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Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

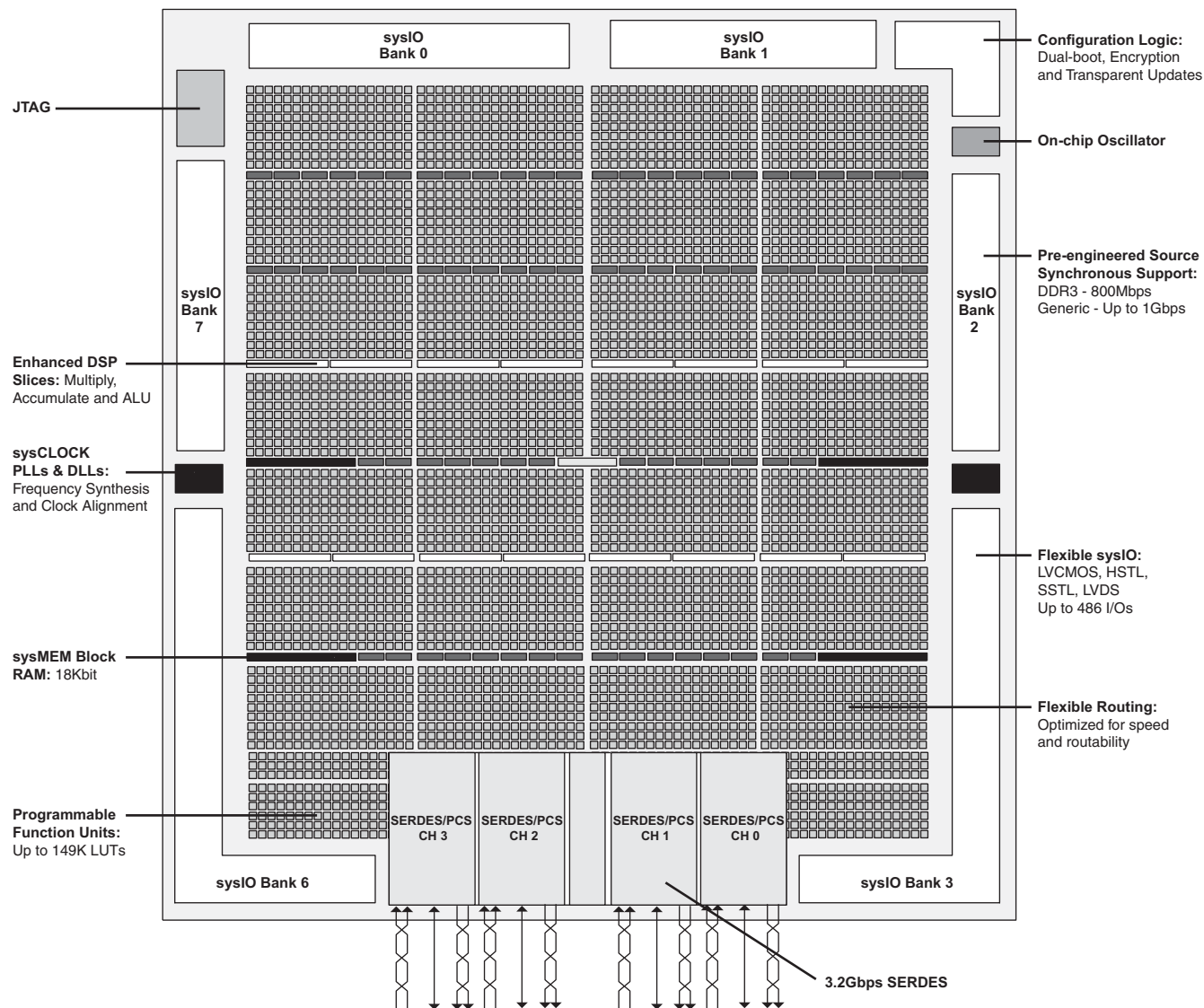
Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details

Product Status	Active
Number of LABs/CLBs	2125
Number of Logic Elements/Cells	17000
Total RAM Bits	716800
Number of I/O	116
Number of Gates	-
Voltage - Supply	1.14V ~ 1.26V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 125°C (TJ)
Package / Case	328-LFBGA, CSBGA
Supplier Device Package	328-CSBGA (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/lattice-semiconductor/lae3-17ea-6mg328e

Figure 2-1. Simplified Block Diagram, LA-LatticeECP3-35 Device (Top Level)



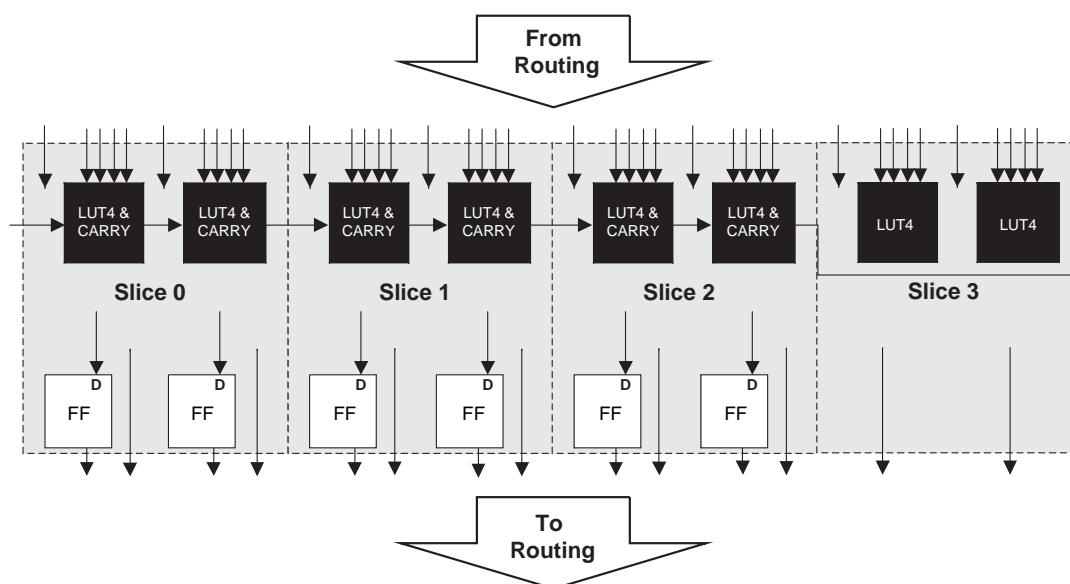
Note: There is no Bank 4 or Bank 5 in LatticeECP3 devices.

PFU Blocks

The core of the LA-LatticeECP3 device consists of PFU blocks, which are provided in two forms, the PFU and PFF. The PFUs can be programmed to perform Logic, Arithmetic, Distributed RAM and Distributed ROM functions. PFF blocks can be programmed to perform Logic, Arithmetic and ROM functions. Except where necessary, the remainder of this data sheet will use the term PFU to refer to both PFU and PFF blocks.

Each PFU block consists of four interconnected slices numbered 0-3 as shown in Figure 2-2. Each slice contains two LUTs. All the interconnections to and from PFU blocks are from routing. There are 50 inputs and 23 outputs associated with each PFU block.

Figure 2-2. PFU Diagram



Slice

Slice 0 through Slice 2 contain two LUT4s feeding two registers, whereas Slice 3 contains two LUT4s only. For PFUs, Slice 0 through Slice 2 can be configured as distributed memory, a capability not available in the PFF. Table 2-1 shows the capability of the slices in both PFF and PFU blocks along with the operation modes they enable. In addition, each PFU contains logic that allows the LUTs to be combined to perform functions such as LUT5, LUT6, LUT7 and LUT8. There is control logic to perform set/reset functions (programmable as synchronous/asynchronous), clock select, chip-select and wider RAM/ROM functions.

