT	;	present data to table latch
	TBLWT+*	ŧ	;	write data, perform a short write
			;	to internal TBLWT holding register.
	DECFSZ	COUNTER WRITE WORD TO HREGG	;	loop until buffers are full
	DICH	WITTE WOILD IN THE		

### 12.1 Timer0 Operation

Timer0 can operate as either a timer or a counter; the mode is selected with the TOCS bit (T0CON<5>). In Timer mode (T0CS = 0), the module increments on every clock by default unless a different prescaler value is selected (see **Section 12.3 "Prescaler"**). If the TMR0 register is written to, the increment is inhibited for the following two instruction cycles. The user can work around this by writing an adjusted value to the TMR0 register.

The Counter mode is selected by setting the TOCS bit (= 1). In this mode, Timer0 increments either on every rising or falling edge of pin RA4/T0CKI. The incrementing edge is determined by the Timer0 Source Edge Select bit, T0SE (T0CON<4>); clearing this bit selects the rising edge. Restrictions on the external clock input are discussed below.

An external clock source can be used to drive Timer0; however, it must meet certain requirements to ensure that the external clock can be synchronized with the internal phase clock (Tosc). There is a delay between synchronization and the onset of incrementing the timer/counter.

### 12.2 Timer0 Reads and Writes in 16-bit Mode

TMR0H is not the actual high byte of Timer0 in 16-bit mode; it is actually a buffered version of the real high byte of Timer0 which is not directly readable nor writable (refer to Figure 12-2). TMR0H is updated with the contents of the high byte of Timer0 during a read of TMR0L. This provides the ability to read all 16 bits of Timer0 without having to verify that the read of the high and low byte were valid, due to a rollover between successive reads of the high and low byte.

Similarly, a write to the high byte of Timer0 must also take place through the TMR0H Buffer register. The high byte is updated with the contents of TMR0H when a write occurs to TMR0L. This allows all 16 bits of Timer0 to be updated at once.

### FIGURE 12-1: TIMER0 BLOCK DIAGRAM (8-BIT MODE)



### FIGURE 12-2: TIMER0 BLOCK DIAGRAM (16-BIT MODE)



### 13.1 Timer1 Operation

Timer1 can operate in one of these modes:

- Timer
- Synchronous Counter
- Asynchronous Counter

The operating mode is determined by the clock select bit, TMR1CS (T1CON<1>). When TMR1CS is cleared (= 0), Timer1 increments on every internal instruction

cycle (Fosc/4). When the bit is set, Timer1 increments on every rising edge of the Timer1 external clock input or the Timer1 oscillator, if enabled.

When Timer1 is enabled, the RC1/T1OSI and RC0/T10SO/T13CKI pins become inputs. This means the values of TRISC<1:0> are ignored and the pins are read as '0'.



### FIGURE 13-2: TIMER1 BLOCK DIAGRAM (16-BIT READ/WRITE MODE)



### 16.0 TIMER4 MODULE

The Timer4 timer module has the following features:

- 8-bit Timer register (TMR4)
- 8-bit Period register (PR4)
- Readable and writable (both registers)
- Software programmable prescaler (1:1, 1:4, 1:16)
- Software programmable postscaler (1:1 to 1:16)
- Interrupt on TMR4 match of PR4

Timer4 has a control register shown in Register 16-1. Timer4 can be shut off by clearing control bit, TMR4ON (T4CON<2>), to minimize power consumption. The prescaler and postscaler selection of Timer4 are also controlled by this register. Figure 16-1 is a simplified block diagram of the Timer4 module.

### 16.1 Timer4 Operation

Timer4 can be used as the PWM time base for the PWM mode of the CCP modules. The TMR4 register is readable and writable and is cleared on any device Reset. The input clock (Fosc/4) has a prescale option of 1:1, 1:4 or 1:16, selected by control bits T4CKPS<1:0> (T4CON<1:0>). The match output of TMR4 goes through a 4-bit postscaler (which gives a 1:1 to 1:16 scaling inclusive) to generate a TMR4 interrupt, latched in flag bit, TMR4IF (PIR3<3>).

