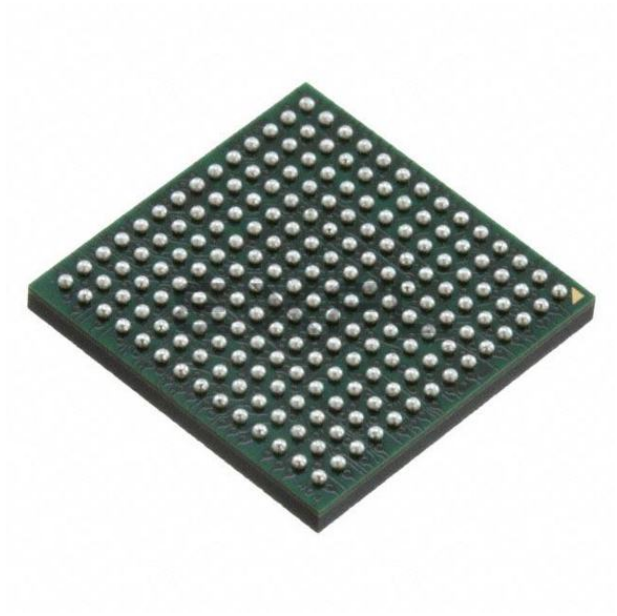

Welcome to E-XFL.COM

Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Obsolete
Type	Floating Point
Interface	DAI, DPI, EBI/EMI, I²C, SPI, SPORT, UART/USART
Clock Rate	266MHz
Non-Volatile Memory	ROM (4Mbit)
On-Chip RAM	5Mbit
Voltage - I/O	3.30V
Voltage - Core	1.20V
Operating Temperature	0°C ~ 70°C
Mounting Type	Surface Mount
Package / Case	196-LFBGA, CSPBGA
Supplier Device Package	196-CSPBGA (12x12)
Purchase URL	https://www.e-xfl.com/product-detail/analog-devices/adsp-21479sbc2-ep

GENERAL DESCRIPTION

The ADSP-2147x SHARC® processors are members of the SIMD SHARC family of DSPs that feature Analog Devices' Super Harvard Architecture. The processors are source code compatible with the ADSP-2126x, ADSP-2136x, ADSP-2137x, ADSP-2146x, and ADSP-2116x DSPs as well as with first generation ADSP-2106x SHARC processors in SISD (single-instruction, single-data) mode. These processors are 32-bit/40-bit floating-point processors optimized for high performance audio applications with a large on-chip SRAM, multiple internal buses to eliminate I/O bottlenecks, and an innovative digital applications interface (DAI).

Table 1 shows performance benchmarks for the ADSP-2147x processors. Table 2 shows the features of the individual product offerings.

Table 1. Processor Benchmarks

Benchmark Algorithm	Speed (at 300 MHz)	Speed (at 200 MHz)
1024 Point Complex FFT (Radix 4, with Reversal)	30.59 µs	45.885 µs
FIR Filter (per Tap)[1]	1.66 ns	2.49 ns
IIR Filter (per Biquad)[1]	6.65 ns	9.975 ns
Matrix Multiply (Pipelined)		
[3 × 3] × [3 × 1]	14.99 ns	22.485 ns
[4 × 4] × [4 × 1]	26.66 ns	39.99 ns
Divide (y/x)	11.61 ns	17.41 ns
Inverse Square Root	18.08 ns	27.12 ns

[1] Assumes two files in multichannel SIMD mode.

Table 2. ADSP-2147x Family Features

Feature	ADSP-21477	ADSP-21478	ADSP-21479
Frequency	200 MHz	Up to 300 MHz	
RAM	2M bits	3M bits	5M bits
ROM	N/A	4M bits	
Pulse-Width Modulation	3	4 units (3 in 100-lead package)	
External Port Interface (SDRAM, AMI)[1]	No	Yes, 16-Bit	
Serial Ports	8		
Direct DMA from SPORTs to External Memory	No	Yes	
FIR, IIR, FFT Accelerator	Yes		
MediaLB Interface	No	Automotive models only	
Watch Dog Timer[2]	No	Yes	
Real-Time Clock[2, 3]	No	Yes	
Shift Register[2]	No	Yes	
IDP/PDAP	Yes		
UART	1		
DAI (SRU)/DPI (SRU2)	20/14 Pins		
S/PDIF Transceiver	1		
SPI	2		
TWI	1		
SRC SNR Performance	–128 dB		
Thermal Diode[4]	Yes		
VISA Support	Yes		
Package[1]	100-Lead LQFP 88-Lead LFCSP_VQ	196-Ball CSP_BGA 100-Lead LQFP 88-lead LFCSP_VQ	

[1] The 100-lead and 88-lead packages of the processors do not contain an external port. The SDRAM controller pins must be disabled when using this package. For more information, see Pin Function Descriptions.
[2] Available on the 196-ball CSP_BGA package only.
[3] Real Time Clock (RTC) is supported only for products with a temperature range of 0°C to +70°C and not supported for all other temperature grades.
[4] Available on the 88-lead and 100-lead packages only.

The diagram on Page 1 shows the two clock domains (core and I/O processor) that make up the ADSP-2147x processors. The core clock domain contains the following features.

- Two processing elements (PEx, PEy), each of which comprises an ALU, multiplier, shifter, and data register file
- Two data address generators (DAG1, DAG2)
- A program sequencer with instruction cache
- PM and DM buses capable of supporting 2 × 64-bit data transfers between memory and the core at every core processor cycle
- One periodic interval timer with pinout
- On-chip SRAM (up to 5M bit)
- A JTAG test access port for emulation and boundary scan. The JTAG provides software debug through user breakpoints, which allows flexible exception handling.

The block diagram of the ADSP-2147x on Page 1 also shows the peripheral clock domain (also known as the I/O processor), which contains the following features:

- IOD0 (peripheral DMA) and IOD1 (external port DMA) buses for 32-bit data transfers
- Peripheral and external port buses for core connection
- External port with an asynchronous memory interface (AMI) and SDRAM controller
- 4 units for pulse width modulation (PWM) control
- 1 memory-to-memory (MTM) unit for internal-to-internal memory transfers
- Digital applications interface that includes four precision clock generators (PCG), an input data port (IDP/PDAP) for serial and parallel interconnect, an S/PDIF receiver/transmitter, four asynchronous sample rate converters, eight serial ports, a shift register, and a flexible signal routing unit (DAI SRU).
- Digital peripheral interface that includes two timers, a 2-wire interface, one UART, two serial peripheral interfaces (SPI), two precision clock generators (PCG), three pulse width modulation (PWM) units, and a flexible signal routing unit (DPI SRU).

As shown in the SHARC core block diagram on Page 5, the processors use two computational units to deliver a significant performance increase over the previous SHARC processors on a range of DSP algorithms. With its SIMD computational hardware, the processors can perform 1.8 GFLOPS running at 300 MHz.

FAMILY CORE ARCHITECTURE

The processors are code compatible at the assembly level with the ADSP-2146x, ADSP-2137x, ADSP-2136x, ADSP-2126x, ADSP-21160, and ADSP-21161, and with the first generation ADSP-2106x SHARC processors. The ADSP-2147x share architectural features with the ADSP-2126x, ADSP-2136x, ADSP-2137x, ADSP-2146x, and ADSP-2116x SIMD SHARC processors, as shown in Figure 2 and detailed in the following sections.

SIMD Computational Engine

The processors contain two computational processing elements that operate as a single-instruction, multiple-data (SIMD) engine. The processing elements are referred to as PEX and PEY and each contains an ALU, multiplier, shifter, and register file. PEX is always active, and PEY may be enabled by setting the PEYEN mode bit in the MODE1 register. SIMD mode allows the processor to execute the same instruction in both processing elements, but each processing element operates on different data. This architecture is efficient at executing math intensive DSP algorithms.

SIMD mode also affects the way data is transferred between memory and the processing elements because twice the data bandwidth is required to sustain computational operation in the processing elements. Therefore, entering SIMD mode also doubles the bandwidth between memory and the processing elements. When using the DAGs to transfer data in SIMD mode, two data values are transferred with each memory or register file access.

SIMD mode is supported from external SDRAM but is not supported in the AMI.

Independent, Parallel Computation Units

Within each processing element is a set of computational units. The computational units consist of an arithmetic/logic unit (ALU), multiplier, and shifter. These units perform all operations in a single cycle. The three units within each processing element are arranged in parallel, maximizing computational throughput. Single multifunction instructions execute parallel ALU and multiplier operations. In SIMD mode, the parallel ALU and multiplier operations occur in both processing elements. These computation units support IEEE 32-bit single-precision floating-point, 40-bit extended precision floating-point, and 32-bit fixed-point data formats.

