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### Understanding [Embedded - DSP \(Digital Signal Processors\)](#)

[Embedded - DSP \(Digital Signal Processors\)](#) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

### Applications of [Embedded - DSP \(Digital Signal Processors\)](#)

#### Details

Product Status	Active
Type	Floating Point
Interface	EBI/EMI, DAI, I <sup>2</sup> C, SPI, SPORT, UART/USART
Clock Rate	400MHz
Non-Volatile Memory	External
On-Chip RAM	5Mbit
Voltage - I/O	3.30V
Voltage - Core	1.10V
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	176-LQFP Exposed Pad
Supplier Device Package	176-LQFP-EP (24x24)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/analog-devices/adsp-21489kswz-4b">https://www.e-xfl.com/product-detail/analog-devices/adsp-21489kswz-4b</a>

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

The diagram on [Page 1](#) shows the two clock domains that make up the ADSP-2148x processors. The core clock domain contains the following features:

- Two processing elements (PE<sub>x</sub>, PE<sub>y</sub>), each of which comprises an ALU, multiplier, shifter, and data register file
- Data address generators (DAG1, DAG2)
- Program sequencer with instruction cache
- PM and DM buses capable of supporting 2x64-bit data transfers between memory and the core at every core processor cycle
- One periodic interval timer with pinout
- On-chip SRAM (5 Mbit) and mask-programmable ROM (4 Mbit)
- JTAG test access port for emulation and boundary scan. The JTAG provides software debug through user breakpoints which allows flexible exception handling.

The block diagram of the ADSP-2148x on [Page 1](#) also shows the peripheral clock domain (also known as the I/O processor) which contains the following features:

- IOD0 (peripheral DMA) and IOD1 (external port DMA) buses for 32-bit data transfers
- Peripheral and external port buses for core connection
- External port with an AMI and SDRAM controller
- 4 units for PWM control
- 1 memory-to-memory (MTM) unit for internal-to-internal memory transfers
- Digital applications interface that includes four precision clock generators (PCG), an input data port (IDP/PDAP) for serial and parallel interconnects, an S/PDIF receiver/transmitter, four asynchronous sample rate converters, eight serial ports, and a flexible signal routing unit (DAI SRU).
- Digital peripheral interface that includes two timers, a 2-wire interface (TWI), one UART, two serial peripheral interfaces (SPI), 2 precision clock generators (PCG), pulse width modulation (PWM), and a flexible signal routing unit (DPI SRU2).

As shown in the SHARC core block diagram on [Page 5](#), the processor uses two computational units to deliver a significant performance increase over the previous SHARC processors on a range of DSP algorithms. With its SIMD computational hardware, the processors can perform 2.7 GFLOPS running at 450 MHz.

## FAMILY CORE ARCHITECTURE

The ADSP-2148x is code compatible at the assembly level with the ADSP-2147x, ADSP-2146x, ADSP-2137x, ADSP-2136x, ADSP-2126x, ADSP-21160, and ADSP-21161, and with the first generation ADSP-2106x SHARC processors. The ADSP-2148x shares architectural features with the ADSP-2126x, ADSP-2136x, ADSP-2137x, ADSP-2146x and ADSP-2116x SIMD SHARC processors, as shown in [Figure 2](#) and detailed in the following sections.

## SIMD Computational Engine

The ADSP-2148x contains two computational processing elements that operate as a single-instruction, multiple-data (SIMD) engine. The processing elements are referred to as PEX and PEY and each contains an ALU, multiplier, shifter, and register file. PEX is always active, and PEY may be enabled by setting the PEYEN mode bit in the MODE1 register. SIMD mode allows the processor to execute the same instruction in both processing elements, but each processing element operates on different data. This architecture is efficient at executing math intensive DSP algorithms.

SIMD mode also affects the way data is transferred between memory and the processing elements because twice the data bandwidth is required to sustain computational operation in the processing elements. Therefore, entering SIMD mode also doubles the bandwidth between memory and the processing elements. When using the DAGs to transfer data in SIMD mode, two data values are transferred with each memory or register file access.

## Independent, Parallel Computation Units

Within each processing element is a set of computational units. The computational units consist of an arithmetic/logic unit (ALU), multiplier, and shifter. These units perform all operations in a single cycle and are arranged in parallel, maximizing computational throughput. Single multifunction instructions execute parallel ALU and multiplier operations. In SIMD mode, the parallel ALU and multiplier operations occur in both processing elements. These computation units support IEEE 32-bit single-precision floating-point, 40-bit extended precision floating-point, and 32-bit fixed-point data formats.

## Timer

The processor contains a core timer that can generate periodic software interrupts. The core timer can be configured to use FLAG3 as a timer expired signal.

## Data Register File

Each processing element contains a general-purpose data register file. The register files transfer data between the computation units and the data buses, and store intermediate results. These 10-port, 32-register (16 primary, 16 secondary) register files, combined with the processor's enhanced Harvard architecture, allow unconstrained data flow between computation units and internal memory. The registers in PEX are referred to as R0–R15 and in PEY as S0–S15.

## Context Switch

Many of the processor's registers have secondary registers that can be activated during interrupt servicing for a fast context switch. The data registers in the register file, the DAG registers, and the multiplier result registers all have secondary registers. The primary registers are active at reset, while the secondary registers are activated by control bits in a mode control register.

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

- Arbitration logic to coordinate core and DMA transfers between internal and external memory over the external port.

Non-SDRAM external memory address space is shown in [Table 5](#).

**Table 5. External Memory for Non-SDRAM Addresses**

Bank	Size in Words	Address Range
Bank 0	6M	0x0020 0000–0x007F FFFF
Bank 1	8M	0x0400 0000–0x047F FFFF
Bank 2	8M	0x0800 0000–0x087F FFFF
Bank 3	8M	0x0C00 0000–0x0C7F FFFF

## External Port

The external port provides a high performance, glueless interface to a wide variety of industry-standard memory devices. The external port, available on the 176-lead LQFP, may be used to interface to synchronous and/or asynchronous memory devices through the use of its separate internal memory controllers. The first is an SDRAM controller for connection of industry-standard synchronous DRAM devices while the second is an asynchronous memory controller intended to interface to a variety of memory devices. Four memory select pins enable up to four separate devices to coexist, supporting any desired combination of synchronous and asynchronous device types.

## Asynchronous Memory Controller

The asynchronous memory controller provides a configurable interface for up to four separate banks of memory or I/O devices. Each bank can be independently programmed with different timing parameters, enabling connection to a wide variety of memory devices including SRAM, flash, and EPROM, as well as I/O devices that interface with standard memory control lines. Bank 0 occupies a 6M word window and banks 1, 2, and 3 occupy a 8M word window in the processor's address space but, if not fully populated, these windows are not made contiguous by the memory controller logic.

## SDRAM Controller

The SDRAM controller provides an interface of up to four separate banks of industry-standard SDRAM devices at speeds up to  $f_{SDCLK}$ . Fully compliant with the SDRAM standard, each bank has its own memory select line ( $\overline{MS0}$ – $\overline{MS3}$ ), and can be configured to contain between 4M bytes and 256M bytes of memory. SDRAM external memory address space is shown in [Table 6](#). NOTE: this feature is not available on the ADSP-21486 model.

**Table 6. External Memory for SDRAM Addresses**

Bank	Size in Words	Address Range
Bank 0	62M	0x0020 0000–0x03FF FFFF
Bank 1	64M	0x0400 0000–0x07FF FFFF
Bank 2	64M	0x0800 0000–0x0BFF FFFF
Bank 3	64M	0x0C00 0000–0x0FFF FFFF

A set of programmable timing parameters is available to configure the SDRAM banks to support slower memory devices. Note that 32-bit wide devices are not supported on the SDRAM and AMI interfaces.

The SDRAM controller address, data, clock, and control pins can drive loads up to distributed 30 pF. For larger memory systems, the SDRAM controller external buffer timing should be selected and external buffering should be provided so that the load on the SDRAM controller pins does not exceed 30 pF.

Note that the external memory bank addresses shown are for normal-word (32-bit) accesses. If 48-bit instructions as well as 32-bit data are both placed in the same external memory bank, care must be taken while mapping them to avoid overlap.

## SIMD Access to External Memory

The SDRAM controller on the processor supports SIMD access on the 64-bit EPD (external port data bus) which allows access to the complementary registers on the PEy unit in the normal word space (NW). This removes the need to explicitly access the complimentary registers when the data is in external SDRAM memory.

## VISA and ISA Access to External Memory

The SDRAM controller on the ADSP-2148x processors supports VISA code operation which reduces the memory load since the VISA instructions are compressed. Moreover, bus fetching is reduced because, in the best case, one 48-bit fetch contains three valid instructions. Code execution from the traditional ISA operation is also supported. Note that code execution is only supported from bank 0 regardless of VISA/ISA. [Table 7](#) shows the address ranges for instruction fetch in each mode.

**Table 7. External Bank 0 Instruction Fetch**

Access Type	Size in Words	Address Range
ISA (NW)	4M	0x0020 0000–0x005F FFFF
VISA (SW)	10M	0x0060 0000–0x00FF FFFF

## Pulse-Width Modulation

The PWM module is a flexible, programmable, PWM waveform generator that can be programmed to generate the required switching patterns for various applications related to motor and engine control or audio power control. The PWM generator can generate either center-aligned or edge-aligned PWM waveforms. In addition, it can generate complementary signals on two outputs in paired mode or independent signals in non-paired mode (applicable to a single group of four PWM waveforms).

The entire PWM module has four groups of four PWM outputs generating 16 PWM outputs in total. Each PWM group produces two pairs of PWM signals on the four PWM outputs.