The prescaler and postscaler counters are cleared when any of the following occurs:

- · a write to the TMR4 register
- a write to the T4CON register
- any device Reset (Power-on Reset, MCLR Reset, Watchdog Timer Reset or Brown-out Reset)

TMR4 is not cleared when T4CON is written.

### REGISTER 16-1: T4CON: TIMER4 CONTROL REGISTER

U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
—	T4OUTPS3	T4OUTPS2	T4OUTPS1	T4OUTPS0	TMR4ON	T4CKPS1	T4CKPS0
bit 7							bit 0

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read	as '0'
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

bit 7	Unimplemented: Read as '0'
bit 6-3	T4OUTPS<3:0>: Timer4 Output Postscale Select bits
	0000 = 1:1 Postscale
	0001 = 1:2 Postscale
	•
	•
	•

1111 = 1:16 Postscale

bit 2 **TMR4ON**: Timer4 On bit

1 = Timer4 is on 0 = Timer4 is off

- bit 1-0 **T4CKPS<1:0>**: Timer4 Clock Prescale Select bits
  - 00 =Prescaler is 1 01 =Prescaler is 4
  - 1x = Prescaler is 16

## PIC18F8722 FAMILY

ECCP Mode	CCP2CON Configuration	RB3	RC1	RE7	RE2	RE1	RE0		
	F	PIC18F6527/66	22/6627/6722	Devices, CCP	2MX = 1:				
Compatible CCP	00xx 11xx	RB3/INT3	ECCP2	RE7	RE2	RE1	RE0		
Dual PWM	10xx 11xx	RB3/INT3	P2A	RE7	P2B	RE1	RE0		
Quad PWM	x1xx 11xx	RB3/INT3	P2A	RE7	P2B	P2C	P2D		
	I	PIC18F6527/60	622/6627/6722	Devices CCP2	2 <b>MX</b> = 0:				
Compatible CCP	00xx 11xx	RB3/INT3	RC1/T1OSI	ECCP2	RE2	RE1	RE0		
Dual PWM	10xx 11xx	RB3/INT3	RC1/T1OSI	P2A	P2B	RE1	RE0		
Quad PWM	x1xx 11xx	RB3/INT3	RC1/T1OSI	P2A	P2B	P2C	P2D		
	PIC18F8527/8622/8627/8722 Devices, CCP2MX = 1, Microcontroller mode:								
Compatible CCP	00xx 11xx	RB3/INT3	ECCP2	RE7	RE2	RE1	RE0		
Dual PWM	10xx 11xx	RB3/INT3	P2A	RE7	P2B	RE1	RE0		
Quad PWM	x1xx 11xx	RB3/INT3	P2A	RE7	P2B	P2C	P2D		
	PIC18F8527/	/8622/8627/87	22 Devices, CO	CP2MX = 0, Mi	crocontroller	mode:			
Compatible CCP	00xx 11xx	RB3/INT3	RC1/T1OSI	ECCP2	RE2	RE1	RE0		
Dual PWM	10xx 11xx	RB3/INT3	RC1/T10SI	P2A	P2B	RE1	RE0		
Quad PWM	x1xx 11xx	RB3/INT3	RC1/T1OSI	P2A	P2B	P2C	P2D		
PI	C18F8527/8622/8	8627/8722 Dev	vices, CCP2M)	( = 1, all other	Program Men	nory modes:			
Compatible CCP	00xx 11xx	RB3/INT3	ECCP2	AD15 <sup>(1)</sup>	AD10 <sup>(1)</sup>	AD9 <sup>(1)</sup>	AD8 <sup>(1)</sup>		
Dual PWM	10xx 11xx	RB3/INT3	P2A	AD15 <sup>(1)</sup>	AD10/P2B <sup>(1)</sup>	AD9 <sup>(1)</sup>	AD8 <sup>(1)</sup>		
Quad PWM	x1xx 11xx	RB3/INT3	P2A	AD15 <sup>(1)</sup>	AD10/P2B <sup>(1)</sup>	AD9/P2C <sup>(1)</sup>	P2D/AD8 <sup>(1)</sup>		
PI	C18F8527/8622/	8627/8722 Dev	vices, CCP2M)	( = 0, all other	Program Men	nory modes:			
Compatible CCP	00xx 11xx	ECCP2	RC1/T10SI	AD15 <sup>(1)</sup>	AD10 <sup>(1)</sup>	AD9 <sup>(1)</sup>	AD8 <sup>(1)</sup>		
Dual PWM	10xx 11xx	P2A	RC1/T10SI	AD15 <sup>(1)</sup>	AD10/P2B <sup>(1)</sup>	AD9 <sup>(1)</sup>	AD8 <sup>(1)</sup>		
Quad PWM	x1xx 11xx	P2A	RC1/T1OSI	AD15 <sup>(1)</sup>	AD10/P2B <sup>(1)</sup>	AD9/P2C <sup>(1)</sup>	P2D/AD8 <sup>(1)</sup>		