Timer

The processor contains a core timer that can generate periodic software interrupts. The core timer can be configured to use FLAG3 as a timer expired signal.

Data Register File

Each processing element contains a general-purpose data register file. The register files transfer data between the computation units and the data buses, and store intermediate results. These 10-port, 32-register (16 primary, 16 secondary) register files, combined with the processor's enhanced Harvard architecture, allow unconstrained data flow between computation units and internal memory. The registers in PEX are referred to as R0–R15 and in PEY as S0–S15.

Context Switch

Many of the processor's registers have secondary registers that can be activated during interrupt servicing for a fast context switch. The data registers in the register file, the DAG registers, and the multiplier result registers all have secondary registers. The primary registers are active at reset, while the secondary registers are activated by control bits in a mode control register.

Universal Registers

Universal registers can be used for general-purpose tasks. The USTAT (4) registers allow easy bit manipulations (Set, Clear, Toggle, Test, XOR) for all peripheral control and status registers.

The data bus exchange register (PX) permits data to be passed between the 64-bit PM data bus and the 64-bit DM data bus, or between the 40-bit register file and the PM/DM data bus. These registers contain hardware to handle the data width difference.

Single-Cycle Fetch of Instruction and Four Operands

The processors feature an enhanced Harvard architecture in which the data memory (DM) bus transfers data and the program memory (PM) bus transfers both instructions and data (see Figure 2). With its separate program and data memory

Shift Register

The shift register can be used as a serial to parallel data converter. The shift register module consists of an 18-stage serial shift register, 18-bit latch, and three-state output buffers. The shift register and latch have separate clocks. Data is shifted into the serial shift register on the positive-going transitions of the shift register serial clock (SR_SCLK) input. The data in each flip-flop is transferred to the respective latch on a positive-going transition of the shift register latch clock (SR_LAT) input.

The shift register's signals can be configured as follows.

- The SR_SCLK can come from any of the SPORT0–7 SCLK outputs, PCGA/B clock, any of the DAI pins (1–8), and one dedicated pin (SR_SCLK).
- The SR_LAT can come from any of SPORT0–7 frame sync outputs, PCGA/B frame sync, any of the DAI pins (1–8), and one dedicated pin (SR_LAT).
- The SR_SDI input can from any of SPORT0–7 serial data outputs, any of the DAI pins (1–8), and one dedicated pin (SR_SDI).

Note that the SR_SCLK, SR_LAT, and SR_SDI inputs must come from same source except in the case of where SR_SCLK comes from PCGA/B or SR_SCLK and SR_LAT come from PCGA/B.

If SR_SCLK comes from PCGA/B, then SPORT0–7 generates the SR_LAT and SR_SDI signals. If SR_SCLK and SR_LAT come from PCGA/B, then SPORT0–7 generates the SR_SDI signal.

I/O PROCESSOR FEATURES

The I/O processor provides up to 65 channels of DMA as well as an extensive set of peripherals.

DMA Controller

The DMA controller operates independently and invisibly to the processor core, allowing DMA operations to occur while the core is simultaneously executing its program instructions. DMA transfers can occur between the processor's internal memory and its serial ports, the SPI-compatible (serial peripheral interface) ports, the IDP (input data port), the parallel data acquisition port (PDAP) or the UART.

Up to 65 channels of DMA are available on the processors as shown in Table 9.

Programs can be downloaded using DMA transfers. Other DMA features include interrupt generation upon completion of DMA transfers, and DMA chaining for automatic linked DMA transfers.

Table 9. DMA Channels

Peripheral	DMA Channels
SPORTs	16
PDAP	8
SPI	2
UART	2
External Port	2
Accelerators	2
Memory-to-Memory	2
MediaLB[1]	31

[1] Automotive models only.

Delay Line DMA

The processor provides delay line DMA functionality. This allows processor reads and writes to external delay line buffers (and therefore to external memory) with limited core interaction.

Scatter/Gather DMA

The processor provides scatter/gather DMA functionality. This allows processor DMA reads/writes to/from noncontiguous memory blocks.

FFT Accelerator

The FFT accelerator implements radix-2 complex/real input, complex output FFTs with no core intervention. The FFT accelerator runs at the peripheral clock frequency.

FIR Accelerator

The FIR (finite impulse response) accelerator consists of a 1024 word coefficient memory, a 1024 word deep delay line for the data, and four MAC units. A controller manages the accelerator. The FIR accelerator runs at the peripheral clock frequency.

IIR Accelerator

The IIR (infinite impulse response) accelerator consists of a 1440 word coefficient memory for storage of biquad coefficients, a data memory for storing the intermediate data and one MAC unit. A controller manages the accelerator. The IIR accelerator runs at the peripheral clock frequency.

Watchdog Timer (WDT)

The processors include a 32-bit watchdog timer that can be used to implement a software watchdog function. A software watchdog can improve system reliability by forcing the processor to a known state through generation of a system reset if the timer expires before being reloaded by software. Software initializes the count value of the timer, and then enables the timer.

The WDT is used to supervise the stability of the system software. When used in this way, software reloads the WDT in a regular manner so that the downward counting timer never expires. An expiring timer then indicates that system software might be out of control.

The WDT resets both the core and the internal peripherals. Software must be able to determine if the watch dog was the source of the hardware reset by interrogating a status bit in the watch dog timer control register.

PIN FUNCTION DESCRIPTIONS

Table 11. Pin Descriptions

Name	Type	State During/ After Reset	Description
$ADDR_{23-0}$	I/O/T (ipu)	High-Z/driven low (boot)	**External Address.** The processor outputs addresses for external memory and peripherals on these pins. The ADDR pins can be multiplexed to support the external memory interface address, FLAGS15–8 (I/O) and PWM (O). After reset, all ADDR pins are in EMIF mode, and FLAG(0–3) pins are in FLAGS mode (default). When configured in the IDP_PDAP_CTL register, IDP channel 0 scans the $ADDR_{23-4}$ pins for parallel input data.
$DATA_{15-0}$	I/O/T (ipu)	High-Z	**External Data.** The data pins can be multiplexed to support the external memory interface data (I/O) and $FLAGS_{7-0}$ (I/O).
AMI_ACK	I (ipu)		**Memory Acknowledge.** External devices can deassert AMI_ACK (low) to add wait states to an external memory access. AMI_ACK is used by I/O devices, memory controllers, or other peripherals to hold off completion of an external memory access.
$\overline{MS}_{0-1}$	O/T (ipu)	High-Z	**Memory Select Lines 0–1.** These lines are asserted (low) as chip selects for the corresponding banks of external memory. The $\overline{MS}_{1-0}$ lines are decoded memory address lines that change at the same time as the other address lines. When no external memory access is occurring the $\overline{MS}_{1-0}$ lines are inactive; they are active however when a conditional memory access instruction is executed, when the condition evaluates as true. The $\overline{MS1}$ pin can be used in EPORT/FLASH boot mode. For more information on processor booting, see the *ADSP-214xx SHARC Processor Hardware Reference*.
$\overline{AMI\_RD}$	O/T (ipu)	High-Z	**AMI Port Read Enable.** $\overline{AMI\_RD}$ is asserted whenever the processor reads a word from external memory.
$\overline{AMI\_WR}$	O/T (ipu)	High-Z	**AMI Port Write Enable.** $\overline{AMI\_WR}$ is asserted when the processor writes a word to external memory.
FLAG0/$\overline{IRQ0}$	I/O (ipu)	FLAG[0] INPUT	**FLAG0/Interrupt Request0.**
FLAG1/$\overline{IRQ1}$	I/O (ipu)	FLAG[1] INPUT	**FLAG1/Interrupt Request1.**
FLAG2/$\overline{IRQ2}$/$\overline{MS2}$	I/O (ipu)	FLAG[2] INPUT	**FLAG2/Interrupt Request2/Memory Select2.** This pin is multiplexed with $\overline{MS2}$ in the 196-ball BGA package only.
FLAG3/TMREXP/$\overline{MS3}$	I/O (ipu)	FLAG[3] INPUT	**FLAG3/Timer Expired/Memory Select3.** This pin is multiplexed with $\overline{MS3}$ in the 196-ball BGA package only.

The following symbols appear in the Type column of Table 11: **A** = asynchronous, **I** = input, **O** = output, **S** = synchronous, **A/D** = active drive, **O/D** = open drain, and **T** = three-state, **ipd** = internal pull-down resistor, **ipu** = internal pull-up resistor.

