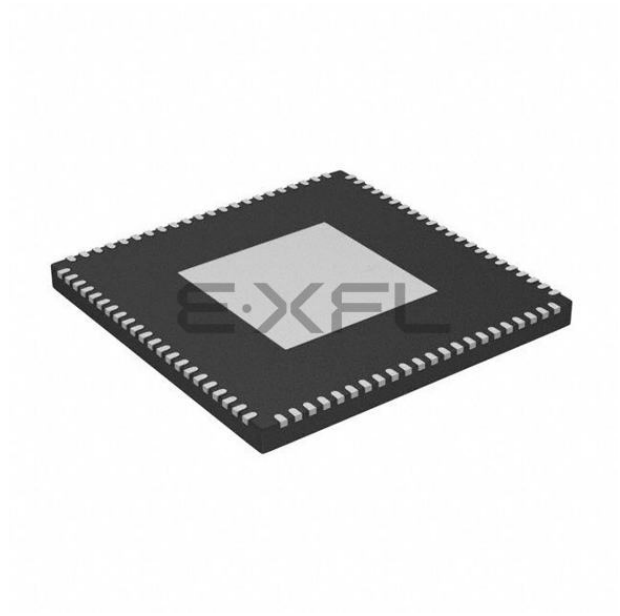

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Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Active
Type	Floating Point
Interface	DAI, DPI, EBI/EMI, I²C, SPI, SPORT, UART/USART
Clock Rate	266MHz
Non-Volatile Memory	ROM (4Mbit)
On-Chip RAM	3Mbit
Voltage - I/O	3.30V
Voltage - Core	1.20V
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	88-VFQFN Exposed Pad, CSP
Supplier Device Package	88-LFCSP-VQ (12x12)
Purchase URL	https://www.e-xfl.com/product-detail/analog-devices/ad21478wybcz2a02

bits within the 48-bit instruction to create more efficient and compact code. The program sequencer supports fetching these 16-bit and 32-bit instructions from both internal and external SDRAM memory. This support is not extended to the asynchronous memory interface (AMI). Source modules need to be built using the VISA option, in order to allow code generation tools to create these more efficient opcodes.

On-Chip Memory

The processors contain varying amounts of internal RAM and internal ROM which is shown in Table 3 through Table 5. Each block can be configured for different combinations of code and data storage. Each memory block supports single-cycle, independent accesses by the core processor and I/O processor.

The processor's SRAM can be configured as a maximum of 160k words of 32-bit data, 320k words of 16-bit data, 106.7k words of 48-bit instructions (or 40-bit data), or combinations of different word sizes up to 5M bits. All of the memory can be accessed as 16-bit, 32-bit, 48-bit, or 64-bit words. A 16-bit floating-point storage format is supported that effectively doubles the amount of data that may be stored on-chip. Conversion between the 32-bit floating-point and 16-bit floating-point formats is performed in a single instruction. While each memory block can store combinations of code and data, accesses are most efficient when one block stores data using the DM bus for transfers, and the other block stores instructions and data using the PM bus for transfers.

Using the DM bus and PM buses, with one bus dedicated to a memory block, assures single-cycle execution with two data transfers. In this case, the instruction must be available in the cache.

The memory maps in Table 3 through Table 5 display the internal memory address space of the processors. The 48-bit space section describes what this address range looks like to an instruction that retrieves 48-bit memory. The 32-bit section describes what this address range looks like to an instruction that retrieves 32-bit memory.

Table 3. ADSP-21477 Internal Memory Space (2M bits)

IOP Registers 0x0000 0000–0x0003 FFFF			
Long Word (64 Bits)	**Extended Precision Normal or Instruction Word (48 Bits)**	**Normal Word (32 Bits)**	**Short Word (16 Bits)**
Block 0 ROM (Reserved) 0x0004 0000–0x0004 7FFF	Block 0 ROM (Reserved) 0x0008 0000–0x0008 AAA9	Block 0 ROM (Reserved) 0x0008 0000–0x0008 FFFF	Block 0 ROM (Reserved) 0x0010 0000–0x0011 FFFF
Reserved 0x0004 8000–0x0004 8FFF	Reserved 0x0008 AAAA–0x0008 BFFF	Reserved 0x0009 0000–0x0009 1FFF	Reserved 0x0012 0000–0x0012 FFFF
Block 0 SRAM 0x0004 9000–0x0004 BFFF	Block 0 SRAM 0x0008 C000–0x0008 FFFF	Block 0 SRAM 0x0009 2000–0x0009 7FFF	Block 0 SRAM 0x0012 4000–0x0012 FFFF
Reserved 0x0004 C000–0x0004 FFFF	Reserved 0x0009 000–0x0009 5554	Reserved 0x0009 8000–0x0009 FFFF	Reserved 0x0013 0000–0x0013 FFFF
Block 1 ROM (Reserved) 0x0005 0000–0x0005 7FFF	Block 1 ROM (Reserved) 0x000A 0000–0x000A AAA9	Block 1 ROM (Reserved) 0x000A 0000–0x000AFFFF	Block 1 ROM (Reserved) 0x0014 0000–0x0015 FFFF
Reserved 0x0005 8000–0x0005 8FFF	Reserved 0x000A AAAA–0x000A BFFF	Reserved 0x000B 0000–0x000B 1FFF	Reserved 0x0016 0000–0x0016 3FFF
Block 1 SRAM 0x0005 9000–0x0005 BFFF	Block 1 SRAM 0x000A C000–0x000A FFFF	Block 1 SRAM 0x000B 2000–0x000B 7FFF	Block 1 SRAM 0x0016 4000–0x0016 FFFF
Reserved 0x0005 C000–0x0005 FFFF	Reserved 0x000B 0000–0x000B 5554	Reserved 0x000B 8000–0x000B FFFF	Reserved 0x0017 0000–0x0017 FFFF
Block 2 SRAM 0x0006 0000–0x0006 0FFF	Block 2 SRAM 0x000C 0000–0x000C 1554	Block 2 SRAM 0x000C 0000–0x000C 1FFF	Block 2 SRAM 0x0018 0000–0x0018 3FFF
Reserved 0x0006 1000– 0x0006 FFFF	Reserved 0x000C 1555–0x000D 5554	Reserved 0x000C 2000–0x000D FFFF	Reserved 0x0018 4000–0x001B FFFF
Block 3 SRAM 0x0007 0000–0x0007 0FFF	Block 3 SRAM 0x000E 0000–0x000E 1554	Block 3 SRAM 0x000E 0000–0x000E 1FFF	Block 3 SRAM 0x001C 0000–0x001C 3FFF
Reserved 0x0007 1000–0x0007 FFFF	Reserved 0x000E 1555–0x000F 5554	Reserved 0x000E 2000–0x000F FFFF	Reserved 0x001C 4000–0x001F FFFF

Shift Register

The shift register can be used as a serial to parallel data converter. The shift register module consists of an 18-stage serial shift register, 18-bit latch, and three-state output buffers. The shift register and latch have separate clocks. Data is shifted into the serial shift register on the positive-going transitions of the shift register serial clock (SR_SCLK) input. The data in each flip-flop is transferred to the respective latch on a positive-going transition of the shift register latch clock (SR_LAT) input.