### TABLE 18-2:PIN CONFIGURATIONS FOR ECCP2

**Legend:** x = Don't care. Shaded cells indicate pin assignments not used by ECCP2 in a given mode.

Note 1: The EMB address bus width will determine whether the pin will perform an EMB or port/peripheral function.

### 18.4.6 PROGRAMMABLE DEAD-BAND DELAY

In half-bridge applications where all power switches are modulated at the PWM frequency at all times, the power switches normally require more time to turn off than to turn on. If both the upper and lower power switches are switched at the same time (one turned on and the other turned off), both switches may be on for a short period of time until one switch completely turns off. During this brief interval, a very high current (*shoot-through current*) may flow through both power switches, shorting the bridge supply. To avoid this potentially destructive shoot-through current from flowing during switching, turning on either of the power switches is normally delayed to allow the other switch to completely turn off.

In the Half-Bridge Output mode, a digitally programmable dead-band delay is available to avoid shoot-through current from destroying the bridge power switches. The delay occurs at the signal transition from the non-active state to the active state. See Figure 18-4 for illustration. The lower seven bits of the ECCP1DEL register (Register 18-2) set the delay period in terms of microcontroller instruction cycles (TcY or 4 Tosc).

### 18.4.7 ENHANCED PWM AUTO-SHUTDOWN

When the ECCP is programmed for any of the Enhanced PWM modes, the active output pins may be configured for auto-shutdown. Auto-shutdown immediately places the Enhanced PWM output pins into a defined shutdown state when a shutdown event occurs.

A shutdown event can be caused by either of the two comparator modules or the FLT0 pin (or any combination of these three sources). The comparators may be used to monitor a voltage input proportional to a current being monitored in the bridge circuit. If the voltage exceeds a threshold, the comparator switches state and triggers a shutdown. Alternatively, a digital signal on the FLT0 pin can also trigger a shutdown. The auto-shutdown feature can be disabled by not selecting any auto-shutdown sources. The auto-shutdown sources to be used are selected using the ECCP1AS<2:0> bits (ECCP1AS<6:4>).

When a shutdown occurs, the output pins are asynchronously placed in their shutdown states, specified by the PSS1AC<1:0> and PSS1BD<1:0> bits (ECCP1AS<3:0>). Each pin pair (P1A/P1C and P1B/P1D) may be set to drive high, drive low or be tri-stated (not driving). The ECCP1ASE bit (ECCP1AS<7>) is also set to hold the Enhanced PWM outputs in their shutdown states.

The ECCP1ASE bit is set by hardware when a shutdown event occurs. If automatic restarts are not enabled, the ECCP1ASE bit is cleared by firmware when the cause of the shutdown clears. If automatic restarts are enabled, the ECCP1ASE bit is automatically cleared when the cause of the auto-shutdown has cleared.

If the ECCP1ASE bit is set when a PWM period begins, the PWM outputs remain in their shutdown state for that entire PWM period. When the ECCP1ASE bit is cleared, the PWM outputs will return to normal operation at the beginning of the next PWM period.