The internal pull-up (ipu) and internal pull-down (ipd) resistors are designed to hold the internal path from the pins at the expected logic levels. To pull-up or pull-down the external pads to the expected logic levels, use external resistors. Internal pull-up/pull-down resistors cannot be enabled/disabled and the value of these resistors cannot be programmed. The range of an ipu resistor can be 26 kΩ to 63 kΩ. The range of an ipd resistor can be 31 kΩ to 85 kΩ. The three-state voltage of ipu pads will not reach to full the V_{DD_EXT} level; at typical conditions the voltage is in the range of 2.3 V to 2.7 V.

In this table, all pins are LVTTL compliant with the exception of the thermal diode, shift register, and real-time clock (RTC) pins.

Not all pins are available in the 88-lead LFCSP_VQ and 100-lead LQFP package. For more information, see Table 2 on Page 3 and Table 62 on Page 70.

Table 11. Pin Descriptions (Continued)

Name	Type	State During/ After Reset	Description
THD_P	I		**Thermal Diode Anode.** When not used, this pin can be left floating.
THD_M	O		**Thermal Diode Cathode.** When not used, this pin can be left floating.
MLBCLK	I		**Media Local Bus Clock.** This clock is generated by the MLB controller that is synchronized to the MOST network and provides the timing for the entire MLB interface at 49.152 MHz at FS = 48 kHz. When the MLB controller is not used, this pin should be grounded.
MLBDAT	I/O/T in 3 pin mode. I in 5 pin mode.	High-Z	**Media Local Bus Data.** The MLBDAT line is driven by the transmitting MLB device and is received by all other MLB devices including the MLB controller. The MLBDAT line carries the actual data. In 5-pin MLB mode, this pin is an input only. When the MLB controller is not used, this pin should be grounded.
MLBSIG	I/O/T in 3 pin mode. I in 5 pin mode	High-Z	**Media Local Bus Signal.** This is a multiplexed signal which carries the Channel/Address generated by the MLB Controller, as well as the Command and RxStatus bytes from MLB devices. In 5-pin mode, this pin is input only. When the MLB controller is not used, this pin should be grounded.
MLBDO	O/T	High-Z	**Media Local Bus Data Output (in 5 Pin Mode).** This pin is used only in 5-pin MLB mode and serves as the output data pin. When the MLB controller is not used, this pin should be grounded.
MLBSO	O/T	High-Z	**Media Local Bus Signal Output (in 5 Pin Mode).** This pin is used only in 5-pin MLB mode and serves as the output signal pin. When the MLB controller is not used, this pin should be grounded.
SR_SCLK	I (ipu)		**Shift Register Serial Clock.** (Active high, rising edge sensitive)
$\overline{\text{SR\_CLR}}$	I (ipu)		**Shift Register Reset.** (Active low)
SR_SDI	I (ipu)		**Shift Register Serial Data Input.**
SR_SDO	O (ipu)	Driven Low	**Shift Register Serial Data Output.**
SR_LAT	I (ipu)		**Shift Register Latch Clock Input.** (Active high, rising edge sensitive)
SR_LDO_{17-0}	O/T (ipu)	High-Z	**Shift Register Parallel Data Output.**
RTXI	I		**RTC Crystal Input.** If RTC is not used, then this pin can be NC (no connect) or grounded and the RTC_PDN and RTC_BUSDIS bits of the RTC_INIT register must be set to 1.
RTXO	O		**RTC Crystal Output.** If RTC is not used, then this pin needs to be NC (No Connect).
RTCLKOUT	O (ipd)		**RTC Clock Output.** For calibration purposes. The clock runs at 1 Hz. If RTC is not used, then this pin needs to be NC (No Connect).

The following symbols appear in the Type column of Table 11: **A** = asynchronous, **I** = input, **O** = output, **S** = synchronous, **A/D** = active drive, **O/D** = open drain, and **T** = three-state, **ipd** = internal pull-down resistor, **ipu** = internal pull-up resistor.

The internal pull-up (ipu) and internal pull-down (ipd) resistors are designed to hold the internal path from the pins at the expected logic levels. To pull-up or pull-down the external pads to the expected logic levels, use external resistors. Internal pull-up/pull-down resistors cannot be enabled/disabled and the value of these resistors cannot be programmed. The range of an ipu resistor can be 26 kΩ to 63 kΩ. The range of an ipd resistor can be 31 kΩ to 85 kΩ. The three-state voltage of ipu pads will not reach to full the V_{DD_EXT} level; at typical conditions the voltage is in the range of 2.3 V to 2.7 V.

In this table, all pins are LVTTL compliant with the exception of the thermal diode, shift register, and real-time clock (RTC) pins.

Not all pins are available in the 88-lead LFCSP_VQ and 100-lead LQFP package. For more information, see Table 2 on Page 3 and Table 62 on Page 70.

SPECIFICATIONS

For information about product specifications, contact your Analog Devices, Inc. representative.

OPERATING CONDITIONS

Parameter[1]	Description	200 MHz			266 MHz			300 MHz			Unit
		Min	Nom	Max	Min	Nom	Max	Min	Nom	Max	
V_{DD_INT}	Internal (Core) Supply Voltage	1.14	1.2	1.26	1.14	1.2	1.26	1.25	1.3	1.35	V
V_{DD_EXT}	External (I/O) Supply Voltage	3.13	3.3	3.47	3.13	3.3	3.47	3.13	3.3	3.47	V
V_{DD_THD}	Thermal Diode Supply Voltage	3.13	3.3	3.47	3.13	3.3	3.47	3.13	3.3	3.47	V
V_{DD_RTC}	Real-Time Clock Power Supply Voltage	2.0	3.0	3.6	2.0	3.0	3.6	2.0	3.0	3.6	V
V_{IH}[2]	High Level Input Voltage @ V_{DD_EXT} = Max	2.0			2.0			2.0			V
V_{IL}[3]	Low Level Input Voltage @ V_{DD_EXT} = Min			0.8			0.8			0.8	V
V_{IH_CLKIN}[3]	High Level Input Voltage @ V_{DD_EXT} = Max	2.2		V_{DD_EXT}	2.2		V_{DD_EXT}	2.2		V_{DD_EXT}	V
V_{IL_CLKIN}	Low Level Input Voltage @ V_{DD_EXT} = Max	–0.3		+0.8	–0.3		+0.8	–0.3		+0.8	V
T_J	Junction Temperature 88-Lead LFCSP_VQ @ $T_{AMBIENT}$ 0°C to +70°C	0		105	N/A		N/A	N/A		N/A	°C
T_J	Junction Temperature 88-Lead LFCSP_VQ @ $T_{AMBIENT}$ –40°C to +85°C	–40		+115	N/A		N/A	N/A		N/A	°C
T_J	Junction Temperature 100-Lead LQFP_EP @ $T_{AMBIENT}$ 0°C to +70°C	0		105	0		105	N/A		N/A	°C
T_J	Junction Temperature 100-Lead LQFP_EP @ $T_{AMBIENT}$ –40°C to +85°C	N/A		N/A	–40		+125	N/A		N/A	°C
T_J[4]	Junction Temperature 196-Ball CSP_BGA @ $T_{AMBIENT}$ 0°C to +70°C	N/A		N/A	0		105	0		100	°C
T_J	Junction Temperature 196-Ball CSP_BGA @ $T_{AMBIENT}$ –40°C to +85°C	N/A		N/A	–40		+125	N/A		N/A	°C
AUTOMOTIVE USE ONLY											
T_J	Junction Temperature 88-Lead LFCSP_VQ @ $T_{AMBIENT}$ –40°C to +105°C (Automotive Grade)	–40		+125[5]	N/A		N/A	N/A		N/A	°C
T_J	Junction Temperature 100-Lead LQFP_EP @ $T_{AMBIENT}$ –40°C to +105°C (Automotive Grade)	–40		+125[5]	–40		+125	N/A		N/A	°C
T_J[4]	Junction Temperature 196-Ball CSP_BGA @ $T_{AMBIENT}$ –40°C to +105°C (Automotive Grade)	–40		+125[5]	N/A		N/A	N/A		N/A	°C

[1] Specifications subject to change without notice.

[2] Applies to input and bidirectional pins: ADDR23–0, DATA15–0, FLAG3–0, DAI_Px, DPI_Px, BOOT_CFGx, CLK_CFGx, $\overline{\text{RUNRSTIN}}$, $\overline{\text{RESET}}$, TCK, TMS, TDI, $\overline{\text{TRST}}$, SDA10, AMI_ACK, MLBCLK, MLBDAT, MLBSIG.

[3] Applies to input pin CLKIN, WDT_CLKIN.

[4] Real Time Clock (RTC) is supported only for products in the BGA package with a temperature range of 0°C to +70°C. For the status of unused RTC pins please see Table 11.

[5] Automotive application use profile only. Not supported for nonautomotive use. Contact Analog Devices for more information.

