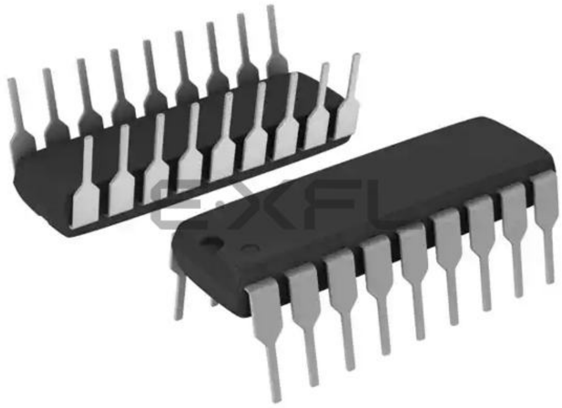

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What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	-
Peripherals	Brown-out Detect/Reset, POR, WDT
Number of I/O	13
Program Memory Size	1.75KB (1K x 14)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	80 x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 6V
Data Converters	-
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Through Hole
Package / Case	18-DIP (0.300", 7.62mm)
Supplier Device Package	18-PDIP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16c621-20i-p

PIC16C62X

EPROM-Based 8-Bit CMOS Microcontrollers

Devices included in this data sheet:

Referred to collectively as PIC16C62X.

- PIC16C620
- PIC16C621
- PIC16C622
- PIC16CR620A
- PIC16C620A
- PIC16C621A
- PIC16C622A

High Performance RISC CPU:

- Only 35 instructions to learn
- All single cycle instructions (200 ns), except for program branches which are two-cycle
- Operating speed:
  - DC - 40 MHz clock input
  - DC - 100 ns instruction cycle

Device	Program Memory	Data Memory
PIC16C620	512	80
PIC16C620A	512	96
PIC16CR620A	512	96
PIC16C621	1K	80
PIC16C621A	1K	96
PIC16C622	2K	128
PIC16C622A	2K	128

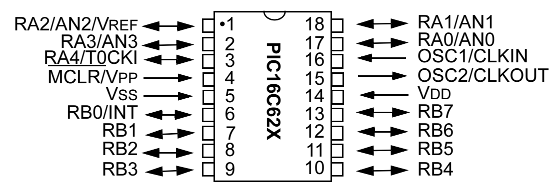
- Interrupt capability
- 16 special function hardware registers
- 8-level deep hardware stack
- Direct, Indirect and Relative addressing modes

Peripheral Features:

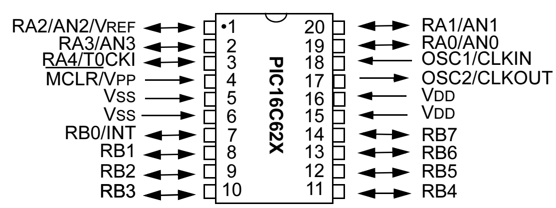
- 13 I/O pins with individual direction control
- High current sink/source for direct LED drive
- Analog comparator module with:
  - Two analog comparators
  - Programmable on-chip voltage reference (V_{REF}) module
  - Programmable input multiplexing from device inputs and internal voltage reference
  - Comparator outputs can be output signals
- Timer0: 8-bit timer/counter with 8-bit programmable prescaler

Pin Diagrams

PDIP, SOIC, Windowed CERDIP



SSOP



Special Microcontroller Features:

- Power-on Reset (POR)
- Power-up Timer (PWRT) and Oscillator Start-up Timer (OST)
- Brown-out Reset
- Watchdog Timer (WDT) with its own on-chip RC oscillator for reliable operation
- Programmable code protection
- Power saving SLEEP mode
- Selectable oscillator options
- Serial in-circuit programming (via two pins)
- Four user programmable ID locations

CMOS Technology:

- Low power, high speed CMOS EPROM technology
- Fully static design
- Wide operating range
  - 2.5V to 5.5V
- Commercial, industrial and extended temperature range
- Low power consumption
  - < 2.0 mA @ 5.0V, 4.0 MHz
  - 15 μA typical @ 3.0V, 32 kHz
  - < 1.0 μA typical standby current @ 3.0V