**Note:** Writing to the ECCP1ASE bit is disabled while a shutdown condition is active.

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
PxRSEN	PxDC6	PxDC5	PxDC4	PxDC3	PxDC2	PxDC1	PxDC0
bit 7							bit 0

### REGISTER 18-2: ECCPxDEL: ENHANCED PWM DEAD-BAND DELAY REGISTER

Legend:							
R = Readable bit	W = Writable bit	U = Unimplemented bit,	U = Unimplemented bit, read as '0'				
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown				

bit 7	PxRSEN: PWM Restart Enable bit
	1 = Upon auto-shutdown, the ECCPxASE bit clears automatically once the shutdown event goes away; the PWM restarts automatically
	0 = Upon auto-shutdown, the ECCPxASE bit must be cleared in software to restart the PWM
bit 6-0	PxDC<6:0>: PWM Delay Count bits
	Delay time, in number of Fosc/4 (4 * Tosc) cycles, between the scheduled and actual time for a PWM signal to transition to active.

### 19.4.6 MASTER MODE

Master mode is enabled by setting and clearing the appropriate SSPM bits in SSPxCON1 and by setting the SSPEN bit. In Master mode, the SCLx and SDAx lines are manipulated by the MSSP hardware if the TRIS bits are set.

Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the  $I^2C$  bus may be taken when the P bit is set, or the bus is Idle, with both the S and P bits clear.

In Firmware Controlled Master mode, user code conducts all  ${\rm I}^2{\rm C}$  bus operations based on Start and Stop bit conditions.

Once Master mode is enabled, the user has six options.

- 1. Assert a Start condition on SDAx and SCLx.
- 2. Assert a Repeated Start condition on SDAx and SCLx.
- 3. Write to the SSPxBUF register initiating transmission of data/address.
- 4. Configure the  $I^2C$  port to receive data.
- 5. Generate an Acknowledge condition at the end of a received byte of data.
- 6. Generate a Stop condition on SDAx and SCLx.

Note: The MSSP module, when configured in I<sup>2</sup>C Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSPxBUF register to initiate transmission before the Start condition is complete. In this case, the SSPxBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPxBUF did not occur.

The following events will cause the SSP Interrupt Flag bit, SSPxIF, to be set (and SSP interrupt, if enabled):

- Start condition
- · Stop condition
- Data transfer byte transmitted/received
- Acknowledge transmit
- Repeated Start



# PIC18F8722 FAMILY

### 19.4.17.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

- a) SDAx or SCLx are sampled low at the beginning of the Start condition (Figure 19-26).
- b) SCLx is sampled low before SDAx is asserted low (Figure 19-27).

During a Start condition, both the SDAx and the SCLx pins are monitored.

If the SDAx pin is already low, or the SCLx pin is already low, then all of the following occur:

- the Start condition is aborted,
- the BCLxIF flag is set and
- the MSSP module is reset to its inactive state (Figure 19-26).

The Start condition begins with the SDAx and SCLx pins deasserted. When the SDAx pin is sampled high, the Baud Rate Generator is loaded from SSPxADD<6:0> and counts down to '0'. If the SCLx pin is sampled low while SDAx is high, a bus collision occurs because it is assumed that another master is attempting to drive a data '1' during the Start condition.

If the SDAx pin is sampled low during this count, the BRG is reset and the SDAx line is asserted early (Figure 19-28). If, however, a '1' is sampled on the SDAx pin, the SDAx pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to '0'. If the SCLx pin is sampled as '0' during this time, a bus collision does not occur. At the end of the BRG count, the SCLx pin is asserted low.

Note: The reason that bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDAx before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.





### EXAMPLE 20-1: CALCULATING BAUD RATE ERROR

For a device with FOSC of 16 MHz, desired baud rate of 9600, Asynchronous mode, 8-bit BRG:								
Desired Baud Rate =	Desired Baud Rate = $FOSC/(64 ([SPBRGHx:SPBRGx] + 1))$							
Solving for SPBRGHx	:SPBRGx:							
X =	((Fosc/Desired Baud Rate)/64) – 1							
=	((16000000/9600)/64) - 1							
=	[25.042] = 25							
Calculated Baud Rate=	= 16000000/(64 (25 + 1))							
=	9615							
Error =	(Calculated Baud Rate - Desired Baud Rate)/Desired Baud Rate							
=	(9615 - 9600)/9600 = 0.16%							

### TABLE 20-2: REGISTERS ASSOCIATED WITH BAUD RATE GENERATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset Values on page
TXSTAx	CSRC	TX9	TXEN	SYNC	SENDB	BRGH	TRMT	TX9D	59
RCSTAx	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	59
BAUDCONx	ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	61
SPBRGHx	EUSARTx Baud Rate Generator Register High Byte								59
SPBRGx	EUSARTx	Baud Rate	Generator I	Register Lo	w Byte				59

**Legend:** — = unimplemented, read as '0'. Shaded cells are not used by the BRG.

The value in the ADRESH:ADRESL registers is not modified for a Power-on Reset. The ADRESH:ADRESL registers will contain unknown data after a Power-on Reset.