Running Reset

The following timing specification applies to $\overline{\text{RESETOUT}}$/$\overline{\text{RUNRSTIN}}$ pin when it is configured as $\overline{\text{RUNRSTIN}}$.

Table 22. Running Reset

Parameter		Min	Max	Unit
Timing Requirements				
$t_{WRUNRST}$	Running $\overline{\text{RESET}}$ Pulse Width Low	$4 \times t_{CK}$		ns
$t_{SRUNRST}$	Running $\overline{\text{RESET}}$ Setup Before CLKIN High	8		ns

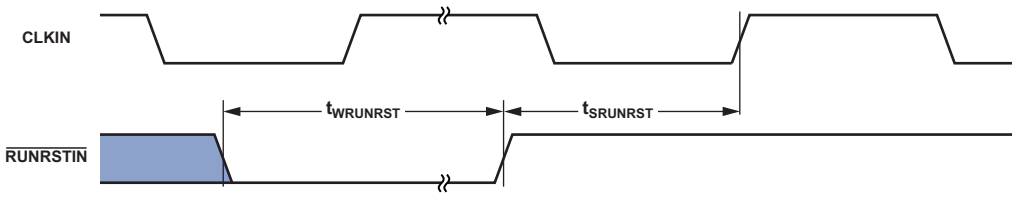

Figure 10. Running Reset

Interrupts

The following timing specification applies to the FLAG0, FLAG1, and FLAG2 pins when they are configured as $\overline{\text{IRQ0}}$, $\overline{\text{IRQ1}}$, and $\overline{\text{IRQ2}}$ interrupts, as well as the DAI_P20–1 and DPI_P14–1 pins when they are configured as interrupts.

Table 23. Interrupts

Parameter		Min	Max	Unit
Timing Requirement				
t_{IPW}	$\overline{\text{IRQx}}$ Pulse Width	$2 \times t_{PCLK} + 2$		ns

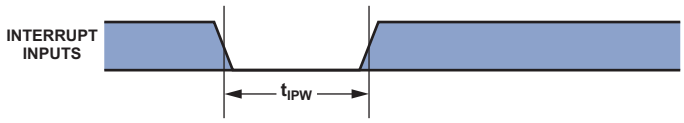

Figure 11. Interrupts

AMI Read

Use these specifications for asynchronous interfacing to memories. Note that timing for AMI_ACK, ADDR, DATA, $\overline{\text{AMI\_RD}}$, $\overline{\text{AMI\_WR}}$, and strobe timing parameters only apply to asynchronous access mode.

Table 32. AMI Read

Parameter		Min	Max	Unit
Timing Requirements				
t_{DAD}[1, 2, 3]	Address Selects Delay to Data Valid		W + t_{SDCLK} – 6.32	ns
t_{DRLD}[1, 3]	$\overline{\text{AMI\_RD}}$ Low to Data Valid		W – 3	ns
t_{SDS}[4, 5]	Data Setup to $\overline{\text{AMI\_RD}}$ High	2.6		ns
t_{HDRH}	Data Hold from $\overline{\text{AMI\_RD}}$ High	0.4		ns
t_{DAAK}[2, 6]	AMI_ACK Delay from Address Selects		t_{SDCLK} – 10 + W	ns
t_{DSAK}[4]	AMI_ACK Delay from $\overline{\text{AMI\_RD}}$ Low		W – 7.0	ns
Switching Characteristics				
t_{DRHA}	Address Selects Hold After $\overline{\text{AMI\_RD}}$ High	RHC + 0.38		ns
t_{DARL}[2]	Address Selects to $\overline{\text{AMI\_RD}}$ Low	t_{SDCLK} – 5		ns
t_{RW}	$\overline{\text{AMI\_RD}}$ Pulse Width	W – 1.4		ns
t_{RWR}	$\overline{\text{AMI\_RD}}$ High to $\overline{\text{AMI\_RD}}$ Low	HI + t_{SDCLK} – 1.2		ns

W = (number of wait states specified in AMICTLx register) × t_{SDCLK}.
RHC = (number of Read Hold Cycles specified in AMICTLx register) × t_{SDCLK}
Where PREDIS = 0
HI = RHC (if IC = 0): Read to Read from same bank
HI = RHC+ t_{SDCLK} (if IC > 0): Read to Read from same bank
HI = RHC + IC: Read to Read from different bank
HI = RHC + Max (IC, (4 × t_{SDCLK})): Read to Write from same or different bank
Where PREDIS = 1
HI = RHC + Max (IC, (4 × t_{SDCLK})): Read to Write from same or different bank
HI = RHC + (3 × t_{SDCLK}): Read to Read from same bank
HI = RHC + Max (IC, (3 × t_{SDCLK})): Read to Read from different bank
IC = (number of idle cycles specified in AMICTLx register) × t_{SDCLK}
H = (number of hold cycles specified in AMICTLx register) × t_{SDCLK}.

[1] Data delay/setup: System must meet t_{DAD}, t_{DRLD}, or t_{SDS}.
[2] The falling edge of $\overline{\text{AMI_MS}}$x, is referenced.
[3] The maximum limit of timing requirement values for t_{DAD} and t_{DRLD} parameters are applicable for the case where AMI_ACK is always high and when the ACK feature is not used.
[4] Note that timing for AMI_ACK, ADDR, DATA, $\overline{\text{AMI\_RD}}$, $\overline{\text{AMI\_WR}}$, and strobe timing parameters only apply to asynchronous access mode.
[5] Data hold: User must meet t_{HDRH} in asynchronous access mode. See Test Conditions for the calculation of hold times given capacitive and dc loads.
[6] AMI_ACK delay/setup: User must meet t_{daak}, or t_{dsak}, for deassertion of AMI_ACK (low).

AMI Write

Use these specifications for asynchronous interfacing to memories. Note that timing for AMI_ACK, ADDR, DATA, $\overline{\text{AMI\_RD}}$, $\overline{\text{AMI\_WR}}$, and strobe timing parameters only apply to asynchronous access mode.

Table 33. AMI Write

Parameter		Min	Max	Unit
Timing Requirements				
t_{DAAK}	AMI_ACK Delay from Address Selects[1, 2]		t_{SDCLK} – 10.1 + W	ns
t_{DSAK}	AMI_ACK Delay from $\overline{\text{AMI\_WR}}$ Low[1, 3]		W – 7.1	ns
Switching Characteristics				
t_{DAWH}	Address Selects to $\overline{\text{AMI\_WR}}$ Deasserted[2]	t_{SDCLK} – 4.4 + W		ns
t_{DAWL}	Address Selects to $\overline{\text{AMI\_WR}}$ Low[2]	t_{SDCLK} – 4.5		ns
t_{WW}	$\overline{\text{AMI\_WR}}$ Pulse Width	W – 1.3		ns
t_{DDWH}	Data Setup Before $\overline{\text{AMI\_WR}}$ High	t_{SDCLK} – 4.3 + W		ns
t_{DWHA}	Address Hold After $\overline{\text{AMI\_WR}}$ Deasserted	H		ns
t_{DWHD}	Data Hold After $\overline{\text{AMI\_WR}}$ Deasserted	H		ns
t_{DATRWH}	Data Disable After $\overline{\text{AMI\_WR}}$ Deasserted[4]	t_{SDCLK} – 1.37 + H	t_{SDCLK} + 6.75 + H	ns
t_{WWR}	$\overline{\text{AMI\_WR}}$ High to $\overline{\text{AMI\_WR}}$ Low[5]	t_{SDCLK} – 1.5 + H		ns
t_{DDWR}	Data Disable Before $\overline{\text{AMI\_RD}}$ Low	2 × t_{SDCLK} – 7.1		ns
t_{WDE}	Data Enabled to $\overline{\text{AMI\_WR}}$ Low	t_{SDCLK} – 4.5		ns

W = (number of wait states specified in AMICTLx register) × t_{SDCLK}
H = (number of hold cycles specified in AMICTLx register) × t_{SDCLK}

[1] AMI_ACK delay/setup: System must meet t_{DAAK}, or t_{DSAK}, for deassertion of AMI_ACK (low).
[2] The falling edge of $\overline{\text{AMI\_MSx}}$ is referenced.
[3] Note that timing for AMI_ACK, ADDR, DATA, $\overline{\text{AMI\_RD}}$, $\overline{\text{AMI\_WR}}$, and strobe timing parameters only applies to asynchronous access mode.
[4] See Test Conditions for calculation of hold times given capacitive and dc loads.
[5] For Write to Write: t_{SDCLK} + H, for both same bank and different bank. For Write to Read: 3 × t_{SDCLK} + H, for the same bank and different banks.

