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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	64MHz
Connectivity	I ² C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, LVD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	32KB (16K x 16)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	2.3V ~ 5.5V
Data Converters	A/D 35x10b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-UFQFN Exposed Pad
Supplier Device Package	28-UQFN (4x4)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18f25k40-i-mv

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REGISTER 3-13	: REVI	SION ID: REVIS	SION ID R	EGISTER			
R	R	R	R	R	R	R	R
1	0	1	0		MJR	REV<5:2>	
bit 15							bit 8
R	R	R	R	R	R	R	R
MJRREV<	MJRREV<1:0>			MNRREV<5:0>			
bit 7							bit 0
Legend:							
R = Readable bit '1' = Bit is set			0' = Bit is clea	ared	x = Bit is unki	nown	
bit 15-12 R	ead as '1	010'					

 bit 10 12
 These bits are fixed with value '1010' for all devices in this family.

 bit 11-6
 MJRREV<5:0>: Major Revision ID bits

 These bits are used to identify a major revision. A major revision is indicated by an all-layer revision (A0, B0, C0, etc.).

 Revision A = 6 'b00_0000

 bit 5-0

 MNRREV<5:0>: Minor Revision ID bits

bit 5-0 **MNRREV<5:0>:** Minor Revision ID bits These bits are used to identify a minor revision.

8.3 Power-on Reset (POR)

The POR circuit holds the device in Reset until VDD has reached an acceptable level for minimum operation. Slow rising VDD, fast operating speeds or analog performance may require greater than minimum VDD. The PWRT, BOR or MCLR features can be used to extend the start-up period until all device operation conditions have been met.

8.4 Brown-out Reset (BOR)

The BOR circuit holds the device in Reset when VDD reaches a selectable minimum level. Between the POR and BOR, complete voltage range coverage for execution protection can be implemented.

The Brown-out Reset module has four operating modes controlled by the BOREN<1:0> bits in Configuration Words. The four operating modes are:

- · BOR is always on
- BOR is off when in Sleep
- · BOR is controlled by software
- BOR is always off

Refer to Table 8-1 for more information.

The Brown-out Reset voltage level is selectable by configuring the BORV<1:0> bits in Configuration Words.

A VDD noise rejection filter prevents the BOR from triggering on small events. If VDD falls below VBOR for a duration greater than parameter TBORDC, the device will reset. See Table 37-11 for more information.

8.4.1 BOR IS ALWAYS ON

When the BOREN bits of Configuration Words are programmed to '11', the BOR is always on. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is active during Sleep. The BOR does not delay wake-up from Sleep.

8.4.2 BOR IS OFF IN SLEEP

When the BOREN bits of Configuration Words are programmed to '10', the BOR is on, except in Sleep. The device start-up will be delayed until the BOR is ready and VDD is higher than the BOR threshold.

BOR protection is not active during Sleep. The device wake-up will be delayed until the BOR is ready.

8.4.3 BOR CONTROLLED BY SOFTWARE

When the BOREN bits of Configuration Words are programmed to '01', the BOR is controlled by the SBOREN bit of the BORCON register. The device start-up is not delayed by the BOR ready condition or the VDD level.

BOR protection begins as soon as the BOR circuit is ready. The status of the BOR circuit is reflected in the BORRDY bit of the BORCON register.

BOR protection is unchanged by Sleep.

8.4.4 BOR AND BULK ERASE

BOR is forced ON during PFM Bulk Erase operations to make sure that the system code protection cannot be compromised by reducing VDD.

During Bulk Erase, the BOR is enabled at 2.45V for F and LF devices, even if it is configured to some other value. If VDD falls, the erase cycle will be aborted, but the device will not be reset.

R/W-1/1	R/W-1/1	U-0	U-0	U-0	U-0	R/W-1/1	R/W-1/1	
HLVDIP	ZCDIP	_	_	_	—	C2IP	C1IP	
bit 7							bit 0	
Legend:								
R = Readable	bit	W = Writable	bit	U = Unimpler	mented bit, read	l as '0'		
-n = Value at F	POR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unkr	nown	
bit 7	HLVDIP: HLV	D Interrupt Pric	ority bit					
	1 = High prio	rity						
	0 = Low prior	ity						
bit 6	ZCDIP: Zero-	Cross Detect Ir	nterrupt Priori	ty bit				
	1 = High prior	rity						
bit 5-2		ted: Read as '	ı'					
bit 1		rator 2 Interrup	t Driority bit					
	1 - High prio	rity	t Fhonty bit					
	$\perp = \operatorname{High} \operatorname{priority}$							
bit 0	C1IP: Comparator 1 Interrupt Priority bit							
	1 = High prio	rity	·) · ·					
	0 = Low prior	ity						

REGISTER 14-20: IPR2: PERIPHERAL INTERRUPT PRIORITY REGISTER 2

18.1 Timer0 Operation

Timer0 can operate as either an 8-bit timer/counter or a 16-bit timer/counter. The mode is selected with the T016BIT bit of the T0CON register.