Table 2-1. Resources and Modes Available per Slice

Slice	PFU BLock		PFF Block	
	Resources	Modes	Resources	Modes
Slice 0	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 1	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 2	2 LUT4s and 2 Registers	Logic, Ripple, RAM, ROM	2 LUT4s and 2 Registers	Logic, Ripple, ROM
Slice 3	2 LUT4s	Logic, ROM	2 LUT4s	Logic, ROM

Figure 2-3 shows an overview of the internal logic of the slice. The registers in the slice can be configured for positive/negative and edge triggered or level sensitive clocks.

Slices 0, 1 and 2 have 14 input signals: 13 signals from routing and one from the carry-chain (from the adjacent slice or PFU). There are seven outputs: six to routing and one to carry-chain (to the adjacent PFU). Slice 3 has 10 input signals from routing and four signals to routing. Table 2-2 lists the signals associated with Slice 0 to Slice 2.

PLL/DLL Cascading

LA-LatticeECP3 devices have been designed to allow certain combinations of PLL and DLL cascading. The allowable combinations are:

- PLL to PLL supported
- PLL to DLL supported

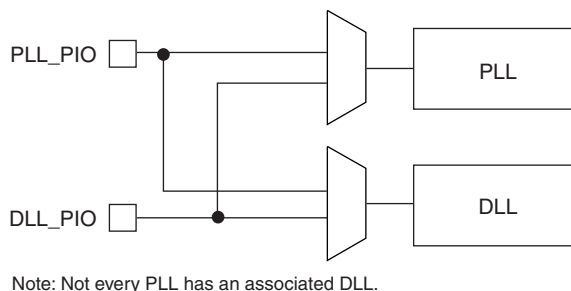
The DLLs in the LA-LatticeECP3 are used to shift the clock in relation to the data for source synchronous inputs. PLLs are used for frequency synthesis and clock generation for source synchronous interfaces. Cascading PLL and DLL blocks allows applications to utilize the unique benefits of both DLLs and PLLs.

For further information about the DLL, please see the list of technical documentation at the end of this data sheet.

PLL/DLL PIO Input Pin Connections

All LA-LatticeECP3 devices contains two DLLs and up to ten PLLs, arranged in quadrants. If a PLL and a DLL are next to each other, they share input pins as shown in the Figure 2-7.

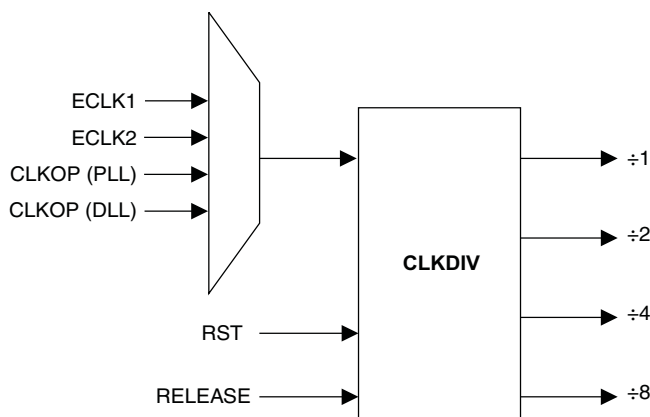
Figure 2-7. Sharing of PIO Pins by PLLs and DLLs in LA-LatticeECP3 Devices



Clock Dividers

LA-LatticeECP3 devices have two clock dividers, one on the left side and one on the right side of the device. These are intended to generate a slower-speed system clock from a high-speed edge clock. The block operates in a $\div 2$, $\div 4$ or $\div 8$ mode and maintains a known phase relationship between the divided down clock and the high-speed clock based on the release of its reset signal. The clock dividers can be fed from selected PLL/DLL outputs, the Slave Delay lines, routing or from an external clock input. The clock divider outputs serve as primary clock sources and feed into the clock distribution network. The Reset (RST) control signal resets input and asynchronously forces all outputs to low. The RELEASE signal releases outputs synchronously to the input clock. For further information on clock dividers, please see TN1178, [LatticeECP3 sysCLOCK PLL/DLL Design and Usage Guide](#). Figure 2-8 shows the clock divider connections.

Figure 2-8. Clock Divider Connections



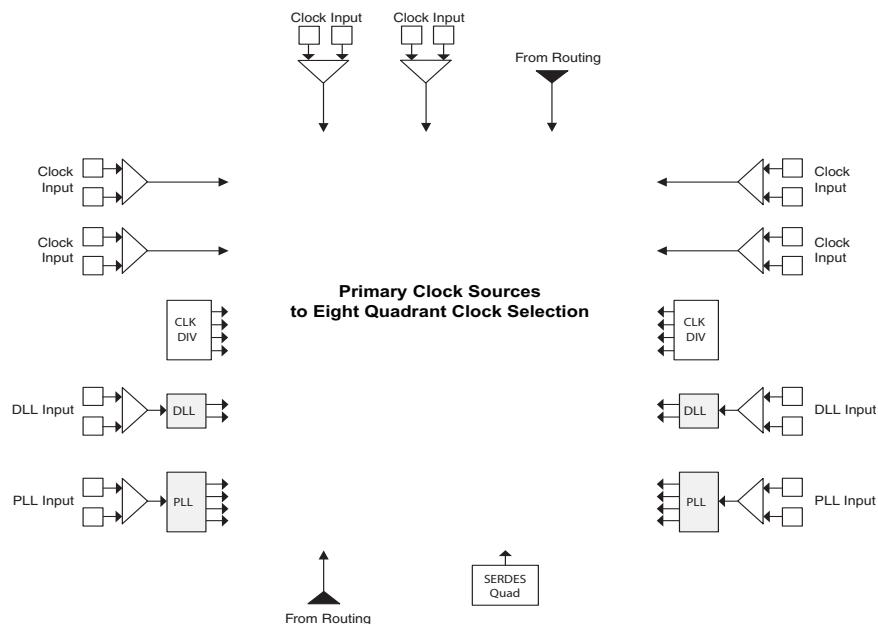
Clock Distribution Network

LA-LatticeECP3 devices have eight quadrant-based primary clocks and eight secondary clock/control sources. Two high performance edge clocks are available on the top, left, and right edges of the device to support high speed interfaces. These clock sources are selected from external I/Os, the sysCLOCK PLLs, DLLs or routing. These clock sources are fed throughout the chip via a clock distribution system.

Primary Clock Sources

LA-LatticeECP3 devices derive clocks from six primary source types: PLL outputs, DLL outputs, CLKDIV outputs, dedicated clock inputs, routing and SERDES Quad. LA-LatticeECP3 devices have two to four sysCLOCK PLLs and two DLLs, located on the left and right sides of the device. There are six dedicated clock inputs: two on the top side, two on the left side and two on the right side of the device. Figures 2-9 and 2-10 show the primary clock sources for LA-LatticeECP3 devices.