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

Details on power consumption and Static and Dynamic current consumption can be found at [Total Power Dissipation on Page 20](#). Also see [Operating Conditions on Page 18](#) for more information.

The following are SVS features.

- SVS is applicable only to 450 MHz models (not applicable to 400 MHz or lower frequency models).
- Each individual SVS device includes a register (SVS\_DAT) containing the unique SVS voltage set at the factory, known as SVS<sub>NOM</sub>.
- The SVS<sub>NOM</sub> value is the intended set voltage for the V<sub>DD\_INT</sub> voltage regulator.
- No dedicated pins are required for SVS. The TWI serial bus is used to communicate SVS<sub>NOM</sub> to the voltage regulator.
- Analog Devices recommends a specific voltage regulator design and initialization code sequence that optimizes the power-up sequence.

The Engineer-to-Engineer Note [Static Voltage Scaling for ADSP-2148x SHARC Processors \(EE-357\)](#) contains the details of the regulator design and the initialization requirements.

- Any differences from the Analog Devices recommended programmable regulator design must be reviewed by Analog Devices to ensure that it meets the voltage accuracy and range requirements.

## Target Board JTAG Emulator Connector

Analog Devices DSP Tools product line of JTAG emulators uses the IEEE 1149.1 JTAG test access port of the ADSP-2148x processors to monitor and control the target board processor during emulation. Analog Devices DSP Tools product line of JTAG emulators provides emulation at full processor speed, allowing inspection and modification of memory, registers, and processor stacks. The processor's JTAG interface ensures that the emulator will not affect target system loading or timing.

For complete information on Analog Devices' SHARC DSP Tools product line of JTAG emulator operation, see the appropriate emulator hardware user's guide.

## DEVELOPMENT TOOLS

Analog Devices supports its processors with a complete line of software and hardware development tools, including integrated development environments (which include CrossCore<sup>®</sup> Embedded Studio and/or VisualDSP++<sup>®</sup>), evaluation products, emulators, and a wide variety of software add-ins.

### Integrated Development Environments (IDEs)

For C/C++ software writing and editing, code generation, and debug support, Analog Devices offers two IDEs.

CrossCore Embedded Studio is based on the Eclipse™ framework. Supporting most Analog Devices processor families, it is the IDE of choice for future processors, including multicore devices. CrossCore Embedded Studio seamlessly integrates available software add-ins to support real time operating sys-

tems, file systems, TCP/IP stacks, USB stacks, algorithmic software modules, and evaluation hardware board support packages. For more information visit [www.analog.com/cces](http://www.analog.com/cces).

The other Analog Devices IDE, VisualDSP++, supports processor families introduced prior to the release of CrossCore Embedded Studio. This IDE includes the Analog Devices VDK real time operating system and an open source TCP/IP stack. For more information visit [www.analog.com/visualdsp](http://www.analog.com/visualdsp). Note that VisualDSP++ will not support future Analog Devices processors.

### EZ-KIT Lite Evaluation Board

For processor evaluation, Analog Devices provides wide range of EZ-KIT Lite<sup>®</sup> evaluation boards. Including the processor and key peripherals, the evaluation board also supports on-chip emulation capabilities and other evaluation and development features. Also available are various EZ-Extenders<sup>®</sup>, which are daughter cards delivering additional specialized functionality, including audio and video processing. For more information visit [www.analog.com](http://www.analog.com) and search on "ezkit" or "ezextender".

### EZ-KIT Lite Evaluation Kits

For a cost-effective way to learn more about developing with Analog Devices processors, Analog Devices offer a range of EZ-KIT Lite evaluation kits. Each evaluation kit includes an EZ-KIT Lite evaluation board, directions for downloading an evaluation version of the available IDE(s), a USB cable, and a power supply. The USB controller on the EZ-KIT Lite board connects to the USB port of the user's PC, enabling the chosen IDE evaluation suite to emulate the on-board processor in-circuit. This permits the customer to download, execute, and debug programs for the EZ-KIT Lite system. It also supports in-circuit programming of the on-board Flash device to store user-specific boot code, enabling standalone operation. With the full version of CrossCore Embedded Studio or VisualDSP++ installed (sold separately), engineers can develop software for supported EZ-KITs or any custom system utilizing supported Analog Devices processors.

### Software Add-Ins for CrossCore Embedded Studio

Analog Devices offers software add-ins which seamlessly integrate with CrossCore Embedded Studio to extend its capabilities and reduce development time. Add-ins include board support packages for evaluation hardware, various middleware packages, and algorithmic modules. Documentation, help, configuration dialogs, and coding examples present in these add-ins are viewable through the CrossCore Embedded Studio IDE once the add-in is installed.

### Board Support Packages for Evaluation Hardware

Software support for the EZ-KIT Lite evaluation boards and EZ-Extender daughter cards is provided by software add-ins called Board Support Packages (BSPs). The BSPs contain the required drivers, pertinent release notes, and select example code for the given evaluation hardware. A download link for a specific BSP is located on the web page for the associated EZ-KIT or EZ-Extender product. The link is found in the Product Download area of the product web page.

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

## PIN FUNCTION DESCRIPTIONS

Table 11. Pin Descriptions

Name	Type	State During/ After Reset	Description
ADDR <sub>23-0</sub>	I/O/T (ipu)	High-Z/ driven low (boot)	<b>External Address.</b> The processor outputs addresses for external memory and peripherals on these pins. The ADDR pins can be multiplexed to support the external memory interface address, and FLAGS <sub>15-8</sub> (I/O) and PWM (O). After reset, all ADDR pins are in external memory interface mode and FLAG(0-3) pins are in FLAGS mode (default). When configured in the IDP_PDAP_CTL register, IDP channel 0 scans the ADDR <sub>23-4</sub> pins for parallel input data.
DATA <sub>15-0</sub>	I/O/T (ipu)	High-Z	<b>External Data.</b> The data pins can be multiplexed to support the external memory interface data (I/O), and FLAGS <sub>7-0</sub> (I/O).
AMI_ACK	I (ipu)		<b>Memory Acknowledge.</b> External devices can deassert AMI_ACK (low) to add wait states to an external memory access. AMI_ACK is used by I/O devices, memory controllers, or other peripherals to hold off completion of an external memory access.
$\overline{MS}_{0-1}$	O/T (ipu)	High-Z	<b>Memory Select Lines 0-1.</b> These lines are asserted (low) as chip selects for the corresponding banks of external memory. The $\overline{MS}_{1-0}$ lines are decoded memory address lines that change at the same time as the other address lines. When no external memory access is occurring the $\overline{MS}_{1-0}$ lines are inactive; they are active however when a conditional memory access instruction is executed, when the condition evaluates as true. The $\overline{MS}_1$ pin can be used in EPORT/FLASH boot mode. For more information, see the hardware reference.
$\overline{AMI\_RD}$	O/T (ipu)	High-Z	<b>AMI Port Read Enable.</b> $\overline{AMI\_RD}$ is asserted whenever the processor reads a word from external memory.
$\overline{AMI\_WR}$	O/T (ipu)	High-Z	<b>AMI Port Write Enable.</b> $\overline{AMI\_WR}$ is asserted when the processor writes a word to external memory.
FLAG0/ $\overline{IRQ0}$	I/O (ipu)	FLAG[0] INPUT	<b>FLAG0/Interrupt Request0.</b>
FLAG1/ $\overline{IRQ1}$	I/O (ipu)	FLAG[1] INPUT	<b>FLAG1/Interrupt Request1.</b>
FLAG2/ $\overline{IRQ2}/\overline{MS2}$	I/O (ipu)	FLAG[2] INPUT	<b>FLAG2/Interrupt Request2/Memory Select2.</b>
FLAG3/TMREXP/ $\overline{MS3}$	I/O (ipu)	FLAG[3] INPUT	<b>FLAG3/Timer Expired/Memory Select3.</b>

The following symbols appear in the Type column of this table: **A** = asynchronous, **I** = input, **O** = output, **S** = synchronous, **A/D** = active drive, **O/D** = open drain, and **T** = three-state, **ipd** = internal pull-down resistor, **ipu** = internal pull-up resistor.

The internal pull-up (ipu) and internal pull-down (ipd) resistors are designed to hold the internal path from the pins at the expected logic levels. To pull-up or pull-down the external pads to the expected logic levels, use external resistors. Internal pull-up/pull-down resistors cannot be enabled/disabled and the value of these resistors cannot be programmed. The range of an ipu resistor can be between 26 k $\Omega$ –63 k $\Omega$ . The range of an ipd resistor can be between 31 k $\Omega$ –85k $\Omega$ . The three-state voltage of ipu pads will not reach to the full V<sub>DD\_EXT</sub> level; at typical conditions the voltage is in the range of 2.3 V to 2.7 V.

In this table, all pins are LVTTTL compliant with the exception of the thermal diode pins.