18.1.1 16-BIT MODE

The register pair TMR0H:TMR0L, increments on the rising edge of the clock source. A 15-bit prescaler on the clock input gives several prescale options (see prescaler control bits, T0CKPS<3:0> in the T0CON1 register).

18.1.1.1 Timer0 Reads and Writes in 16-Bit Mode

In 16-bit mode, to avoid rollover between reading high and low registers, the TMR0H register is a buffered copy of the actual high byte of Timer0, which is neither directly readable nor writable (see Figure 18-1). TMR0H is updated with the contents of the high byte of Timer0 during a read of TMR0L. This provides the ability to read all 16 bits of Timer0 without having to verify that the read of the high and low byte was valid, due to a rollover between successive reads of the high and low byte.

Similarly, a write to the high byte of Timer0 must also take place through the TMR0H Buffer register. The high byte is updated with the contents of TMR0H when a write occurs to TMR0L. This allows all 16 bits of Timer0 to be updated at once.

18.1.2 8-BIT MODE

In normal operation, TMR0 increments on the rising edge of the clock source. A 15-bit prescaler on the clock input gives several prescale options (see prescaler control bits, T0CKPS<3:0> in the T0CON1 register).

In 8-bit mode, the value of TMR0L is compared to that of the Period buffer, a copy of TMR0H, on each clock cycle. When the two values match, the following events happen:

- TMR0_out goes high for one prescaled clock period
- TMR0L is reset
- The contents of TMR0H are copied to the period buffer

In 8-bit mode, the TMR0L and TMR0H registers are both directly readable and writable. The TMR0L register is cleared on any device Reset, while the TMR0H register initializes at FFh. Both the prescaler and postscaler counters are cleared on the following events:

- A write to the TMR0L register
- A write to either the T0CON0 or T0CON1 registers
- Any device Reset Power-on Reset (POR), MCLR Reset, Watchdog Timer Reset (WDTR) or
- Brown-out Reset (BOR)

18.1.3 COUNTER MODE

In Counter mode, the prescaler is normally disabled by setting the T0CKPS bits of the T0CON1 register to '0000'. Each rising edge of the clock input (or the output of the prescaler if the prescaler is used) increments the counter by '1'.

18.1.4 TIMER MODE

In Timer mode, the Timer0 module will increment every instruction cycle as long as there is a valid clock signal and the T0CKPS bits of the T0CON1 register (Register 18-2) are set to '0000'. When a prescaler is added, the timer will increment at the rate based on the prescaler value.

18.1.5 ASYNCHRONOUS MODE

When the TOASYNC bit of the TOCON1 register is set (TOASYNC = '1'), the counter increments with each rising edge of the input source (or output of the prescaler, if used). Asynchronous mode allows the counter to continue operation during Sleep mode provided that the clock also continues to operate during Sleep.

18.1.6 SYNCHRONOUS MODE

When the T0ASYNC bit of the T0CON1 register is clear (T0ASYNC = 0), the counter clock is synchronized to the system clock (Fosc/4). When operating in Synchronous mode, the counter clock frequency cannot exceed Fosc/4.

18.2 Clock Source Selection

The T0CS<2:0> bits of the T0CON1 register are used to select the clock source for Timer0. Register 18-2 displays the clock source selections.

18.2.1 INTERNAL CLOCK SOURCE

When the internal clock source is selected, Timer0 operates as a timer and will increment on multiples of the clock source, as determined by the Timer0 prescaler.

18.2.2 EXTERNAL CLOCK SOURCE

When an external clock source is selected, Timer0 can operate as either a timer or a counter. Timer0 will increment on multiples of the rising edge of the external clock source, as determined by the Timer0 prescaler.

19.1 Register Definitions: Timer1/3/5

Long bit name prefixes for the Timer1/3/5 are shown in Table 20-1. Refer to **Section 1.4.2.2 "Long Bit Names"** for more information.