Figure 2-9. Primary Clock Sources for LA-LatticeECP3-17

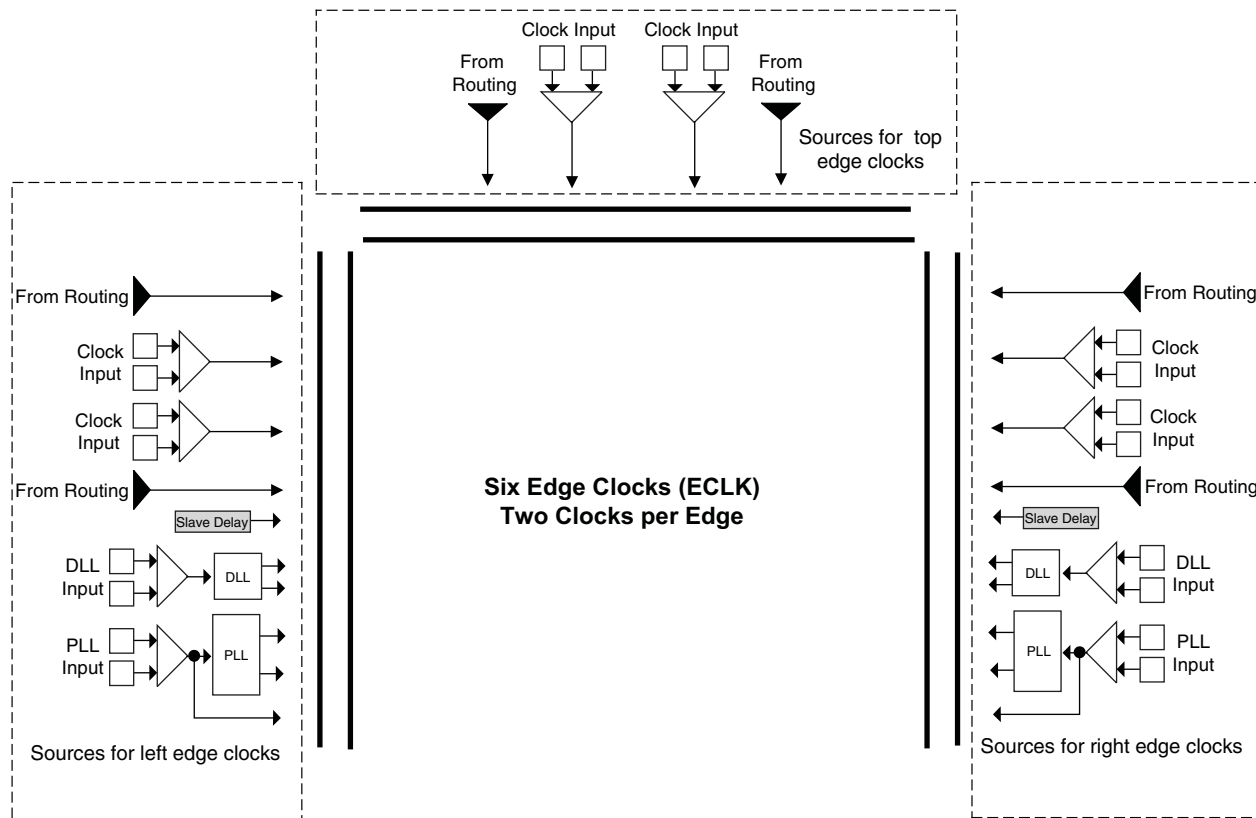


Note: Clock inputs can be configured in differential or single-ended mode.

Edge Clock Sources

Edge clock resources can be driven from a variety of sources at the same edge. Edge clock resources can be driven from adjacent edge clock PIOs, primary clock PIOs, PLLs, DLLs, Slave Delay and clock dividers as shown in Figure 2-18.

Figure 2-18. Edge Clock Sources



Notes:

1. Clock inputs can be configured in differential or single ended mode.
2. The two DLLs can also drive the two top edge clocks.
3. The top left and top right PLL can also drive the two top edge clocks.

Edge Clock Routing

LA-LatticeECP3 devices have a number of high-speed edge clocks that are intended for use with the PIOs in the implementation of high-speed interfaces. There are six edge clocks per device: two edge clocks on each of the top, left, and right edges. Different PLL and DLL outputs are routed to the two muxes on the left and right sides of the device. In addition, the CLKINDEL signal (generated from the DLL Slave Delay Line block) is routed to all the edge clock muxes on the left and right sides of the device. Figure 2-19 shows the selection muxes for these clocks.

Figure 2-19. Sources of Edge Clock (Left and Right Edges)

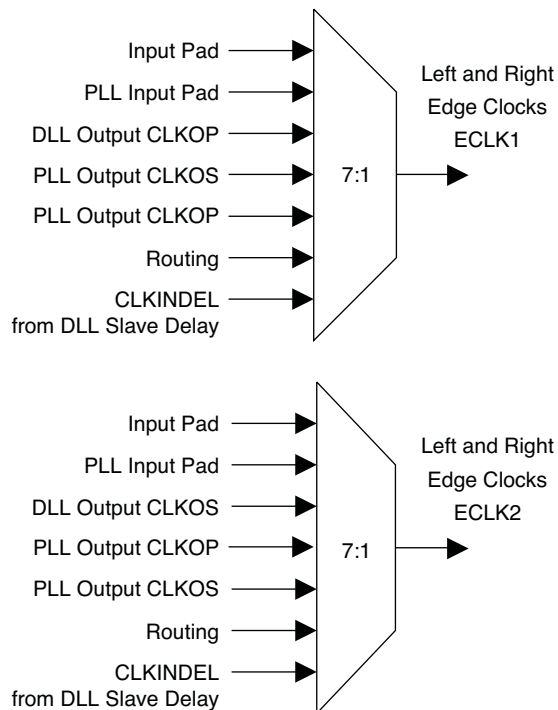
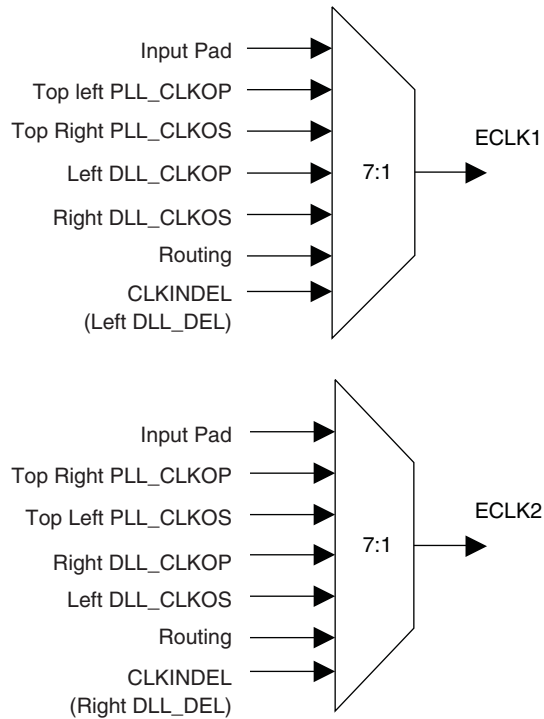


Figure 2-20. Sources of Edge Clock (Top Edge)