After the A/D module has been configured as desired, the selected channel must be acquired before the conversion is started. The analog input channels must have their corresponding TRIS bits selected as an input. To determine acquisition time, see **Section 21.1 "A/D Acquisition Requirements"**. After this acquisition time has elapsed, the A/D conversion can be started. An acquisition time can be programmed to occur between setting the GO/DONE bit and the actual start of the conversion.

The following steps should be followed to perform an A/D conversion:

- 1. Configure the A/D module:
  - Configure analog pins, voltage reference and digital I/O (ADCON1)
  - Select A/D input channel (ADCON0)
  - Select A/D acquisition time (ADCON2)
  - Select A/D conversion clock (ADCON2)
  - Turn on A/D module (ADCON0)
- 2. Configure A/D interrupt (if desired):
  - Clear ADIF bit
  - Set ADIE bit
  - Set GIE bit
- 3. Wait the required acquisition time (if required).
- 4. Start conversion:
  - Set GO/DONE bit (ADCON0 register)

- 5. Wait for A/D conversion to complete, by either:
  - Polling for the GO/DONE bit to be cleared OR
  - Waiting for the A/D interrupt
- 6. Read A/D Result registers (ADRESH:ADRESL); clear bit ADIF, if required.
- 7. For next conversion, go to step 1 or step 2, as required. The A/D conversion time per bit is defined as TAD. A minimum wait of 2 TAD is required before the next acquisition starts.

### FIGURE 21-2: A/D TRANSFER FUNCTION





### FIGURE 21-3: ANALOG INPUT MODEL

### 21.4 Operation in Power-Managed Modes

The selection of the automatic acquisition time and A/D conversion clock is determined in part by the clock source and frequency while in a power-managed mode.

If the A/D is expected to operate while the device is in a power-managed mode, the ACQT<2:0> and ADCS<2:0> bits in ADCON2 should be updated in accordance with the clock source to be used in that mode. After entering the mode, an A/D acquisition or conversion may be started. Once started, the device should continue to be clocked by the same clock source until the conversion has been completed.

If desired, the device may be placed into the corresponding Idle mode during the conversion. If the device clock frequency is less than 1 MHz, the A/D RC clock source should be selected.

Operation in the Sleep mode requires the A/D FRC clock to be selected. If bits ACQT<2:0> are set to '000' and a conversion is started, the conversion will be delayed one instruction cycle to allow execution of the SLEEP instruction and entry to Sleep mode. The IDLEN bit (OSCCON<7>) must have already been cleared prior to starting the conversion.

### 21.5 Configuring Analog Port Pins

The ADCON1, TRISA, TRISF and TRISH registers all configure the A/D port pins. The port pins needed as analog inputs must have their corresponding TRIS bits set (input). If the TRIS bit is cleared (output), the digital output level (VOH or VOL) will be converted.

The A/D operation is independent of the state of the CHS<3:0> bits and the TRIS bits.

- Note 1: When reading the Port register, all pins configured as analog input channels will read as cleared (a low level). Pins configured as digital inputs will convert as analog inputs. Analog levels on a digitally configured input will be accurately converted.
  - 2: Analog levels on any pin defined as a digital input may cause the digital input buffer to consume current out of the device's specification limits.