TABLE 19-1:

Peripheral	Bit Name Prefix
Timer1	T1
Timer3	Т3
Timer5	T5

REGISTER 19-1: TxCON: TIMERx CONTROL REGISTER

U-0	U-0	R/W-0/u	R/W-0/u	U-0	R/W-0/u	R/W-0/0	R/W-0/u
—	—	CKPS<1:0>		—	SYNC	RD16	ON
bit 7							bit 0

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	u = unchanged	

bit 7-6	Unimplemented: Read as '0'
---------	----------------------------

bit 5-4 CKPS<1:0>: Timerx Input Clock Prescale Select bits

- 11 = 1:8 Prescale value
- 10 = 1:4 Prescale value
- 01 = 1:2 Prescale value
- 00 = 1:1 Prescale value
- bit 3 Unimplemented: Read as '0'
- bit 2 SYNC: Timerx External Clock Input Synchronization Control bit TMRxCLK = Fosc/4 or Fosc:
 - This bit is ignored. Timer1 uses the incoming clock as is.

Else:

- 1 = Do not synchronize external clock input
- 0 = Synchronize external clock input with system clock
- bit 1 RD16: 16-Bit Read/Write Mode Enable bit
 - 1 = Enables register read/write of Timer in one 16-bit operation
 - 0 = Enables register read/write of Timer in two 8-bit operations
- bit 0 ON: Timerx On bit
 - 1 = Enables Timerx
 - 0 = Disables Timerx

REGISTER 19-3: TMRxCLK: TIMERx CLOCK REGISTER

U-0	U-0	U-0	U-0	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u
—	—	—	—	CS<3:0>			
bit 7							bit 0

Legend:					
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'			
-n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	u = unchanged		

bit 7-4 Unimplemented: Read as '0'

bit 3-0 **CS<3:0>:** Timerx Clock Source Selection bits

66	Timer1	Timer3	Timer5
	Clock Source	Clock Source	Clock Source
1111-1100	Reserved	Reserved	Reserved
1011	TMR5 overflow	TMR5 overflow	Reserved
1010	TMR3 overflow	Reserved	TMR3 overflow
1001	Reserved	TMR1 overflow	TMR1 overflow
1000	TMR0 overflow	TMR0 overflow	TMR0 overflow
0111	CLKREF	CLKREF	CLKREF
0110	SOSC	SOSC	SOSC
0101	MFINTOSC (500 kHz)	MFINTOSC (500 kHz)	MFINTOSC (500 kHz)
0100	LFINTOSC	LFINTOSC	LFINTOSC
0011	HFINTOSC	HFINTOSC	HFINTOSC
0010	Fosc	Fosc	Fosc
0001	Fosc/4	Fosc/4	Fosc/4
0000	T1CKIPPS	T3CKIPPS	T5CKIPPS



FIGURE 20-12: RISING EDGE-TRIGGERED MONOSTABLE MODE TIMING DIAGRAM (MODE = 10001)

R/W-x/x	R/W-x/x	R/W-x/x	R/W-x/x	R/W-x/x	R/W-x/x	R/W-x/x	R/W-x/x			
CCPRx<15:8>										
bit 7							bit 0			
Legend:										
R = Readable	lable bit W = Writable bit U = Unimplemented bit, read as '0'			l as '0'						
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is cle	ared	x = Bit is unknown				

REGISTER 21-5: CCPRxH: CCPx REGISTER HIGH BYTE

bit 7-0
MODE = Capture Mode:
CCPRxH<7:0>: MSB of captured TMR1 value
MODE = Compare Mode:
CCPRxH<7:0>: MSB compared to TMR1 value
MODE = PWM Mode && FMT = 0:
CCPRxH<7:2>: Not used
CCPRxH<7:2>: Not used
CCPRxH<1:0>: CCPW<9:8> - Pulse-Width MS 2 bits
MODE = PWM Mode && FMT = 1:
CCPRxH<7:0>: CCPW<9:2> - Pulse-Width MS 8 bits

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
INTCON	GIE/GIEH	PEIE/GIEL	IPEN	—	_	INT2EDG	INT1EDG	INT0EDG	166
PIE6	—	_	_	_	—	_	CCP2IE	CCP1IE	181
PIR6	_	_	_	_	_	_	CCP2IF	CCP1IF	173
IPR6	_	_	_	_	_	_	CCP2IP	CCP1IP	189
PMD3	—	_	_	_	PWM4MD	PWM3MD	CCP2MD	CCP1MD	67
CCPxCON	EN	—	OUT	FMT		MODE	E<3:0>		262
CCPxCAP	—	—		—	_		CTS<	<1:0>	265
CCPRxL	CCPRx<7:0>								265
CCPRxH		CCPRx<15:8>						266	
CCPTMRS	P4TSE	L<1:0>	P3TSE	L<1:0>	<1:0> C2TSEL<1:0> C1TSEL<1:0>				264
CCPxPPS	—	—				CCPxPPS<4:0	>		211
RxyPPS	—	—				RxyPPS<4:0>			213
T1CON	—	—	T1CKP	S<1:0>	_	T1SYNC	T1RD16	TMR10N	223
T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GO/DONE	T1GVAL		—	224
T1CLK	—	—		—		CS<	:3:0>		225
T1GATE	—	—		—		GSS	<3:0>		226
TMR1L	TMR1L7	TMR1L6	TMR1L5	TMR1L4	TMR1L3	TMR1L2	TMR1L1	TMR1L0	227
TMR1H	TMR1H7	TMR1H6	TMR1H5	TMR1H4	TMR1H3	TMR1H2	TMR1H1	TMR1H0	227
TMR2					TMR2<7:0>				238*
T2PR					PR2<7:0>				238*
T2CON	ON		CKPS<2:0>			OUTP	S<3:0>		256
T2HLT	PSYNC	CPOL	CSYNC			MODE<4:0>			257
T2CLKCON	_	_	_	_		CS<	:3:0>		258
T2RST	_	_	_	_		RSEL	<3:0>		259