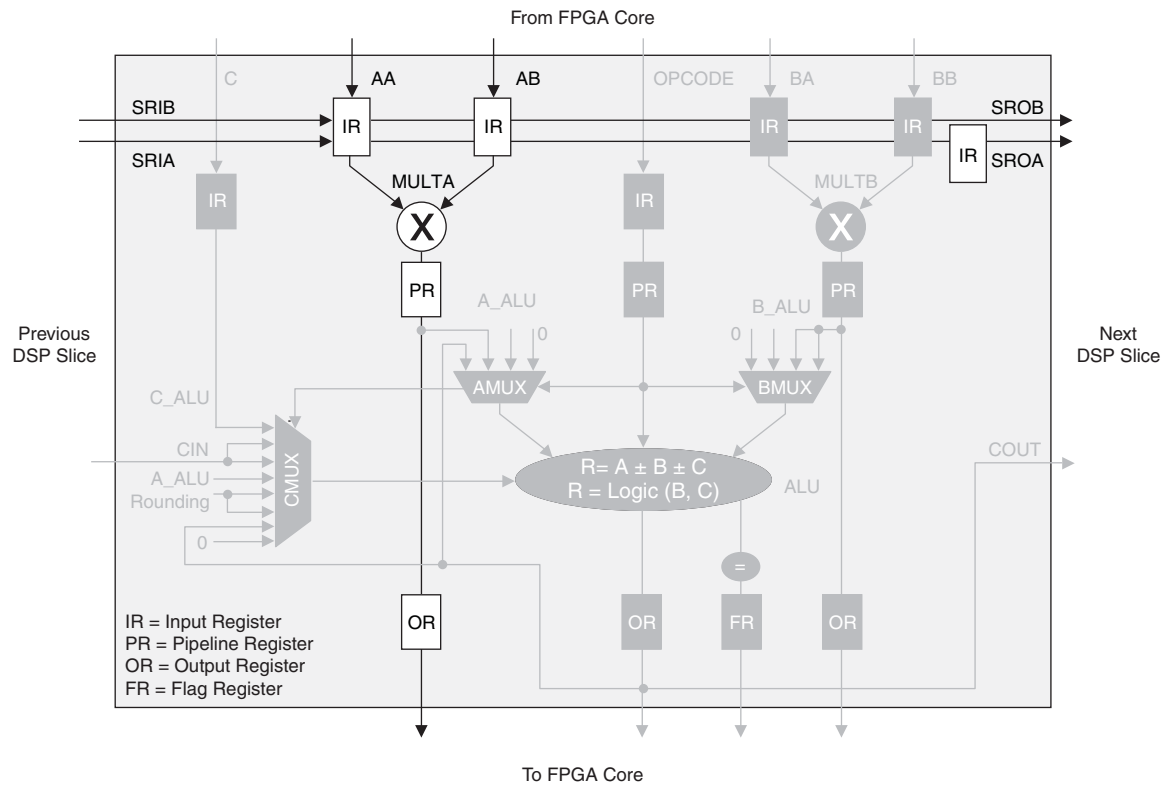
The edge clocks have low injection delay and low skew. They are used to clock the I/O registers and thus are ideal for creating I/O interfaces with a single clock signal and a wide data bus. They are also used for DDR Memory or Generic DDR interfaces.

For further information, please refer to TN1182, [LatticeECP3 sysDSP Usage Guide](#).

MULT DSP Element

This multiplier element implements a multiply with no addition or accumulator nodes. The two operands, AA and AB, are multiplied and the result is available at the output. The user can enable the input/output and pipeline registers. Figure 2-25 shows the MULT sysDSP element.

Figure 2-25. MULT sysDSP Element



MULTADDSUB DSP Element

In this case, the operands AA and AB are multiplied and the result is added/subtracted with the result of the multiplier operation of operands BA and BB. The user can enable the input, output and pipeline registers. Figure 2-28 shows the MULTADDSUB sysDSP element.

Figure 2-28. MULTADDSUB

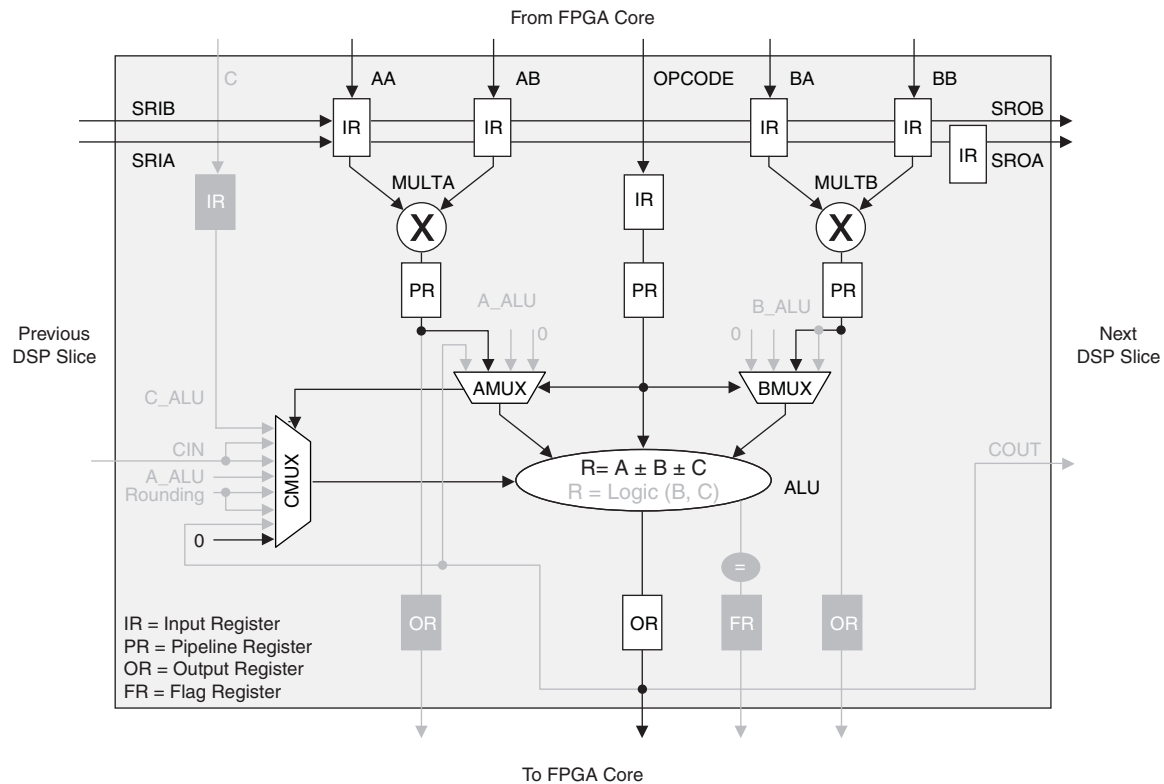


Figure 2-39. SERDES/PCS Quads (LA-LatticeECP3-35)

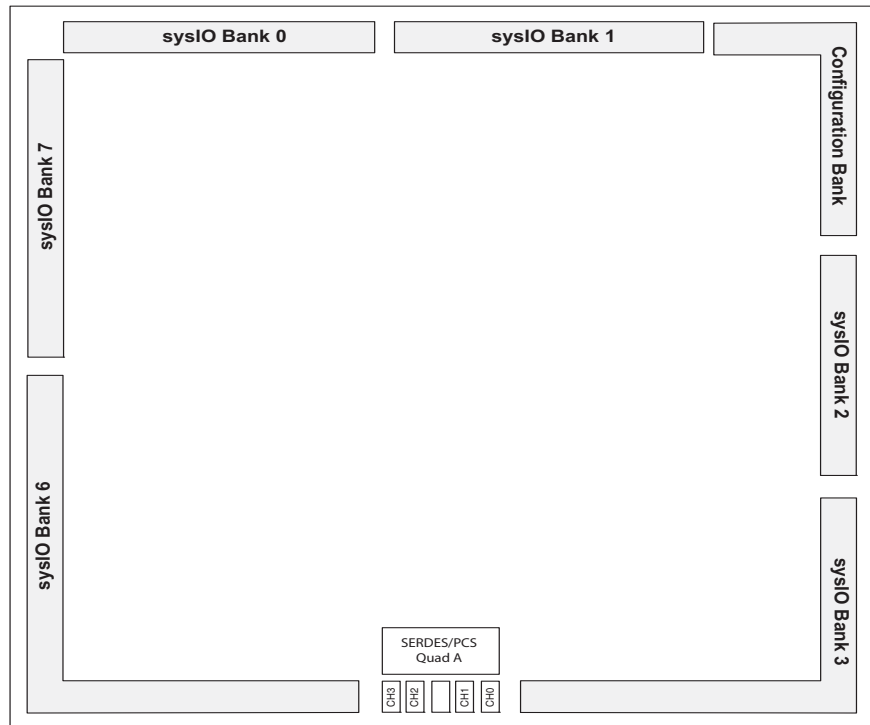


Table 2-13. LA-LatticeECP3 SERDES Standard Support

Standard	Data Rate (Mbps)	Number of General/Link Width	Encoding Style
PCI Express 1.1	2500	x1, x2, x4	8b10b
Gigabit Ethernet	1250, 2500	x1	8b10b
SGMII	1250	x1	8b10b
XAUI	3125	x4	8b10b
Serial RapidIO Type I, Serial RapidIO Type II, Serial RapidIO Type III	1250, 2500, 3125	x1, x4	8b10b
CPRI-1, CPRI-2, CPRI-3, CPRI-4	614.4, 1228.8, 2457.6, 3072.0	x1	8b10b
SD-SDI (259M, 344M)	143 ¹ , 177 ¹ , 270, 360, 540	x1	NRZI/Scrambled
HD-SDI (292M)	1483.5, 1485	x1	NRZI/Scrambled
3G-SDI (424M)	2967, 2970	x1	NRZI/Scrambled
SONET-STS-3 ²	155.52	x1	N/A
SONET-STS-12 ²	622.08	x1	N/A
SONET-STS-48 ²	2488	x1	N/A

1. For slower rates, the SERDES are bypassed and CML signals are directly connected to the FPGA routing.

2. The SONET protocol is supported in 8-bit SERDES mode. See TN1176 [Lattice ECP3 SERDES/PCS Usage Guide](#) for more information.