FIGURE 4-6: DATA MEMORY MAP FOR THE PIC16C620A/CR620A/621A

File Address	Bank 0	Bank 1	File Address
00h	INDF(1)	INDF(1)	80h
01h	TMR0	OPTION	81h
02h	PCL	PCL	82h
03h	STATUS	STATUS	83h
04h	FSR	FSR	84h
05h	PORTA	TRISA	85h
06h	PORTB	TRISB	86h
07h			87h
08h			88h
09h			89h
0Ah	PCLATH	PCLATH	8Ah
0Bh	INTCON	INTCON	8Bh
0Ch	PIR1	PIE1	8Ch
0Dh			8Dh
0Eh		PCON	8Eh
0Fh			8Fh
10h			90h
11h			91h
12h			92h
13h			93h
14h			94h
15h			95h
16h			96h
17h			97h
18h			98h
19h			99h
1Ah			9Ah
1Bh			9Bh
1Ch			9Ch
1Dh			9Dh
1Eh			9Eh
1Fh	CMCON	VRCON	9Fh
20h–6Fh	General Purpose Register		A0h
70h–7Fh	General Purpose Register	Accesses 70h-7Fh	F0h–FFh

Unimplemented data memory locations, read as '0'. (shaded/empty cells)

Note 1: Not a physical register.

FIGURE 4-7: DATA MEMORY MAP FOR THE PIC16C622A

File Address	Bank 0	Bank 1	File Address
00h	INDF(1)	INDF(1)	80h
01h	TMR0	OPTION	81h
02h	PCL	PCL	82h
03h	STATUS	STATUS	83h
04h	FSR	FSR	84h
05h	PORTA	TRISA	85h
06h	PORTB	TRISB	86h
07h			87h
08h			88h
09h			89h
0Ah	PCLATH	PCLATH	8Ah
0Bh	INTCON	INTCON	8Bh
0Ch	PIR1	PIE1	8Ch
0Dh			8Dh
0Eh		PCON	8Eh
0Fh			8Fh
10h			90h
11h			91h
12h			92h
13h			93h
14h			94h
15h			95h
16h			96h
17h			97h
18h			98h
19h			99h
1Ah			9Ah
1Bh			9Bh
1Ch			9Ch
1Dh			9Dh
1Eh			9Eh
1Fh	CMCON	VRCON	9Fh
20h–6Fh	General Purpose Register	General Purpose Register	A0h–BFh
			C0h
70h–7Fh	General Purpose Register	Accesses 70h-7Fh	F0h–FFh

Unimplemented data memory locations, read as '0'. (shaded/empty cells)

Note 1: Not a physical register.

5.3 I/O Programming Considerations

5.3.1 BI-DIRECTIONAL I/O PORTS

Any instruction which writes, operates internally as a read followed by a write operation. The `BCF` and `BSF` instructions, for example, read the register into the CPU, execute the bit operation and write the result back to the register. Caution must be used when these instructions are applied to a port with both inputs and outputs defined. For example, a `BSF` operation on bit5 of PORTB will cause all eight bits of PORTB to be read into the CPU. Then the `BSF` operation takes place on bit5 and PORTB is written to the output latches. If another bit of PORTB is used as a bi-directional I/O pin (e.g., bit0) and it is defined as an input at this time, the input signal present on the pin itself would be read into the CPU and re-written to the data latch of this particular pin, overwriting the previous content. As long as the pin stays in the Input mode, no problem occurs. However, if bit0 is switched into Output mode later on, the content of the data latch may now be unknown.

Reading the port register reads the values of the port pins. Writing to the port register writes the value to the port latch. When using read-modify-write instructions (ex. `BCF, BSF`, etc.) on a port, the value of the port pins is read, the desired operation is done to this value, and this value is then written to the port latch.

Example 5-2 shows the effect of two sequential read-modify-write instructions (ex., `BCF, BSF`, etc.) on an I/O port.

A pin actively outputting a Low or High should not be driven from external devices at the same time in order to change the level on this pin (“wired-or”, “wired-and”). The resulting high output currents may damage the chip.