TABLE 21-5: SUMMARY OF REGISTERS ASSOCIATED WITH CCPx

Legend: — = Unimplemented location, read as '0'. Shaded cells are not used by the CCP module.

Not a physical register.

R/W-0/0	R/W-1/1	R/W-0/0	R/W-1/1	R/W-0/0	R/W-1/1	R/W-0/0	R/W-1/1
P4TSE	L<1:0>	P3TSEL<1:0>		C2TSE	L<1:0>	C1TSEI	_<1:0>
bit 7							bit 0
Legend:							
R = Readable	bit	W = Writable I	oit	U = Unimplem	nented bit, read	as '0'	
-n = Value at P	OR	'1' = Bit is set		'0' = Bit is clea	ared	x = Bit is unkn	own
bit 7-6	P4TSEL<1:0> 11 = PWM4 10 = PWM4 01 = PWM4 00 = Reserve	PWM4 Timer based on TMR based on TMR based on TMR based on TMR based on TMR	⁻ Selection bit 6 4 2	S			
bit 5-4	5-4 P3TSEL<1:0>: PWM3 Timer Selection bits 11 = PWM3 based on TMR6 10 = PWM3 based on TMR4 01 = PWM3 based on TMR2 00 = Reserved						
bit 3-2	C2TSEL<1:0>: CCP2 Timer Selection bits 11 = CCP2 is based off Timer5 in Capture/Compare mode and Timer6 in PWM mode 10 = CCP2 is based off Timer3 in Capture/Compare mode and Timer4 in PWM mode 01 = CCP2 is based off Timer1 in Capture/Compare mode and Timer2 in PWM mode 00 = Reserved						
bit 1-0	1-0 C1TSEL<1:0>: CCP1 Timer Selection bits 11 = CCP1 is based off Timer5 in Capture/Compare mode and Timer6 in PWM mode 10 = CCP1 is based off Timer3 in Capture/Compare mode and Timer4 in PWM mode 01 = CCP1 is based off Timer1 in Capture/Compare mode and Timer2 in PWM mode 00 = Reserved						

REGISTER 22-2: CCPTMRS: CCP TIMERS CONTROL REGISTER

26.5.1 SPI MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK line. The master determines when the slave (Processor 2, Figure 26-3) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPxBUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPxBUF register as if a normal received byte (interrupts and Status bits appropriately set). The clock polarity is selected by appropriately programming the CKP bit of the SSPxCON1 register and the CKE bit of the SSPxSTAT register. This then, would give waveforms for SPI communication as shown in Figure 26-4, Figure 26-6, Figure 26-7 and Figure 26-8, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 * Tcy)
- Fosc/64 (or 16 * Tcy)
- Timer2 output/2
- Fosc/(4 * (SSPxADD + 1))

Figure 26-4 shows the waveforms for Master mode.

When the CKE bit is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPxBUF is loaded with the received data is shown.

Note: In Master mode the clock signal output to the SCK pin is also the clock signal input to the peripheral. The pin selected for output with the RxyPPS register must also be selected as the peripheral input with the SSPxCLKPPS register. The pin that is selected using the SSPxCLKPPS register should also be made a digital I/O. This is done by clearing the corresponding ANSEL bit.

26.10.5 I²C MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition (Figure 26-27) occurs when the RSEN bit of the SSPxCON2 register is programmed high and the master state machine is no longer active. When the RSEN bit is set, the SCL pin is asserted low. When the SCL pin is sampled low, the Baud Rate Generator is loaded and begins counting. The SDA pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, if SDA is sampled high, the SCL pin will be deasserted (brought high). When SCL is sampled high, the Baud Rate Generator is reloaded and begins counting. SDA and SCL must be sampled high for one TBRG. This action is then followed by assertion of the SDA pin (SDA = 0) for one TBRG while SCL is high. SCL is asserted low. Following this, the RSEN bit of the SSPxCON2 register will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDA pin held low. As soon as a Start condition is detected on the SDA and SCL pins, the S bit of the SSPxSTAT register will be set. The SSPxIF bit will not be set until the Baud Rate Generator has timed out.