The Diamond design tool supports all modes of the PCS. Most modes are dedicated to applications associated with a specific industry standard data protocol. Other more general purpose modes allow users to define their own operation. With these tools, the user can define the mode for each quad in a design.

Popular standards such as 10Gb Ethernet, x4 PCI Express and 4x Serial RapidIO can be implemented using IP (available through Lattice), a single quad (Four SERDES channels and PCS) and some additional logic from the core.

The LA-LatticeECP3 family also supports a wide range of primary and secondary protocols. Within the same quad, the LA-LatticeECP3 family can support mixed protocols with semi-independent clocking as long as the required clock frequencies are integer x1, x2, or x11 multiples of each other. Table 2-15 lists the allowable combination of primary and secondary protocol combinations.

Flexible Quad SERDES Architecture

The LA-LatticeECP3 family SERDES architecture is a quad-based architecture. For most SERDES settings and standards, the whole quad (consisting of four SERDES) is treated as a unit. This helps in silicon area savings, better utilization and overall lower cost.

However, for some specific standards, the LA-LatticeECP3 quad architecture provides flexibility; more than one standard can be supported within the same quad.

Table 2-15 shows the standards can be mixed and matched within the same quad. In general, the SERDES standards whose nominal data rates are either the same or a defined subset of each other, can be supported within the same quad. In Table 2-15, the Primary Protocol column refers to the standard that determines the reference clock and PLL settings. The Secondary Protocol column shows the other standard that can be supported within the same quad.

Furthermore, Table 2-15 also implies that more than two standards in the same quad can be supported, as long as they conform to the data rate and reference clock requirements. For example, a quad may contain PCI Express 1.1, SGMII, Serial RapidIO Type I and Serial RapidIO Type II, all in the same quad.

Table 2-15. LA-LatticeECP3 Primary and Secondary Protocol Support

Primary Protocol	Secondary Protocol
PCI Express 1.1	SGMII
PCI Express 1.1	Gigabit Ethernet
PCI Express 1.1	Serial RapidIO Type I
PCI Express 1.1	Serial RapidIO Type II
Serial RapidIO Type I	SGMII
Serial RapidIO Type I	Gigabit Ethernet
Serial RapidIO Type II	SGMII
Serial RapidIO Type II	Gigabit Ethernet
Serial RapidIO Type II	Serial RapidIO Type I
CPRI-3	CPRI-2 and CPRI-1
3G-SDI	HD-SDI and SD-SDI

There are some restrictions to be aware of when using spread spectrum. When a quad shares a PCI Express x1 channel with a non-PCI Express channel, ensure that the reference clock for the quad is compatible with all protocols within the quad. For example, a PCI Express spread spectrum reference clock is not compatible with most Gigabit Ethernet applications because of tight CTC ppm requirements.

While the LA-LatticeECP3 architecture will allow the mixing of a PCI Express channel and a Gigabit Ethernet, Serial RapidIO or SGMII channel within the same quad, using a PCI Express spread spectrum clocking as the

transmit reference clock will cause a violation of the Gigabit Ethernet, Serial RapidIO and SGMII transmit jitter specifications.

For further information on SERDES, please see TN1176, [LatticeECP3 SERDES/PCS Usage Guide](#).

IEEE 1149.1-Compliant Boundary Scan Testability

All LA-LatticeECP3 devices have boundary scan cells that are accessed through an IEEE 1149.1 compliant Test Access Port (TAP). This allows functional testing of the circuit board on which the device is mounted through a serial scan path that can access all critical logic nodes. Internal registers are linked internally, allowing test data to be shifted in and loaded directly onto test nodes, or test data to be captured and shifted out for verification. The test access port consists of dedicated I/Os: TDI, TDO, TCK and TMS. The test access port has its own supply voltage V_{CCJ} and can operate with LVCMOS3.3, 2.5, 1.8, 1.5 and 1.2 standards.

For more information, please see TN1169, [LatticeECP3 sysCONFIG Usage Guide](#).

Device Configuration

All LA-LatticeECP3 devices contain two ports that can be used for device configuration. The Test Access Port (TAP), which supports bit-wide configuration, and the sysCONFIG port, support dual-byte, byte and serial configuration. The TAP supports both the IEEE Standard 1149.1 Boundary Scan specification and the IEEE Standard 1532 In-System Configuration specification. The sysCONFIG port includes seven I/Os used as dedicated pins with the remaining pins used as dual-use pins. See TN1169, [LatticeECP3 sysCONFIG Usage Guide](#) for more information about using the dual-use pins as general purpose I/Os.

There are various ways to configure a LA-LatticeECP3 device:

1. JTAG
2. Standard Serial Peripheral Interface (SPI and SPI_{MEM} modes) - interface to boot PROM memory
3. System microprocessor to drive a x8 CPU port (PCM mode)
4. System microprocessor to drive a serial slave SPI port (SSPI mode)
5. Generic byte wide flash with a MachXO™ device, providing control and addressing

On power-up, the FPGA SRAM is ready to be configured using the selected sysCONFIG port. Once a configuration port is selected, it will remain active throughout that configuration cycle. The IEEE 1149.1 port can be activated any time after power-up by sending the appropriate command through the TAP port.

LA-LatticeECP3 devices also support the Slave SPI Interface. In this mode, the FPGA behaves like a SPI Flash device (slave mode) with the SPI port of the FPGA to perform read-write operations.

Enhanced Configuration Options

LA-LatticeECP3 devices have enhanced configuration features such as: decryption support, TransFR™ I/O and dual-boot image support.

1. TransFR (Transparent Field Reconfiguration)

TransFR I/O (TFR) is a unique Lattice technology that allows users to update their logic in the field without interrupting system operation using a single ispVM command. TransFR I/O allows I/O states to be frozen during device configuration. This allows the device to be field updated with a minimum of system disruption and downtime. See TN1087, [Minimizing System Interruption During Configuration Using TransFR Technology](#) for details.