The shift register's signals can be configured as follows.

- The SR_SCLK can come from any of the SPORT0–7 SCLK outputs, PCGA/B clock, any of the DAI pins (1–8), and one dedicated pin (SR_SCLK).
- The SR_LAT can come from any of SPORT0–7 frame sync outputs, PCGA/B frame sync, any of the DAI pins (1–8), and one dedicated pin (SR_LAT).
- The SR_SDI input can from any of SPORT0–7 serial data outputs, any of the DAI pins (1–8), and one dedicated pin (SR_SDI).

Note that the SR_SCLK, SR_LAT, and SR_SDI inputs must come from same source except in the case of where SR_SCLK comes from PCGA/B or SR_SCLK and SR_LAT come from PCGA/B.

If SR_SCLK comes from PCGA/B, then SPORT0–7 generates the SR_LAT and SR_SDI signals. If SR_SCLK and SR_LAT come from PCGA/B, then SPORT0–7 generates the SR_SDI signal.

I/O PROCESSOR FEATURES

The I/O processor provides up to 65 channels of DMA as well as an extensive set of peripherals.

DMA Controller

The DMA controller operates independently and invisibly to the processor core, allowing DMA operations to occur while the core is simultaneously executing its program instructions. DMA transfers can occur between the processor's internal memory and its serial ports, the SPI-compatible (serial peripheral interface) ports, the IDP (input data port), the parallel data acquisition port (PDAP) or the UART.

Up to 65 channels of DMA are available on the processors as shown in Table 9.

Programs can be downloaded using DMA transfers. Other DMA features include interrupt generation upon completion of DMA transfers, and DMA chaining for automatic linked DMA transfers.

Table 9. DMA Channels

Peripheral	DMA Channels
SPORTs	16
PDAP	8
SPI	2
UART	2
External Port	2
Accelerators	2
Memory-to-Memory	2
MediaLB[1]	31

[1] Automotive models only.

Delay Line DMA

The processor provides delay line DMA functionality. This allows processor reads and writes to external delay line buffers (and therefore to external memory) with limited core interaction.

Scatter/Gather DMA

The processor provides scatter/gather DMA functionality. This allows processor DMA reads/writes to/from noncontiguous memory blocks.

FFT Accelerator

The FFT accelerator implements radix-2 complex/real input, complex output FFTs with no core intervention. The FFT accelerator runs at the peripheral clock frequency.

FIR Accelerator

The FIR (finite impulse response) accelerator consists of a 1024 word coefficient memory, a 1024 word deep delay line for the data, and four MAC units. A controller manages the accelerator. The FIR accelerator runs at the peripheral clock frequency.

IIR Accelerator

The IIR (infinite impulse response) accelerator consists of a 1440 word coefficient memory for storage of biquad coefficients, a data memory for storing the intermediate data and one MAC unit. A controller manages the accelerator. The IIR accelerator runs at the peripheral clock frequency.

Watchdog Timer (WDT)

The processors include a 32-bit watchdog timer that can be used to implement a software watchdog function. A software watchdog can improve system reliability by forcing the processor to a known state through generation of a system reset if the timer expires before being reloaded by software. Software initializes the count value of the timer, and then enables the timer.

The WDT is used to supervise the stability of the system software. When used in this way, software reloads the WDT in a regular manner so that the downward counting timer never expires. An expiring timer then indicates that system software might be out of control.

The WDT resets both the core and the internal peripherals. Software must be able to determine if the watch dog was the source of the hardware reset by interrogating a status bit in the watch dog timer control register.

Table 11. Pin Descriptions (Continued)

Name	Type	State During/ After Reset	Description
$\overline{\text{SDRAS}}$	O/T (ipu)	High-Z/ driven high	**SDRAM Row Address Strobe.** Connect to SDRAM's RAS pin. In conjunction with other SDRAM command pins, defines the operation for the SDRAM to perform.
$\overline{\text{SDCAS}}$	O/T (ipu)	High-Z/ driven high	**SDRAM Column Address Select.** Connect to SDRAM's CAS pin. In conjunction with other SDRAM command pins, defines the operation for the SDRAM to perform.
$\overline{\text{SDWE}}$	O/T (ipu)	High-Z/ driven high	**SDRAM Write Enable.** Connect to SDRAM's WE or W buffer pin.
SDCKE	O/T (ipu)	High-Z/ driven high	**SDRAM Clock Enable.** Connect to SDRAM's CKE pin. Enables and disables the CLK signal. For details, see the data sheet supplied with the SDRAM device.
SDA10	O/T (ipu)	High-Z/ driven high	**SDRAM A10 Pin.** Enables applications to refresh an SDRAM in parallel with non-SDRAM accesses. This pin replaces the DSP's ADDR10 pin only during SDRAM accesses.
SDDQM	O/T (ipu)	High-Z/ driven high	**DQM Data Mask.** SDRAM input mask signal for write accesses and output enable signal for read accesses. Input data is masked when DQM is sampled high during a write cycle. The SDRAM output buffers are placed in a High-Z state when DQM is sampled high during a read cycle. SDDQM is driven high from reset de-assertion until SDRAM initialization completes. Afterwards, it is driven low irrespective of whether any SDRAM accesses occur or not.
SDCLK	O/T (ipd)	High-Z/ driving	**SDRAM Clock Output.** Clock driver for this pin differs from all other clock drivers. See Figure 47. For models in the 100-lead package, the SDRAM interface should be disabled to avoid unnecessary power switching by setting the DSDCTL bit in SDCTL register. For more information, see the *ADSP-214xx SHARC Processor Hardware Reference*.
DAI _P_{20-1}	I/O/T (ipu)	High-Z	**Digital Applications Interface**. These pins provide the physical interface to the DAI SRU. The DAI SRU configuration registers define the combination of on-chip audio-centric peripheral inputs or outputs connected to the pin and to the pin's output enable. The configuration registers of these peripherals then determines the exact behavior of the pin. Any input or output signal present in the DAI SRU may be routed to any of these pins.
DPI _P_{14-1}	I/O/T (ipu)	High-Z	**Digital Peripheral Interface.** These pins provide the physical interface to the DPI SRU. The DPI SRU configuration registers define the combination of on-chip peripheral inputs or outputs connected to the pin and to the pin's output enable. The configuration registers of these peripherals then determine the exact behavior of the pin. Any input or output signal present in the DPI SRU may be routed to any of these pins.
WDT_CLKIN	I		**Watch Dog Timer Clock Input.** This pin should be pulled low when not used.
WDT_CLKO	O		**Watch Dog Resonator Pad Output.**
$\overline{\text{WDTRSTO}}$	O (ipu)		**Watch Dog Timer Reset Out.**

The following symbols appear in the Type column of Table 11: **A** = asynchronous, **I** = input, **O** = output, **S** = synchronous, **A/D** = active drive, **O/D** = open drain, and **T** = three-state, **ipd** = internal pull-down resistor, **ipu** = internal pull-up resistor.