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

**Table 11. Pin Descriptions (Continued)**

Name	Type	State During/ After Reset	Description
$\overline{\text{SDRAS}}$	O/T (ipu)	High-Z/ driven high	<b>SDRAM Row Address Strobe.</b> Connect to SDRAM's RAS pin. In conjunction with other SDRAM command pins, defines the operation for the SDRAM to perform.
$\overline{\text{SDCAS}}$	O/T (ipu)	High-Z/ driven high	<b>SDRAM Column Address Select.</b> Connect to SDRAM's CAS pin. In conjunction with other SDRAM command pins, defines the operation for the SDRAM to perform.
$\overline{\text{SDWE}}$	O/T (ipu)	High-Z/ driven high	<b>SDRAM Write Enable.</b> Connect to SDRAM's WE or W buffer pin. In conjunction with other SDRAM command pins, defines the operation for the SDRAM to perform.
SDCKE	O/T (ipu)	High-Z/ driven high	<b>SDRAM Clock Enable.</b> Connect to SDRAM's CE pin. Enables and disables the CLK signal. For details, see the data sheet supplied with the SDRAM device.
SDA10	O/T (ipu)	High-Z/ driven high	<b>SDRAM A10 Pin.</b> Enables applications to refresh an SDRAM in parallel with non-SDRAM accesses. This pin replaces the DSP's ADDR10 pin only during SDRAM accesses.
SDDQM	O/T (ipu)	High-Z/ driven high	<b>DQM Data Mask.</b> SDRAM Input mask signal for write accesses and output mask signal for read accesses. Input data is masked when DQM is sampled high during a write cycle. The SDRAM output buffers are placed in a High-Z state when DQM is sampled high during a read cycle. SDDQM is driven high from reset de-assertion until SDRAM initialization completes. Afterwards it is driven low irrespective of whether any SDRAM accesses occur or not.
SDCLK	O/T (ipd)	High-Z/ driving	<b>SDRAM Clock Output.</b> Clock driver for this pin differs from all other clock drivers. See <a href="#">Figure 41 on Page 55</a> . For models in the 100-lead package, the SDRAM interface should be disabled to avoid unnecessary power switching by setting the DSDCTL bit in SDCTL register. For more information, see the hardware reference.
DAI_P20-1	I/O/T (ipu)	High-Z	<b>Digital Applications Interface.</b> These pins provide the physical interface to the DAI SRU. The DAI SRU configuration registers define the combination of on-chip audio-centric peripheral inputs or outputs connected to the pin and to the pin's output enable. The configuration registers of these peripherals then determines the exact behavior of the pin. Any input or output signal present in the DAI SRU may be routed to any of these pins.
DPI_P14-1	I/O/T (ipu)	High-Z	<b>Digital Peripheral Interface.</b> These pins provide the physical interface to the DPI SRU. The DPI SRU configuration registers define the combination of on-chip peripheral inputs or outputs connected to the pin and to the pin's output enable. The configuration registers of these peripherals then determines the exact behavior of the pin. Any input or output signal present in the DPI SRU may be routed to any of these pins.
WDT_CLKIN	I		<b>Watchdog Timer Clock Input.</b> This pin should be pulled low when not used.
WDT_CLKO	O		<b>Watchdog Resonator Pad Output.</b>
$\overline{\text{WDRSTO}}$	O (ipu)		<b>Watchdog Timer Reset Out.</b>
THD_P	I		<b>Thermal Diode Anode.</b> When not used, this pin can be left floating.
THD_M	O		<b>Thermal Diode Cathode.</b> When not used, this pin can be left floating.

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The internal pull-up (ipu) and internal pull-down (ipd) resistors are designed to hold the internal path from the pins at the expected logic levels. To pull-up or pull-down the external pads to the expected logic levels, use external resistors. Internal pull-up/pull-down resistors cannot be enabled/disabled and the value of these resistors cannot be programmed. The range of an ipu resistor can be between 26 k $\Omega$ –63 k $\Omega$ . The range of an ipd resistor can be between 31 k $\Omega$ –85k $\Omega$ . The three-state voltage of ipu pads will not reach to the full V<sub>DD\_EXT</sub> level; at typical conditions the voltage is in the range of 2.3 V to 2.7 V.

In this table, all pins are LVTTTL compliant with the exception of the thermal diode pins.

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

Table 11. Pin Descriptions (Continued)

Name	Type	State During/ After Reset	Description
MLBCLK <sup>1</sup>	I		<b>Media Local Bus Clock.</b> This clock is generated by the MLB controller that is synchronized to the MOST network and provides the timing for the entire MLB interface at 49.152 MHz at FS=48 kHz. When the MLB controller is not used, this pin should be grounded.
MLBDAT <sup>1</sup>	I/O/T in 3 pin mode. I in 5 pin mode.	High-Z	<b>Media Local Bus Data.</b> The MLBDAT line is driven by the transmitting MLB device and is received by all other MLB devices including the MLB controller. The MLBDAT line carries the actual data. In 5-pin MLB mode, this pin is an input only. When the MLB controller is not used, this pin should be grounded.
MLBSIG <sup>1</sup>	I/O/T in 3 pin mode. I in 5 pin mode	High-Z	<b>Media Local Bus Signal.</b> This is a multiplexed signal which carries the Channel/Address generated by the MLB Controller, as well as the Command and RxStatus bytes from MLB devices. In 5-pin mode, this pin is input only. When the MLB controller is not used, this pin should be grounded.
MLBDO <sup>1</sup>	O/T	High-Z	<b>Media Local Bus Data Output (in 5 pin mode).</b> This pin is used only in 5-pin MLB mode. This serves as the output data pin in 5-pin mode. When the MLB controller is not used, this pin should be connected to ground.
MLBSO <sup>1</sup>	O/T	High-Z	<b>Media Local Bus Signal Output (in 5 pin mode).</b> This pin is used only in 5-pin MLB mode. This serves as the output signal pin in 5-pin mode. When the MLB controller is not used, this pin should be connected to ground.
TDI	I (ipu)	High-Z	<b>Test Data Input (JTAG).</b> Provides serial data for the boundary scan logic.
TDO	O/T		<b>Test Data Output (JTAG).</b> Serial scan output of the boundary scan path.
TMS	I (ipu)		<b>Test Mode Select (JTAG).</b> Used to control the test state machine.
TCK	I		<b>Test Clock (JTAG).</b> Provides a clock for JTAG boundary scan. TCK must be asserted (pulsed low) after power-up or held low for proper operation of the device.
$\overline{\text{TRST}}$	I (ipu)		<b>Test Reset (JTAG).</b> Resets the test state machine. $\overline{\text{TRST}}$ must be asserted (pulsed low) after power-up or held low for proper operation of the processor.
$\overline{\text{EMU}}$	O (O/D, ipu)	High-Z	<b>Emulation Status.</b> Must be connected to the ADSP-2148x Analog Devices DSP Tools product line of JTAG emulators target board connector only.

The following symbols appear in the Type column of this table: **A** = asynchronous, **I** = input, **O** = output, **S** = synchronous, **A/D** = active drive, **O/D** = open drain, and **T** = three-state, **ipd** = internal pull-down resistor, **ipu** = internal pull-up resistor.

The internal pull-up (ipu) and internal pull-down (ipd) resistors are designed to hold the internal path from the pins at the expected logic levels. To pull-up or pull-down the external pads to the expected logic levels, use external resistors. Internal pull-up/pull-down resistors cannot be enabled/disabled and the value of these resistors cannot be programmed. The range of an ipu resistor can be between 26 k $\Omega$ –63 k $\Omega$ . The range of an ipd resistor can be between 31 k $\Omega$ –85k $\Omega$ . The three-state voltage of ipu pads will not reach to the full V<sub>DD\_EXT</sub> level; at typical conditions the voltage is in the range of 2.3 V to 2.7 V.

In this table, all pins are LVTTTL compliant with the exception of the thermal diode pins.

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

## ELECTRICAL CHARACTERISTICS

Parameter <sup>1</sup>	Description	Test Conditions	300 MHz / 350 MHz / 400 MHz / 450 MHz			Unit
			Min	Typ	Max	
V <sub>OH</sub> <sup>2</sup>	High Level Output Voltage	@ V <sub>DD_EXT</sub> = Min, I <sub>OH</sub> = -1.0 mA <sup>3</sup>	2.4			V
V <sub>OL</sub> <sup>2</sup>	Low Level Output Voltage	@ V <sub>DD_EXT</sub> = Min, I <sub>OL</sub> = 1.0 mA <sup>3</sup>			0.4	V
I <sub>IH</sub> <sup>4,5</sup>	High Level Input Current	@ V <sub>DD_EXT</sub> = Max, V <sub>IN</sub> = V <sub>DD_EXT</sub> Max			10	μA
I <sub>IL</sub> <sup>4</sup>	Low Level Input Current	@ V <sub>DD_EXT</sub> = Max, V <sub>IN</sub> = 0 V			10	μA
I <sub>ILPU</sub> <sup>5</sup>	Low Level Input Current Pull-up	@ V <sub>DD_EXT</sub> = Max, V <sub>IN</sub> = 0 V			200	μA
I <sub>OZH</sub> <sup>6,7</sup>	Three-State Leakage Current	@ V <sub>DD_EXT</sub> = Max, V <sub>IN</sub> = V <sub>DD_EXT</sub> Max			10	μA
I <sub>OZL</sub> <sup>6</sup>	Three-State Leakage Current	@ V <sub>DD_EXT</sub> = Max, V <sub>IN</sub> = 0 V			10	μA
I <sub>OZLPU</sub> <sup>7</sup>	Three-State Leakage Current Pull-up	@ V <sub>DD_EXT</sub> = Max, V <sub>IN</sub> = 0 V			200	μA
I <sub>OZHPD</sub> <sup>8</sup>	Three-State Leakage Current Pull-down	@ V <sub>DD_EXT</sub> = Max, V <sub>IN</sub> = V <sub>DD_EXT</sub> Max			200	μA
I <sub>DD_INT</sub> <sup>9</sup>	Supply Current (Internal)	f <sub>CCLK</sub> > 0 MHz			Table 14 + Table 15 × ASF	mA
I <sub>DD_INT</sub>	Supply Current (Internal)	V <sub>DDINT</sub> = 1.1 V, ASF = 1, T <sub>J</sub> = 25°C		410 / 450 / 500 / 550		mA
C <sub>IN</sub> <sup>10, 11</sup>	Input Capacitance	T <sub>CASE</sub> = 25°C			5	pF

<sup>1</sup> Specifications subject to change without notice.