- Note 1: If RSEN is programmed while any other event is in progress, it will not take effect.
 - **2:** A bus collision during the Repeated Start condition occurs if:
 - SDA is sampled low when SCL goes from low-to-high.
 - SCL goes low before SDA is asserted low. This may indicate that another master is attempting to transmit a data '1'.

26.10.6 I²C MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address is accomplished by simply writing a value to the SSPxBUF register. This action will set the Buffer Full flag bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDA pin after the falling edge of SCL is asserted. SCL is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCL is released high. When the SCL pin is released high, it is held that way for TBRG. The data on the SDA pin must remain stable for that duration and some hold time after the next falling edge of SCL. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDA. This allows the slave device being addressed to respond with an \overline{ACK} bit during the ninth bit time if an address match occurred, or if data was received properly. The status of ACK is written into the ACKSTAT bit on the rising edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared. If not, the bit is set. After the ninth clock, the SSPxIF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSPxBUF, leaving SCL low and SDA unchanged (Figure 26-28).

FIGURE 26-27: REPEATED START CONDITION WAVEFORM



26.10.13.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

- a) SDA or SCL are sampled low at the beginning of the Start condition (Figure 26-33).
- b) SCL is sampled low before SDA is asserted low (Figure 26-34).

During a Start condition, both the SDA and the SCL pins are monitored.

If the SDA pin is already low, or the SCL pin is already low, then all of the following occur:

- · the Start condition is aborted,
- the BCLxIF flag is set and
- the MSSP module is reset to its Idle state (Figure 26-33).

The Start condition begins with the SDA and SCL pins deasserted. When the SDA pin is sampled high, the Baud Rate Generator is loaded and counts down. If the SCL pin is sampled low while SDA is high, a bus collision occurs because it is assumed that another master is attempting to drive a data '1' during the Start condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 26-35). If, however, a '1' is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to zero; if the SCL pin is sampled as '0' during this time, a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

Note: The reason that bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.





29.0 TEMPERATURE INDICATOR MODULE

This family of devices is equipped with a temperature circuit designed to measure the operating temperature of the silicon die. The circuit's range of operating temperature falls between -40°C and +85°C. The output is a voltage that is proportional to the device temperature. The output of the temperature indicator is internally connected to the device ADC.

The circuit may be used as a temperature threshold detector or a more accurate temperature indicator, depending on the level of calibration performed. A one-point calibration allows the circuit to indicate a temperature closely surrounding that point. A two-point calibration allows the circuit to sense the entire range of temperature more accurately. Reference Application Note AN1333, "Use and Calibration of the Internal Temperature Indicator" (DS00001333) for more details regarding the calibration process.

29.1 Circuit Operation

Figure 29-1 shows a simplified block diagram of the temperature circuit. The proportional voltage output is achieved by measuring the forward voltage drop across multiple silicon junctions.

Equation 29-1 describes the output characteristics of the temperature indicator.

EQUATION 29-1: VOUT RANGES

High Range: VOUT = VDD - 4VT

Low Range: VOUT = VDD - 2VT

The temperature sense circuit is integrated with the Fixed Voltage Reference (FVR) module. See **Section 28.0 "Fixed Voltage Reference (FVR)"** for more information.

The circuit is enabled by setting the TSEN bit of the FVRCON register. When disabled, the circuit draws no current.

The circuit operates in either high or low range. The high range, selected by setting the TSRNG bit of the FVRCON register, provides a wider output voltage. This provides more resolution over the temperature range, but may be less consistent from part to part. This range requires a higher bias voltage to operate and thus, a higher VDD is needed.

The low range is selected by clearing the TSRNG bit of the FVRCON register. The low range generates a lower voltage drop and thus, a lower bias voltage is needed to operate the circuit. The low range is provided for low voltage operation.

FIGURE 29-1: TEMPERATURE CIRCUIT DIAGRAM



29.2 Minimum Operating VDD

When the temperature circuit is operated in low range, the device may be operated at any operating voltage that is within specifications.

When the temperature circuit is operated in high range, the device operating voltage, VDD, must be high enough to ensure that the temperature circuit is correctly biased.

Table 29-1 shows the recommended minimum $\mathsf{V}\mathsf{D}\mathsf{D}$ vs. range setting.

TABLE 29-1: RECOMMENDED VDD VS. RANGE

Min. VDD, TSRNG = 1	Min. VDD, TSRNG = 0
3.6V	1.8V

29.3 Temperature Output

The output of the circuit is measured using the internal Analog-to-Digital Converter. A channel is reserved for the temperature circuit output. Refer to **Section 31.0 "Analog-to-Digital Converter with Computation (ADC2) Module"** for detailed information.

