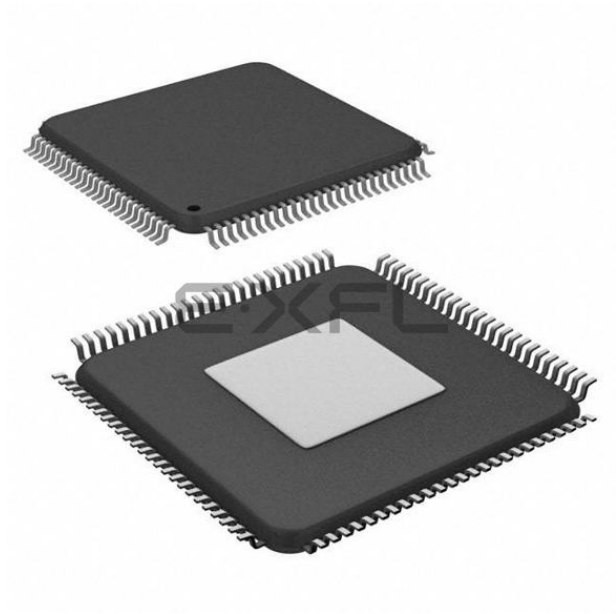

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Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Active
Type	Floating Point
Interface	EBI/EMI, DAI, I²C, SPI, SPORT, UART/USART
Clock Rate	350MHz
Non-Volatile Memory	External
On-Chip RAM	3Mbit
Voltage - I/O	3.30V
Voltage - Core	1.10V
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP Exposed Pad
Supplier Device Package	100-LQFP-EP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/analog-devices/adsp-21488kswz-3a

Delay Line DMA

The processor provides delay line DMA functionality. This allows processor reads and writes to external delay line buffers (and hence to external memory) with limited core interaction.

Scatter/Gather DMA

The processor provides scatter/gather DMA functionality. This allows processor DMA reads/writes to/from non contiguous memory blocks.

FFT Accelerator

The FFT accelerator implements a radix-2 complex/real input, complex output FFT with no core intervention. The FFT accelerator runs at the peripheral clock frequency.

FIR Accelerator

The FIR (finite impulse response) accelerator consists of a 1024 word coefficient memory, a 1024 word deep delay line for the data, and four MAC units. A controller manages the accelerator. The FIR accelerator runs at the peripheral clock frequency.

IIR Accelerator

The IIR (infinite impulse response) accelerator consists of a 1440 word coefficient memory for storage of biquad coefficients, a data memory for storing the intermediate data, and one MAC unit. A controller manages the accelerator. The IIR accelerator runs at the peripheral clock frequency.

Watchdog Timer

The watchdog timer is used to supervise the stability of the system software. When used in this way, software reloads the watchdog timer in a regular manner so that the downward counting timer never expires. An expiring timer then indicates that system software might be out of control.

The 32-bit watchdog timer that can be used to implement a software watchdog function. A software watchdog can improve system reliability by forcing the processor to a known state through generation of a system reset, if the timer expires before being reloaded by software. Software initializes the count value of the timer, and then enables the timer. The watchdog timer resets both the core and the internal peripherals. Note that this feature is available on the 176-lead package only.

SYSTEM DESIGN

The following sections provide an introduction to system design options and power supply issues.

Program Booting

The internal memory of the ADSP-2148x boots at system power-up from an 8-bit EPROM via the external port, an SPI master, or an SPI slave. Booting is determined by the boot configuration (BOOT_CFG2–0) pins in Table 9 for the 176-lead package and Table 10 for the 100-lead package.

Table 9. Boot Mode Selection, 176-Lead Package

BOOT_CFG2–0	Booting Mode
000	SPI Slave Boot
001	SPI Master Boot
010	AMI User Boot (for 8-bit Flash Boot)
011	No boot (processor executes from internal ROM after reset)
1xx	Reserved

Table 10. Boot Mode Selection, 100-Lead Package

BOOT_CFG1–0	Booting Mode
00	SPI Slave Boot
01	SPI Master Boot
10	Reserved
11	No boot (processor executes from internal ROM after reset)

The "Running Reset" feature allows a user to perform a reset of the processor core and peripherals, but without resetting the PLL and SDRAM controller, or performing a boot. The functionality of the RESETOUT/RUNRSTIN pin has now been extended to also act as the input for initiating a Running Reset. For more information, see the hardware reference.

Power Supplies

The processors have separate power supply connections for the internal (V_{DD_INT}) and external (V_{DD_EXT}) power supplies. The internal supply must meet the V_{DD_INT} specifications. The external supply must meet the V_{DD_EXT} specification. All external supply pins must be connected to the same power supply.

To reduce noise coupling, the PCB should use a parallel pair of power and ground planes for V_{DD_INT} and GND.

Static Voltage Scaling (SVS)

Some models of the ADSP-2148x feature Static Voltage Scaling (SVS) on the V_{DD_INT} power supply. (See the Ordering Guide on Page 66 for model details.) This voltage specification technique can provide significant performance benefits including 450 MHz core frequency operation without a significant increase in power.

SVS optimizes the required V_{DD_INT} voltage for each individual device to enable enhanced operating frequency up to 450 MHz. The optimized SVS voltage results in a reduction of maximum I_{DD_INT} which enables 450 MHz operation at the same or lower maximum power than 400 MHz operation at a fixed voltage supply. Implementation of SVS requires a specific voltage regulator circuit design and initialization code.

Refer to the Engineer-to-Engineer Note Static Voltage Scaling for ADSP-2148x SHARC Processors (EE-357) for further information. The EE-Note details the requirements and process to implement a SVS power supply system to enable operation up to 450 MHz. This applies only to specific products within the ADSP-2148x family which are capable of supporting 450 MHz operation.

Details on power consumption and Static and Dynamic current consumption can be found at Total Power Dissipation on Page 20. Also see Operating Conditions on Page 18 for more information.

The following are SVS features.

- SVS is applicable only to 450 MHz models (not applicable to 400 MHz or lower frequency models).
- Each individual SVS device includes a register (SVS_DAT) containing the unique SVS voltage set at the factory, known as SVS_{NOM}.
- The SVS_{NOM} value is the intended set voltage for the V_{DD_INT} voltage regulator.
- No dedicated pins are required for SVS. The TWI serial bus is used to communicate SVS_{NOM} to the voltage regulator.
- Analog Devices recommends a specific voltage regulator design and initialization code sequence that optimizes the power-up sequence.

  The Engineer-to-Engineer Note Static Voltage Scaling for ADSP-2148x SHARC Processors (EE-357) contains the details of the regulator design and the initialization requirements.
- Any differences from the Analog Devices recommended programmable regulator design must be reviewed by Analog Devices to ensure that it meets the voltage accuracy and range requirements.

Target Board JTAG Emulator Connector

Analog Devices DSP Tools product line of JTAG emulators uses the IEEE 1149.1 JTAG test access port of the ADSP-2148x processors to monitor and control the target board processor during emulation. Analog Devices DSP Tools product line of JTAG emulators provides emulation at full processor speed, allowing inspection and modification of memory, registers, and processor stacks. The processor's JTAG interface ensures that the emulator will not affect target system loading or timing.

For complete information on Analog Devices' SHARC DSP Tools product line of JTAG emulator operation, see the appropriate emulator hardware user's guide.

DEVELOPMENT TOOLS

Analog Devices supports its processors with a complete line of software and hardware development tools, including integrated development environments (which include CrossCore® Embedded Studio and/or VisualDSP++®), evaluation products, emulators, and a wide variety of software add-ins.

Integrated Development Environments (IDEs)

For C/C++ software writing and editing, code generation, and debug support, Analog Devices offers two IDEs.

CrossCore Embedded Studio is based on the Eclipse™ framework. Supporting most Analog Devices processor families, it is the IDE of choice for future processors, including multicore devices. CrossCore Embedded Studio seamlessly integrates available software add-ins to support real time operating systems, file systems, TCP/IP stacks, USB stacks, algorithmic software modules, and evaluation hardware board support packages. For more information visit www.analog.com/cces.