Parallel Data Acquisition Port (PDAP)

The timing requirements for the PDAP are provided in Table 40. PDAP is the parallel mode operation of Channel 0 of the IDP. For details on the operation of the PDAP, see the PDAP chapter of the *ADSP-214xx SHARC Processor Hardware Reference*. Note that the 20 bits of external PDAP data can be provided through the ADDR23–0 pins or over the DAI pins.

Table 40. Parallel Data Acquisition Port (PDAP)

Parameter		88-Lead LFCSP Package Min	88-Lead LFCSP Package Max	All Other Packages Min	All Other Packages Max	Unit
Timing Requirements						
t_{SPHOLD}[1]	PDAP_HOLD Setup Before PDAP_CLK Sample Edge	4		2.5		ns
t_{HPHOLD}[1]	PDAP_HOLD Hold After PDAP_CLK Sample Edge	4		2.5		ns
t_{PDSD}[1]	PDAP_DAT Setup Before SCLK PDAP_CLK Sample Edge	5		3.85		ns
t_{PDHD}[1]	PDAP_DAT Hold After SCLK PDAP_CLK Sample Edge	4		2.5		ns
t_{PDCLKW}	Clock Width	$(t_{PCLK} \times 4) \div 2 - 3$		$(t_{PCLK} \times 4) \div 2 - 3$		ns
t_{PDCLK}	Clock Period	$t_{PCLK} \times 4$		$t_{PCLK} \times 4$		ns
Switching Characteristics						
t_{PDHLDD}	Delay of PDAP Strobe After Last PDAP_CLK Capture Edge for a Word	$2 \times t_{PCLK} + 3$		$2 \times t_{PCLK} + 3$		ns
t_{PDSTRB}	PDAP Strobe Pulse Width	$2 \times t_{PCLK} - 1.5$		$2 \times t_{PCLK} - 1.5$		ns

[1] Source pins of DATA and control are ADDR23–0 or DAI pins. Source pins for SCLK and FS are: 1) DAI pins, 2) CLKIN through PCG, or 3) DAI pins through PCG.

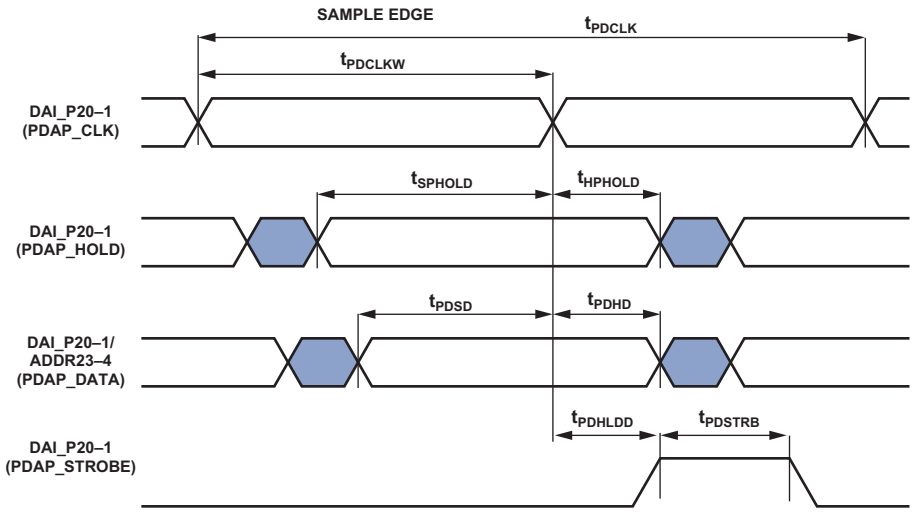

Figure 27. PDAP Timing

Pulse-Width Modulation Generators (PWM)

The following timing specifications apply when the ADDR23–8/DPI_14–1 pins are configured as PWM.

Table 43. Pulse-Width Modulation (PWM) Timing

Parameter		88-Lead LFCSP Package Min	88-Lead LFCSP Package Max	All Other Packages Min	All Other Packages Max	Unit
Switching Characteristics						
t_{PWMW}	PWM Output Pulse Width	$t_{PCLK} - 2$	$(2^{16} - 2) \times t_{PCLK}$	$t_{PCLK} - 2$	$(2^{16} - 2) \times t_{PCLK}$	ns
t_{PWMP}	PWM Output Period	$2 \times t_{PCLK} - 2$	$(2^{16} - 1) \times t_{PCLK}$	$2 \times t_{PCLK} - 1.5$	$(2^{16} - 1) \times t_{PCLK}$	ns

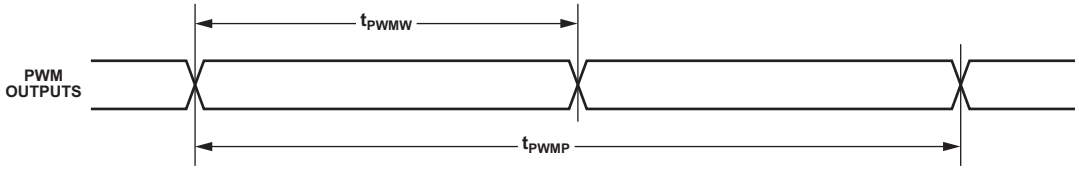

Figure 30. PWM Timing

S/PDIF Transmitter

Serial data input to the S/PDIF transmitter can be formatted as left-justified, I²S, or right-justified with word widths of 16, 18, 20, or 24 bits. The following sections provide timing for the transmitter.

S/PDIF Transmitter-Serial Input Waveforms

Figure 31 shows the right-justified mode. Frame sync is high for the left channel and low for the right channel. Data is valid on the rising edge of serial clock. The MSB is delayed the minimum in 24-bit output mode or the maximum in 16-bit output mode from a frame sync transition, so that when there are 64 serial clock periods per frame sync period, the LSB of the data is right-justified to the next frame sync transition.

Figure 32 shows the default I²S-justified mode. The frame sync is low for the left channel and high for the right channel. Data is valid on the rising edge of serial clock. The MSB is left-justified to the frame sync transition but with a delay.

Table 44. S/PDIF Transmitter Right-Justified Mode

Parameter		Nominal	Unit
Timing Requirement			
t_{RJD}	FS to MSB Delay in Right-Justified Mode		
	16-Bit Word Mode	16	SCLK
	18-Bit Word Mode	14	SCLK
	20-Bit Word Mode	12	SCLK
	24-Bit Word Mode	8	SCLK

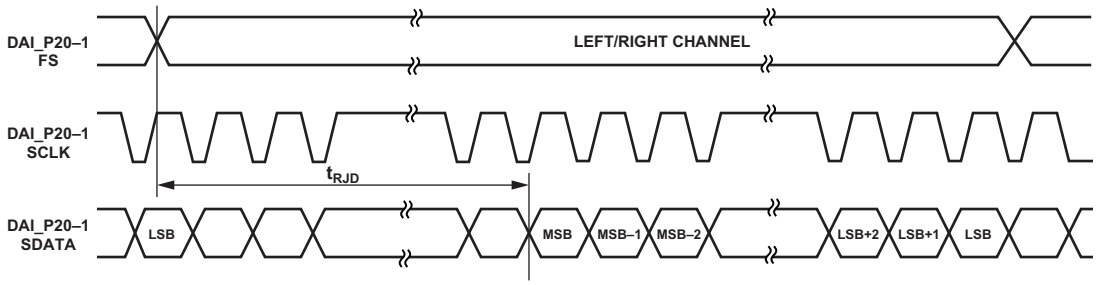

Figure 31. Right-Justified Mode

Table 45. S/PDIF Transmitter I²S Mode

Parameter		Nominal	Unit
Timing Requirement			
t_{I2SD}	FS to MSB Delay in I²S Mode	1	SCLK

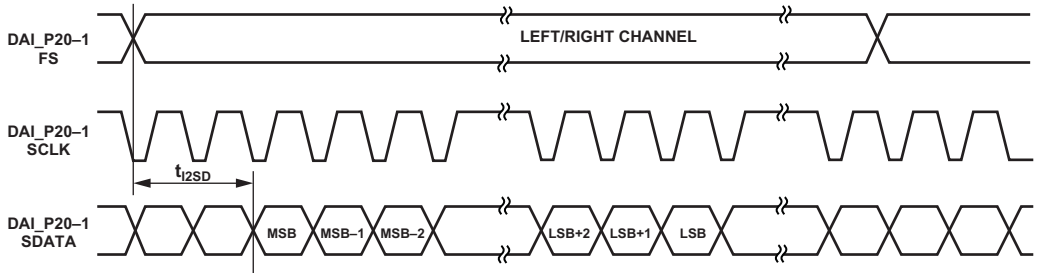

Figure 32. I²S-Justified Mode

Figure 33 shows the left-justified mode. The frame sync is high for the left channel and low for the right channel. Data is valid on the rising edge of serial clock. The MSB is left-justified to the frame sync transition with no delay.