### REGISTER 25-9: CONFIG6L: CONFIGURATION REGISTER 6 LOW (BYTE ADDRESS 30000Ah)

					``		,	
R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	R/C-1	
WRT7 <sup>(</sup>	<sup>1)</sup> WRT6 <sup>(1)</sup>	WRT5 <sup>(2)</sup>	WRT4 <sup>(2)</sup>	WRT3 <sup>(3)</sup>	WRT2	WRT1	WRT0	
bit 7							bit (	
Legend:								
R = Reada	able bit	W = Writable	bit	U = Unimplemented bit, read as '0'				
-n = Value at POR		'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unknown		
bit 7	WRT7: Write	Protection bit <sup>(1</sup>	)					
	1 = Block 7 (	01C000-01FFF	Fh) not write-p	protected				
1.11.0	0 = BIOCK / (		rn) write-prote	ected				
DIT 6	1 Plack 6 (		() () has the second se	rataatad				
	$\perp = \text{Block 6}$ ( 0 = Block 6 (	01BFFF-01800	0h) write-prote	acted				
bit 5	WRT5: Write	Protection bit <sup>(2</sup>	2)					
	1 = Block 5 (	014000-017FFI	-h) not write-p	orotected				
	0 = Block 5 (	014000-017FFI	h) write-prote	ected				
bit 4	WRT4: Write	Protection bit <sup>(2</sup>	2)					
	1 = Block 4 (	010000-013FFI	<sup>-</sup> h) not write-p	orotected				
	0 = Block 4 (	010000-013FFI	h) write-prote	ected				
bit 3	WRT3: Write	Protection bit <sup>(C</sup>	5)					
	1 = Block 3(	00C000-00FFF	Fh) not write-p	protected				
<b>h</b> it 0	0 = BIOCK 3 (		rn) write-prote	ected				
DIT Z	1 = Plock 2		Eh) not write r	rotootod				
	1 = Block  2 ( 0 = Block 2 (	008000-00BFF	Fh) write-prote	ected				
bit 1	WRT1: Write	Protection bit	,					
	1 = Block 1 (	004000-007FFI	-h) not write-p	orotected				
	0 = Block 1 (	004000-007FFI	h) write-prote	ected				
bit 0	WRT0: Write	Protection bit						
	1 = Block  0 (	000800, 00100	0 or 002000 <sup>(4)</sup>	-003FFFh) no	t write-protecte	d		
	0 = Block  0 (	000800, 00100	0 or 002000 <sup>(4)</sup>	-003FFFh) wri	te-protected			
Note 1:	Unimplemented ir	n PIC18F6527/6	622/6627/852	27/8622/8627	devices; mainta	in this bit set.		
2:	Unimplemented in	PIC18F6527/6	622/8527/862	22 devices; ma	intain this bit s	et.		
3:	Unimplemented in	n PIC18F6527/8	527 devices;	maintain this b	it set.			

4: Boot block size is determined by the BBSIZ<1:0> bits in CONFIG4L.

### 25.2 Watchdog Timer (WDT)

For the PIC18F8722 family of devices, the WDT is driven by the INTRC source. When the WDT is enabled, the clock source is also enabled. The nominal WDT period is 4 ms and has the same stability as the INTRC oscillator.

The 4 ms period of the WDT is multiplied by a 16-bit postscaler. Any output of the WDT postscaler is selected by a multiplexor, controlled by bits in Configuration Register 2H. Available periods range from 4 ms to 131.072 seconds (2.18 minutes). The WDT and postscaler are cleared when any of the following events occur: a SLEEP or CLRWDT instruction is executed, the IRCF bits (OSCCON<6:4>) are changed or a clock failure has occurred.

- Note 1: The CLRWDT and SLEEP instructions clear the WDT and postscaler counts when executed.
  - 2: Changing the setting of the IRCF bits (OSCCON<6:4>) clears the WDT and postscaler counts.
  - **3:** When a CLRWDT instruction is executed, the postscaler count will be cleared.

### 25.2.1 CONTROL REGISTER

Register 25-15 shows the WDTCON register. This is a readable and writable register which contains a control bit that allows software to override the WDT enable Configuration bit, but only if the Configuration bit has disabled the WDT.



### FIGURE 25-1: WDT BLOCK DIAGRAM

### 26.0 INSTRUCTION SET SUMMARY

The PIC18F8722 family of devices incorporates the standard set of 75 PIC18 core instructions, as well as an extended set of 8 new instructions for the optimization of code that is recursive or that utilizes a software stack. The extended set is discussed later in this section.

### 26.1 Standard Instruction Set

The standard PIC18 instruction set adds many enhancements to the previous PIC<sup>®</sup> MCU instruction sets, while maintaining an easy migration from these PIC MCU instruction sets. Most instructions are a single program memory word (16 bits), but there are four instructions that require two program memory locations.