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TABLE 35-2: INSTRUCTION SET

Mnemonic, Operands		Deperimtion	Qualas	16-Bit Instruction Word			/ord	Status	Natas
		Description	Cycles	MSb			LSb	Affected	Notes
BYTE-ORI	ENTED (OPERATIONS							
ADDWF	f, d, a	Add WREG and f	1	0010	01da	ffff	ffff	C, DC, Z, OV, N	1, 2
ADDWFC	f, d, a	Add WREG and CARRY bit to f	1	0010	00da	ffff	ffff	C, DC, Z, OV, N	1, 2
ANDWF	f, d, a	AND WREG with f	1	0001	01da	ffff	ffff	Z, N	1,2
CLRF	f, a	Clear f	1	0110	101a	ffff	ffff	Z	2
COMF	f, d, a	Complement f	1	0001	11da	ffff	ffff	Z, N	1, 2
CPFSEQ	f, a	Compare f with WREG, skip =	1 (2 or 3)	0110	001a	ffff	ffff	None	4
CPFSGT	f, a	Compare f with WREG, skip >	1 (2 or 3)	0110	010a	ffff	ffff	None	4
CPFSLT	f, a	Compare f with WREG, skip <	1 (2 or 3)	0110	000a	ffff	ffff	None	1, 2
DECF	f, d, a	Decrement f	1	0000	01da	ffff	ffff	C, DC, Z, OV, N	1, 2, 3, 4
DECFSZ	f, d, a	Decrement f, Skip if 0	1 (2 or 3)	0010	11da	ffff	ffff	None	1, 2, 3, 4
DCFSNZ	f, d, a	Decrement f, Skip if Not 0	1 (2 or 3)	0100	11da	ffff	ffff	None	1, 2
INCF	f, d, a	Increment f	1	0010	10da	ffff	ffff	C, DC, Z, OV, N	1, 2, 3, 4
INCFSZ	f, d, a	Increment f, Skip if 0	1 (2 or 3)	0011	11da	ffff	ffff	None	4
INFSNZ	f, d, a	Increment f, Skip if Not 0	1 (2 or 3)	0100	10da	ffff	ffff	None	1, 2
IORWF	f, d, a	Inclusive OR WREG with f	1	0001	00da	ffff	ffff	Z, N	1, 2
MOVF	f, d, a	Move f	1	0101	00da	ffff	ffff	Z, N	1
MOVFF	f _s , f _d	Move f _s (source) to 1st word	2	1100	ffff	ffff	ffff	None	
	0 u	f _d (destination) 2nd word		1111	ffff	ffff	ffff		
MOVWF	f, a	Move WREG to f	1	0110	111a	ffff	ffff	None	
MULWF	f, a	Multiply WREG with f	1	0000	001a	ffff	ffff	None	1, 2
NEGF	f, a	Negate f	1	0110	110a	ffff	ffff	C, DC, Z, OV, N	
RLCF	f, d, a	Rotate Left f through Carry	1	0011	01da	ffff	ffff	C, Z, N	1, 2
RLNCF	f, d, a	Rotate Left f (No Carry)	1	0100	01da	ffff	ffff	Z, N	
RRCF	f, d, a	Rotate Right f through Carry	1	0011	00da	ffff	ffff	C, Z, N	
RRNCF	f, d, a	Rotate Right f (No Carry)	1	0100	00da	ffff	ffff	Z, N	
SETF	f, a	Set f	1	0110	100a	ffff	ffff	None	1, 2
SUBFWB	f, d, a	Subtract f from WREG with	1	0101	01da	ffff	ffff	C, DC, Z, OV, N	
		borrow							
SUBWF	f, d, a	Subtract WREG from f	1	0101	11da	ffff	ffff	C, DC, Z, OV, N	1, 2
SUBWFB	f, d, a	Subtract WREG from f with	1	0101	10da	ffff	ffff	C, DC, Z, OV, N	
		borrow						,	
SWAPF	f, d, a	Swap nibbles in f	1	0011	10da	ffff	ffff	None	4
TSTFSZ	f, a	Test f, skip if 0	1 (2 or 3)	0110	011a	ffff	ffff	None	1, 2
XORWF	f, d, a	Exclusive OR WREG with f	1	0001	10da	ffff	ffff	Z, N	

Note 1: When a PORT register is modified as a function of itself (e.g., MOVF PORTB, 1, 0), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and where applicable, 'd' = 1), the prescaler will be cleared if assigned.