<sup>2</sup> Applies to output and bidirectional pins: ADDR23-0, DATA15-0, AMI\_RD, AMI\_WR, FLAG3-0, DAL\_Px, DPI\_Px, EMU, TDO, RESETOUT, MLBSIG, MLBDAT, MLBDO, MLBSO, SDRAS, SDCAS, SDWE, SDCKE, SDA10, SDDQM, MS0-I.

<sup>3</sup> See [Output Drive Currents on Page 55](#) for typical drive current capabilities.

<sup>4</sup> Applies to input pins: BOOT\_CFGx, CLK\_CFGx, TCK, RESET, CLKIN.

<sup>5</sup> Applies to input pins with internal pull-ups: TRST, TMS, TDI.

<sup>6</sup> Applies to three-statable pin: TDO.

<sup>7</sup> Applies to three-statable pins with pull-ups: DAL\_Px, DPI\_Px, EMU.

<sup>8</sup> Applies to three-statable pin with pull-down: SDCLK.

<sup>9</sup> See Engineer-to-Engineer Note [Estimating Power for ADSP-214xx SHARC Processors \(EE-348\)](#) for further information.

<sup>10</sup> Applies to all signal pins.

<sup>11</sup> Guaranteed, but not tested.

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

## Total Power Dissipation

The information in this section should be augmented with the Engineer-to-Engineer Note [Estimating Power for ADSP-214xx SHARC Processors \(EE-348\)](#).

Total power dissipation has two components:

- Internal power consumption is additionally comprised of two components:
  - Static current due to leakage. [Table 14](#) shows the static current consumption ( $I_{DD\_INT\_STATIC}$ ) as a function of junction temperature ( $T_J$ ) and core voltage ( $V_{DD\_INT}$ ).
  - Dynamic current ( $I_{DD\_INT\_DYNAMIC}$ ), due to transistor switching characteristics and activity level of the processor. The activity level is reflected by the Activity Scaling Factor (ASF), which represents the activity level of the application code running on the processor core and having various levels of peripheral and external port activity ([Table 13](#)).

Dynamic current consumption is calculated by selecting the ASF that corresponds most closely with the user application and then multiplying that with the dynamic current consumption ([Table 15](#)).

- External power consumption is due to the switching activity of the external pins.

**Table 13. Activity Scaling Factors (ASF)<sup>1</sup>**

Activity	Scaling Factor (ASF)
Idle	0.29
Low	0.53
Medium Low	0.61
Medium High	0.77
Peak Typical (50:50) <sup>2</sup>	0.85
Peak Typical (60:40) <sup>2</sup>	0.93
Peak Typical (70:30) <sup>2</sup>	1.00
High Typical	1.16
High	1.25
Peak	1.31

<sup>1</sup> See the Engineer-to-Engineer Note [Estimating Power for ADSP-214xx SHARC Processors \(EE-348\)](#) for more information on the explanation of the power vectors specific to the ASF table.

<sup>2</sup> Ratio of continuous instruction loop (core) to SDRAM control code reads and writes.

**Table 14. Static Current— $I_{DD\_INT\_STATIC}$  (mA)<sup>1</sup>**

$T_J$ (°C)	$V_{DD\_INT}$ (V)								
	0.975 V	1.0 V	1.025 V	1.05 V	1.075 V	1.10 V	1.125 V	1.15 V	1.175 V
-45	68	77	86	96	107	118	131	144	159
-35	74	83	92	103	114	126	140	154	170
-25	82	92	101	113	125	138	153	168	185
-15	94	104	115	127	140	155	171	187	205
-5	109	121	133	147	161	177	194	212	233
+5	129	142	156	171	188	206	225	245	268
+15	152	168	183	201	219	240	261	285	309
+25	182	199	216	237	257	280	305	331	360
+35	217	237	256	279	303	329	358	388	420
+45	259	282	305	331	359	389	421	455	492
+55	309	334	361	391	423	458	495	533	576
+65	369	398	429	464	500	539	582	626	675
+75	437	471	506	547	588	633	682	731	789
+85	519	559	599	645	693	746	802	860	926
+95	615	662	707	761	816	877	942	1007	1083
+105	727	779	833	897	958	1026	1103	1179	1266
+115	853	914	975	1047	1119	1198	1285	1372	1473
+125	997	1067	1138	1219	1305	1397	1498	1601	1716

<sup>1</sup> Valid temperature and voltage ranges are model-specific. See [Operating Conditions on Page 18](#).

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

Table 15. Dynamic Current in CCLK Domain— $I_{DD\_INT\_DYNAMIC}$  (mA, with ASF = 1.0)<sup>1, 2</sup>

f <sub>CCLK</sub> (MHz)	V <sub>DD\_INT</sub> (V)								
	0.975 V	1.0 V	1.025 V	1.05 V	1.075 V	1.10 V	1.125 V	1.15 V	1.175 V
100	76	77	81	84	87	88	90	92	95
150	117	119	123	126	130	133	136	139	144
200	153	156	161	165	170	174	179	183	188
250	190	195	201	207	212	217	223	229	235
300	227	233	240	246	253	260	266	273	280
350	263	272	278	286	294	302	309	318	325
400	300	309	317	326	335	344	352	361	370
450	339	349	356	365	374	385	394	405	415

<sup>1</sup>The values are not guaranteed as standalone maximum specifications. They must be combined with static current per the equations of [Electrical Characteristics on Page 19](#).

<sup>2</sup>Valid frequency and voltage ranges are model-specific. See [Operating Conditions on Page 18](#).

## ABSOLUTE MAXIMUM RATINGS

Stresses at or above those listed in [Table 16](#) may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

Table 16. Absolute Maximum Ratings

Parameter	Rating
Internal (Core) Supply Voltage (V <sub>DD\_INT</sub> )	-0.3 V to +1.32 V
External (I/O) Supply Voltage (V <sub>DD\_EXT</sub> )	-0.3 V to +3.6 V
Thermal Diode Supply Voltage (V <sub>DD\_THD</sub> )	-0.3 V to +3.6 V
Input Voltage	-0.5 V to +3.6 V
Output Voltage Swing	-0.5 V to V <sub>DD\_EXT</sub> + 0.5 V
Storage Temperature Range	-65°C to +150°C
Junction Temperature While Biased	125°C

## ESD SENSITIVITY



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PACKAGE INFORMATION

The information presented in [Figure 3](#) provides details about the package branding for the ADSP-2148x processors. For a complete listing of product availability, see [Ordering Guide on Page 66](#).



Figure 3. Typical Package Brand

Table 17. Package Brand Information<sup>1</sup>

Brand Key	Field Description
t	Temperature Range
pp	Package Type
Z	RoHS Compliant Option
cc	See Ordering Guide
vvvvvv.x	Assembly Lot Code
n.n	Silicon Revision
#	RoHS Compliant Designation
yyww	Date Code

<sup>1</sup>Non automotive only. For branding information specific to automotive products, contact Analog Devices Inc.

## MAXIMUM POWER DISSIPATION

See Engineer-to-Engineer Note [Estimating Power for ADSP-214xx SHARC Processors \(EE-348\)](#) for detailed thermal and power information regarding maximum power dissipation. For information on package thermal specifications, see [Thermal Characteristics on Page 56](#).

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

**Table 37. Serial Ports—Enable and Three-State**

Parameter	Min	Max	Unit
<i>Switching Characteristics</i>			
$t_{DDTEN}^1$ Data Enable from External Transmit SCLK	2		ns
$t_{DDTTE}^1$ Data Disable from External Transmit SCLK		11.5	ns
$t_{DDTIN}^1$ Data Enable from Internal Transmit SCLK	-1.5		ns

<sup>1</sup>Referenced to drive edge.

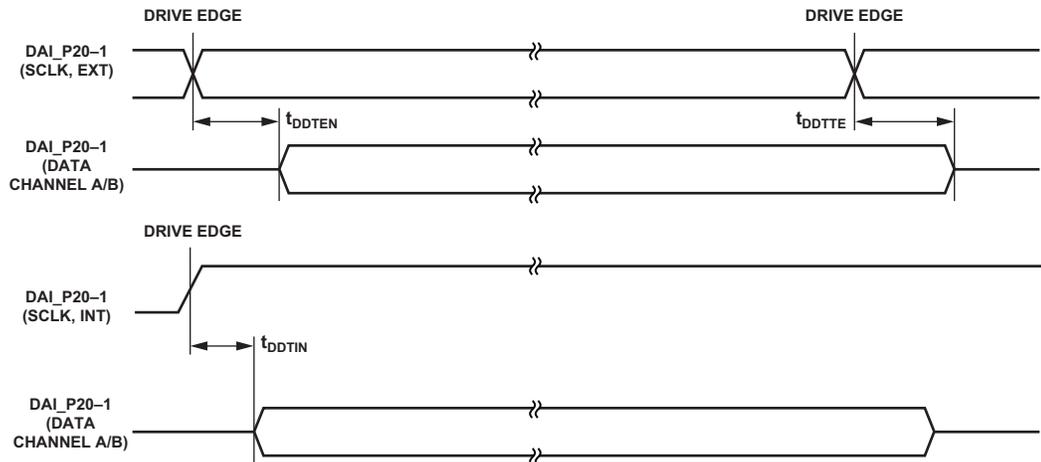


Figure 23. Serial Ports—Enable and Three-State

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

## Parallel Data Acquisition Port (PDAP)

The timing requirements for the PDAP are provided in [Table 40](#). PDAP is the parallel mode operation of Channel 0 of the IDP. For details on the operation of the PDAP, see the

PDAP chapter of the hardware reference. Note that the 20 bits of external PDAP data can be provided through the ADDR23–4 pins or over the DAI pins.