The internal pull-up (ipu) and internal pull-down (ipd) resistors are designed to hold the internal path from the pins at the expected logic levels. To pull-up or pull-down the external pads to the expected logic levels, use external resistors. Internal pull-up/pull-down resistors cannot be enabled/disabled and the value of these resistors cannot be programmed. The range of an ipu resistor can be 26 kΩ to 63 kΩ. The range of an ipd resistor can be 31 kΩ to 85 kΩ. The three-state voltage of ipu pads will not reach to full the V_{DD_EXT} level; at typical conditions the voltage is in the range of 2.3 V to 2.7 V.

In this table, all pins are LVTTL compliant with the exception of the thermal diode, shift register, and real-time clock (RTC) pins.

Not all pins are available in the 88-lead LFCSP_VQ and 100-lead LQFP package. For more information, see Table 2 on Page 3 and Table 62 on Page 70.

Table 11. Pin Descriptions (Continued)

Name	Type	State During/ After Reset	Description
TDI	I (ipu)		**Test Data Input (JTAG).** Provides serial data for the boundary scan logic.
TDO	O/T	High-Z	**Test Data Output (JTAG).** Serial scan output of the boundary scan path.
TMS	I (ipu)		**Test Mode Select (JTAG).** Used to control the test state machine.
TCK	I		**Test Clock (JTAG).** Provides a clock for JTAG boundary scan. TCK must be asserted (pulsed low) after power-up or held low for proper operation of the device.
$\overline{\text{TRST}}$	I (ipu)		**Test Reset (JTAG).** Resets the test state machine. $\overline{\text{TRST}}$ must be asserted (pulsed low) after power-up or held low for proper operation of the processor.
$\overline{\text{EMU}}$	O (O/D, ipu)	High-Z	**Emulation Status.** Must be connected to the Analog Devices DSP Tools product line of JTAG emulators target board connector only.
CLK_CFG_{1-0}	I		**Core to CLKIN Ratio Control.** These pins set the startup clock frequency. Note that the operating frequency can be changed by programming the PLL multiplier and divider in the PMCTL register at any time after the core comes out of reset. The allowed values are: 00 = 8:1 01 = 32:1 10 = 16:1 11 = reserved
CLKIN	I		**Local Clock In.** Used in conjunction with XTAL. CLKIN is the clock input. It configures the processors to use either its internal clock generator or an external clock source. Connecting the necessary components to CLKIN and XTAL enables the internal clock generator. Connecting the external clock to CLKIN while leaving XTAL unconnected configures the processors to use the external clock source such as an external clock oscillator. CLKIN may not be halted, changed, or operated below the specified frequency.
XTAL	O		**Crystal Oscillator Terminal.** Used in conjunction with CLKIN to drive an external crystal.
$\overline{\text{RESET}}$	I		**Processor Reset.** Resets the processor to a known state. Upon deassertion, there is a 4096 CLKIN cycle latency for the PLL to lock. After this time, the core begins program execution from the hardware reset vector address. The $\overline{\text{RESET}}$ input must be asserted (low) at power-up.
$\overline{\text{RESETOUT}}/\overline{\text{RUNRSTIN}}$	I/O (ipu)		**Reset Out/Running Reset In.** The default setting on this pin is reset out. This pin also has a second function as RUNRSTIN which is enabled by setting bit 0 of the RUNRSTCTL register. For more information, see the *ADSP-214xx SHARC Processor Hardware Reference*.
BOOT_CFG_{2-0}	I		**Boot Configuration Select.** These pins select the boot mode for the processor. The BOOT_CFG pins must be valid before $\overline{\text{RESET}}$ (hardware and software) is de-asserted. The BOOT_CFG2 pin is only available on the 196-lead package.

The following symbols appear in the Type column of Table 11: **A** = asynchronous, **I** = input, **O** = output, **S** = synchronous, **A/D** = active drive, **O/D** = open drain, and **T** = three-state, **ipd** = internal pull-down resistor, **ipu** = internal pull-up resistor.

The internal pull-up (ipu) and internal pull-down (ipd) resistors are designed to hold the internal path from the pins at the expected logic levels. To pull-up or pull-down the external pads to the expected logic levels, use external resistors. Internal pull-up/pull-down resistors cannot be enabled/disabled and the value of these resistors cannot be programmed. The range of an ipu resistor can be 26 kΩ to 63 kΩ. The range of an ipd resistor can be 31 kΩ to 85 kΩ. The three-state voltage of ipu pads will not reach to full the V_{DD_EXT} level; at typical conditions the voltage is in the range of 2.3 V to 2.7 V.

In this table, all pins are LVTTL compliant with the exception of the thermal diode, shift register, and real-time clock (RTC) pins.

Not all pins are available in the 88-lead LFCSP_VQ and 100-lead LQFP package. For more information, see Table 2 on Page 3 and Table 62 on Page 70.

ELECTRICAL CHARACTERISTICS

Parameter[1]	Description	Test Conditions	200 MHz Min	200 MHz Max	266 MHz Min	266 MHz Max	300 MHz Min	300 MHz Max	Unit
V_{OH}[2]	High Level Output Voltage	@ V_{DD_EXT} = Min, I_{OH} = –1.0 mA[3]	2.4		2.4		2.4		V
V_{OL}[2]	Low Level Output Voltage	@ V_{DD_EXT} = Min, I_{OL} = 1.0 mA[3]		0.4		0.4		0.4	V
I_{IH}[4, 5]	High Level Input Current	@ V_{DD_EXT} = Max, V_{IN} = V_{DD_EXT} Max		10		10		10	µA
I_{IL}[4]	Low Level Input Current	@ V_{DD_EXT} = Max, V_{IN} = 0 V		–10		–10		–10	µA
I_{ILPU}[5]	Low Level Input Current Pull-up	@ V_{DD_EXT} = Max, V_{IN} = 0 V		200		200		200	µA
I_{OZH}[6, 7]	Three-State Leakage Current	@ V_{DD_EXT} = Max, V_{IN} = V_{DD_EXT} Max		10		10		10	µA
I_{OZL}[6]	Three-State Leakage Current	@ V_{DD_EXT} = Max, V_{IN} = 0 V		–10		–10		–10	µA
I_{OZLPU}[7]	Three-State Leakage Current Pull-up	@ V_{DD_EXT} = Max, V_{IN} = 0 V		200		200		200	µA
I_{OZHPD}[8]	Three-State Leakage Current Pull-down	@ V_{DD_EXT} = Max, V_{IN} = V_{DD_EXT} Max		200		200		200	µA
I_{DD_RTC}	V_{DD_RTC} Current	@ V_{DD_RTC} = 3.0, T_J = 25°C		0.76		0.76		0.76	µA
I_{DD_INT}[9]	Supply Current (Internal)	f_{CCLK} > 0 MHz		Table 14 + Table 15 × ASF		Table 14 + Table 15 × ASF		Table 14 + Table 15 × ASF	mA
C_{IN}[10, 11]	Input Capacitance	T_{CASE} = 25°C		5		5		5	pF