3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

4: Some instructions are two-word instructions. The second word of these instructions will be executed as a NOP unless the first word of the instruction retrieves the information embedded in these 16 bits. This ensures that all program memory locations have a valid instruction.

35.2.2 EXTENDED INSTRUCTION SET

ADD	DFSR	Add Lite	Add Literal to FSR					
Synta	ax:	ADDFSR	f, k					
Oper	ands:	0 ≤ k ≤ 63 f ∈ [0, 1, 2						
Oper	ation:	FSR(f) + k	$x \rightarrow FSR($	f)				
Statu	s Affected:	None	None					
Enco	oding:	1110	1000	ffkk	kkkk			
Description:		The 6-bit I contents c	The 6-bit literal 'k' is added to the contents of the FSR specified by 'f'.					
Word	ls:	1	1					
Cycle	es:	1						
QC	ycle Activity:							
	Q1	Q2	Q3		Q4			
	Decode	Read	Proce	SS	Write to			
		literal 'k'	Data	3	FSR			

Example:	ADDFSR	2,	23h

Before Instru		
FSR2	=	03FFh
After Instruct		
FSR2	=	0422h

ADDULNK	Add Literal to FSR2 and Return						
Syntax:	ADDULN	Kk					
Operands:	$0 \le k \le 63$	3					
Operation:	FSR2 + k	\rightarrow FSR2	,				
	$(TOS) \rightarrow$	PC					
Status Affected:	None						
Encoding:	1110	1000	11kk	kkkk			
Description:	The 6-bit literal 'k' is added to the contents of FSR2. A RETURN is then executed by loading the PC with the TOS. The instruction takes two cycles to execute; a NOP is performed during the second cycle. This may be thought of as a special case of the ADDFSR instruction, where $f = 3$ (binary '11'); it operates						
Words:	1						
Cycles:	2						
O Cuelo A stivitur							

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read	Process	Write to
	literal 'k'	Data	FSR
No	No	No	No
Operation	Operation	Operation	Operation

Example: ADDULNK 23h

_			

Before Instru	ction	
FSR2	=	03FFh
PC	=	0100h
After Instruct	ion	
FSR2	=	0422h
PC	=	(TOS)

Note: All PIC18 instructions may take an optional label argument preceding the instruction mnemonic for use in symbolic addressing. If a label is used, the instruction syntax then becomes: {label} instruction argument(s).

PIC18(L)F24/25K40

MO\	/SS	Move Indexed to Indexed							
Synta	ax:	MOVSS [z	z _s], [z _d]						
Oper	ands:	0 ≤ z _s ≤ 127 0 ≤ z _d ≤ 127	$0 \le z_s \le 127$ $0 \le z_d \le 127$						
Oper	ation:	((FSR2) + z	$((FSR2) + z_s) \rightarrow ((FSR2) + z_d)$						
Statu	s Affected:	None							
Enco	ding:								
1st w	ord (source)	1110	1011	1zz	z	ZZZZS			
2nd v	vord (dest.)	1111	xxxx	XZZ	Z	zzzzd			
Worc	Is:	addresses of registers ar 7-bit literal respectively registers ca the 4096-by (000h to FF The MOVSS PCL, TOSL destination If the result an indirect a value return resultant de an indirect a instruction v 2	moved to the destination register. The addresses of the source and destination registers are determined by adding the 7-bit literal offsets ' z_s ' or ' z_d ', respectively, to the value of FSR2. Both registers can be located anywhere in the 4096-byte data memory space (000h to FFFh). The MOVSS instruction cannot use the PCL, TOSU, TOSH or TOSL as the destination register. If the resultant source address points to an indirect addressing register, the value returned will be 00h. If the resultant destination address points to an indirect addressing register, the instruction will execute as a NOP.						
Cycle	26.	2	- 2						
QC	ycle Activity:	£							
	Q1	Q2	Q3			Q4			
	Decode	Determine	Determi	ine		Read			
		source addr	source a	ıddr	SO	urce reg			
	Decode	Determine dest addr	Determi dest ad	ine Idr	to	Write dest reg			

Example:	MOVSS	[05h],	[06h]
Before Instructio	on	0.01-	
FSR2 Contents	=	80h	
of 85h	=	33h	
of 86h	=	11h	
After Instruction			
FSR2	=	80h	
Contents	_	22h	
Contents	-	3311	
of 86h	=	33h	