Each single-word instruction is a 16-bit word divided into an opcode, which specifies the instruction type and one or more operands, which further specify the operation of the instruction.

The instruction set is highly orthogonal and is grouped into four basic categories:

- Byte-oriented operations
- Bit-oriented operations
- Literal operations
- Control operations

The PIC18 instruction set summary in Table 26-2 lists **byte-oriented**, **bit-oriented**, **literal** and **control** operations. Table 26-1 shows the opcode field descriptions.

Most byte-oriented instructions have three operands:

- 1. The file register (specified by 'f')
- 2. The destination of the result (specified by 'd')
- 3. The accessed memory (specified by 'a')

The file register designator 'f' specifies which file register is to be used by the instruction. The destination designator 'd' specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the WREG register. If 'd' is one, the result is placed in the file register specified in the instruction.

All bit-oriented instructions have three operands:

- 1. The file register (specified by 'f')
- 2. The bit in the file register (specified by 'b')
- 3. The accessed memory (specified by 'a')

The bit field designator 'b' selects the number of the bit affected by the operation, while the file register designator 'f' represents the number of the file in which the bit is located. The **literal** instructions may use some of the following operands:

- A literal value to be loaded into a file register (specified by 'k')
- The desired FSR register to load the literal value into (specified by 'f')
- No operand required (specified by '—')

The **control** instructions may use some of the following operands:

- A program memory address (specified by 'n')
- The mode of the CALL or RETURN instructions (specified by 's')
- The mode of the table read and table write instructions (specified by 'm')
- No operand required (specified by '—')

All instructions are a single word, except for four double-word instructions. These instructions were made double-word to contain the required information in 32 bits. In the second word, the 4 MSbs are 1's. If this second word is executed as an instruction (by itself), it will execute as a NOP.

All single-word instructions are executed in a single instruction cycle, unless a conditional test is true or the program counter is changed as a result of the instruction. In these cases, the execution takes two instruction cycles with the additional instruction cycle(s) executed as a NOP.

The double word instructions execute in two instruction cycles.

One instruction cycle consists of four oscillator periods. Thus, for an oscillator frequency of 4 MHz, the normal instruction execution time is 1  $\mu$ s. If a conditional test is true, or the program counter is changed as a result of an instruction, the instruction execution time is 2  $\mu$ s. Two-word branch instructions (if true) would take 3  $\mu$ s.

Figure 26-1 shows the general formats that the instructions can have. All examples use the convention 'nnh' to represent a hexadecimal number.

The Instruction Set Summary, shown in Table 26-2, lists the standard instructions recognized by the Microchip MPASM<sup>™</sup> Assembler.

Section 26.1.1 "Standard Instruction Set" provides a description of each instruction.

## PIC18F8722 FAMILY

RCALL Relative Call								
Synta	x:	RCALL n						
Opera	ands:	-1024 ≤ n ≤	$-1024 \le n \le 1023$					
Opera	ation:	$(PC) + 2 \rightarrow (PC) + 2 + 2$	$(PC) + 2 \rightarrow TOS,$ (PC) + 2 + 2n $\rightarrow$ PC					
Status	s Affected:	None	None					
Enco	ding:	1101	1nnn	nnn	n	nnnn		
Descr	iption:	Subroutine from the cu address (Pt stack. Ther number '2n have increr instruction, PC + 2 + 2r two-cycle ir	Subroutine call with a jump up to 1K from the current location. First, return address (PC + 2) is pushed onto the stack. Then, add the 2's complement number '2n' to the PC. Since the PC will have incremented to fetch the next instruction, the new address will be PC + 2 + 2n. This instruction is a two-cycle instruction.					
Words:		1	1					
Cycles:		2	2					
Q Cycle Activity:								
-	Q1	Q2	Q	3		Q4		
	Decode	Read literal 'n' PUSH PC to stack	Proce Data	ess a	Wri	te to PC		
No		No	No			No		

operation

operation

Example: HERE RCALL Jump

operation

Before Instruction

operation

PC = Address (HERE) After Instruction PC = TOS = Address (Jump) Address (HERE + 2)