[1] Specifications subject to change without notice.
[2] Applies to output and bidirectional pins: ADDR23-0, DATA15-0, $\overline{\text{AMI_RD}}$, $\overline{\text{AMI_WR}}$, FLAG3–0, DAI_Px, DPI_Px, $\overline{\text{EMU}}$, TDO, $\overline{\text{RESETOUT}}$,MLBSIG, MLBDAT, MLBDO, MLBSO, $\overline{\text{SDRAS}}$, $\overline{\text{SDCAS}}$, $\overline{\text{SDWE}}$, SDCKE, SDA10, SDDQM, MS0-1.
[3] See Output Drive Currents for typical drive current capabilities.
[4] Applies to input pins: BOOT_CFGx, CLK_CFGx, TCK, $\overline{\text{RESET}}$, CLKIN.
[5] Applies to input pins with internal pull-ups: $\overline{\text{TRST}}$, TMS, TDI.
[6] Applies to three-statable pins: TDO, MLBDAT, MLBSIG, MLBDO, and MLBSO.
[7] Applies to three-statable pins with pull-ups: DAI_Px, DPI_Px, $\overline{\text{EMU}}$.
[8] Applies to three-statable pin with pull-down: SDCLK.
[9] See Engineer-to-Engineer Note “Estimating Power Dissipation for ADSP-214xx SHARC Processors” for further information.
[10] Applies to all signal pins.
[11] Guaranteed, but not tested.

TIMING SPECIFICATIONS

Use the exact timing information given. Do not attempt to derive parameters from the addition or subtraction of others. While addition or subtraction would yield meaningful results for an individual device, the values given in this data sheet reflect statistical variations and worst cases. Consequently, it is not meaningful to add parameters to derive longer times. See Figure 49 under Test Conditions for voltage reference levels.

Switching Characteristics specify how the processor changes its signals. Circuitry external to the processor must be designed for compatibility with these signal characteristics. Switching characteristics describe what the processor will do in a given circumstance. Use switching characteristics to ensure that any timing requirement of a device connected to the processor (such as memory) is satisfied.

Timing Requirements apply to signals that are controlled by circuitry external to the processor, such as the data input for a read operation. Timing requirements guarantee that the processor operates correctly with other devices.

Core Clock Requirements

The processor's internal clock (a multiple of CLKIN) provides the clock signal for timing internal memory, processor core, and serial ports. During reset, program the ratio between the processor's internal clock frequency and external (CLKIN) clock frequency with the CLK_CFG1–0 pins.

The processor's internal clock switches at higher frequencies than the system input clock (CLKIN). To generate the internal clock, the processor uses an internal phase-locked loop (PLL, see Figure 5). This PLL-based clocking minimizes the skew between the system clock (CLKIN) signal and the processor's internal clock.

Voltage Controlled Oscillator (VCO)

In application designs, the PLL multiplier value should be selected in such a way that the VCO frequency never exceeds f_{VCO} specified in Table 20.

- The product of CLKIN and PLLM must never exceed 1/2 of f_{VCO} (max) in Table 20 if the input divider is not enabled (INDIV = 0).
- The product of CLKIN and PLLM must never exceed f_{VCO} (max) in Table 20 if the input divider is enabled (INDIV = 1).

The VCO frequency is calculated as follows:

$$f_{VCO} = 2 \times PLLM \times f_{INPUT}$$
$$f_{CCLK} = (2 \times PLLM \times f_{INPUT}) \div PLLD$$

where:

f_{VCO} = VCO output

PLLM = Multiplier value programmed in the PMCTL register. During reset, the PLLM value is derived from the ratio selected using the CLK_CFG pins in hardware.

PLLD = 2, 4, 8, or 16 based on the divider value programmed on the PMCTL register. During reset this value is 2.

f_{INPUT} is the input frequency to the PLL.

f_{INPUT} = CLKIN when the input divider is disabled, or CLKIN ÷ 2 when the input divider is enabled.

Note the definitions of the clock periods that are a function of CLKIN and the appropriate ratio control shown in Table 20. All of the timing specifications for the peripherals are defined in relation to t_{PCLK}. See the peripheral specific section for each peripheral's timing information.

Table 18. Clock Periods

Timing Requirements	Description
t_{CK}	CLKIN Clock Period
t_{CCLK}	Processor Core Clock Period
t_{PCLK}	Peripheral Clock Period = 2 × t_{CCLK}
t_{SDCLK}	SDRAM Clock Period = (t_{CCLK}) × SDCKR

Figure 5 shows core to CLKIN relationships with an external oscillator or crystal. The shaded divider/multiplier blocks denote where clock ratios can be set through hardware or software using the power management control register (PMCTL). For more information, see the *ADSP-214xx SHARC Processor Hardware Reference*.