2. Dual-Boot Image Support

Dual-boot images are supported for applications requiring reliable remote updates of configuration data for the

sysI/O Single-Ended DC Electrical Characteristics

Input/Output Standard	V_{IL}		V_{IH}		V_{OL} Max. (V)	V_{OH} Min. (V)	I_{OL}^1 (mA)	I_{OH}^1 (mA)
	Min. (V)	Max. (V)	Min. (V)	Max. (V)				
LVCMOS33	-0.3	0.8	2.0	3.6	0.4	$V_{CCIO} - 0.4$	20, 16, 12, 8, 4	-20, -16, -12, -8, -4
					0.2	$V_{CCIO} - 0.2$	0.1	-0.1
LVCMOS25	-0.3	0.7	1.7	3.6	0.4	$V_{CCIO} - 0.4$	20, 16, 12, 8, 4	-20, -16, -12, -8, -4
					0.2	$V_{CCIO} - 0.2$	0.1	-0.1
LVCMOS18	-0.3	$0.35 V_{CCIO}$	$0.65 V_{CCIO}$	3.6	0.4	$V_{CCIO} - 0.4$	16, 12, 8, 4	-16, -12, -8, -4
					0.2	$V_{CCIO} - 0.2$	0.1	-0.1
LVCMOS15	-0.3	$0.35 V_{CCIO}$	$0.65 V_{CCIO}$	3.6	0.4	$V_{CCIO} - 0.4$	8, 4	-8, -4
					0.2	$V_{CCIO} - 0.2$	0.1	-0.1
LVCMOS12	-0.3	$0.35 V_{CC}$	$0.65 V_{CC}$	3.6	0.4	$V_{CCIO} - 0.4$	6, 2	-6, -2
					0.2	$V_{CCIO} - 0.2$	0.1	-0.1
LVTTL33	-0.3	0.8	2.0	3.6	0.4	$V_{CCIO} - 0.4$	20, 16, 12, 8, 4	-20, -16, -12, -8, -4
					0.2	$V_{CCIO} - 0.2$	0.1	-0.1
PCI33	-0.3	$0.3 V_{CCIO}$	$0.5 V_{CCIO}$	3.6	$0.1 V_{CCIO}$	$0.9 V_{CCIO}$	1.5	-0.5
SSTL18_I	-0.3	$V_{REF} - 0.125$	$V_{REF} + 0.125$	3.6	0.4	$V_{CCIO} - 0.4$	6.7	-6.7
SSTL18_II (DDR2 Memory)	-0.3	$V_{REF} - 0.125$	$V_{REF} + 0.125$	3.6	0.28	$V_{CCIO} - 0.28$	8	-8
							11	-11
SSTL2_I	-0.3	$V_{REF} - 0.18$	$V_{REF} + 0.18$	3.6	0.54	$V_{CCIO} - 0.62$	7.6	-7.6
							12	-12
SSTL2_II (DDR Memory)	-0.3	$V_{REF} - 0.18$	$V_{REF} + 0.18$	3.6	0.35	$V_{CCIO} - 0.43$	15.2	-15.2
							20	-20
SSTL3_I	-0.3	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	0.7	$V_{CCIO} - 1.1$	8	-8
SSTL3_II	-0.3	$V_{REF} - 0.2$	$V_{REF} + 0.2$	3.6	0.5	$V_{CCIO} - 0.9$	16	-16
SSTL15 (DDR3 Memory)	-0.3	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.3	$V_{CCIO} - 0.3$	7.5	-7.5
						$V_{CCIO} * 0.8$	9	-9
HSTL15_I	-0.3	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.4	$V_{CCIO} - 0.4$	4	-4
							8	-8
HSTL18_I	-0.3	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.4	$V_{CCIO} - 0.4$	8	-8
							12	-12
HSTL18_II	-0.3	$V_{REF} - 0.1$	$V_{REF} + 0.1$	3.6	0.4	$V_{CCIO} - 0.4$	16	-16

1. For electromigration, the average DC current drawn by I/Os between GND connections, or between the last GND in an I/O bank and the end of an I/O bank, as shown in the logic signal connections table shall not exceed $n * 8\text{mA}$, where n is the number of I/Os between bank GND connections or between the last VCCIO and GND in a bank and the end of a bank.

LVPECL33

The LA-LatticeECP3 devices support the differential LVPECL standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The LVPECL input standard is supported by the LVDS differential input buffer. The scheme shown in Figure 3-3 is one possible solution for point-to-point signals.

Figure 3-3. Differential LVPECL33

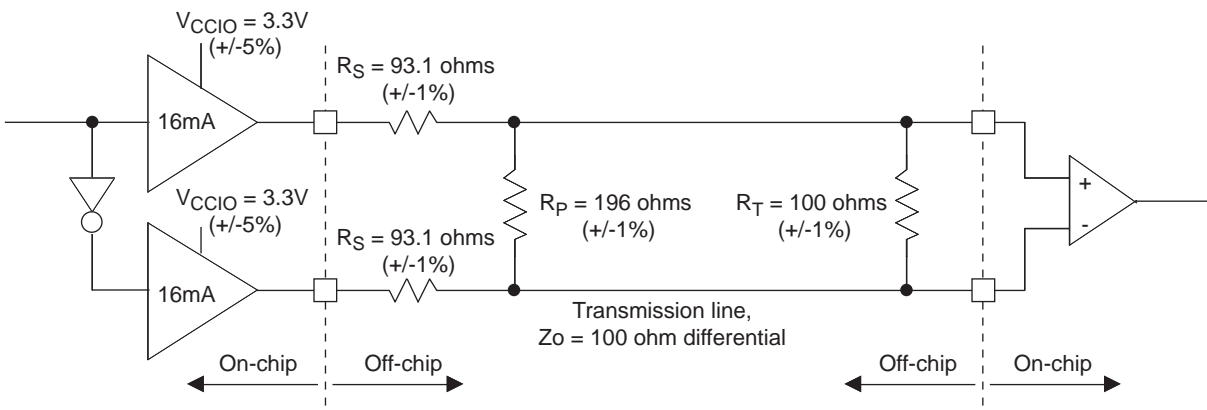


Table 3-5. LVPECL33 DC Conditions¹

Over Recommended Operating Conditions

Parameter	Description	Typical	Units
V_{CCIO}	Output Driver Supply (+/-5%)	3.30	V
Z_{OUT}	Driver Impedance	10	Ω
R_S	Driver Series Resistor (+/-1%)	93	Ω
R_P	Driver Parallel Resistor (+/-1%)	196	Ω
R_T	Receiver Termination (+/-1%)	100	Ω
V_{OH}	Output High Voltage	2.05	V
V_{OL}	Output Low Voltage	1.25	V
V_{OD}	Output Differential Voltage	0.80	V
V_{CM}	Output Common Mode Voltage	1.65	V
Z_{BACK}	Back Impedance	100.5	Ω
I_{DC}	DC Output Current	12.11	mA

1. For input buffer, see LVDS table.

MLVDS25

The LA-LatticeECP3 devices support the differential MLVDS standard. This standard is emulated using complementary LVCMOS outputs in conjunction with a parallel resistor across the driver outputs. The MLVDS input standard is supported by the LVDS differential input buffer. The scheme shown in Figure 3-5 is one possible solution for MLVDS standard implementation. Resistor values in Figure 3-5 are industry standard values for 1% resistors.

Figure 3-5. MLVDS25 (Multipoint Low Voltage Differential Signaling)