Table 46. S/PDIF Transmitter Left-Justified Mode

Parameter		Nominal	Unit
Timing Requirement			
t_{LJD}	FS to MSB Delay in Left-Justified Mode	0	SCLK

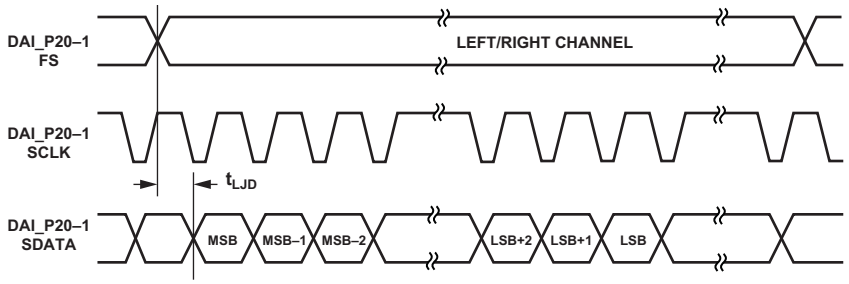

Figure 33. Left-Justified Mode

S/PDIF Receiver

The following section describes timing as it relates to the S/PDIF receiver.

Internal Digital PLL Mode

In the internal digital phase-locked loop mode the internal PLL (digital PLL) generates the 512 × FS clock.

Table 49. S/PDIF Receiver Internal Digital PLL Mode Timing

Parameter		Min	Max	Unit
Switching Characteristics				
t_{DFSI}	FS Delay After Serial Clock		5	ns
t_{HOFSI}	FS Hold After Serial Clock	–2		ns
t_{DDTI}	Transmit Data Delay After Serial Clock		5	ns
t_{HDTI}	Transmit Data Hold After Serial Clock	–2		ns
t_{SCLKIW}[1]	Transmit Serial Clock Width	38.5		ns

[1] The serial clock frequency is 64 × frame sync (FS) where FS = the frequency of LRCLK.

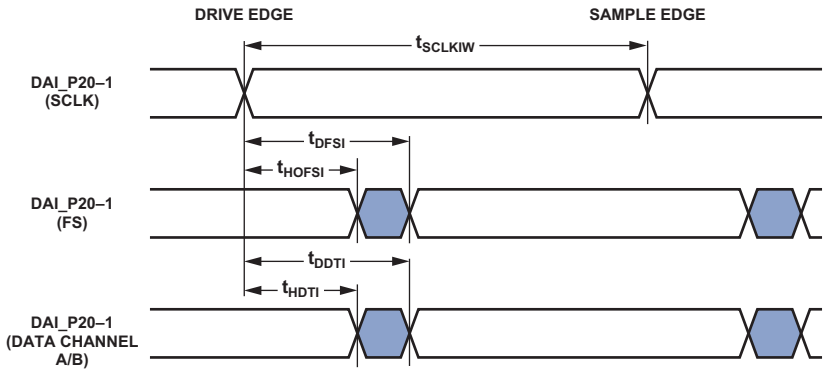

Figure 35. S/PDIF Receiver Internal Digital PLL Mode Timing

SPI Interface—Master

Both the primary and secondary SPIs are available through DPI only. The timing provided in Table 50 and Table 51 applies to both.

Table 50. SPI Interface Protocol—Master Switching and Timing Specifications

Parameter		88-Lead LFCSP Package Min	88-Lead LFCSP Package Max	All Other Packages Min	All Other Packages Max	Unit
Timing Requirements						
t_{SSPIDM}	Data Input Valid to SPICLK Edge (Data Input Setup Time)	10		8.6		ns
t_{HSPIDM}	SPICLK Last Sampling Edge to Data Input Not Valid	2		2		ns
Switching Characteristics						
$t_{SPICLKM}$	Serial Clock Cycle	$8 \times t_{PCLK} - 2$		$8 \times t_{PCLK} - 2$		ns
t_{SPICHM}	Serial Clock High Period	$4 \times t_{PCLK} - 2$		$4 \times t_{PCLK} - 2$		ns
t_{SPICLM}	Serial Clock Low Period	$4 \times t_{PCLK} - 2$		$4 \times t_{PCLK} - 2$		ns
$t_{DDSPIDM}$	SPICLK Edge to Data Out Valid (Data Out Delay time)		2.5		2.5	
$t_{HDSPIDM}$	SPICLK Edge to Data Out Not Valid (Data Out Hold time)	$4 \times t_{PCLK} - 2$		$4 \times t_{PCLK} - 2$		ns
t_{SDSCIM}	DPI Pin (SPI Device Select) Low to First SPICLK Edge	$4 \times t_{PCLK} - 2$		$4 \times t_{PCLK} - 2$		ns
t_{HDSM}	Last SPICLK Edge to DPI Pin (SPI Device Select) High	$4 \times t_{PCLK} - 2$		$4 \times t_{PCLK} - 2$		ns
t_{SPITDM}	Sequential Transfer Delay	$4 \times t_{PCLK} - 2$		$4 \times t_{PCLK} - 1.4$		ns

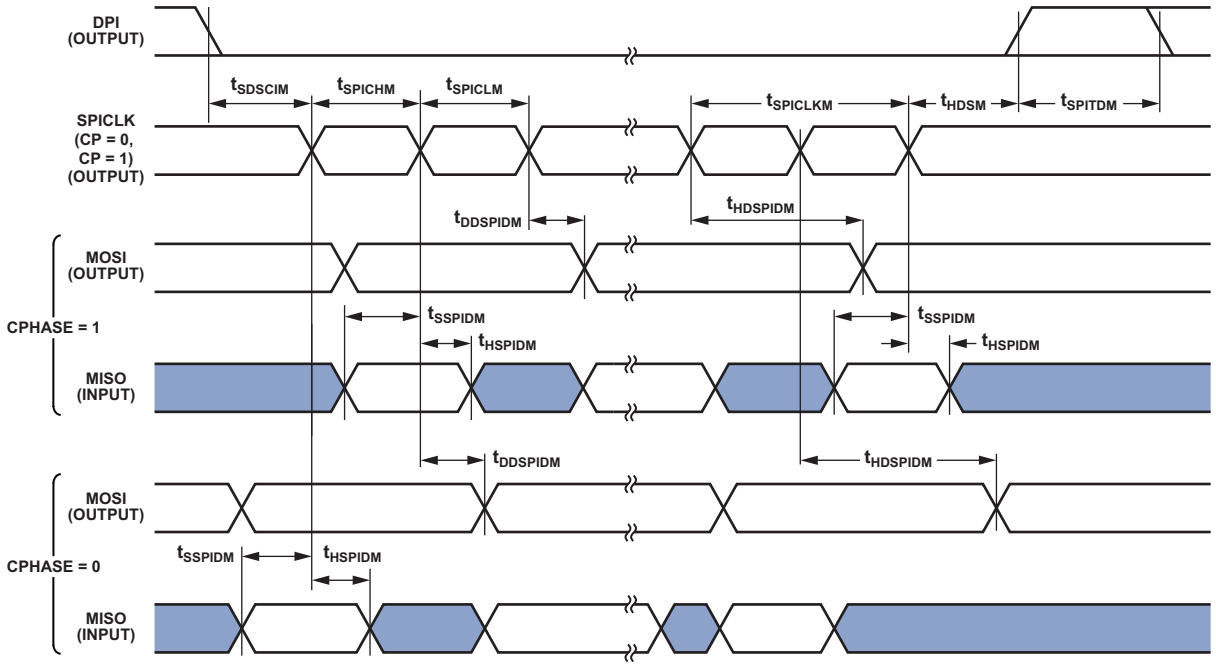

Figure 36. SPI Master Timing

Media Local Bus

All the numbers given are applicable for all speed modes (1024 FS, 512 FS, and 256 FS for 3-pin; 512 FS and 256 FS for 5-pin) unless otherwise specified. Please refer to MediaLB specification document rev 3.0 for more details.