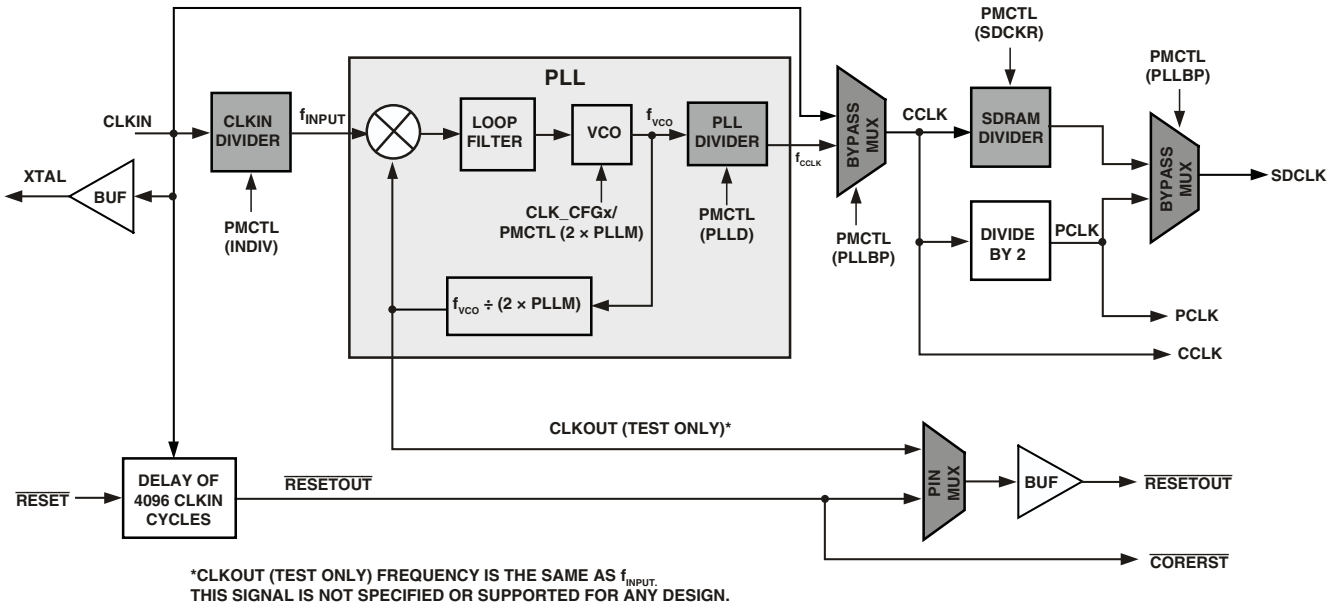

Figure 5. Core Clock and System Clock Relationship to CLKIN

Clock Signals

The processors can use an external clock or a crystal. See the CLKIN pin description in Table 11. Programs can configure the processor to use its internal clock generator by connecting the necessary components to CLKIN and XTAL. Figure 8 shows the component connections used for a crystal operating in fundamental mode. Note that the clock rate is achieved using a 16.67 MHz crystal and a PLL multiplier ratio 16:1 (CCLK:CLKIN achieves a clock speed of 266 MHz). To achieve the full core clock rate, programs need to configure the multiplier bits in the PMCTL register.

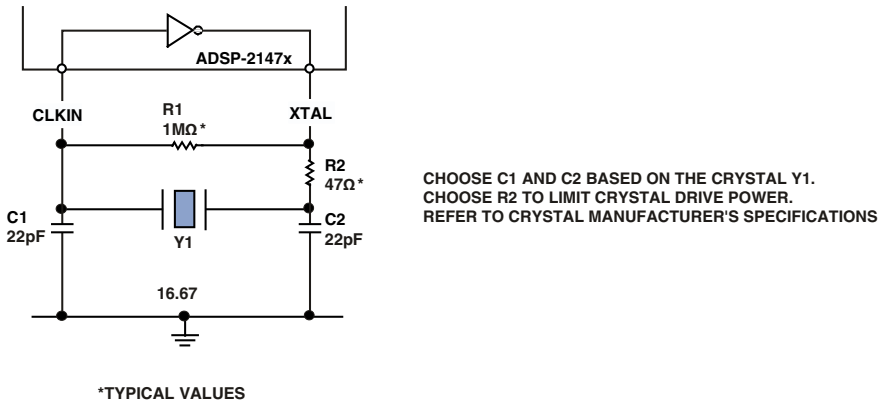

Figure 8. 266 MHz Operation (Fundamental Mode Crystal)

Reset

Table 21. Reset

Parameter		Min	Max	Unit
Timing Requirements				
t_{WRST}[1]	$\overline{\text{RESET}}$ Pulse Width Low	$4 \times t_{CK}$		ns
t_{SRST}	$\overline{\text{RESET}}$ Setup Before CLKIN Low	8		ns

[1] Applies after the power-up sequence is complete. At power-up, the processor's internal phase-locked loop requires no more than 100 µs while $\overline{\text{RESET}}$ is low, assuming stable V_{DD} and CLKIN (not including start-up time of external clock oscillator).

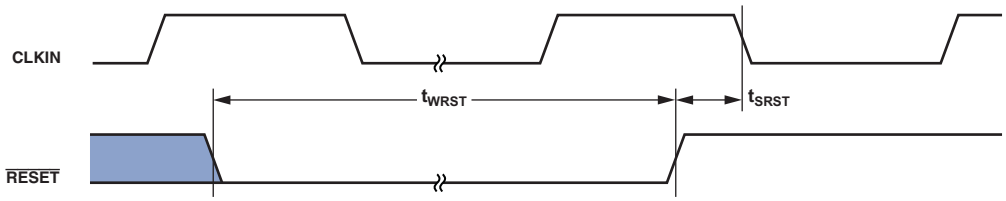

Figure 9. Reset

Pin to Pin Direct Routing (DAI and DPI)

For direct pin connections only (for example, DAI_PB01_I to DAI_PB02_O).

Table 28. DAI/DPI Pin to Pin Routing

Parameter		Min	Max	Unit
Timing Requirement				
t_{DPIO}	Delay DAI/DPI Pin Input Valid to DAI/DPI Output Valid	1.5	10	ns

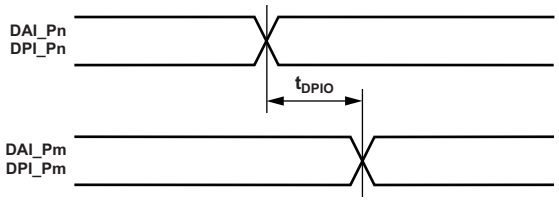

Figure 16. DAI Pin to Pin Direct Routing

Precision Clock Generator (Direct Pin Routing)

This timing is only valid when the SRU is configured such that the precision clock generator (PCG) takes its inputs directly from the DAI pins (via pin buffers) and sends its outputs directly to the DAI pins. For the other cases, where the PCG's inputs and outputs are not directly routed to/from DAI pins (via pin buffers) there is no timing data available. All timing parameters and switching characteristics apply to external DAI pins (DAI_P01 – DAI_P20).

Table 29. Precision Clock Generator (Direct Pin Routing)

Parameter		88-Lead LFCSP Package Min	88-Lead LFCSP Package Max	All Other Packages Min	All Other Packages Max	Unit
Timing Requirements						
t_{PCGIP}	Input Clock Period	$t_{PCLK} \times 4$		$t_{PCLK} \times 4$		ns
t_{STRIG}	PCG Trigger Setup Before Falling Edge of PCG Input Clock	4.5		4.5		ns
t_{HTRIG}	PCG Trigger Hold After Falling Edge of PCG Input Clock	3		3		ns
Switching Characteristics						
t_{DPCGIO}	PCG Output Clock and Frame Sync Active Edge Delay After PCG Input Clock	2.5	$2 \times t_{PCLK}$	2.5	12.5	ns
$t_{DTRIGCLK}$	PCG Output Clock Delay After PCG Trigger	$2.5 + (2.5 \times t_{PCGIP})$	$2 \times t_{PCLK} + (2.5 \times t_{PCGIP})$	$2.5 + (2.5 \times t_{PCGIP})$	$12.5 + (2.5 \times t_{PCGIP})$	ns
$t_{DTRIGFS}$	PCG Frame Sync Delay After PCG Trigger	$2.5 + ((2.5 + D - PH) \times t_{PCGIP})$	$2 \times t_{PCLK} + ((2.5 + D - PH) \times t_{PCGIP})$	$2.5 + ((2.5 + D - PH) \times t_{PCGIP})$	$12.5 + ((2.5 + D - PH) \times t_{PCGIP})$	ns
t_{PCGOW}[1]	Output Clock Period	$2 \times t_{PCGIP} - 1$		$2 \times t_{PCGIP} - 1$		ns

D = FSxDIV, PH = FSxPHASE. For more information, see the ADSP-214xx SHARC Processor Hardware Reference, "Precision Clock Generators" chapter.