**Table 40. Parallel Data Acquisition Port (PDAP)**

Parameter	Min	Max	Unit
<i>Timing Requirements</i>			
$t_{SPHOLD}^1$	PDAP_HOLD Setup Before PDAP_CLK Sample Edge	2.5	ns
$t_{HPHOLD}^1$	PDAP_HOLD Hold After PDAP_CLK Sample Edge	2.5	ns
$t_{PDS}^1$	PDAP_DAT Setup Before PDAP_CLK Sample Edge	3.85	ns
$t_{PDHD}^1$	PDAP_DAT Hold After PDAP_CLK Sample Edge	2.5	ns
$t_{PDCLKW}$	Clock Width	$(t_{PCLK} \times 4) \div 2 - 3$	ns
$t_{PDCLK}$	Clock Period	$t_{PCLK} \times 4$	ns
<i>Switching Characteristics</i>			
$t_{PDHLDD}$	Delay of PDAP Strobe After Last PDAP_CLK Capture Edge for a Word	$2 \times t_{PCLK} + 3$	ns
$t_{PDSTRB}$	PDAP Strobe Pulse Width	$2 \times t_{PCLK} - 1.5$	ns

<sup>1</sup> Source pins of PDAP\_DATA are ADDR23–4 or DAI pins. Source pins for PDAP\_CLK and PDAP\_HOLD are 1) DAI pins; 2) CLKIN through PCG; 3) DAI pins through PCG; or 4) ADDR3–2 pins.

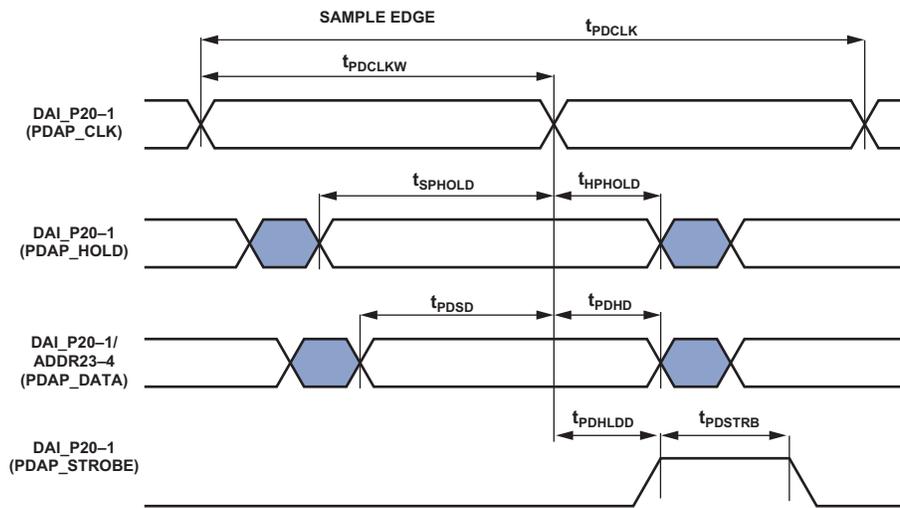


Figure 26. PDAP Timing

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

## Sample Rate Converter—Serial Input Port

The ASRC input signals are routed from the DAI\_P20–1 pins using the SRU. Therefore, the timing specifications provided in Table 41 are valid at the DAI\_P20–1 pins.

**Table 41. ASRC, Serial Input Port**

Parameter	Min	Max	Unit
<i>Timing Requirements</i>			
$t_{SRCSFS}^1$	4		ns
$t_{SRCHFS}^1$	5.5		ns
$t_{SRCSD}^1$	4		ns
$t_{SRCHD}^1$	5.5		ns
$t_{SRCLKW}$	$(t_{PCLK} \times 4) \div 2 - 1$		ns
$t_{SRCLK}$	$t_{PCLK} \times 4$		ns

<sup>1</sup> The serial clock, data, and frame sync signals can come from any of the DAI pins. The serial clock and frame sync signals can also come via PCG or SPORTs. PCG's input can be either CLKIN or any of the DAI pins.

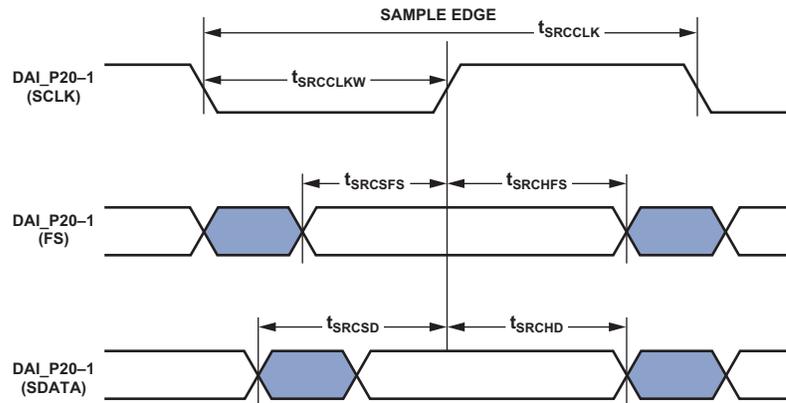


Figure 27. ASRC Serial Input Port Timing

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

## Pulse-Width Modulation Generators (PWM)

The following timing specifications apply when the ADDR23–8/DPI\_14–1 pins are configured as PWM.

**Table 43. Pulse-Width Modulation (PWM) Timing**

Parameter	Min	Max	Unit
<i>Switching Characteristics</i>			
$t_{PWMW}$ PWM Output Pulse Width	$t_{PCLK} - 2$	$(2^{16} - 2) \times t_{PCLK}$	ns
$t_{PWMP}$ PWM Output Period	$2 \times t_{PCLK} - 1.5$	$(2^{16} - 1) \times t_{PCLK}$	ns

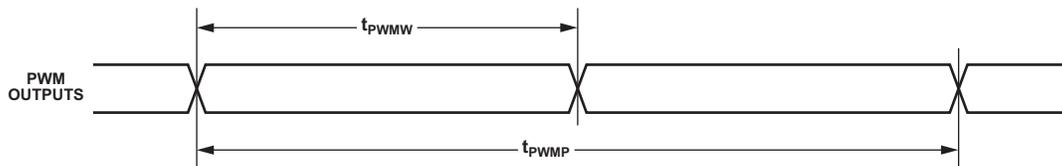


Figure 29. PWM Timing

## S/PDIF Transmitter

Serial data input to the S/PDIF transmitter can be formatted as left-justified, I<sup>2</sup>S, or right-justified with word widths of 16, 18, 20, or 24 bits. The following sections provide timing for the transmitter.

### S/PDIF Transmitter-Serial Input Waveforms

Figure 30 shows the right-justified mode. Frame sync is high for the left channel and low for the right channel. Data is valid on the rising edge of serial clock. The MSB is delayed the minimum in 24-bit output mode or the maximum in 16-bit output mode

from a frame sync transition, so that when there are 64 serial clock periods per frame sync period, the LSB of the data is right-justified to the next frame sync transition.

**Table 44. S/PDIF Transmitter Right-Justified Mode**

Parameter	Nominal	Unit
<i>Timing Requirement</i>		
$t_{RJD}$ Frame Sync to MSB Delay in Right-Justified Mode		
16-Bit Word Mode	16	SCLK
18-Bit Word Mode	14	SCLK
20-Bit Word Mode	12	SCLK
24-Bit Word Mode	8	SCLK

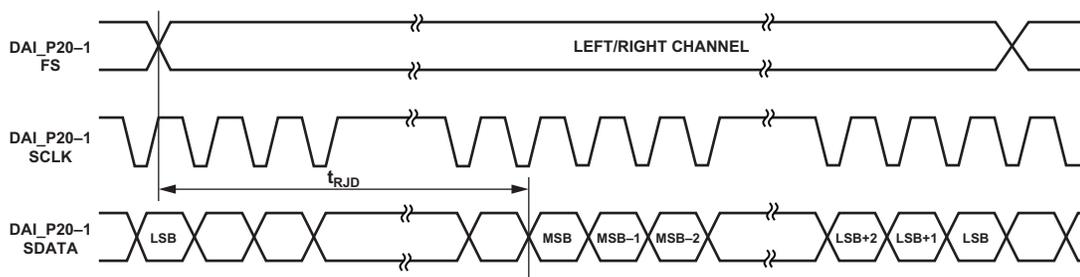


Figure 30. Right-Justified Mode

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

## SPI Interface—Slave

Table 51. SPI Interface Protocol—Slave Switching and Timing Specifications

Parameter		Min	Max	Unit
<i>Timing Requirements</i>				
$t_{SPICLK}$	Serial Clock Cycle	$4 \times t_{PCLK} - 2$		ns
$t_{SPICHS}$	Serial Clock High Period	$2 \times t_{PCLK} - 2$		ns
$t_{SPICLS}$	Serial Clock Low Period	$2 \times t_{PCLK} - 2$		ns
$t_{SDSCO}$	$\overline{SPIDS}$ Assertion to First SPICLK Edge CPHASE = 0	$2 \times t_{PCLK}$		ns
	CPHASE = 1			
$t_{HDS}$	Last SPICLK Edge to $\overline{SPIDS}$ Not Asserted, CPHASE = 0	$2 \times t_{PCLK}$		ns
$t_{SSPIDS}$	Data Input Valid to SPICLK edge (Data Input Set-up Time)	2		ns
$t_{HSPIDS}$	SPICLK Last Sampling Edge to Data Input Not Valid	2		ns
$t_{SDPPW}$	$\overline{SPIDS}$ Deassertion Pulse Width (CPHASE=0)	$2 \times t_{PCLK}$		ns
<i>Switching Characteristics</i>				
$t_{DSOE}$	$\overline{SPIDS}$ Assertion to Data Out Active	0	7.5	ns
$t_{DSOE}^1$	$\overline{SPIDS}$ Assertion to Data Out Active (SPI2)	0	7.5	ns
$t_{DSDHI}$	$\overline{SPIDS}$ Deassertion to Data High Impedance	0	10.5	ns
$t_{DSDHI}^1$	$\overline{SPIDS}$ Deassertion to Data High Impedance (SPI2)	0	10.5	ns
$t_{DDSPIDS}$	SPICLK Edge to Data Out Valid (Data Out Delay Time)		9.5	ns
$t_{HDSPIDS}$	SPICLK Edge to Data Out Not Valid (Data Out Hold Time)	$2 \times t_{PCLK}$		ns
$t_{DSOV}$	$\overline{SPIDS}$ Assertion to Data Out Valid (CPHASE = 0)		$5 \times t_{PCLK}$	ns