The other Analog Devices IDE, VisualDSP++, supports processor families introduced prior to the release of CrossCore Embedded Studio. This IDE includes the Analog Devices VDK real time operating system and an open source TCP/IP stack. For more information visit www.analog.com/visualdsp. Note that VisualDSP++ will not support future Analog Devices processors.

EZ-KIT Lite Evaluation Board

For processor evaluation, Analog Devices provides wide range of EZ-KIT Lite® evaluation boards. Including the processor and key peripherals, the evaluation board also supports on-chip emulation capabilities and other evaluation and development features. Also available are various EZ-Extenders®, which are daughter cards delivering additional specialized functionality, including audio and video processing. For more information visit www.analog.com and search on "ezkit" or "ezextender".

EZ-KIT Lite Evaluation Kits

For a cost-effective way to learn more about developing with Analog Devices processors, Analog Devices offer a range of EZ-KIT Lite evaluation kits. Each evaluation kit includes an EZ-KIT Lite evaluation board, directions for downloading an evaluation version of the available IDE(s), a USB cable, and a power supply. The USB controller on the EZ-KIT Lite board connects to the USB port of the user's PC, enabling the chosen IDE evaluation suite to emulate the on-board processor in-circuit. This permits the customer to download, execute, and debug programs for the EZ-KIT Lite system. It also supports in-circuit programming of the on-board Flash device to store user-specific boot code, enabling standalone operation. With the full version of CrossCore Embedded Studio or VisualDSP++ installed (sold separately), engineers can develop software for supported EZ-KITs or any custom system utilizing supported Analog Devices processors.

Software Add-Ins for CrossCore Embedded Studio

Analog Devices offers software add-ins which seamlessly integrate with CrossCore Embedded Studio to extend its capabilities and reduce development time. Add-ins include board support packages for evaluation hardware, various middleware packages, and algorithmic modules. Documentation, help, configuration dialogs, and coding examples present in these add-ins are viewable through the CrossCore Embedded Studio IDE once the add-in is installed.

Board Support Packages for Evaluation Hardware

Software support for the EZ-KIT Lite evaluation boards and EZ-Extender daughter cards is provided by software add-ins called Board Support Packages (BSPs). The BSPs contain the required drivers, pertinent release notes, and select example code for the given evaluation hardware. A download link for a specific BSP is located on the web page for the associated EZ-KIT or EZ-Extender product. The link is found in the Product Download area of the product web page.

PIN FUNCTION DESCRIPTIONS

Table 11. Pin Descriptions

Name	Type	State During/ After Reset	Description
$ADDR_{23–0}$	I/O/T (ipu)	High-Z/ driven low (boot)	**External Address.** The processor outputs addresses for external memory and peripherals on these pins. The ADDR pins can be multiplexed to support the external memory interface address, and FLAGS15–8 (I/O) and PWM (O). After reset, all ADDR pins are in external memory interface mode and FLAG(0–3) pins are in FLAGS mode (default). When configured in the IDP_PDAP_CTL register, IDP channel 0 scans the $ADDR_{23–4}$ pins for parallel input data.
$DATA_{15–0}$	I/O/T (ipu)	High-Z	**External Data.** The data pins can be multiplexed to support the external memory interface data (I/O), and $FLAGS_{7–0}$ (I/O).
AMI_ACK	I (ipu)		**Memory Acknowledge.** External devices can deassert AMI_ACK (low) to add wait states to an external memory access. AMI_ACK is used by I/O devices, memory controllers, or other peripherals to hold off completion of an external memory access.
$\overline{MS}_{0–1}$	O/T (ipu)	High-Z	**Memory Select Lines 0–1.** These lines are asserted (low) as chip selects for the corresponding banks of external memory. The $\overline{MS}_{1-0}$ lines are decoded memory address lines that change at the same time as the other address lines. When no external memory access is occurring the $\overline{MS}_{1-0}$ lines are inactive; they are active however when a conditional memory access instruction is executed, when the condition evaluates as true. The $\overline{MS1}$ pin can be used in EPORT/FLASH boot mode. For more information, see the hardware reference.
$\overline{AMI\_RD}$	O/T (ipu)	High-Z	**AMI Port Read Enable.** $\overline{AMI\_RD}$ is asserted whenever the processor reads a word from external memory.
$\overline{AMI\_WR}$	O/T (ipu)	High-Z	**AMI Port Write Enable.** $\overline{AMI\_WR}$ is asserted when the processor writes a word to external memory.
FLAG0/$\overline{IRQ0}$	I/O (ipu)	FLAG[0] INPUT	**FLAG0/Interrupt Request0.**
FLAG1/$\overline{IRQ1}$	I/O (ipu)	FLAG[1] INPUT	**FLAG1/Interrupt Request1.**
FLAG2/$\overline{IRQ2}$/$\overline{MS2}$	I/O (ipu)	FLAG[2] INPUT	**FLAG2/Interrupt Request2/Memory Select2.**
FLAG3/TMREXP/$\overline{MS3}$	I/O (ipu)	FLAG[3] INPUT	**FLAG3/Timer Expired/Memory Select3.**

The following symbols appear in the Type column of this table: **A** = asynchronous, **I** = input, **O** = output, **S** = synchronous, **A/D** = active drive, **O/D** = open drain, and **T** = three-state, **ipd** = internal pull-down resistor, **ipu** = internal pull-up resistor.
The internal pull-up (ipu) and internal pull-down (ipd) resistors are designed to hold the internal path from the pins at the expected logic levels. To pull-up or pull-down the external pads to the expected logic levels, use external resistors. Internal pull-up/pull-down resistors cannot be enabled/disabled and the value of these resistors cannot be programmed. The range of an ipu resistor can be between 26 kΩ–63 kΩ. The range of an ipd resistor can be between 31 kΩ–85kΩ. The three-state voltage of ipu pads will not reach to the full V_{DD_EXT} level; at typical conditions the voltage is in the range of 2.3 V to 2.7 V.
In this table, all pins are LVTTL compliant with the exception of the thermal diode pins.

Table 11. Pin Descriptions (Continued)

Name	Type	State During/ After Reset	Description
MLBCLK[1]	I		**Media Local Bus Clock.** This clock is generated by the MLB controller that is synchronized to the MOST network and provides the timing for the entire MLB interface at 49.152 MHz at FS=48 kHz. When the MLB controller is not used, this pin should be grounded.
MLBDAT[1]	I/O/T in 3 pin mode. I in 5 pin mode.	High-Z	**Media Local Bus Data.** The MLBDAT line is driven by the transmitting MLB device and is received by all other MLB devices including the MLB controller. The MLBDAT line carries the actual data. In 5-pin MLB mode, this pin is an input only. When the MLB controller is not used, this pin should be grounded.
MLBSIG[1]	I/O/T in 3 pin mode. I in 5 pin mode	High-Z	**Media Local Bus Signal.** This is a multiplexed signal which carries the Channel/Address generated by the MLB Controller, as well as the Command and RxStatus bytes from MLB devices. In 5-pin mode, this pin is input only. When the MLB controller is not used, this pin should be grounded.
MLBDO[1]	O/T	High-Z	**Media Local Bus Data Output (in 5 pin mode).** This pin is used only in 5-pin MLB mode. This serves as the output data pin in 5-pin mode. When the MLB controller is not used, this pin should be connected to ground.
MLBSO[1]	O/T	High-Z	**Media Local Bus Signal Output (in 5 pin mode).** This pin is used only in 5-pin MLB mode. This serves as the output signal pin in 5-pin mode. When the MLB controller is not used, this pin should be connected to ground.
TDI	I (ipu)		**Test Data Input (JTAG).** Provides serial data for the boundary scan logic.
TDO	O/T	High-Z	**Test Data Output (JTAG).** Serial scan output of the boundary scan path.
TMS	I (ipu)		**Test Mode Select (JTAG).** Used to control the test state machine.
TCK	I		**Test Clock (JTAG).** Provides a clock for JTAG boundary scan. TCK must be asserted (pulsed low) after power-up or held low for proper operation of the device.
$\overline{\text{TRST}}$	I (ipu)		**Test Reset (JTAG).** Resets the test state machine. $\overline{\text{TRST}}$ must be asserted (pulsed low) after power-up or held low for proper operation of the processor.
$\overline{\text{EMU}}$	O (O/D, ipu)	High-Z	**Emulation Status.** Must be connected to the ADSP-2148x Analog Devices DSP Tools product line of JTAG emulators target board connector only.