PUSHL	Store Liter	al at FSR	2, Decr	ement FSR2	
Syntax:	PUSHL k				
Operands:	$0 \le k \le 255$				
Operation:	$k \rightarrow (FSR2)$ FSR2 – 1 –), → FSR2			
Status Affected:	None				
Encoding:	1111	1010	kkkk	kkkk	
2000194011	memory address specified by FSR2. FSR2 is decremented by 1 after the operation. This instruction allows users to push values onto a software stack.				
Words:	1				
Cycles:	1				
Q Cycle Activity	y:				
Q1	Q2		Q3	Q4	
Decode	Read '	k' Pro	ocess lata	Write to destination	
Example:	PUSHL	08h			
Before Inst FSR2 Memo	ruction H:FSR2L pry (01ECh)	= =	01ECh 00h		
After Instru	iction				

rinstruction		
FSR2H:FSR2L	=	01EBh
Memory (01ECh)	=	08h

35.2.3 BYTE-ORIENTED AND BIT-ORIENTED INSTRUCTIONS IN INDEXED LITERAL OFFSET MODE

Note:	Enabling	the	PIC18	instruction	set	
	extension	may	cause lee	gacy applicat	tions	
	to behave erratically or fail entirely.					

In addition to eight new commands in the extended set, enabling the extended instruction set also enables Indexed Literal Offset Addressing mode (Section 10.7.1 "Indexed Addressing with Literal Offset"). This has a significant impact on the way that many commands of the standard PIC18 instruction set are interpreted.

When the extended set is disabled, addresses embedded in opcodes are treated as literal memory locations: either as a location in the Access Bank ('a' = 0), or in a GPR bank designated by the BSR ('a' = 1). When the extended instruction set is enabled and 'a' = 0, however, a file register argument of 5Fh or less is interpreted as an offset from the pointer value in FSR2 and not as a literal address. For practical purposes, this means that all instructions that use the Access RAM bit as an argument – that is, all byte-oriented and bitoriented instructions, or almost half of the core PIC18 instructions – may behave differently when the extended instruction set is enabled.

When the content of FSR2 is 00h, the boundaries of the Access RAM are essentially remapped to their original values. This may be useful in creating backward compatible code. If this technique is used, it may be necessary to save the value of FSR2 and restore it when moving back and forth between C and assembly routines in order to preserve the Stack Pointer. Users must also keep in mind the syntax requirements of the extended instruction set (see Section 35.2.3.1 "Extended Instruction Syntax with Standard PIC18 Commands").

Although the Indexed Literal Offset Addressing mode can be very useful for dynamic stack and pointer manipulation, it can also be very annoying if a simple arithmetic operation is carried out on the wrong register. Users who are accustomed to the PIC18 programming must keep in mind that, when the extended instruction set is enabled, register addresses of 5Fh or less are used for Indexed Literal Offset Addressing.

Representative examples of typical byte-oriented and bit-oriented instructions in the Indexed Literal Offset Addressing mode are provided on the following page to show how execution is affected. The operand conditions shown in the examples are applicable to all instructions of these types.

35.2.3.1 Extended Instruction Syntax with Standard PIC18 Commands

When the extended instruction set is enabled, the file register argument, 'f', in the standard byte-oriented and bit-oriented commands is replaced with the literal offset value, 'k'. As already noted, this occurs only when 'f' is less than or equal to 5Fh. When an offset value is used, it must be indicated by square brackets ("[]"). As with the extended instructions, the use of brackets indicates to the compiler that the value is to be interpreted as an index or an offset. Omitting the brackets, or using a value greater than 5Fh within brackets, will generate an error in the MPASM assembler.

If the index argument is properly bracketed for Indexed Literal Offset Addressing, the Access RAM argument is never specified; it will automatically be assumed to be '0'. This is in contrast to standard operation (extended instruction set disabled) when 'a' is set on the basis of the target address. Declaring the Access RAM bit in this mode will also generate an error in the MPASM assembler.

The destination argument, 'd', functions as before.

In the latest versions of the MPASMTM assembler, language support for the extended instruction set must be explicitly invoked. This is done with either the command line option, $/_Y$, or the PE directive in the source listing.

35.2.4 CONSIDERATIONS WHEN ENABLING THE EXTENDED INSTRUCTION SET

It is important to note that the extensions to the instruction set may not be beneficial to all users. In particular, users who are not writing code that uses a software stack may not benefit from using the extensions to the instruction set.

Additionally, the Indexed Literal Offset Addressing mode may create issues with legacy applications written to the PIC18 assembler. This is because instructions in the legacy code may attempt to address registers in the Access Bank below 5Fh. Since these addresses are interpreted as literal offsets to FSR2 when the instruction set extension is enabled, the application may read or write to the wrong data addresses.

When porting an application to the PIC18(L)F2x/ 4xK40, it is very important to consider the type of code. A large, re-entrant application that is written in 'C' and would benefit from efficient compilation will do well when using the instruction set extensions. Legacy applications that heavily use the Access Bank will most likely not benefit from using the extended instruction set.

37.4 AC Characteristics