<sup>1</sup>The timing for these parameters applies when the SPI is routed through the signal routing unit. For more information, see the “Serial Peripheral Interface Port” chapter of the hardware reference.

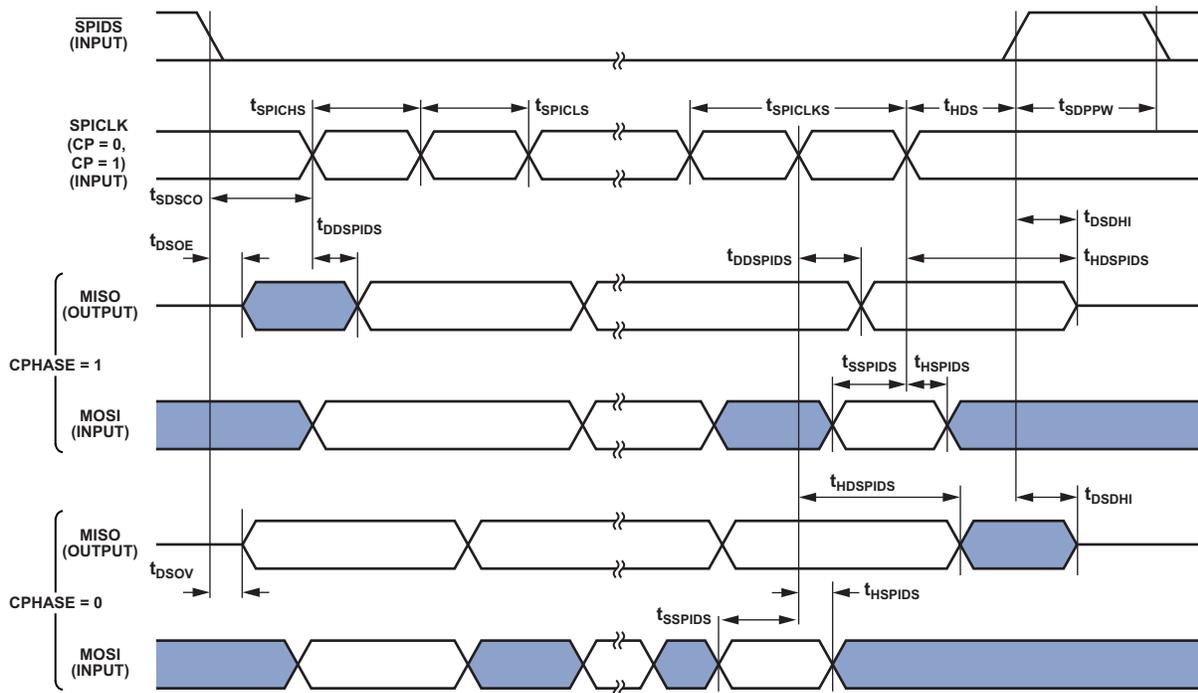


Figure 36. SPI Slave Timing

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

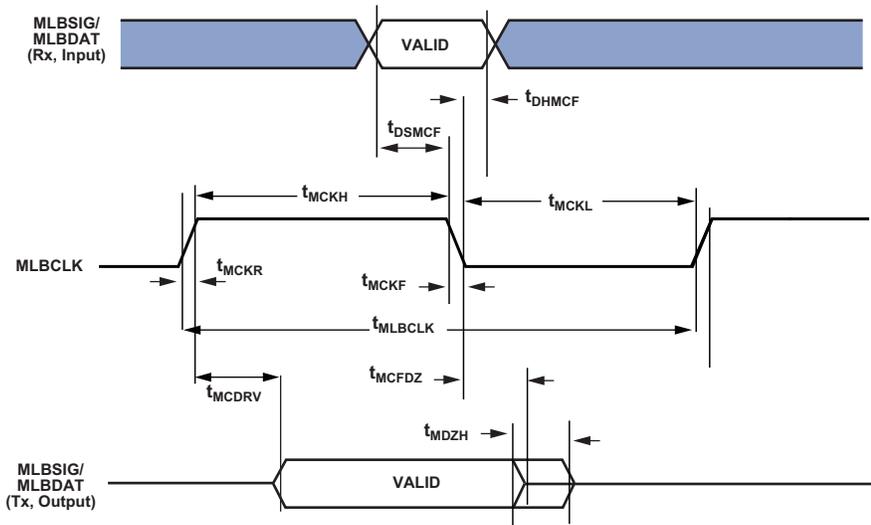


Figure 37. MLB Timing (3-Pin Interface)

Table 53. MLB Interface, 5-Pin Specifications

Parameter	Min	Typ	Max	Unit
<i>5-Pin Characteristics</i>				
$t_{MLBCLK}$	MLB Clock Period			
		40		ns
		81		ns
$t_{MCKL}$	MLBCLK Low Time			
	15			ns
	30			ns
$t_{MCKH}$	MLBCLK High Time			
	15			ns
	30			ns
$t_{MCKR}$	MLBCLK Rise Time ( $V_{IL}$ to $V_{IH}$ )		6	ns
$t_{MCKF}$	MLBCLK Fall Time ( $V_{IH}$ to $V_{IL}$ )		6	ns
$t_{MPWV}^1$	MLBCLK Pulse Width Variation		2	nspp
$t_{DSMCF}^2$	DAT/SIG Input Setup Time			ns
$t_{DHMCf}$	DAT/SIG Input Hold Time			ns
$t_{MCDRV}$	DS/DO Output Data Delay From MLBCLK Rising Edge		8	ns
$t_{MCRDL}^3$	DO/SO Low From MLBCLK High			
			10	ns
			20	ns
$C_{MLB}$	DS/DO Pin Load		40	pf

<sup>1</sup>Pulse width variation is measured at 1.25 V by triggering on one edge of MLBCLK and measuring the spread on the other edge, measured in ns peak-to-peak (pp).

<sup>2</sup>Gate Delays due to OR'ing logic on the pins must be accounted for.

<sup>3</sup>When a node is not driving valid data onto the bus, the MLBSO and MLBDO output lines shall remain low. If the output lines can float at anytime, including while in reset, external pull-down resistors are required to keep the outputs from corrupting the MediaLB signal lines when not being driven.

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

## OUTPUT DRIVE CURRENTS

Figure 41 shows typical I-V characteristics for the output drivers of the ADSP-2148x, and Table 55 shows the pins associated with each driver. The curves represent the current drive capability of the output drivers as a function of output voltage.

Table 55. Driver Types

Driver Type	Associated Pins
A	FLAG[0-3], AMI_ADDR[0-23], DATA[0-15], AMI_RD, AMI_WR, AMI_ACK, MS[1-0], SDRAS, SDCAS, SDWE, SDDQM, SDCKE, SDA10, EMU, TDO, RESETOUT, DPI[1-14], DAI[1-20], WDTRSTO, MLBDAT, MLBSIG, MLBSO, MLBDO, MLBCLK
B	SDCLK

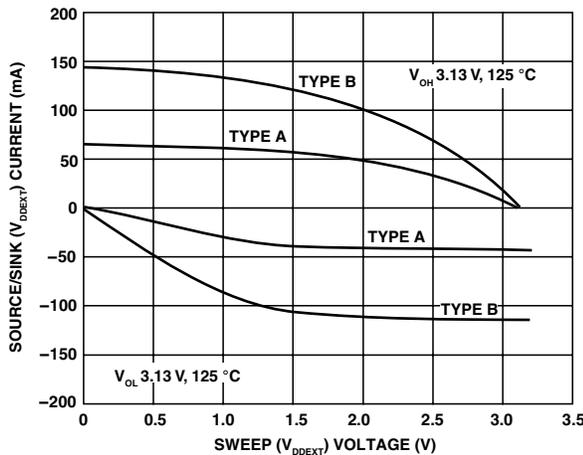


Figure 41. Typical Drive at Junction Temperature

## TEST CONDITIONS

The ac signal specifications (timing parameters) appear in Table 21 on Page 26 through Table 54 on Page 54. These include output disable time, output enable time, and capacitive loading. The timing specifications for the SHARC apply for the voltage reference levels in Figure 42.