The following symbols appear in the Type column of this table: **A** = asynchronous, **I** = input, **O** = output, **S** = synchronous, **A/D** = active drive, **O/D** = open drain, and **T** = three-state, **ipd** = internal pull-down resistor, **ipu** = internal pull-up resistor.
The internal pull-up (ipu) and internal pull-down (ipd) resistors are designed to hold the internal path from the pins at the expected logic levels. To pull-up or pull-down the external pads to the expected logic levels, use external resistors. Internal pull-up/pull-down resistors cannot be enabled/disabled and the value of these resistors cannot be programmed. The range of an ipu resistor can be between 26 kΩ–63 kΩ. The range of an ipd resistor can be between 31 kΩ–85kΩ. The three-state voltage of ipu pads will not reach to the full V_{DD_EXT} level; at typical conditions the voltage is in the range of 2.3 V to 2.7 V.
In this table, all pins are LVTTL compliant with the exception of the thermal diode pins.

ELECTRICAL CHARACTERISTICS

Parameter[1]	Description	Test Conditions	300 MHz / 350 MHz / 400 MHz / 450 MHz Min	Typ	Max	Unit
V_{OH}[2]	High Level Output Voltage	@ V_{DD_EXT} = Min, I_{OH} = –1.0 mA[3]	2.4			V
V_{OL}[2]	Low Level Output Voltage	@ V_{DD_EXT} = Min, I_{OL} = 1.0 mA[3]			0.4	V
I_{IH}[4, 5]	High Level Input Current	@ V_{DD_EXT} = Max, V_{IN} = V_{DD_EXT} Max			10	µA
I_{IL}[4]	Low Level Input Current	@ V_{DD_EXT} = Max, V_{IN} = 0 V			10	µA
I_{ILPU}[5]	Low Level Input Current Pull-up	@ V_{DD_EXT} = Max, V_{IN} = 0 V			200	µA
I_{OZH}[6, 7]	Three-State Leakage Current	@ V_{DD_EXT} = Max, V_{IN} = V_{DD_EXT} Max			10	µA
I_{OZL}[6]	Three-State Leakage Current	@ V_{DD_EXT} = Max, V_{IN} = 0 V			10	µA
I_{OZLPU}[7]	Three-State Leakage Current Pull-up	@ V_{DD_EXT} = Max, V_{IN} = 0 V			200	µA
I_{OZHPD}[8]	Three-State Leakage Current Pull-down	@ V_{DD_EXT} = Max, V_{IN} = V_{DD_EXT} Max			200	µA
I_{DD_INT}[9]	Supply Current (Internal)	f_{CCLK} > 0 MHz			Table 14 + Table 15 × ASF	mA
I_{DD_INT}	Supply Current (Internal)	V_{DDINT} = 1.1 V, ASF = 1, T_J = 25°C		410 / 450 / 500 / 550		mA
C_{IN}[10, 11]	Input Capacitance	T_{CASE} = 25°C			5	pF

[1] Specifications subject to change without notice.
[2] Applies to output and bidirectional pins: ADDR23–0, DATA15–0, $\overline{AMI\_RD}$, $\overline{AMI\_WR}$, FLAG3–0, DAI_Px, DPI_Px, $\overline{EMU}$, TDO, $\overline{RESETOUT}$ MLBSIG, MLBDAT, MLBDO, MLBSO, $\overline{SDRAS}$, $\overline{SDCAS}$, $\overline{SDWE}$, SDCKE, SDA10, SDDQM, $\overline{MS0\text{-}1}$.
[3] See Output Drive Currents on Page 55 for typical drive current capabilities.
[4] Applies to input pins: BOOT_CFGx, CLK_CFGx, TCK, $\overline{RESET}$, CLKIN.
[5] Applies to input pins with internal pull-ups: $\overline{TRST}$, TMS, TDI.
[6] Applies to three-statable pin: TDO.
[7] Applies to three-statable pins with pull-ups: DAI_Px, DPI_Px, $\overline{EMU}$.
[8] Applies to three-statable pin with pull-down: SDCLK.
[9] See Engineer-to-Engineer Note Estimating Power for ADSP-214xx SHARC Processors (EE-348) for further information.
[10] Applies to all signal pins.
[11] Guaranteed, but not tested.

Table 15. Dynamic Current in CCLK Domain—$I_{DD\_INT\_DYNAMIC}$ (mA, with ASF = 1.0)[1, 2]

f_{CCLK} (MHz)	V_{DD_INT} (V)								
	0.975 V	1.0 V	1.025 V	1.05 V	1.075 V	1.10 V	1.125 V	1.15 V	1.175 V
100	76	77	81	84	87	88	90	92	95
150	117	119	123	126	130	133	136	139	144
200	153	156	161	165	170	174	179	183	188
250	190	195	201	207	212	217	223	229	235
300	227	233	240	246	253	260	266	273	280
350	263	272	278	286	294	302	309	318	325
400	300	309	317	326	335	344	352	361	370
450	339	349	356	365	374	385	394	405	415

[1] The values are not guaranteed as standalone maximum specifications. They must be combined with static current per the equations of Electrical Characteristics on Page 19.
[2] Valid frequency and voltage ranges are model-specific. See Operating Conditions on Page 18.

ABSOLUTE MAXIMUM RATINGS

Stresses at or above those listed in Table 16 may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

Table 16. Absolute Maximum Ratings

Parameter	Rating
Internal (Core) Supply Voltage (V_{DD_INT})	–0.3 V to +1.32 V
External (I/O) Supply Voltage (V_{DD_EXT})	–0.3 V to +3.6 V
Thermal Diode Supply Voltage (V_{DD_THD})	–0.3 V to +3.6 V
Input Voltage	–0.5 V to +3.6 V
Output Voltage Swing	–0.5 V to V_{DD_EXT} +0.5 V
Storage Temperature Range	–65°C to +150°C
Junction Temperature While Biased	125°C

ESD SENSITIVITY

ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

MAXIMUM POWER DISSIPATION

See Engineer-to-Engineer Note Estimating Power for ADSP-214xx SHARC Processors (EE-348) for detailed thermal and power information regarding maximum power dissipation. For information on package thermal specifications, see Thermal Characteristics on Page 56.

PACKAGE INFORMATION

The information presented in Figure 3 provides details about the package branding for the ADSP-2148x processors. For a complete listing of product availability, see Ordering Guide on Page 66.

Figure 3. Typical Package Brand

Table 17. Package Brand Information[1]

Brand Key	Field Description
t	Temperature Range
pp	Package Type
Z	RoHS Compliant Option
cc	See Ordering Guide
vvvvvv.x	Assembly Lot Code
n.n	Silicon Revision
#	RoHS Compliant Designation
yyww	Date Code

[1] Non automotive only. For branding information specific to automotive products, contact Analog Devices Inc.

TIMING SPECIFICATIONS

Use the exact timing information given. Do not attempt to derive parameters from the addition or subtraction of others. While addition or subtraction would yield meaningful results for an individual device, the values given in this data sheet reflect statistical variations and worst cases. Consequently, it is not meaningful to add parameters to derive longer times. See Figure 43 on Page 55 for voltage reference levels.

Switching characteristics specify how the processor changes its signals. Circuitry external to the processor must be designed for compatibility with these signal characteristics. Switching characteristics describe what the processor will do in a given circumstance. Use switching characteristics to ensure that any timing requirement of a device connected to the processor (such as memory) is satisfied.

Timing requirements apply to signals that are controlled by circuitry external to the processor, such as the data input for a read operation. Timing requirements guarantee that the processor operates correctly with other devices.

Core Clock Requirements

The processor's internal clock (a multiple of CLKIN) provides the clock signal for timing internal memory, the processor core, and the serial ports. During reset, program the ratio between the processor's internal clock frequency and external (CLKIN) clock frequency with the CLK_CFG1–0 pins.

The processor's internal clock switches at higher frequencies than the system input clock (CLKIN). To generate the internal clock, the processor uses an internal phase-locked loop (PLL, see Figure 4). This PLL-based clocking minimizes the skew between the system clock (CLKIN) signal and the processor's internal clock.