[1] Normal mode of operation.

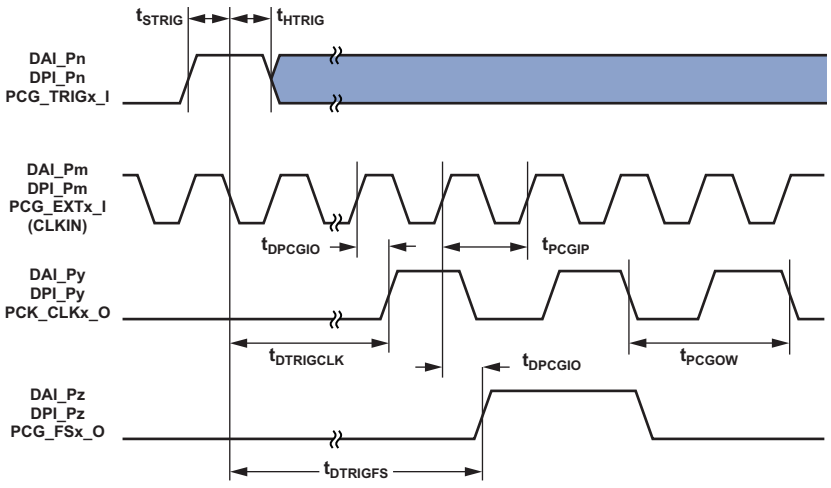

Figure 17. Precision Clock Generator (Direct Pin Routing)

AMI Write

Use these specifications for asynchronous interfacing to memories. Note that timing for AMI_ACK, ADDR, DATA, $\overline{\text{AMI_RD}}$, $\overline{\text{AMI_WR}}$, and strobe timing parameters only apply to asynchronous access mode.

Table 33. AMI Write

Parameter		Min	Max	Unit
Timing Requirements				
t_{DAAK}	AMI_ACK Delay from Address Selects[1, 2]		t_{SDCLK} – 10.1 + W	ns
t_{DSAK}	AMI_ACK Delay from $\overline{\text{AMI_WR}}$ Low[1, 3]		W – 7.1	ns
Switching Characteristics				
t_{DAWH}	Address Selects to $\overline{\text{AMI_WR}}$ Deasserted[2]	t_{SDCLK} – 4.4 + W		ns
t_{DAWL}	Address Selects to $\overline{\text{AMI_WR}}$ Low[2]	t_{SDCLK} – 4.5		ns
t_{WW}	$\overline{\text{AMI_WR}}$ Pulse Width	W – 1.3		ns
t_{DDWH}	Data Setup Before $\overline{\text{AMI_WR}}$ High	t_{SDCLK} – 4.3 + W		ns
t_{DWHA}	Address Hold After $\overline{\text{AMI_WR}}$ Deasserted	H		ns
t_{DWHD}	Data Hold After $\overline{\text{AMI_WR}}$ Deasserted	H		ns
t_{DATRWH}	Data Disable After $\overline{\text{AMI_WR}}$ Deasserted[4]	t_{SDCLK} – 1.37 + H	t_{SDCLK} + 6.75 + H	ns
t_{WWR}	$\overline{\text{AMI_WR}}$ High to $\overline{\text{AMI_WR}}$ Low[5]	t_{SDCLK} – 1.5 + H		ns
t_{DDWR}	Data Disable Before $\overline{\text{AMI_RD}}$ Low	2 × t_{SDCLK} – 7.1		ns
t_{WDE}	Data Enabled to $\overline{\text{AMI_WR}}$ Low	t_{SDCLK} – 4.5		ns

W = (number of wait states specified in AMICTLx register) × t_{SDCLK}
H = (number of hold cycles specified in AMICTLx register) × t_{SDCLK}

[1] AMI_ACK delay/setup: System must meet t_{DAAK}, or t_{DSAK}, for deassertion of AMI_ACK (low).
[2] The falling edge of $\overline{\text{AMI_MSx}}$ is referenced.
[3] Note that timing for AMI_ACK, ADDR, DATA, $\overline{\text{AMI_RD}}$, $\overline{\text{AMI_WR}}$, and strobe timing parameters only applies to asynchronous access mode.
[4] See Test Conditions for calculation of hold times given capacitive and dc loads.
[5] For Write to Write: t_{SDCLK} + H, for both same bank and different bank. For Write to Read: 3 × t_{SDCLK} + H, for the same bank and different banks.

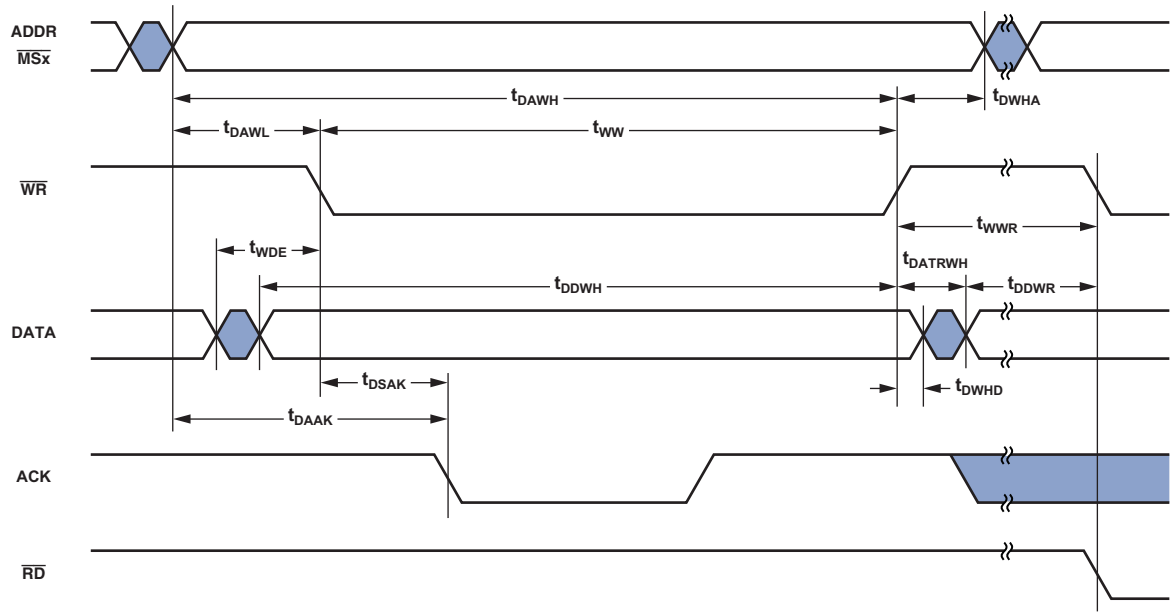

Figure 21. AMI Write

Serial Ports

In slave transmitter mode and master receiver mode, the maximum serial port frequency is $f_{PCLK}/8$. In master transmitter mode and slave receiver mode, the maximum serial port clock frequency is $f_{PCLK}/4$.