Timing is measured on signals when they cross the 1.5 V level as described in Figure 43. All delays (in nanoseconds) are measured between the point that the first signal reaches 1.5 V and the point that the second signal reaches 1.5 V.

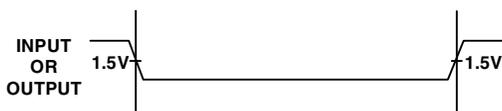
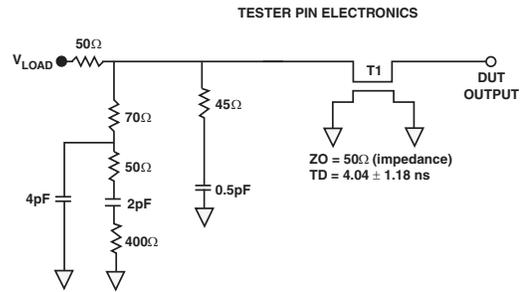


Figure 43. Voltage Reference Levels for AC Measurements



NOTES:  
THE WORST CASE TRANSMISSION LINE DELAY IS SHOWN AND CAN BE USED FOR THE OUTPUT TIMING ANALYSIS TO REFLECT THE TRANSMISSION LINE EFFECT AND MUST BE CONSIDERED. THE TRANSMISSION LINE (TD) IS FOR LOAD ONLY AND DOES NOT AFFECT THE DATA SHEET TIMING SPECIFICATIONS.  
ANALOG DEVICES RECOMMENDS USING THE IBIS MODEL TIMING FOR A GIVEN SYSTEM REQUIREMENT. IF NECESSARY, A SYSTEM MAY INCORPORATE EXTERNAL DRIVERS TO COMPENSATE FOR ANY TIMING DIFFERENCES.

Figure 42. Equivalent Device Loading for AC Measurements (Includes All Fixtures)

## CAPACITIVE LOADING

Output delays and holds are based on standard capacitive loads: 30 pF on all pins (see Figure 42). Figure 46 and Figure 47 show graphically how output delays and holds vary with load capacitance. The graphs of Figure 44 through Figure 47 may not be linear outside the ranges shown for Typical Output Delay vs. Load Capacitance and Typical Output Rise Time (20% to 80%, V = Min) vs. Load Capacitance.

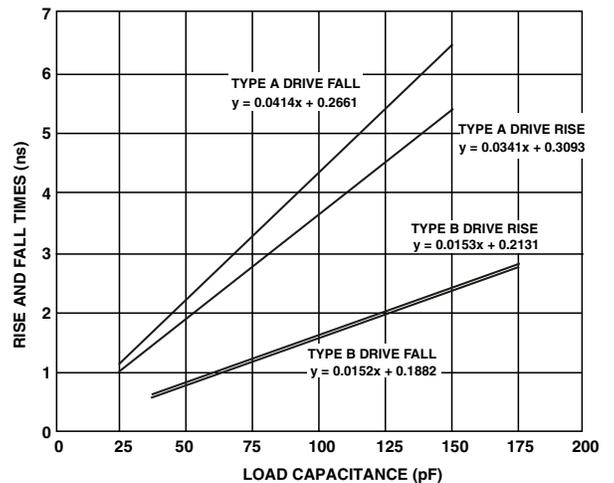


Figure 44. Typical Output Rise/Fall Time (20% to 80%,  $V_{DD\_EXT} = Max$ )

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

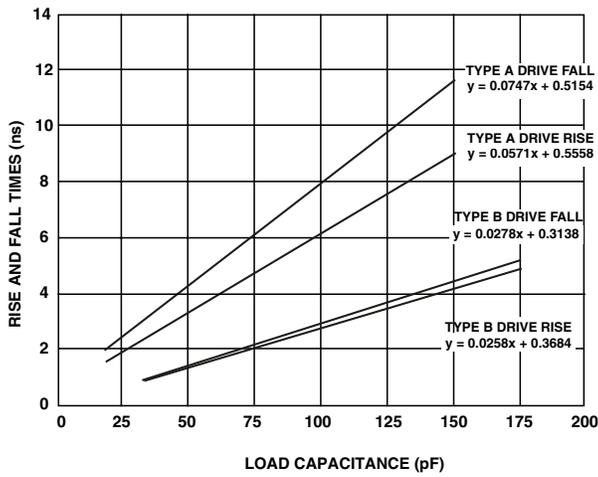


Figure 45. Typical Output Rise/Fall Time  
(20% to 80%,  $V_{DD\_EXT} = \text{Min}$ )

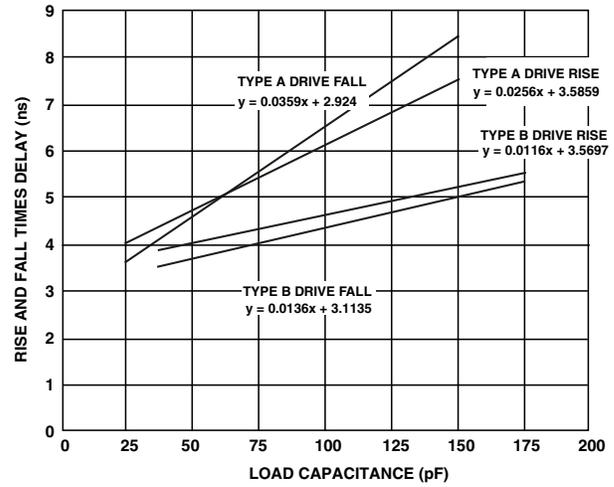


Figure 47. Typical Output Rise/Fall Delay  
( $V_{DD\_EXT} = \text{Min}$ )

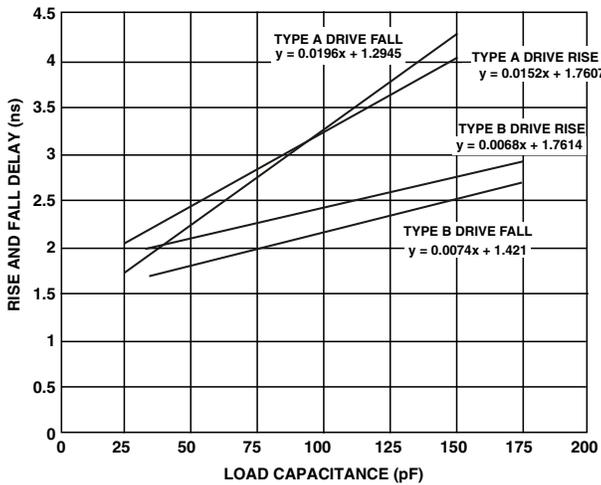


Figure 46. Typical Output Rise/Fall Delay  
( $V_{DD\_EXT} = \text{Max}$ )

## THERMAL CHARACTERISTICS

The ADSP-2148x processor is rated for performance over the temperature range specified in [Operating Conditions on Page 18](#).

[Table 57](#) airflow measurements comply with JEDEC standards JESD51-2 and JESD51-6, and the junction-to-board measurement complies with JESD51-8. Test board design complies with JEDEC standards JESD51-7 (LQFP\_EP). The junction-to-case measurement complies with MIL-STD-883. All measurements use a 2S2P JEDEC test board.

To determine the junction temperature of the device while on the application PCB, use:

$$T_J = T_{CASE} + (\Psi_{JT} \times P_D)$$

where:

$T_J$  = junction temperature °C

$T_{CASE}$  = case temperature (°C) measured at the top center of the package

$\Psi_{JT}$  = junction-to-top (of package) characterization parameter is the Typical value from [Table 57](#).

$P_D$  = power dissipation

Values of  $\theta_{JA}$  are provided for package comparison and PCB design considerations.  $\theta_{JA}$  can be used for a first order approximation of  $T_J$  by the equation:

$$T_J = T_A + (\theta_{JA} \times P_D)$$

where:

$T_A$  = ambient temperature °C

Values of  $\theta_{JC}$  are provided for package comparison and PCB design considerations when an external heatsink is required.

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

## 176-LEAD LQFP\_EP LEAD ASSIGNMENT

Table 60. ADSP-21486 176-Lead LQFP\_EP Lead Assignment (Numerical by Lead Number)