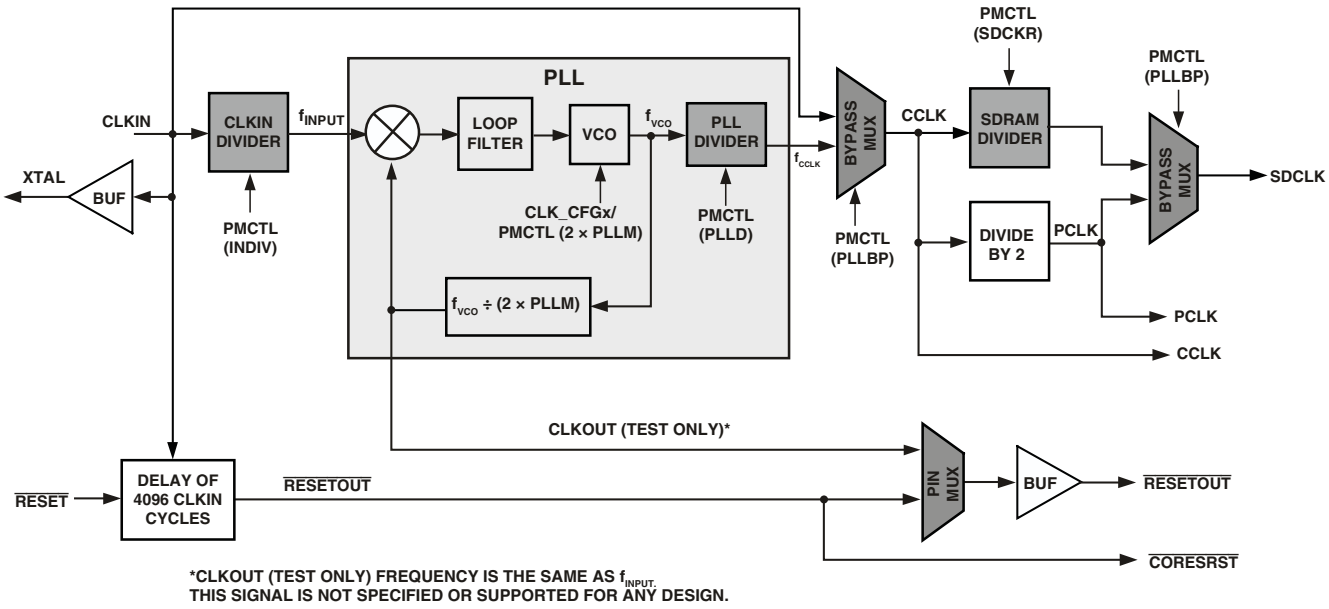

Figure 4. Core Clock and System Clock Relationship to CLKIN

Voltage Controlled Oscillator (VCO)

In application designs, the PLL multiplier value should be selected in such a way that the VCO frequency never exceeds f_{VCO} specified in Table 20.

- The product of CLKIN and PLLM must never exceed 1/2 of f_{VCO} (max) in Table 20 if the input divider is not enabled (INDIV = 0).
- The product of CLKIN and PLLM must never exceed f_{VCO} (max) in Table 20 if the input divider is enabled (INDIV = 1).

The VCO frequency is calculated as follows:

$$f_{VCO} = 2 \times PLLM \times f_{INPUT}$$
$$f_{CCLK} = (2 \times PLLM \times f_{INPUT}) \div PLLD$$

where:

f_{VCO} = VCO output

PLLM = Multiplier value programmed in the PMCTL register. During reset, the PLLM value is derived from the ratio selected using the CLK_CFG pins in hardware.

PLLD = 2, 4, 8, or 16 based on the divider value programmed on the PMCTL register. During reset this value is 2.

f_{INPUT} = is the input frequency to the PLL.

f_{INPUT} = CLKIN when the input divider is disabled or

f_{INPUT} = CLKIN ÷ 2 when the input divider is enabled

Reset

Table 21. Reset

Parameter		Min	Max	Unit
Timing Requirements				
t_{WRST}[1]	$\overline{\text{RESET}}$ Pulse Width Low	$4 \times t_{CK}$		ns
t_{SRST}	$\overline{\text{RESET}}$ Setup Before CLKIN Low	8		ns

[1] Applies after the power-up sequence is complete. At power-up, the processor's internal phase-locked loop requires no more than 100 μσ while $\overline{\text{RESET}}$ is low, assuming stable V_{DD} and CLKIN (not including start-up time of external clock oscillator).

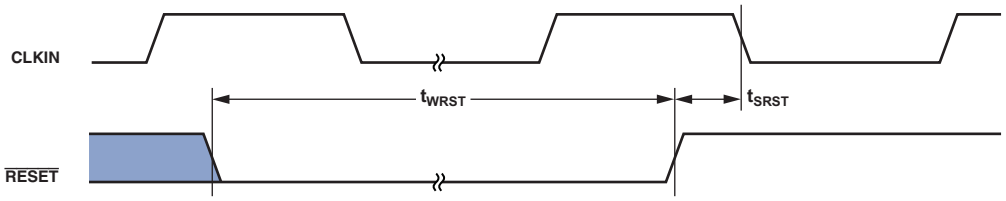

Figure 8. Reset

Running Reset

The following timing specification applies to $\overline{\text{RESETOUT}}$/$\overline{\text{RUNRSTIN}}$ pin when it is configured as $\overline{\text{RUNRSTIN}}$.

Table 22. Running Reset

Parameter		Min	Max	Unit
Timing Requirements				
$t_{WRUNRST}$	Running $\overline{\text{RESET}}$ Pulse Width Low	$4 \times t_{CK}$		ns
$t_{SRUNRST}$	Running $\overline{\text{RESET}}$ Setup Before CLKIN High	8		ns

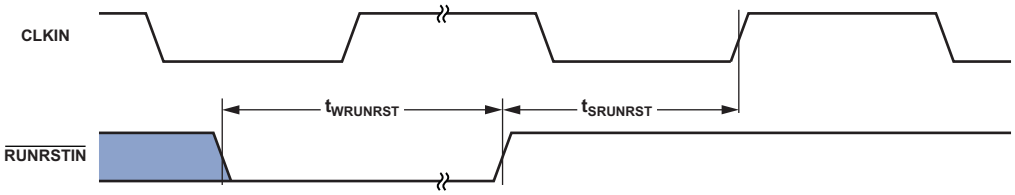

Figure 9. Running Reset

Precision Clock Generator (Direct Pin Routing)

This timing is only valid when the SRU is configured such that the precision clock generator (PCG) takes its inputs directly from the DAI pins (via pin buffers) and sends its outputs directly to the DAI pins. For the other cases, where the PCG's inputs and outputs are not directly routed to/from DAI pins (via pin buffers), there is no timing data available. All timing parameters and switching characteristics apply to external DAI pins (DAI_P01 – DAI_P20).

Table 29. Precision Clock Generator (Direct Pin Routing)

Parameter		Min	Max	Unit
Timing Requirements				
t_{PCGIW}	Input Clock Period	$t_{PCLK} \times 4$		ns
t_{STRIG}	PCG Trigger Setup Before Falling Edge of PCG Input Clock	4.5		ns
t_{HTRIG}	PCG Trigger Hold After Falling Edge of PCG Input Clock	3		ns
Switching Characteristics				
t_{DPCGIO}	PCG Output Clock and Frame Sync Active Edge Delay After PCG Input Clock	2.5	10	ns
$t_{DTRIGCLK}$	PCG Output Clock Delay After PCG Trigger	$2.5 + (2.5 \times t_{PCGIP})$	$10 + (2.5 \times t_{PCGIP})$	ns
$t_{DTRIGFS}$	PCG Frame Sync Delay After PCG Trigger	$2.5 + ((2.5 + D - PH) \times t_{PCGIP})$	$10 + ((2.5 + D - PH) \times t_{PCGIP})$	ns
t_{PCGOW}[1]	Output Clock Period	$2 \times t_{PCGIP} - 1$		ns

D = FSxDIV, PH = FSxPHASE. For more information, see the "Precision Clock Generators" chapter in the hardware reference.

[1] Normal mode of operation.