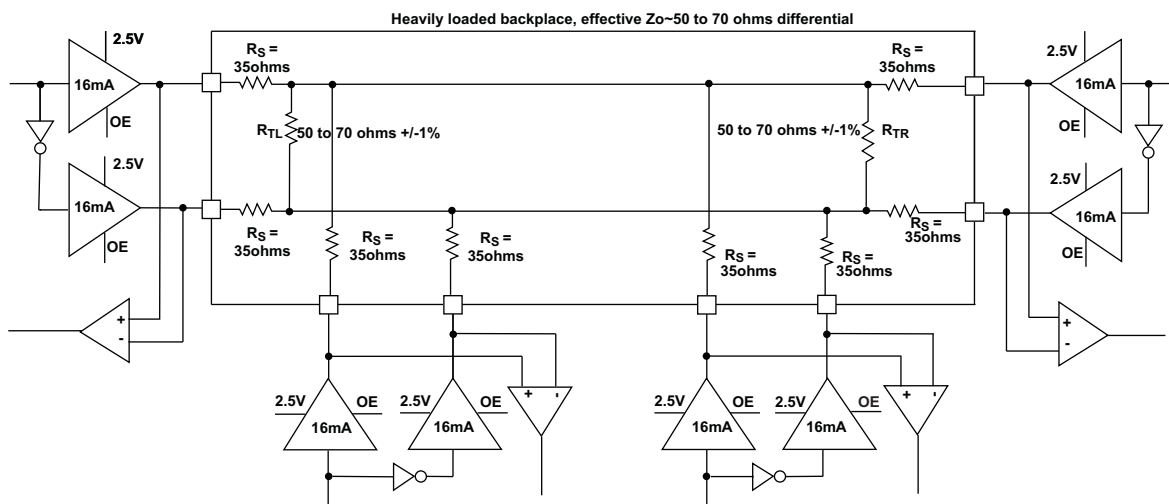


Table 3-7. MLVDS25 DC Conditions¹

Parameter	Description	Typical		Units
		Zo=50Ω	Zo=70Ω	
V _{CCIO}	Output Driver Supply (+/-5%)	2.50	2.50	V
Z _{OUT}	Driver Impedance	10.00	10.00	Ω
R _S	Driver Series Resistor (+/-1%)	35.00	35.00	Ω
R _{TL}	Driver Parallel Resistor (+/-1%)	50.00	70.00	Ω
R _{TR}	Receiver Termination (+/-1%)	50.00	70.00	Ω
V _{OH}	Output High Voltage	1.52	1.60	V
V _{OL}	Output Low Voltage	0.98	0.90	V
V _{OD}	Output Differential Voltage	0.54	0.70	V
V _{CM}	Output Common Mode Voltage	1.25	1.25	V
I _{DC}	DC Output Current	21.74	20.00	mA

1. For input buffer, see LVDS table.

Table 3-13. Periodic Receiver Jitter Tolerance Specification

Description	Frequency	Condition	Min.	Typ.	Max.	Units
Periodic	2.97 Gbps	600 mV differential eye	—	—	0.24	UI, p-p
Periodic	2.5 Gbps	600 mV differential eye	—	—	0.22	UI, p-p
Periodic	1.485 Gbps	600 mV differential eye	—	—	0.24	UI, p-p
Periodic	622 Mbps	600 mV differential eye	—	—	0.15	UI, p-p
Periodic	150 Mbps	600 mV differential eye	—	—	0.5	UI, p-p

Note: Values are measured with PRBS 2^7-1 , all channels operating, FPGA Logic active, I/Os around SERDES pins quiet, voltages are nominal, room temperature.

SERDES External Reference Clock

The external reference clock selection and its interface are a critical part of system applications for this product. Table 3-14 specifies reference clock requirements, over the full range of operating conditions.

Table 3-14. External Reference Clock Specification (refclkp/refclkn)

Symbol	Description	Min.	Typ.	Max.	Units
F_{REF}	Frequency range	15	—	320	MHz
$F_{REF-PPM}$	Frequency tolerance ¹	-1000	—	1000	ppm
$V_{REF-IN-SE}$	Input swing, single-ended clock ²	200	—	V_{CCA}	mV, p-p
$V_{REF-IN-DIFF}$	Input swing, differential clock	200	—	$2 \cdot V_{CCA}$	mV, p-p differential
V_{REF-IN}	Input levels	0	—	$V_{CCA} + 0.3$	V
D_{REF}	Duty cycle ³	40	—	60	%
T_{REF-R}	Rise time (20% to 80%)	200	500	1000	ps
T_{REF-F}	Fall time (80% to 20%)	200	500	1000	ps
$Z_{REF-IN-TERM-DIFF}$	Differential input termination	-20%	100/2K	+20%	Ohms
$C_{REF-IN-CAP}$	Input capacitance	—	—	7	pF

1. Depending on the application, the PLL_LOL_SET and CDR_LOL_SET control registers may be adjusted for other tolerance values as described in TN1176, [LatticeECP3 SERDES/PCS Usage Guide](#).
2. The signal swing for a single-ended input clock must be as large as the p-p differential swing of a differential input clock to get the same gain at the input receiver. Lower swings for the clock may be possible, but will tend to increase jitter.
3. Measured at 50% amplitude.

Figure 3-13. SERDES External Reference Clock Waveforms

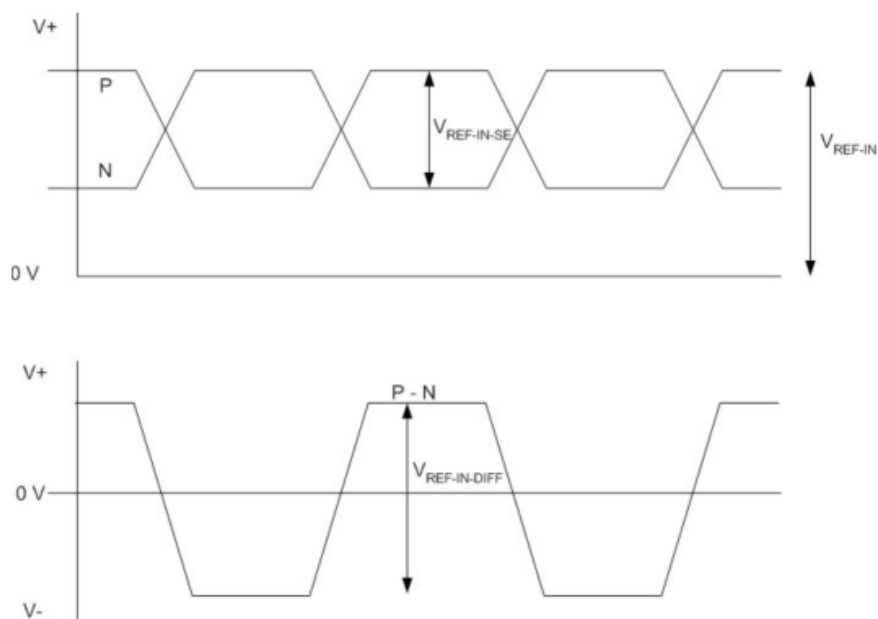
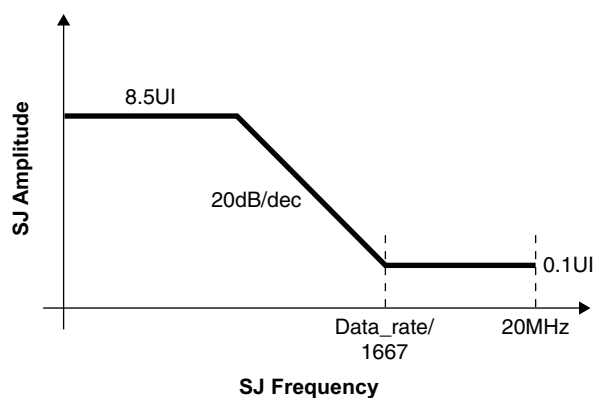


Figure 3-18. XAUI Sinusoidal Jitter Tolerance Mask



Note: The sinusoidal jitter tolerance is measured with at least 0.37UIpp of Deterministic jitter (Dj) and the sum of Dj and Rj (random jitter) is at least 0.55UIpp. Therefore, the sum of Dj, Rj and Sj (sinusoidal jitter) is at least 0.65UIpp (Dj = 0.37, Rj = 0.18, Sj = 0.1).