To determine whether communication is possible between two devices at clock speed, n, the following specifications must be confirmed: 1) frame sync delay and frame sync setup and hold, 2) data delay and data setup and hold, and 3) SCLK width.

Serial port signals (SCLK, FS, Data Channel A, Data Channel B) are routed to the DAI_P20–1 pins using the SRU. Therefore, the timing specifications provided below are valid at the DAI_P20–1 pins.

Table 34. Serial Ports—External Clock

Parameter		88-Lead LFCSP Package Min	88-Lead LFCSP Package Max	All Other Packages Min	All Other Packages Max	Unit
Timing Requirements						
t_{SFSE}[1]	Frame Sync Setup Before SCLK (Externally Generated Frame Sync in Either Transmit or Receive Mode)	4		2.5		ns
t_{HFSE}[1]	Frame Sync Hold After SCLK (Externally Generated Frame Sync in Either Transmit or Receive Mode)	4		2.5		ns
t_{SDRE}[1]	Receive Data Setup Before Receive SCLK	4		2.5		ns
t_{HDRE}[1]	Receive Data Hold After SCLK	4		2.5		ns
t_{SCLKW}	SCLK Width	$(t_{PCLK} \times 4) \div 2 - 1.5$		$(t_{PCLK} \times 4) \div 2 - 1.5$		ns
t_{SCLK}	SCLK Period	$t_{PCLK} \times 4$		$t_{PCLK} \times 4$		ns
Switching Characteristics						
t_{DFSE}[2]	Frame Sync Delay After SCLK (Internally Generated Frame Sync in Either Transmit or Receive Mode)		15		15	ns
t_{HOFSE}[2]	Frame Sync Hold After SCLK (Internally Generated Frame Sync in Either Transmit or Receive Mode)	2		2		ns
t_{DDTE}[2]	Transmit Data Delay After Transmit SCLK		15		15	ns
t_{HDTE}[2]	Transmit Data Hold After Transmit SCLK	2		2		ns

[1] Referenced to sample edge.
[2] Referenced to drive edge.

Table 36. Serial Ports—External Late Frame Sync

Parameter		88-Lead LFCSP Package Min	88-Lead LFCSP Package Max	All Other Packages Min	All Other Packages Max	Unit
Switching Characteristics						
$t_{DDTLFSE}$[1]	Data Delay from Late External Transmit Frame Sync or External Receive Frame Sync with MCE = 1, MFD = 0		2 × t_{PCLK}		13.5	ns
$t_{DDTENFS}$[1]	Data Enable for MCE = 1, MFD = 0	0.5		0.5		ns

[1] The $t_{DDTLFSE}$ and $t_{DDTENFS}$ parameters apply to left-justified as well as DSP serial mode, and MCE = 1, MFD = 0.

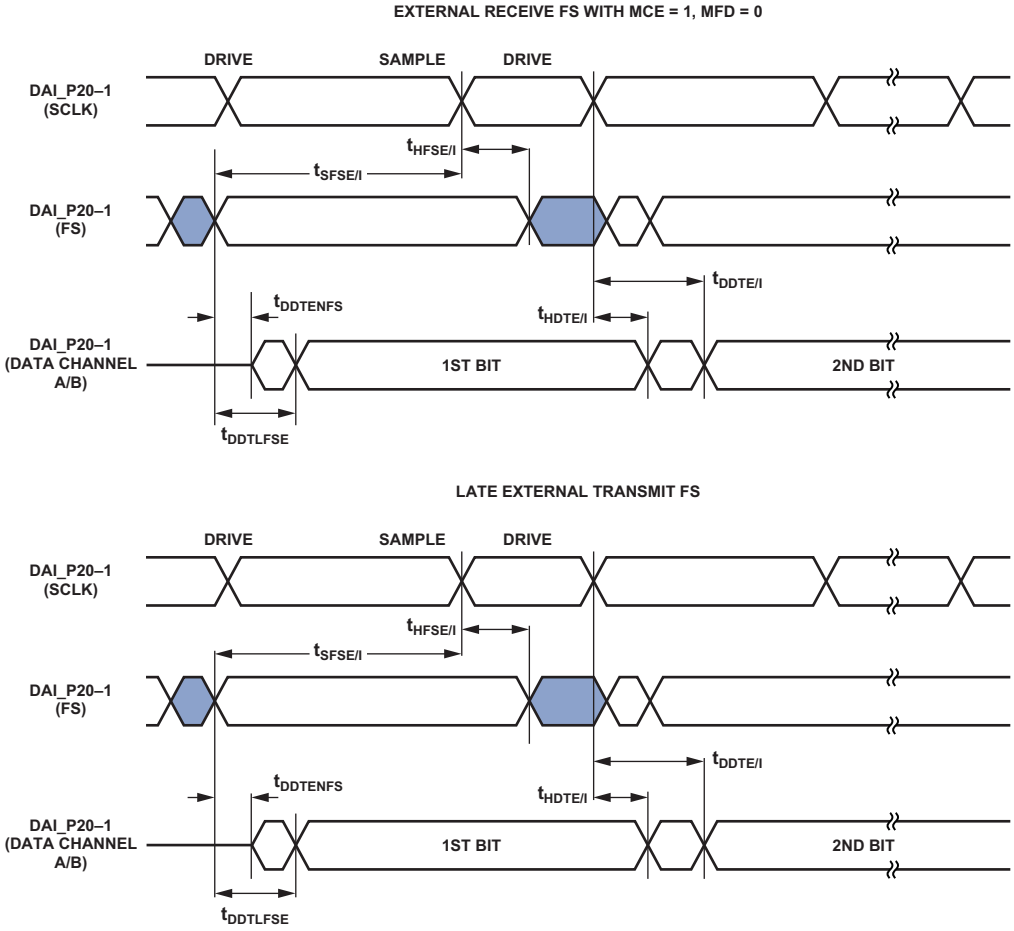

Figure 23. External Late Frame Sync[1]

[1] This figure reflects changes made to support left-justified mode.

Sample Rate Converter—Serial Input Port

The ASRC input signals are routed from the DAI_P20–1 pins using the SRU. Therefore, the timing specifications provided in Table 41 are valid at the DAI_P20–1 pins.