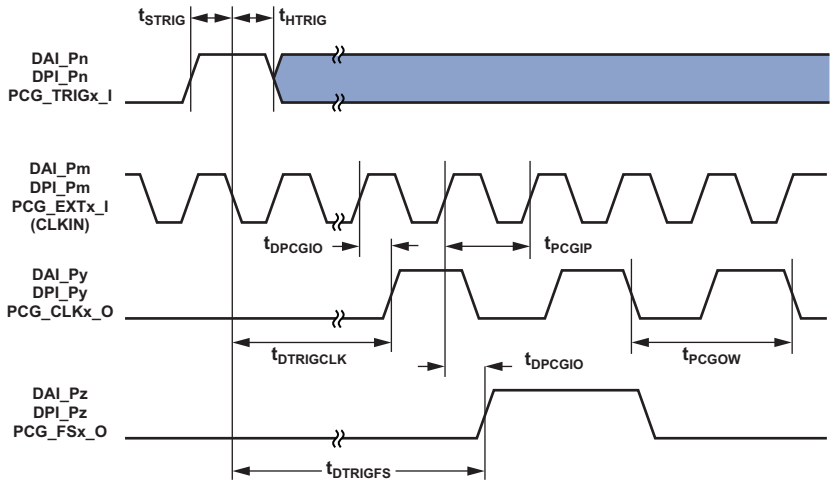

Figure 16. Precision Clock Generator (Direct Pin Routing)

Flags

The timing specifications provided below apply to the DPI_P14–1, ADDR7–0, ADDR23–8, DATA7–0, and FLAG3–0 pins when configured as FLAGS. See Table 11 on Page 14 for more information on flag use.

Table 30. Flags

Parameter		Min	Max	Unit
Timing Requirement				
t_{FIPW}[1]	FLAGs IN Pulse Width	$2 \times t_{PCLK} + 3$		ns
Switching Characteristic				
t_{FOPW}[1]	FLAGs OUT Pulse Width	$2 \times t_{PCLK} - 3$		ns

[1] This is applicable when the Flags are connected to DPI_P14–1, ADDR7–0, ADDR23–8, DATA7–0 and FLAG3–0 pins.

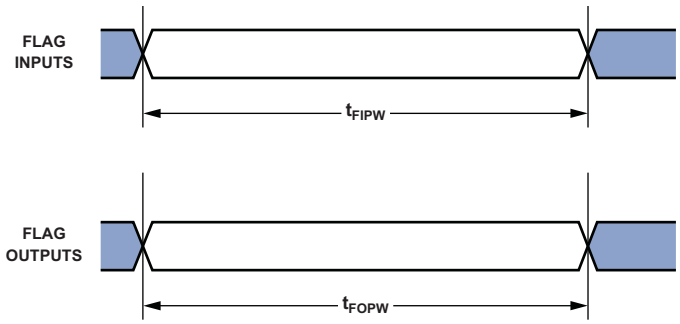

Figure 17. Flags

Table 36. Serial Ports—External Late Frame Sync

Parameter		Min	Max	Unit
Switching Characteristics				
$t_{DDTLFSE}$[1]	Data Delay from Late External Transmit Frame Sync or External Receive Frame Sync with MCE = 1, MFD = 0		8.5	ns
$t_{DDTENFS}$[1]	Data Enable for MCE = 1, MFD = 0	0.5		ns

[1] The $t_{DDTLFSE}$ and $t_{DDTENFS}$ parameters apply to left-justified, as well as DSP serial mode, and MCE = 1, MFD = 0.

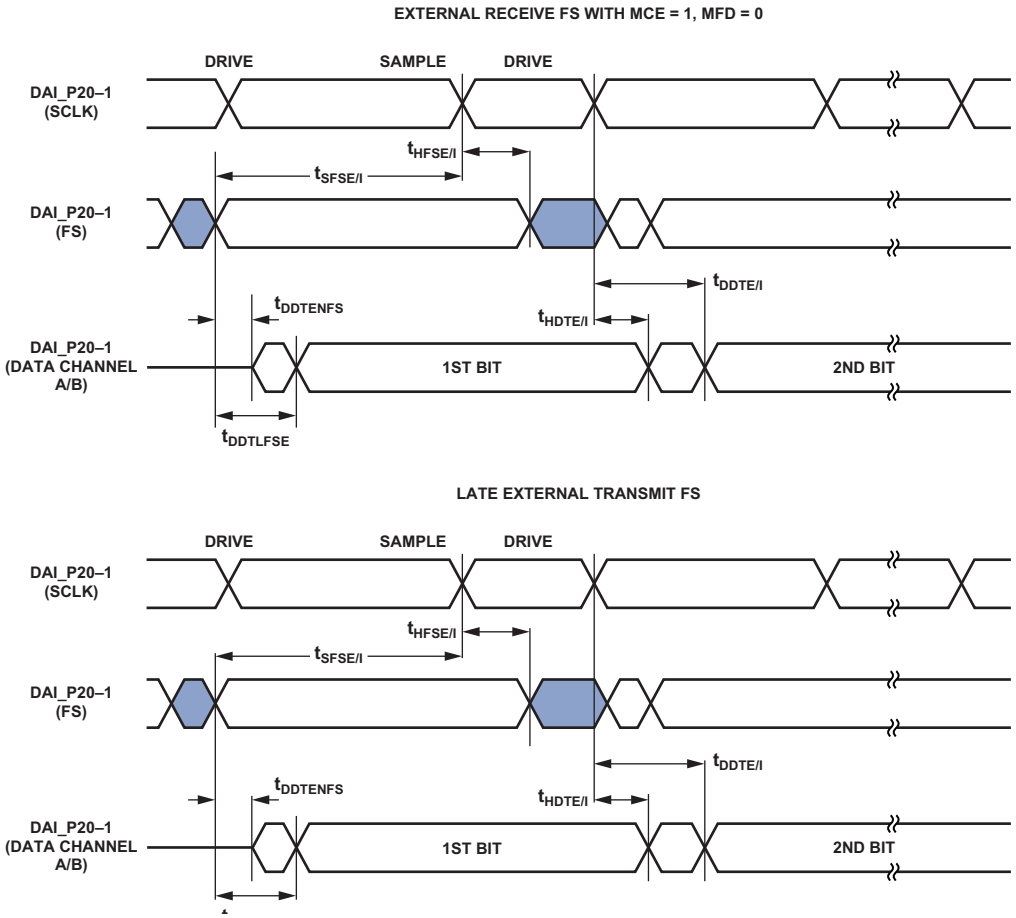

Figure 22. External Late Frame Sync[1]

[1] This figure reflects changes made to support left-justified mode.

Parallel Data Acquisition Port (PDAP)

The timing requirements for the PDAP are provided in Table 40. PDAP is the parallel mode operation of Channel 0 of the IDP. For details on the operation of the PDAP, see the PDAP chapter of the hardware reference. Note that the 20 bits of external PDAP data can be provided through the ADDR23–4 pins or over the DAI pins.

Table 40. Parallel Data Acquisition Port (PDAP)

Parameter		Min	Max	Unit
Timing Requirements				
t_{SPHOLD}[1]	PDAP_HOLD Setup Before PDAP_CLK Sample Edge	2.5		ns
t_{HPHOLD}[1]	PDAP_HOLD Hold After PDAP_CLK Sample Edge	2.5		ns
t_{PDSD}[1]	PDAP_DAT Setup Before PDAP_CLK Sample Edge	3.85		ns
t_{PDHD}[1]	PDAP_DAT Hold After PDAP_CLK Sample Edge	2.5		ns
t_{PDCLKW}	Clock Width	$(t_{PCLK} \times 4) \div 2 - 3$		ns
t_{PDCLK}	Clock Period	$t_{PCLK} \times 4$		ns
Switching Characteristics				
t_{PDHLDD}	Delay of PDAP Strobe After Last PDAP_CLK Capture Edge for a Word	$2 \times t_{PCLK} + 3$		ns
t_{PDSTRB}	PDAP Strobe Pulse Width	$2 \times t_{PCLK} - 1.5$		ns

[1] Source pins of PDAP_DATA are ADDR23–4 or DAI pins. Source pins for PDAP_CLK and PDAP_HOLD are 1) DAI pins; 2) CLKIN through PCG; 3) DAI pins through PCG; or 4) ADDR3–2 pins.