Table 52. MLB Interface, 3-Pin Specifications

Parameter		Min	Typ	Max	Unit
3-Pin Characteristics					
t_{MLBCLK}	MLB Clock Period				
	1024 FS		20.3		ns
	512 FS		40		ns
	256 FS		81		ns
t_{MCKL}	MLBCLK Low Time				
	1024 FS	6.1			ns
	512 FS	14			ns
	256 FS	30			ns
t_{MCKH}	MLBCLK High Time				
	1024 FS	9.3			ns
	512 FS	14			ns
	256 FS	30			ns
t_{MCKR}	MLBCLK Rise Time (V_{IL} to V_{IH})				
	1024 FS			1	ns
	512 FS/256 FS			3	ns
t_{MCKF}	MLBCLK Fall Time (V_{IH} to V_{IL})				
	1024 FS			1	ns
	512 FS/256 FS			3	ns
t_{MPWV}[1]	MLBCLK Pulse Width Variation				
	1024 FS			0.7	ns p-p
	512 FS/256			2.0	ns p-p
t_{DSMCF}	DAT/SIG Input Setup Time	1			ns
t_{DHMCF}	DAT/SIG Input Hold Time	1.2			ns
t_{MCFDZ}	DAT/SIG Output Time to Three-State	0		15	ns
t_{MCDRV}	DAT/SIG Output Data Delay From MLBCLK Rising Edge			8	ns
t_{MDZH}[2]	Bus Hold Time				
	1024 FS	2			ns
	512 FS/256	4			ns
C_{MLB}	DAT/SIG Pin Load				
	1024 FS			40	pf
	512 FS/256			60	pf

[1] Pulse width variation is measured at 1.25 V by triggering on one edge of MLBCLK and measuring the spread on the other edge, measured in ns peak-to-peak (p-p).
[2] The board must be designed to ensure that the high impedance bus does not leave the logic state of the final driven bit for this time period. Therefore, coupling must be minimized while meeting the maximum capacitive load listed.

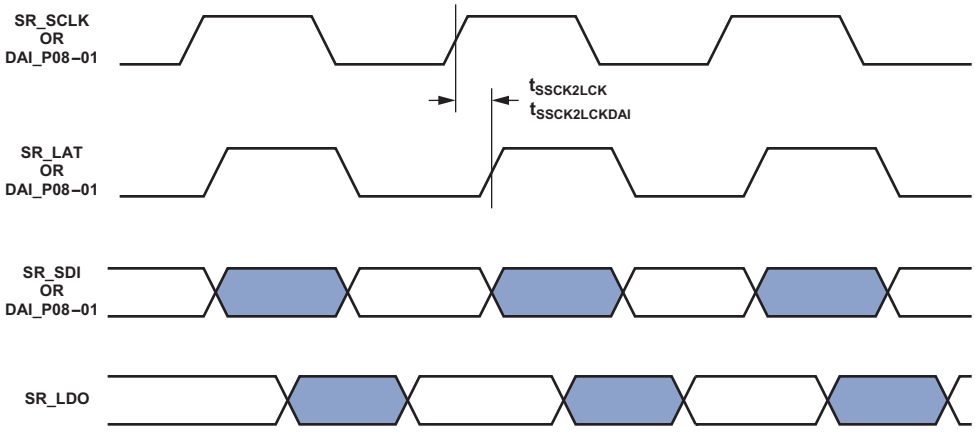

Figure 44. SR_SCLK to SR_LAT Setup, Clocks Pulse Width and Maximum Frequency

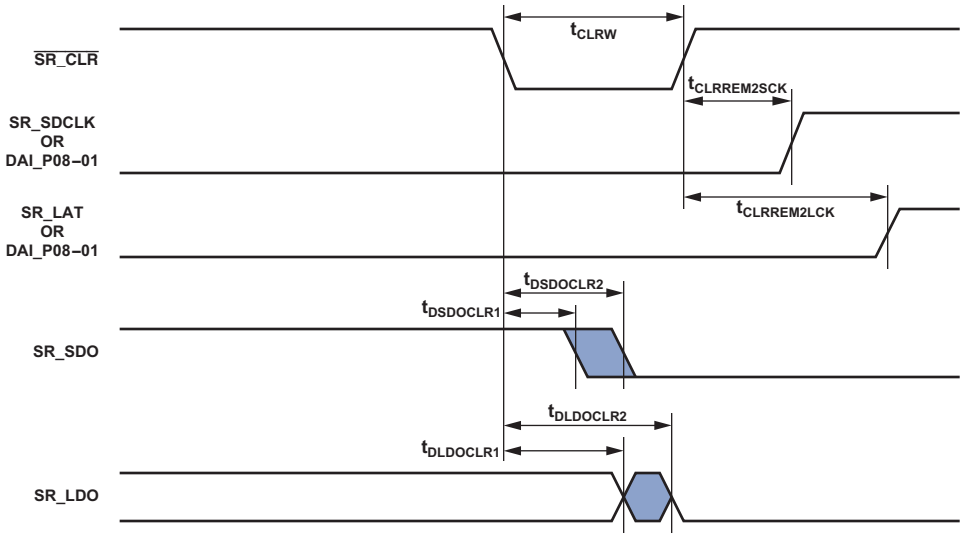

Figure 45. Shift Register Reset Timing

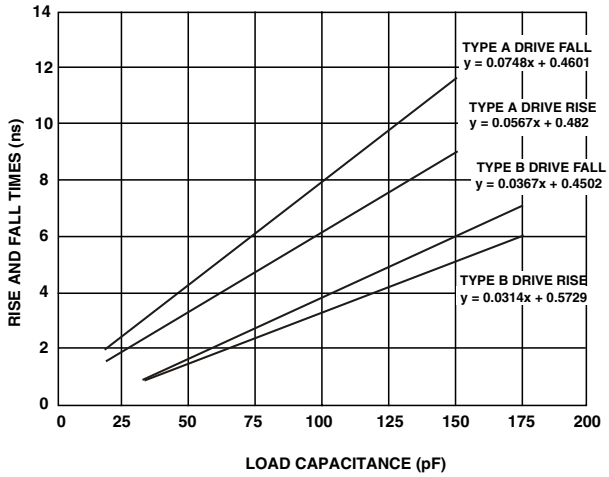

Figure 51. Typical Output Rise/Fall Time (20% to 80%, V_{DD_EXT} = Min)

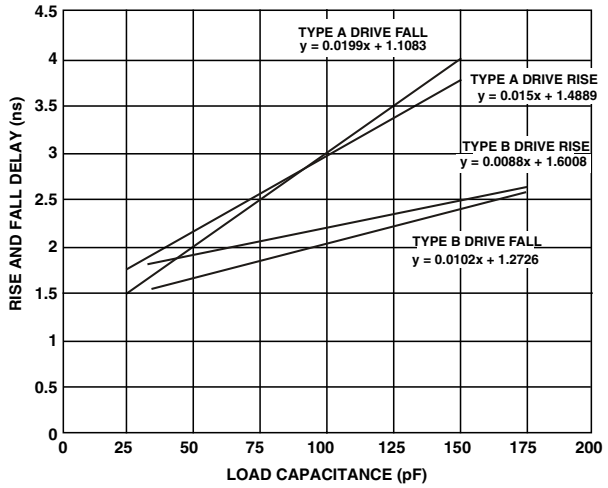

Figure 52. Typical Output Delay or Hold vs. Load Capacitance (at Ambient Temperature)

THERMAL CHARACTERISTICS

The processor is rated for performance over the temperature range specified in Operating Conditions.

Table 58 airflow measurements comply with JEDEC standards JESD51-2 and JESD51-6 and the junction-to-board measurement complies with JESD51-8. Test board design complies with JEDEC standards JESD51-7 (PBGA). The junction-to-case measurement complies with MIL- STD-883. All measurements use a 2S2P JEDEC test board.

To determine the junction temperature of the device while on the application PCB, use:

$$T_J = T_{CASE} + (\Psi_{JT} \times P_D)$$

where:

T_J = junction temperature (°C)

T_{CASE} = case temperature (°C) measured at the top center of the package

Ψ_{JT} = junction-to-top (of package) characterization parameter is the typical value from Table 58

P_D = power dissipation

Values of θ_{JA} are provided for package comparison and PCB design considerations. θ_{JA} can be used for a first order approximation of T_J by the equation:

$$T_J = T_A + (\theta_{JA} \times P_D)$$

where:

T_A = ambient temperature °C

Values of θ_{JC} are provided for package comparison and PCB design considerations when an external heatsink is required.

Note that the thermal characteristics values provided in Table 58 are modeled values.