Table 41. ASRC, Serial Input Port

Parameter		Min	Max	Unit
Timing Requirements				
t_{SRCSFS}[1]	Frame Sync Setup Before Serial Clock Rising Edge	4		ns
t_{SRCHFS}[1]	Frame Sync Hold After Serial Clock Rising Edge	5.5		ns
t_{SRCSD}[1]	Data Setup Before Serial Clock Rising Edge	4		ns
t_{SRCHD}[1]	Data Hold After Serial Clock Rising Edge	5.5		ns
$t_{SRCCLKW}$	Clock Width	$(t_{PCLK} \times 4) \div 2 - 1$		ns
t_{SRCCLK}	Clock Period	$t_{PCLK} \times 4$		ns

[1] The serial clock, data, and frame sync signals can come from any of the DAI pins. The serial clock and frame sync signals can also come via PCG or SPORTs. PCG's input can be either CLKIN or any of the DAI pins.

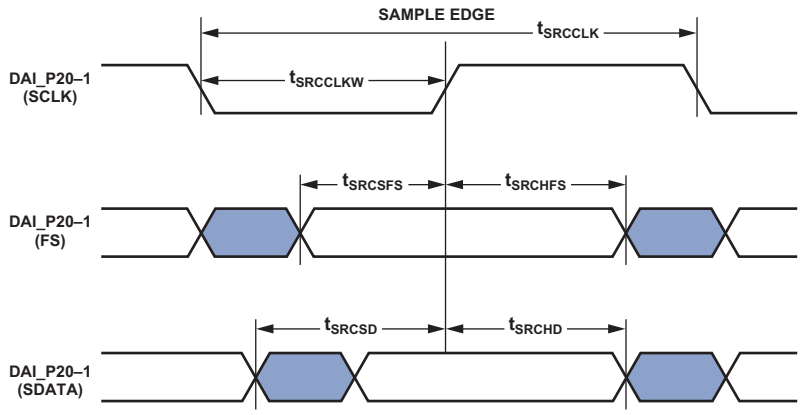

Figure 28. ASRC Serial Input Port Timing

S/PDIF Transmitter

Serial data input to the S/PDIF transmitter can be formatted as left-justified, I²S, or right-justified with word widths of 16, 18, 20, or 24 bits. The following sections provide timing for the transmitter.

S/PDIF Transmitter-Serial Input Waveforms

Figure 31 shows the right-justified mode. Frame sync is high for the left channel and low for the right channel. Data is valid on the rising edge of serial clock. The MSB is delayed the minimum in 24-bit output mode or the maximum in 16-bit output mode from a frame sync transition, so that when there are 64 serial clock periods per frame sync period, the LSB of the data is right-justified to the next frame sync transition.

Figure 32 shows the default I²S-justified mode. The frame sync is low for the left channel and high for the right channel. Data is valid on the rising edge of serial clock. The MSB is left-justified to the frame sync transition but with a delay.

Table 44. S/PDIF Transmitter Right-Justified Mode

Parameter		Nominal	Unit
Timing Requirement			
t_{RJD}	FS to MSB Delay in Right-Justified Mode		
	16-Bit Word Mode	16	SCLK
	18-Bit Word Mode	14	SCLK
	20-Bit Word Mode	12	SCLK
	24-Bit Word Mode	8	SCLK

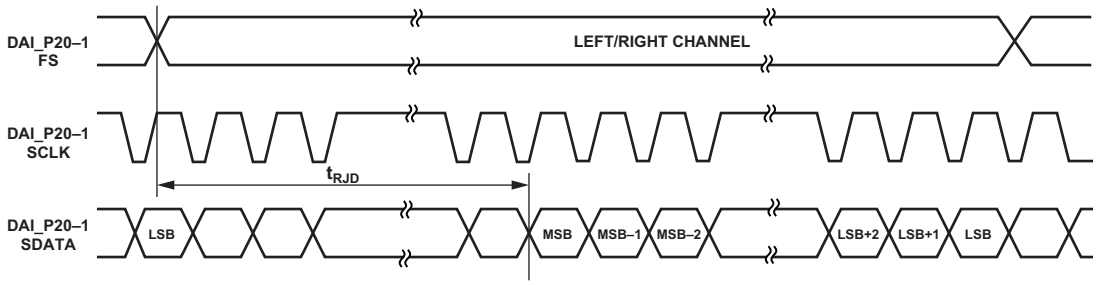

Figure 31. Right-Justified Mode

Table 45. S/PDIF Transmitter I²S Mode

Parameter		Nominal	Unit
Timing Requirement			
t_{I2SD}	FS to MSB Delay in I²S Mode	1	SCLK

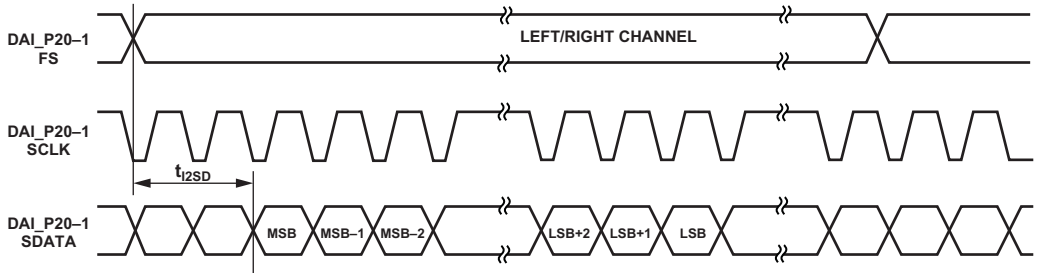

Figure 32. I²S-Justified Mode

Figure 33 shows the left-justified mode. The frame sync is high for the left channel and low for the right channel. Data is valid on the rising edge of serial clock. The MSB is left-justified to the frame sync transition with no delay.

Table 46. S/PDIF Transmitter Left-Justified Mode

Parameter		Nominal	Unit
Timing Requirement			
t_{LJD}	FS to MSB Delay in Left-Justified Mode	0	SCLK

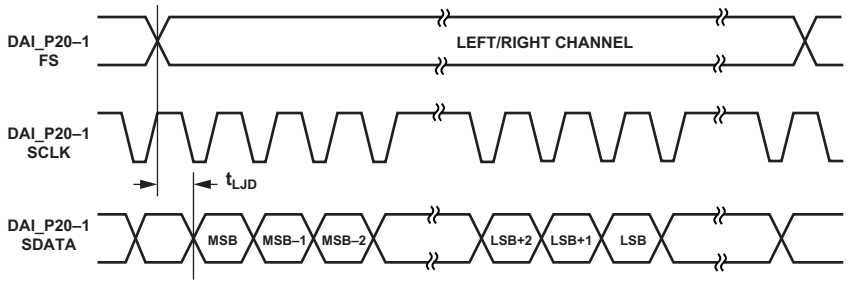

Figure 33. Left-Justified Mode