Lead Name	Lead No.						
NC	1	V <sub>DD_EXT</sub>	45	DAI_P10	89	V <sub>DD_INT</sub>	133
MS0	2	DPI_P08	46	V <sub>DD_INT</sub>	90	FLAG0	134
NC	3	DPI_P07	47	V <sub>DD_EXT</sub>	91	FLAG1	135
V <sub>DD_INT</sub>	4	V <sub>DD_INT</sub>	48	DAI_P20	92	FLAG2	136
CLK_CFG1	5	DPI_P09	49	V <sub>DD_INT</sub>	93	GND	137
ADDR0	6	DPI_P10	50	DAI_P08	94	FLAG3	138
BOOT_CFG0	7	DPI_P11	51	DAI_P14	95	GND	139
V <sub>DD_EXT</sub>	8	DPI_P12	52	DAI_P04	96	GND	140
ADDR1	9	DPI_P13	53	DAI_P18	97	V <sub>DD_EXT</sub>	141
ADDR2	10	DPI_P14	54	DAI_P17	98	GND	142
ADDR3	11	DAI_P03	55	DAI_P16	99	V <sub>DD_INT</sub>	143
ADDR4	12	NC	56	DAI_P12	100	TRST	144
ADDR5	13	V <sub>DD_EXT</sub>	57	DAI_P15	101	GND	145
BOOT_CFG1	14	NC	58	V <sub>DD_INT</sub>	102	EMU	146
GND	15	NC	59	DAI_P11	103	DATA0	147
ADDR6	16	NC	60	V <sub>DD_EXT</sub>	104	DATA1	148
ADDR7	17	NC	61	V <sub>DD_INT</sub>	105	DATA2	149
NC	18	V <sub>DD_INT</sub>	62	BOOT_CFG2	106	DATA3	150
NC	19	NC	63	V <sub>DD_INT</sub>	107	TDO	151
ADDR8	20	NC	64	AMI_ACK	108	DATA4	152
ADDR9	21	V <sub>DD_INT</sub>	65	GND	109	V <sub>DD_EXT</sub>	153
CLK_CFG0	22	NC	66	THD_M	110	DATA5	154
V <sub>DD_INT</sub>	23	NC	67	THD_P	111	DATA6	155
CLKIN	24	V <sub>DD_INT</sub>	68	V <sub>DD_THD</sub>	112	V <sub>DD_INT</sub>	156
XTAL	25	NC	69	V <sub>DD_INT</sub>	113	DATA7	157
ADDR10	26	WDTRSTO	70	V <sub>DD_INT</sub>	114	TDI	158
NC	27	NC	71	MS1	115	NC	159*
V <sub>DD_EXT</sub>	28	V <sub>DD_EXT</sub>	72	V <sub>DD_INT</sub>	116	V <sub>DD_EXT</sub>	160
V <sub>DD_INT</sub>	29	DAI_P07	73	WDT_CLKO	117	DATA8	161
ADDR11	30	DAI_P13	74	WDT_CLKIN	118	DATA9	162
ADDR12	31	DAI_P19	75	V <sub>DD_EXT</sub>	119	DATA10	163
ADDR17	32	DAI_P01	76	ADDR23	120	TCK	164
ADDR13	33	DAI_P02	77	ADDR22	121	DATA11	165
V <sub>DD_INT</sub>	34	V <sub>DD_INT</sub>	78	ADDR21	122	DATA12	166
ADDR18	35	NC	79	V <sub>DD_INT</sub>	123	DATA14	167
RESETOUT/RUNRSTIN	36	NC	80	ADDR20	124	DATA13	168
V <sub>DD_INT</sub>	37	NC	81	ADDR19	125	V <sub>DD_INT</sub>	169
DPI_P01	38	NC	82	V <sub>DD_EXT</sub>	126	DATA15	170
DPI_P02	39	NC	83	ADDR16	127	NC	171
DPI_P03	40	V <sub>DD_EXT</sub>	84	ADDR15	128	NC	172
V <sub>DD_INT</sub>	41	V <sub>DD_INT</sub>	85	V <sub>DD_INT</sub>	129	RESET	173
DPI_P05	42	DAI_P06	86	ADDR14	130	TMS	174
DPI_P04	43	DAI_P05	87	AMI_WR	131	NC	175
DPI_P06	44	DAI_P09	88	AMI_RD	132	V <sub>DD_INT</sub>	176
						GND	177**

\*No external connection should be made to this pin. Use as NC only.

\*\* Lead no. 177 (exposed pad) is the GND supply (see Figure 50 and Figure 51) for the processor; this pad must be **robustly** connected to GND.

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

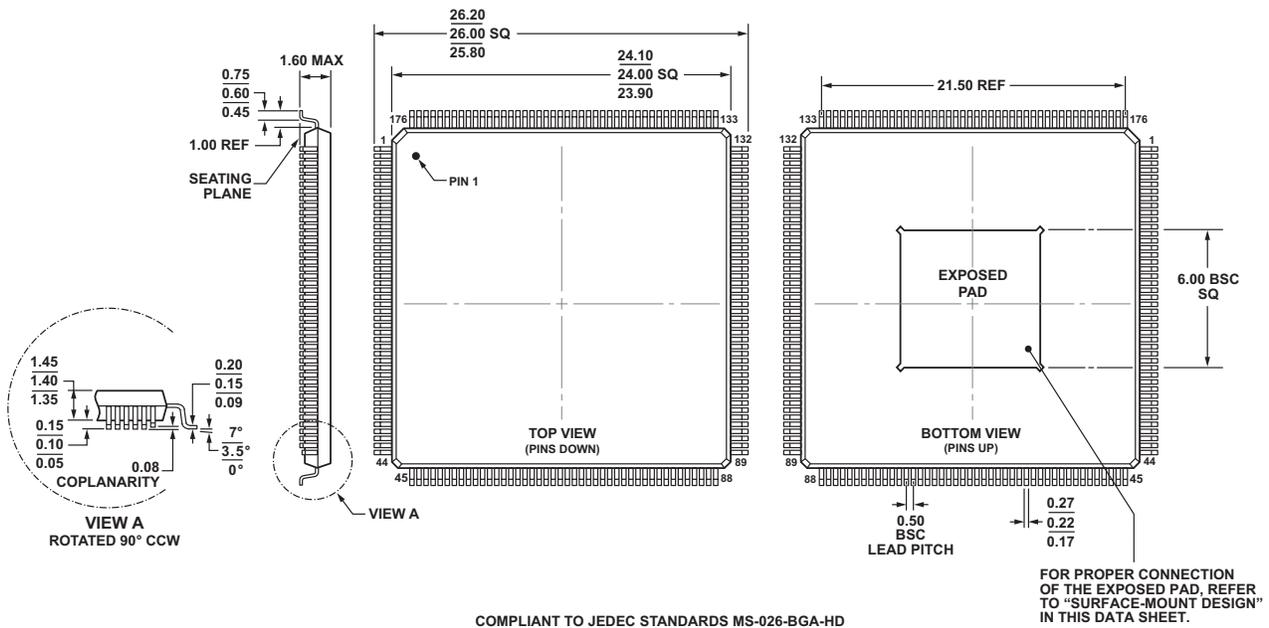


Figure 53. 176-Lead Low Profile Quad Flat Package, Exposed Pad [LQFP\_EP]<sup>1</sup>  
(SW-176-2)

Dimensions shown in millimeters

<sup>1</sup>For information relating to the exposed pad on the SW-176-2 package, see the table endnote on Page 60.

## SURFACE-MOUNT DESIGN

The exposed pad is required to be electrically and thermally connected to GND. Implement this by soldering the exposed pad to a GND PCB land that is the same size as the exposed pad. The GND PCB land should be robustly connected to the GND plane in the PCB for best electrical and thermal performance. No separate GND pins are provided in the package.

# ADSP-21483/ADSP-21486/ADSP-21487/ADSP-21488/ADSP-21489

Model <sup>1</sup>	Notes	Temperature Range <sup>2</sup>	RAM	Processor Instruction Rate (Max)	Package Description	Package Option
ADSP-21487KSWZ-2B	3	0°C to +70°C	5 Mbit	300 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-2BB	3	0°C to +70°C	5 Mbit	300 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-3B	3	0°C to +70°C	5 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-3BB	3	0°C to +70°C	5 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-4B	3	0°C to +70°C	5 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-4BB	3	0°C to +70°C	5 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-5B	3, 4	0°C to +70°C	5 Mbit	450 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21487KSWZ-5BB	3, 4	0°C to +70°C	5 Mbit	450 MHz	176-Lead LQFP_EP	SW-176-2
ADSP21487KSWZ5BBRL	5	0°C to +70°C	5 Mbit	450 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21488BSWZ-3A		-40°C to +85°C	3 Mbit	350 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21488KSWZ-3A		0°C to +70°C	3 Mbit	350 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21488KSWZ-3A1	6	0°C to +70°C	3 Mbit	350 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21488KSWZ-3B		0°C to +70°C	3 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21488BSWZ-3B		-40°C to +85°C	3 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21488KSWZ-4A		0°C to +70°C	3 Mbit	400 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21488BSWZ-4A		-40°C to +85°C	3 Mbit	400 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21488KSWZ-4B		0°C to +70°C	3 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21488BSWZ-4B		-40°C to +85°C	3 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21488KSWZ-4B1	6	0°C to +70°C	3 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21489KSWZ-3A		0°C to +70°C	5 Mbit	350 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21489BSWZ-3A		-40°C to +85°C	5 Mbit	350 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21489KSWZ-3B		0°C to +70°C	5 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21489BSWZ-3B		-40°C to +85°C	5 Mbit	350 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21489KSWZ-4A		0°C to +70°C	5 Mbit	400 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21489BSWZ-4A		-40°C to +85°C	5 Mbit	400 MHz	100-Lead LQFP_EP	SW-100-2
ADSP-21489KSWZ-4B		0°C to +70°C	5 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21489BSWZ-4B		-40°C to +85°C	5 Mbit	400 MHz	176-Lead LQFP_EP	SW-176-2
ADSP-21489KSWZ-5B	4	0°C to +70°C	5 Mbit	450 MHz	176-Lead LQFP_EP	SW-176-2

<sup>1</sup>Z = RoHS compliant part.

<sup>2</sup>Referenced temperature is ambient temperature. The ambient temperature is not a specification. Please see [Operating Conditions on Page 18](#) for junction temperature (T<sub>j</sub>) specification, which is the only temperature specification.

<sup>3</sup>The ADSP-21483, ADSP-21486, and ADSP-21487 models are available with factory programmed ROM including the latest multichannel audio decoding and post-processing algorithms from Dolby Labs and DTS. ROM contents may vary depending on chip version and silicon revision. Please visit [www.analog.com](http://www.analog.com) for complete information.

<sup>4</sup>See Engineer-to-Engineer Note [Static Voltage Scaling for ADSP-2148x SHARC Processors \(EE-357\)](#) for operating ADSP-2148x processors at 450 MHz.

<sup>5</sup>RL = Tape and Reel.

<sup>6</sup>This product contains a -140 dB sample rate converter.