RESET		Reset							
Synta	ax:	RESET	RESET						
Oper	ands:	None	None						
Oper	ation:	Reset all re affected by	Reset all registers and flags that are affected by a MCLR Reset.						
Statu	s Affected:	All	All						
Enco	ding:	0000	0000	1111	1111				
Description:		This instruction provides a way to execute a MCLR Reset in software.							
Words:		1	1						
Cycles:		1	1						
QC	ycle Activity:								
	Q1	Q2	Q3		Q4				
	Decode	Start	No		No				
		reset	operat	ion op	peration				

Example:

After Instruction

Registers =	Reset Value
Flags* =	Reset Value

RESET

### 28.2 DC Characteristics: Power-Down and Supply Current PIC18F6X27/6X22/8X27/8X22 (Industrial, Extended) PIC18LF6X27/6X22/8X27/8X22 (Industrial) (Continued)

PIC18LF6X27/6X22/8X27/8X22 (Industrial)		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial					
PIC18F6X27/6X22/8X27/8X22 (Industrial, Extended)							
Param No.	Device	Тур	Max	Units	Conditions		
	Module Differential Currents (2	NWDT, Albor, Allvd, Aloscb, Alad)					
D022	Watchdog Timer	1.5	2.2	μΑ	-40°C		
$(\Delta I W D T)$		1.6	2.2	μA	+25°C	VDD = 2.0V	
		1.7	2.3	μΑ	+85°C		
		2.3	3.5	μΑ	-40°C		
		2.4	3.5	μA	+25°C	VDD = 3.0V	
		3.4	3.5	μΑ	+85°C		
		4.8	7.5	μΑ	-40°C		
		6.0	7.5	μΑ	+25°C	Vpp = 5.0V	
		6.1	7.8	μA	+85°C	VDD = 5.0V	
		8	10	μA	+125°C		
D022A	Brown-out Reset <sup>(4)</sup>	4.2	50	μΑ	-40°C to +85°C	VDD = 3.0V	
(∆IBOR)		48	55	μΑ	-40°C to +85°C	Vdd = 5.0V	
		66	55	μΑ	-40°C to +125°C		
		0	2.4	μΑ	-40°C to +85°C		Sleep mode,
		0	6.0	μA	-40°C to +125°C		BOREN<1:0> = 10
D022B	High/Low-Voltage Detect <sup>(4)</sup>	2.7	38	μA	-40°C to +85°C	VDD = 2.0V	
(ΔILVD)		30	40	μA	-40°C to +85°C	VDD = 3.0V	
		35	45	μA	-40°C to +85°C	VDD = 5.0V	
		36	45	μA	-40°C to +125°C		
D025	Timer1 Oscillator	4.5	9	μA	-40°C(3)	Vdd = 2.0V	32 kHz on Timer1
(AIOSCB)		.9	1.7	μΑ	-10°C		
		.9	2.2	μΑ	+25°C		
		.9	2.2	μΑ	+85°C		
		4.8	10	μΑ	-40°C(3)	Vdd = 3.0V	
		1	1.8	μΑ	-10°C		32 kHz on Timer1
		1	2.3	μA	+25°C		
		1	2.3	μA	+85°C		
		6	11	μA	-40°C(3)		
		1.6	6	μA	-10°C	VDD = 5.0V	32 kHz on Timer1
		1.6	6	μΑ	+25°C		
		1.6	6	μΑ	+85°C		

Legend: Shading of rows is to assist in readability of the table.

**Note 1:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD or VSS and all features that add delta current disabled (such as WDT, Timer1 Oscillator, BOR, etc.).

2: The supply current is mainly a function of operating voltage, frequency and mode. Other factors, such as I/O pin loading and switching rate, oscillator type and circuit, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD OR VSS;

MCLR = VDD; WDT enabled/disabled as specified.

**3:** When operation below -10°C is expected, use T1OSC High-Power mode, where LPT1OSC (CONFIG3H<2>) = 0. When operation will always be above -10°C, then the low-power Timer1 oscillator may be selected.

4: BOR and HLVD enable internal band gap reference. With both modules enabled, current consumption will be less than the sum of both specifications.



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