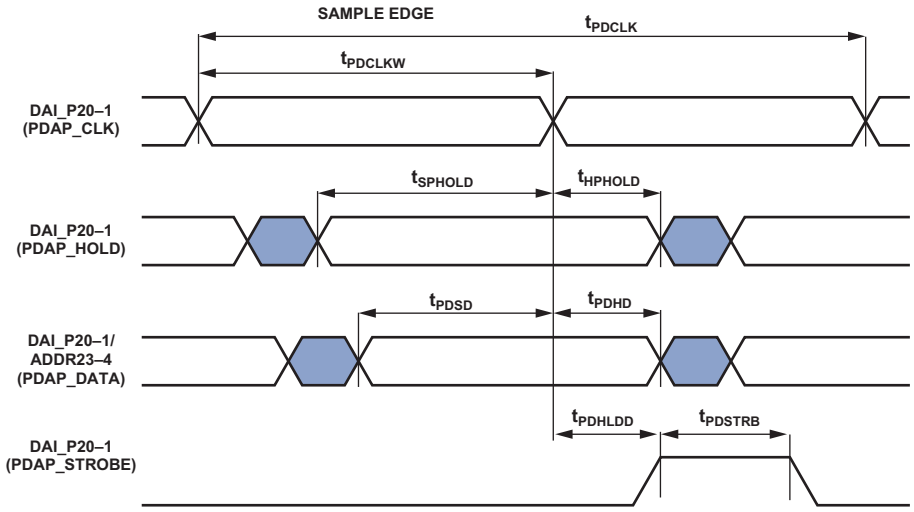

Figure 26. PDAP Timing

Media Local Bus

All the numbers given are applicable for all speed modes (1024 FS, 512 FS and 256 FS for 3-pin; 512 FS and 256 FS for 5-pin), unless otherwise specified. Please refer to the MediaLB specification document revision 3.0 for more details.

Table 52. MLB Interface, 3-Pin Specifications

Parameter		Min	Typ	Max	Unit
3-Pin Characteristics					
t_{MLBCLK}	MLB Clock Period				
	1024 FS		20.3		ns
	512 FS		40		ns
	256 FS		81		ns
t_{MCKL}	MLBCLK Low Time				
	1024 FS	6.1			ns
	512 FS	14			ns
	256 FS	30			ns
t_{MCKH}	MLBCLK High Time				
	1024 FS	9.3			ns
	512 FS	14			ns
	256 FS	30			ns
t_{MCKR}	MLBCLK Rise Time (V_{IL} to V_{IH})				
	1024 FS			1	ns
	512 FS/256 FS			3	ns
t_{MCKF}	MLBCLK Fall Time (V_{IH} to V_{IL})				
	1024 FS			1	ns
	512 FS/256 FS			3	ns
t_{MPWV}[1]	MLBCLK Pulse Width Variation				
	1024 FS			0.7	nspp
	512 FS/256			2.0	nspp
t_{DSMCF}	DAT/SIG Input Setup Time	1			ns
t_{DHMCF}	DAT/SIG Input Hold Time	2			ns
t_{MCFDZ}	DAT/SIG Output Time to Three-state	0		15	ns
t_{MCDRV}	DAT/SIG Output Data Delay From MLBCLK Rising Edge			8	ns
t_{MDZH}[2]	Bus Hold Time				
	1024 FS	2			ns
	512 FS/256	4			ns
C_{MLB}	DAT/SIG Pin Load				
	1024 FS			40	pf
	512 FS/256			60	pf

[1] Pulse width variation is measured at 1.25 V by triggering on one edge of MLBCLK and measuring the spread on the other edge, measured in ns peak-to-peak (pp).
[2] The board must be designed to ensure that the high-impedance bus does not leave the logic state of the final driven bit for this time period. Therefore, coupling must be minimized while meeting the maximum capacitive load listed.

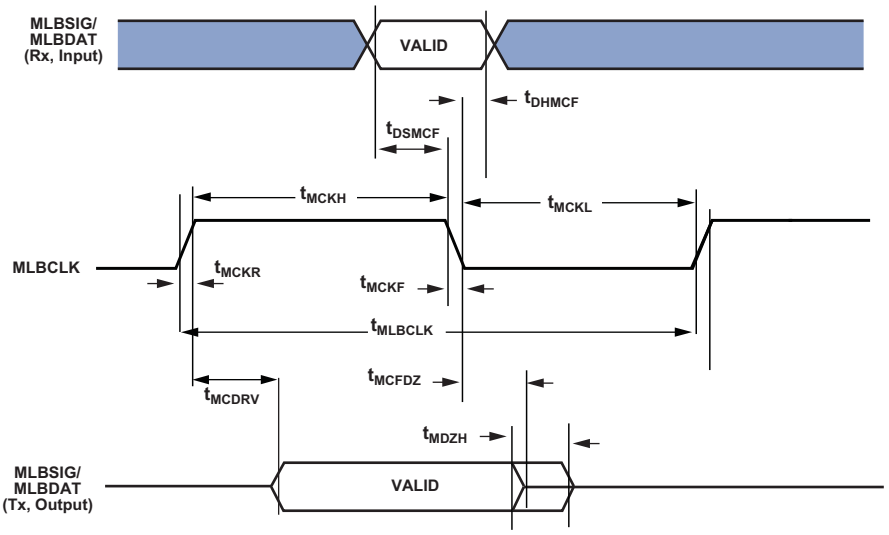

Figure 37. MLB Timing (3-Pin Interface)

Table 53. MLB Interface, 5-Pin Specifications

Parameter		Min	Typ	Max	Unit
5-Pin Characteristics					
t_{MLBCLK}	MLB Clock Period				
	512 FS		40		ns
	256 FS		81		ns
t_{MCKL}	MLBCLK Low Time				
	512 FS	15			ns
	256 FS	30			ns
t_{MCKH}	MLBCLK High Time				
	512 FS	15			ns
	256 FS	30			ns
t_{MCKR}	MLBCLK Rise Time (V_{IL} to V_{IH})			6	ns
t_{MCKF}	MLBCLK Fall Time (V_{IH} to V_{IL})			6	ns
t_{MPWV}[1]	MLBCLK Pulse Width Variation			2	nspp
t_{DSMCF}[2]	DAT/SIG Input Setup Time	3			ns
t_{DHMCF}	DAT/SIG Input Hold Time	5			ns
t_{MCDRV}	DS/DO Output Data Delay From MLBCLK Rising Edge			8	ns
t_{MCRDL}[3]	DO/SO Low From MLBCLK High				
	512 FS			10	ns
	256 FS			20	ns
C_{MLB}	DS/DO Pin Load			40	pf

[1] Pulse width variation is measured at 1.25 V by triggering on one edge of MLBCLK and measuring the spread on the other edge, measured in ns peak-to-peak (pp).
[2] Gate Delays due to OR'ing logic on the pins must be accounted for.
[3] When a node is not driving valid data onto the bus, the MLBSO and MLBDO output lines shall remain low. If the output lines can float at anytime, including while in reset, external pull-down resistors are required to keep the outputs from corrupting the MediaLB signal lines when not being driven.