Serial Rapid I/O Type 2/CPRI LV E.24 Electrical and Timing Characteristics

AC and DC Characteristics

Table 3-17. Transmit

Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
T_{RF}^1	Differential rise/fall time	20%-80%	—	80	—	ps
$Z_{TX_DIFF_DC}$	Differential impedance		80	100	120	Ohms
$J_{TX_DDJ}^{3, 4, 5}$	Output data deterministic jitter		—	—	0.17	UI
$J_{TX_TJ}^{2, 3, 4, 5}$	Total output data jitter		—	—	0.35	UI

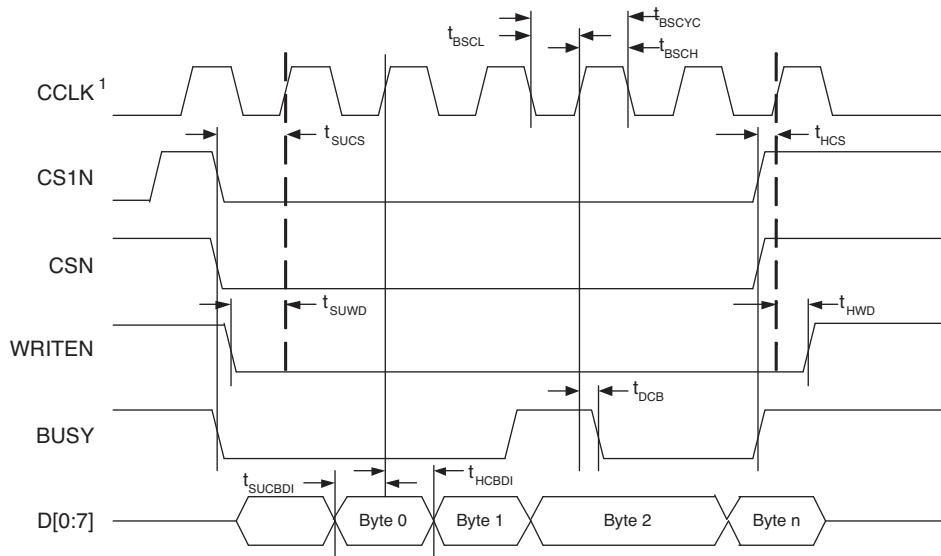
1. Rise and Fall times measured with board trace, connector and approximately 2.5pf load.
2. Total jitter includes both deterministic jitter and random jitter. The random jitter is the total jitter minus the actual deterministic jitter.
3. Jitter values are measured with each CML output AC coupled into a 50-ohm impedance (100-ohm differential impedance).
4. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
5. Values are measured at 2.5 Gbps.

Table 3-18. Receive and Jitter Tolerance

Symbol	Description	Test Conditions	Min.	Typ.	Max.	Units
RL_{RX_DIFF}	Differential return loss	From 100 MHz to 2.5 GHz	10	—	—	dB
RL_{RX_CM}	Common mode return loss	From 100 MHz to 2.5 GHz	6	—	—	dB
Z_{RX_DIFF}	Differential termination resistance		80	100	120	Ohms
$J_{RX_DJ}^{2, 3, 4, 5}$	Deterministic jitter tolerance (peak-to-peak)		—	—	0.37	UI
$J_{RX_RJ}^{2, 3, 4, 5}$	Random jitter tolerance (peak-to-peak)		—	—	0.18	UI
$J_{RX_SJ}^{2, 3, 4, 5}$	Sinusoidal jitter tolerance (peak-to-peak)		—	—	0.10	UI
$J_{RX_TJ}^{1, 2, 3, 4, 5}$	Total jitter tolerance (peak-to-peak)		—	—	0.65	UI
T_{RX_EYE}	Receiver eye opening		0.35	—	—	UI

1. Total jitter includes deterministic jitter, random jitter and sinusoidal jitter. The sinusoidal jitter tolerance mask is shown in Figure 3-18.
2. Jitter values are measured with each high-speed input AC coupled into a 50-ohm impedance.
3. Jitter and skew are specified between differential crossings of the 50% threshold of the reference signal.
4. Jitter tolerance, Differential Input Sensitivity and Receiver Eye Opening parameters are characterized when Full Rx Equalization is enabled.
5. Values are measured at 2.5 Gbps.

Figure 3-21. sysCONFIG Parallel Port Write Cycle



1. In Master Parallel Mode the FPGA provides CCLK (MCLK). In Slave Parallel Mode the external device provides CCLK.

Figure 3-22. sysCONFIG Master Serial Port Timing

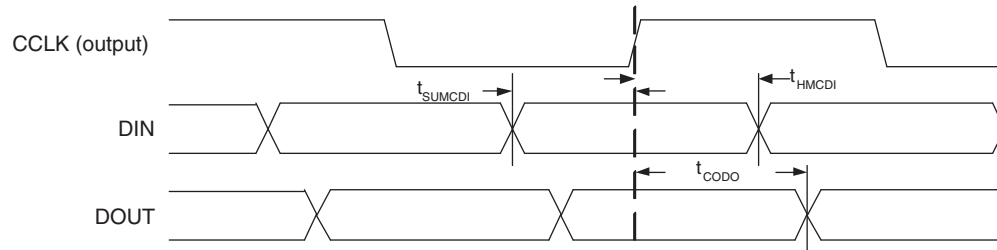
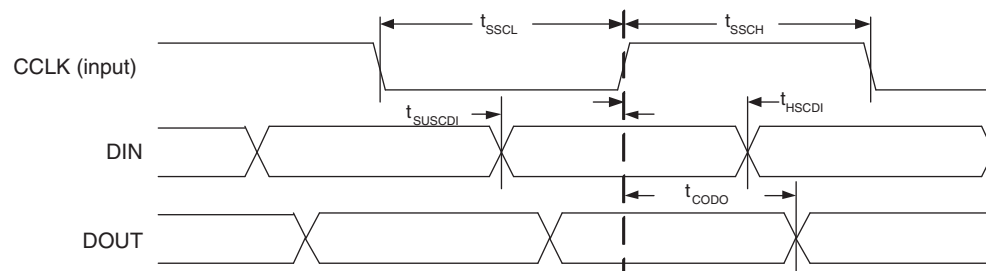


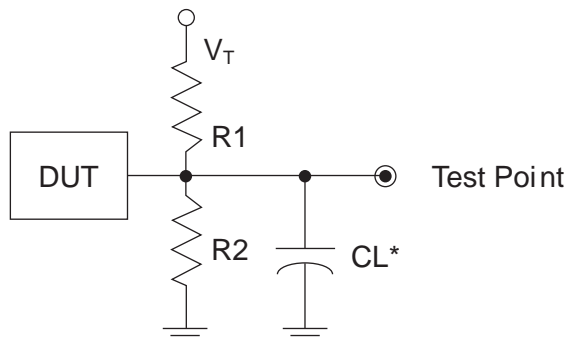
Figure 3-23. sysCONFIG Slave Serial Port Timing



Switching Test Conditions

Figure 3-29 shows the output test load that is used for AC testing. The specific values for resistance, capacitance, voltage, and other test conditions are shown in Table 3-26.

Figure 3-29. Output Test Load, LVTTTL and LVCMOS Standards



*CL Includes Test Fixture and Probe Capacitance

Table 3-26. Test Fixture Required Components, Non-Terminated Interfaces

Test Condition	R ₁	R ₂	C _L	Timing Ref.	V _T
LVTTTL and other LVCMOS settings (L -> H, H -> L)	∞	∞	0pF	LVCMOS 3.3 = 1.5V	—
				LVCMOS 2.5 = V _{CCIO} /2	—
				LVCMOS 1.8 = V _{CCIO} /2	—
				LVCMOS 1.5 = V _{CCIO} /2	—
				LVCMOS 1.2 = V _{CCIO} /2	—
LVCMOS 2.5 I/O (Z -> H)	∞	1M Ω	0pF	V _{CCIO} /2	—
LVCMOS 2.5 I/O (Z -> L)	1M Ω	∞	0pF	V _{CCIO} /2	V _{CCIO}
LVCMOS 2.5 I/O (H -> Z)	∞	100	0pF	V _{OH} - 0.10	—
LVCMOS 2.5 I/O (L -> Z)	100	∞	0pF	V _{OL} + 0.10	V _{CCIO}

Note: Output test conditions for all other interfaces are determined by the respective standards.