Table 57. Thermal Characteristics for 88-Lead LFCSP_VQ

Parameter	Condition	Typical	Unit
θ_{JA}	Airflow = 0 m/s	22.6	°C/W
θ_{JMA}	Airflow = 1 m/s	18.2	°C/W
θ_{JMA}	Airflow = 2 m/s	17.3	°C/W
θ_{JC}		7.9	°C/W
Ψ_{JT}	Airflow = 0 m/s	0.22	°C/W
Ψ_{JMT}	Airflow = 1 m/s	0.36	°C/W
Ψ_{JMT}	Airflow = 2 m/s	0.44	°C/W

Table 58. Thermal Characteristics for 100-Lead LQFP_EP

Parameter	Condition	Typical	Unit
θ_{JA}	Airflow = 0 m/s	18.1	°C/W
θ_{JMA}	Airflow = 1 m/s	15.5	°C/W
θ_{JMA}	Airflow = 2 m/s	14.6	°C/W
θ_{JC}		2.4	°C/W
Ψ_{JT}	Airflow = 0 m/s	0.22	°C/W
Ψ_{JMT}	Airflow = 1 m/s	0.36	°C/W
Ψ_{JMT}	Airflow = 2 m/s	0.50	°C/W

Table 59. Thermal Characteristics for 196-Ball CSP_BGA

Parameter	Condition	Typical	Unit
θ_{JA}	Airflow = 0 m/s	29.0	°C/W
θ_{JMA}	Airflow = 1 m/s	26.1	°C/W
θ_{JMA}	Airflow = 2 m/s	25.1	°C/W
θ_{JC}		8.8	°C/W
Ψ_{JT}	Airflow = 0 m/s	0.23	°C/W
Ψ_{JMT}	Airflow = 1 m/s	0.42	°C/W
Ψ_{JMT}	Airflow = 2 m/s	0.52	°C/W

88-LFCSP_VQ LEAD ASSIGNMENT

Table 61 lists the 88-Lead LFCSP_VQ package lead names.

Table 61. 88-Lead LFCSP_VQ Lead Assignments (Numerical by Lead Number)

Lead Name	Lead No.	Lead Name	Lead No.	Lead Name	Lead No.	Lead Name	Lead No.
CLK_CFG1	1	V_{DD_EXT}	23	DAI_P10	45	V_{DD_INT}	67
BOOT_CFG0	2	DPI_P08	24	V_{DD_INT}	46	FLAG0	68
V_{DD_EXT}	3	DPI_P07	25	V_{DD_EXT}	47	V_{DD_INT}	69
V_{DD_INT}	4	DPI_P09	26	DAI_P20	48	FLAG1	70
BOOT_CFG1	5	DPI_P10	27	V_{DD_INT}	49	FLAG2	71
GND	6	DPI_P11	28	DAI_P08	50	FLAG3	72
CLK_CFG0	7	DPI_P12	29	DAI_P04	51	GND	73
V_{DD_INT}	8	DPI_P13	30	DAI_P14	52	GND	74
CLKIN	9	DAI_P03	31	DAI_P18	53	V_{DD_EXT}	75
XTAL	10	DPI_P14	32	DAI_P17	54	GND	76
V_{DD_EXT}	11	V_{DD_INT}	33	DAI_P16	55	V_{DD_INT}	77
V_{DD_INT}	12	DAI_P13	34	DAI_P15	56	$\overline{TRST}$	78
V_{DD_INT}	13	DAI_P07	35	DAI_P12	57	$\overline{EMU}$	79
$\overline{RESETOUT}$/$\overline{RUNRSTIN}$	14	DAI_P19	36	DAI_P11	58	TDO	80
V_{DD_INT}	15	DAI_P01	37	V_{DD_INT}	59	V_{DD_EXT}	81
DPI_P01	16	DAI_P02	38	GND	60	V_{DD_INT}	82
DPI_P02	17	V_{DD_INT}	39	THD_M	61	TDI	83
DPI_P03	18	V_{DD_EXT}	40	THD_P	62	TCK	84
V_{DD_INT}	19	V_{DD_INT}	41	V_{DD_THD}	63	V_{DD_INT}	85
DPI_P05	20	DAI_P06	42	V_{DD_INT}	64	$\overline{RESET}$	86
DPI_P04	21	DAI_P05	43	V_{DD_INT}	65	TMS	87
DPI_P06	22	DAI_P09	44	V_{DD_INT}	66	V_{DD_INT}	88
						GND	89*

* Lead no. 89 is the GND supply (see Figure 53 and Figure 54) for the processor; this pad must be **robustly** connected to GND in order for the processor to function.

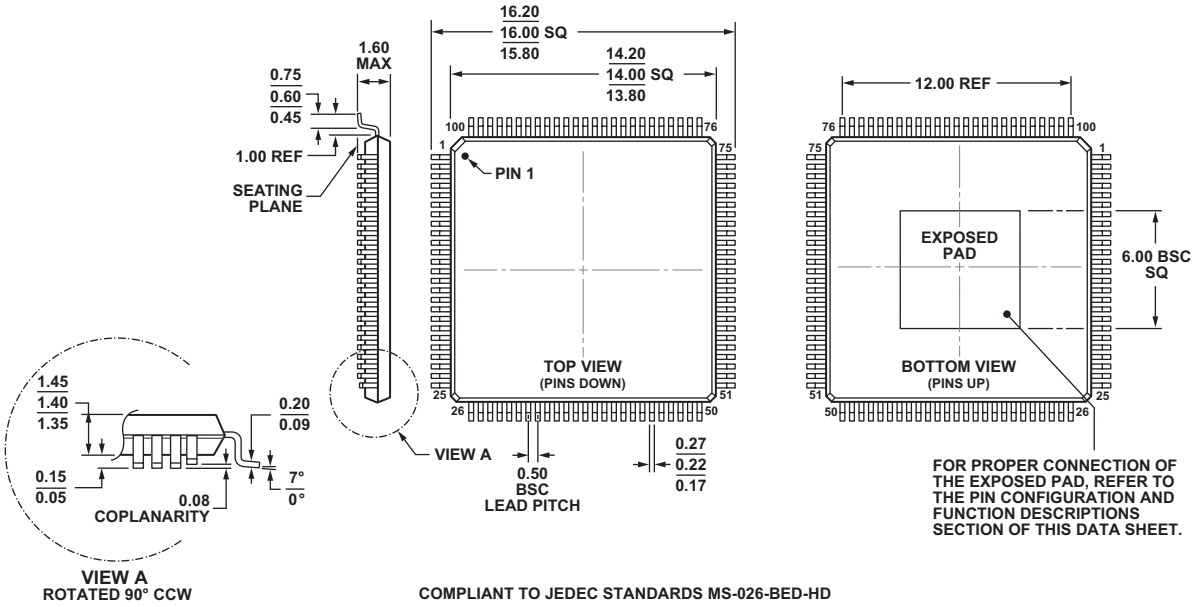

Figure 58. 100-Lead Low Profile Quad Flat Package, Exposed Pad [LQFP_EP[1]]
(SW-100-2)
Dimensions shown in millimeters

[1] For information relating to the exposed pad on the SW-100-2 package, see the table endnote on Page 70.

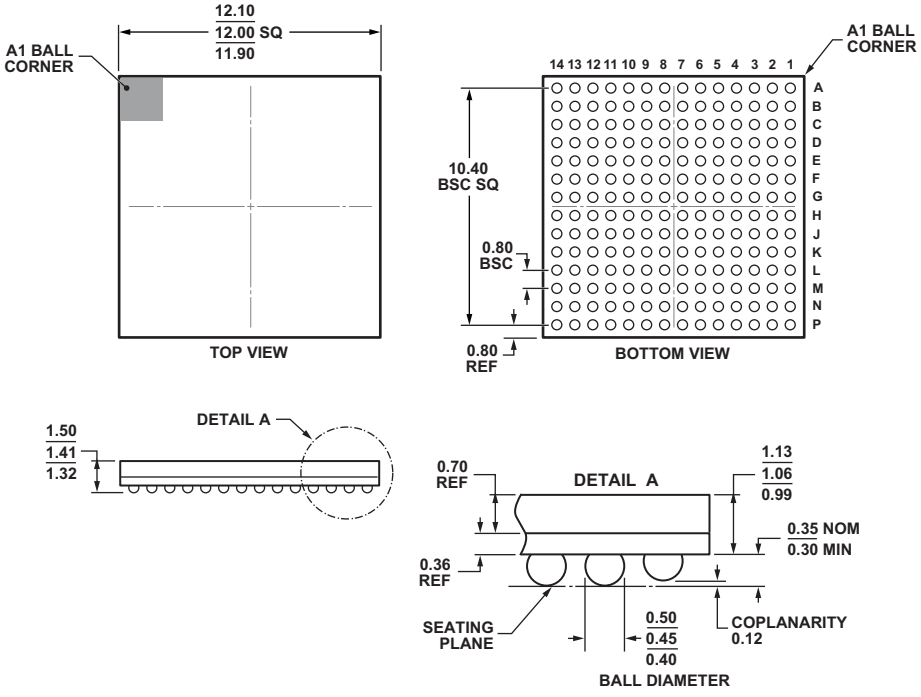

Figure 59. 196-Ball Chip Scale Package, Ball Grid Array [CSP_BGA]
(BC-196-8)
Dimensions shown in millimeters