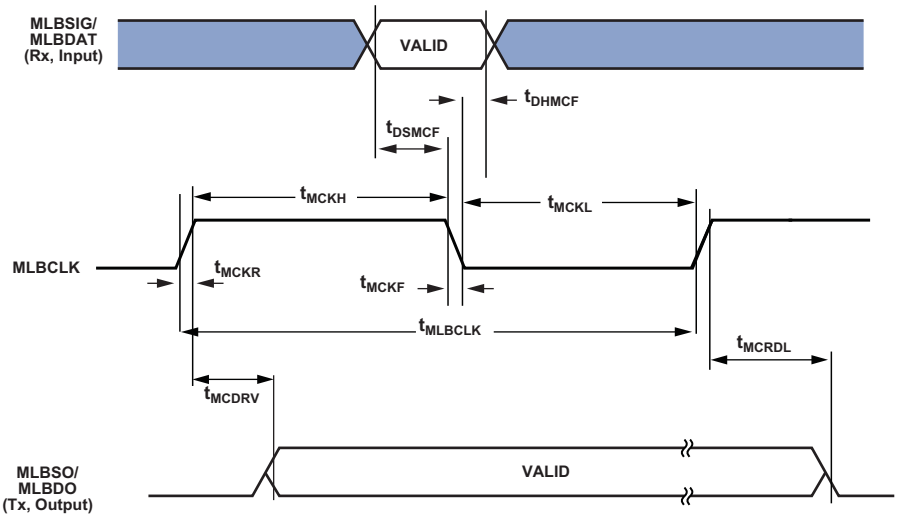

Figure 38. MLB Timing (5-Pin Interface)

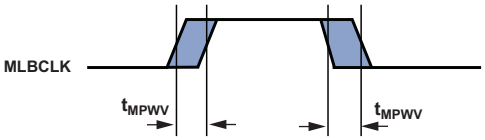

Figure 39. MLB 3-Pin and 5-Pin MLBCLK Pulse Width Variation Timing

Universal Asynchronous Receiver-Transmitter (UART) Ports—Receive and Transmit Timing

For information on the UART port receive and transmit operations, see the hardware reference.

2-Wire Interface (TWI)—Receive and Transmit Timing

For information on the TWI receive and transmit operations, see the hardware reference.

Figure 48 shows the top view of the 100-lead LQFP_EP lead configuration. Figure 49 shows the bottom view of the 100-lead LQFP_EP lead configuration.

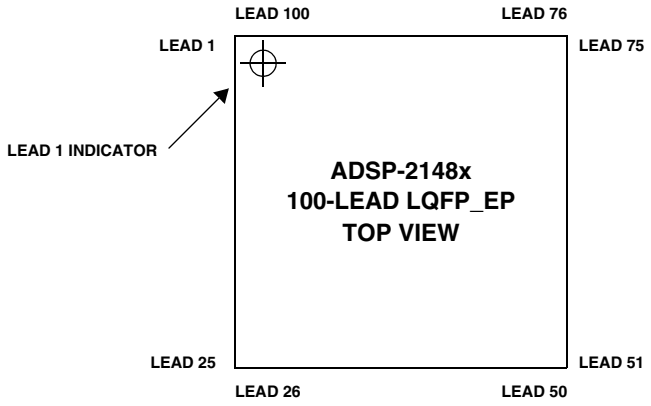

Figure 48. 100-Lead LQFP_EP Lead Configuration (Top View)

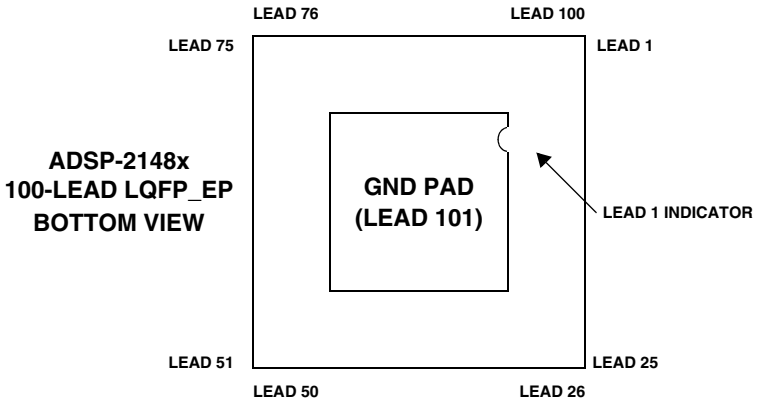

Figure 49. 100-Lead LQFP_EP Lead Configuration (Bottom View)

Table 61. ADSP-21483, ADSP-21487, ADSP-21488, and ADSP-21489 176-Lead LQFP_EP Lead Assignment (Numerical by Lead Number)

Lead Name	Lead No.	Lead Name	Lead No.	Lead Name	Lead No.	Lead Name	Lead No.
SDDQM	1	V_{DD_EXT}	45	DAI_P10	89	V_{DD_INT}	133
$\overline{MS0}$	2	DPI_P08	46	V_{DD_INT}	90	FLAG0	134
SDCKE	3	DPI_P07	47	V_{DD_EXT}	91	FLAG1	135
V_{DD_INT}	4	V_{DD_INT}	48	DAI_P20	92	FLAG2	136
CLK_CFG1	5	DPI_P09	49	V_{DD_INT}	93	GND	137
ADDR0	6	DPI_P10	50	DAI_P08	94	FLAG3	138
BOOT_CFG0	7	DPI_P11	51	DAI_P14	95	GND	139
V_{DD_EXT}	8	DPI_P12	52	DAI_P04	96	GND	140
ADDR1	9	DPI_P13	53	DAI_P18	97	V_{DD_EXT}	141
ADDR2	10	DPI_P14	54	DAI_P17	98	GND	142
ADDR3	11	DAI_P03	55	DAI_P16	99	V_{DD_INT}	143
ADDR4	12	NC	56	DAI_P12	100	$\overline{TRST}$	144
ADDR5	13	V_{DD_EXT}	57	DAI_P15	101	GND	145
BOOT_CFG1	14	NC	58	V_{DD_INT}	102	$\overline{EMU}$	146
GND	15	NC	59	DAI_P11	103	DATA0	147
ADDR6	16	NC	60	V_{DD_EXT}	104	DATA1	148
ADDR7	17	NC	61	V_{DD_INT}	105	DATA2	149
NC	18	V_{DD_INT}	62	BOOT_CFG2	106	DATA3	150
NC	19	NC	63	V_{DD_INT}	107	TDO	151
ADDR8	20	NC	64	AMI_ACK	108	DATA4	152
ADDR9	21	V_{DD_INT}	65	GND	109	V_{DD_EXT}	153
CLK_CFG0	22	NC	66	THD_M	110	DATA5	154
V_{DD_INT}	23	NC	67	THD_P	111	DATA6	155
CLKIN	24	V_{DD_INT}	68	V_{DD_THD}	112	V_{DD_INT}	156
XTAL	25	NC	69	V_{DD_INT}	113	DATA7	157
ADDR10	26	$\overline{WDTRSTO}$	70	V_{DD_INT}	114	TDI	158
SDA10	27	NC	71	$\overline{MS1}$	115	SDCLK	159
V_{DD_EXT}	28	V_{DD_EXT}	72	V_{DD_INT}	116	V_{DD_EXT}	160
V_{DD_INT}	29	DAI_P07	73	WDT_CLKO	117	DATA8	161
ADDR11	30	DAI_P13	74	WDT_CLKIN	118	DATA9	162
ADDR12	31	DAI_P19	75	V_{DD_EXT}	119	DATA10	163
ADDR17	32	DAI_P01	76	ADDR23	120	TCK	164
ADDR13	33	DAI_P02	77	ADDR22	121	DATA11	165
V_{DD_INT}	34	V_{DD_INT}	78	ADDR21	122	DATA12	166
ADDR18	35	NC	79	V_{DD_INT}	123	DATA14	167
$\overline{RESETOUT}$/$\overline{RUNRSTIN}$	36	NC	80	ADDR20	124	DATA13	168
V_{DD_INT}	37	NC	81	ADDR19	125	V_{DD_INT}	169
DPI_P01	38	NC	82	V_{DD_EXT}	126	DATA15	170
DPI_P02	39	NC	83	ADDR16	127	$\overline{SDWE}$	171
DPI_P03	40	V_{DD_EXT}	84	ADDR15	128	$\overline{SDRAS}$	172
V_{DD_INT}	41	V_{DD_INT}	85	V_{DD_INT}	129	$\overline{RESET}$	173
DPI_P05	42	DAI_P06	86	ADDR14	130	TMS	174
DPI_P04	43	DAI_P05	87	$\overline{AMI\_WR}$	131	$\overline{SDCAS}$	175
DPI_P06	44	DAI_P09	88	$\overline{AMI\_RD}$	132	V_{DD_INT}	176
						GND	177*

* Lead no. 177 (exposed pad) is the GND supply (see Figure 50 and Figure 51) for the processor; this pad must be **robustly** connected to GND.

Figure 50 shows the top view of the 176-lead LQFP_EP lead configuration. Figure 51 shows the bottom view of the 176-lead LQFP_EP lead configuration.

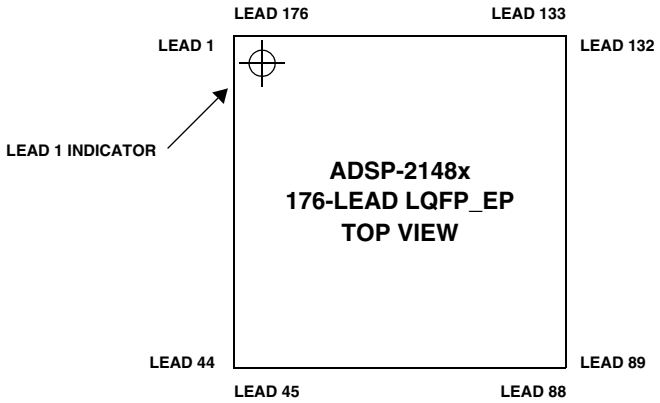

Figure 50. 176-Lead LQFP_EP Lead Configuration (Top View)

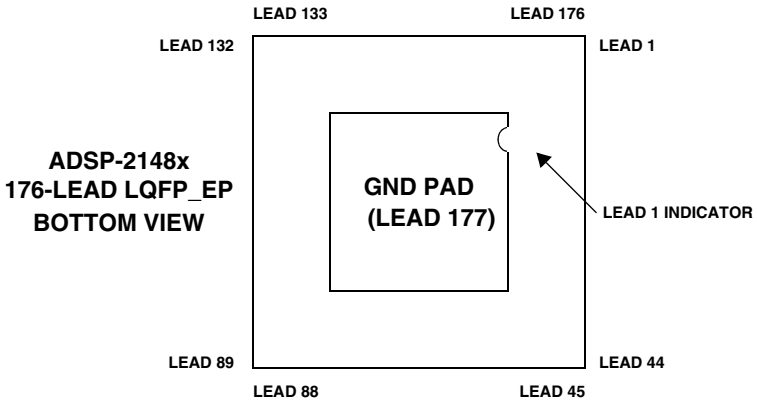

Figure 51. 176-Lead LQFP_EP Lead Configuration (Bottom View)

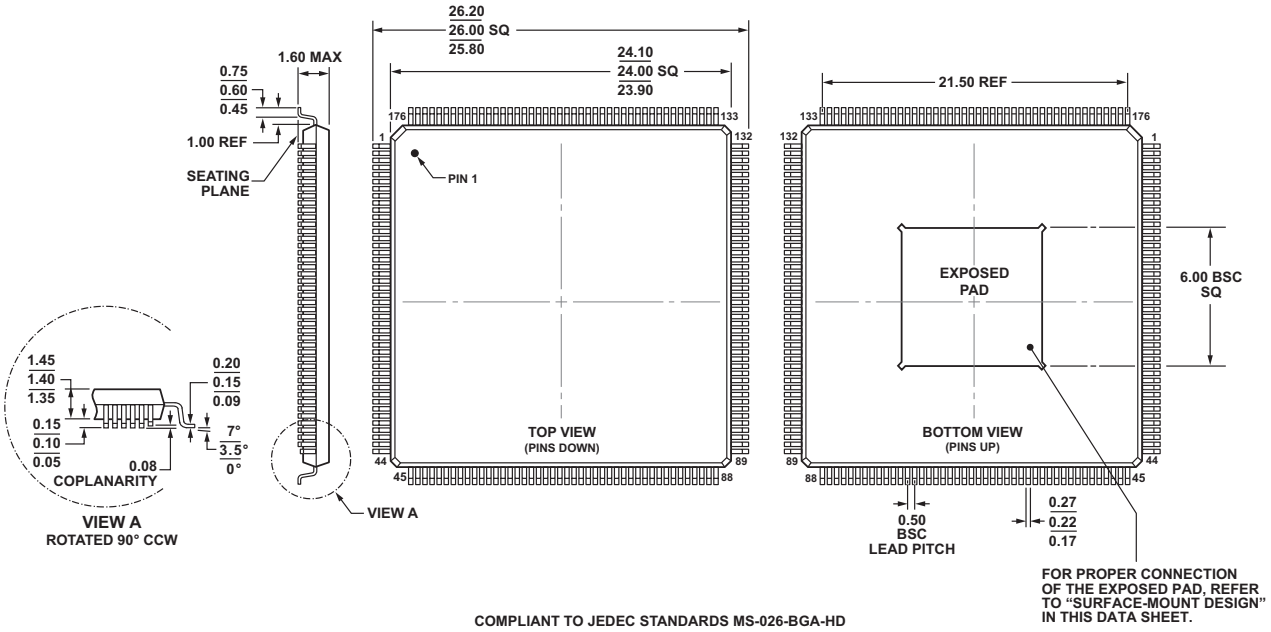

Figure 53. 176-Lead Low Profile Quad Flat Package, Exposed Pad [LQFP_EP][1]
(SW-176-2)
Dimensions shown in millimeters

[1] For information relating to the exposed pad on the SW-176-2 package, see the table endnote on Page 60.

SURFACE-MOUNT DESIGN

The exposed pad is required to be electrically and thermally connected to GND. Implement this by soldering the exposed pad to a GND PCB land that is the same size as the exposed pad. The GND PCB land should be robustly connected to the GND plane in the PCB for best electrical and thermal performance. No separate GND pins are provided in the package.

Model[1]	Notes	Temperature Range[2]	RAM	Processor Instruction Rate (Max)	Package Description	Package Option
ADSP-21487KSWZ-2B	3	0°C to +70°C	5 Mbit	300 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-2BB	3	0°C to +70°C	5 Mbit	300 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-3B	3	0°C to +70°C	5 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-3BB	3	0°C to +70°C	5 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-4B	3	0°C to +70°C	5 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-4BB	3	0°C to +70°C	5 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-5B	3, 4	0°C to +70°C	5 Mbit	450 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-5BB	3, 4	0°C to +70°C	5 Mbit	450 MHz	176-Lead LQFP_EP	SW-176-2
ADSP21487KSWZ5BBRL	5	0°C to +70°C	5 Mbit	450 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21488BSWZ-3A		–40°C to +85°C	3 Mbit	350 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21488KSWZ-3A		0°C to +70°C	3 Mbit	350 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21488KSWZ-3A1	6	0°C to +70°C	3 Mbit	350 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21488KSWZ-3B		0°C to +70°C	3 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21488BSWZ-3B		–40°C to +85°C	3 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21488KSWZ-4A		0°C to +70°C	3 Mbit	400 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21488BSWZ-4A		–40°C to +85°C	3 Mbit	400 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21488KSWZ-4B		0°C to +70°C	3 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21488BSWZ-4B		–40°C to +85°C	3 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21488KSWZ-4B1	6	0°C to +70°C	3 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21489KSWZ-3A		0°C to +70°C	5 Mbit	350 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21489BSWZ-3A		–40°C to +85°C	5 Mbit	350 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21489KSWZ-3B		0°C to +70°C	5 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21489BSWZ-3B		–40°C to +85°C	5 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21489KSWZ-4A		0°C to +70°C	5 Mbit	400 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21489BSWZ-4A		–40°C to +85°C	5 Mbit	400 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21489KSWZ-4B		0°C to +70°C	5 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21489BSWZ-4B		–40°C to +85°C	5 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21489KSWZ-5B	4	0°C to +70°C	5 Mbit	450 MHz	176-Lead LQFP_EP	SW-176-2

[1] Z = RoHS compliant part.

[2] Referenced temperature is ambient temperature. The ambient temperature is not a specification. Please see Operating Conditions on Page 18 for junction temperature (T_J) specification, which is the only temperature specification.

[3] The ADSP-21483, ADSP-21486, and ADSP-21487 models are available with factory programmed ROM including the latest multichannel audio decoding and post-processing algorithms from Dolby Labs and DTS. ROM contents may vary depending on chip version and silicon revision. Please visit www.analog.com for complete information.

[4] See Engineer-to-Engineer Note Static Voltage Scaling for ADSP-2148x SHARC Processors (EE-357) for operating ADSP-2148x processors at 450 MHz.

[5] RL = Tape and Reel.

[6] This product contains a –140 dB sample rate converter.