EXAMPLE 5-2: READ-MODIFY-WRITE INSTRUCTIONS ON AN I/O PORT

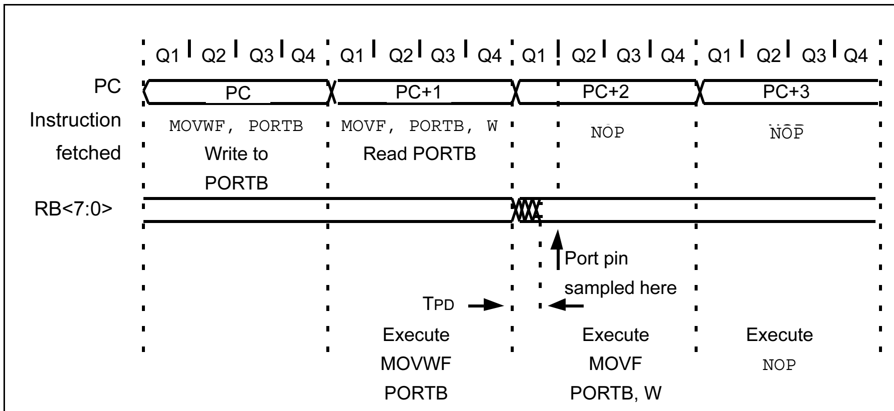
```
; Initial PORT settings:    PORTB<7:4> Inputs
;
;                           PORTB<3:0> Outputs
; PORTB<7:6> have external pull-up and are not
; connected to other circuitry
;
;                           PORT latch   PORT pins
;                           ----------   ---------
                                         -

    BCF  PORTB, 7         ; 01pp pppp    11pp pppp
    BCF  PORTB, 6         ; 10pp pppp    11pp pppp
    BSF  STATUS,RP0       ;
    BCF  TRISB, 7         ; 10pp pppp    11pp pppp
    BCF  TRISB, 6         ; 10pp pppp    10pp pppp
;
; Note that the user may have expected the pin
; values to be 00pp pppp. The 2nd BCF caused
; RB7 to be latched as the pin value (High).
```

5.3.2 SUCCESSIVE OPERATIONS ON I/O PORTS

The actual write to an I/O port happens at the end of an instruction cycle, whereas for reading, the data must be valid at the beginning of the instruction cycle (Figure 5-7). Therefore, care must be exercised if a write followed by a read operation is carried out on the same I/O port. The sequence of instructions should be such to allow the pin voltage to stabilize (load dependent) before the next instruction which causes that file to be read into the CPU is executed. Otherwise, the previous state of that pin may be read into the CPU rather than the new state. When in doubt, it is better to separate these instructions with a `NOP` or another instruction not accessing this I/O port.

FIGURE 5-7: SUCCESSIVE I/O OPERATION



Note:

This example shows write to PORTB followed by a read from PORTB.

Note that:

data setup time = (0.25 T_{CY} - T_{PD}) where T_{CY} = instruction cycle and TPD = propagation delay of Q1 cycle to output valid.

Therefore, at higher clock frequencies, a write followed by a read may be problematic.

FIGURE 6-3: TIMER0 TIMING: INTERNAL CLOCK/PRESCALE 1:2

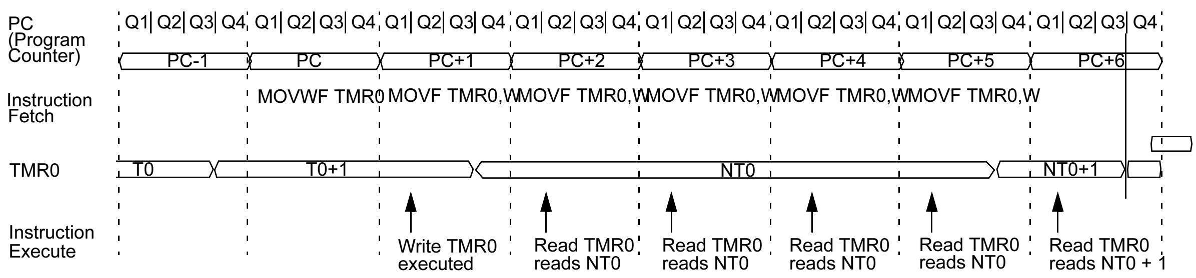
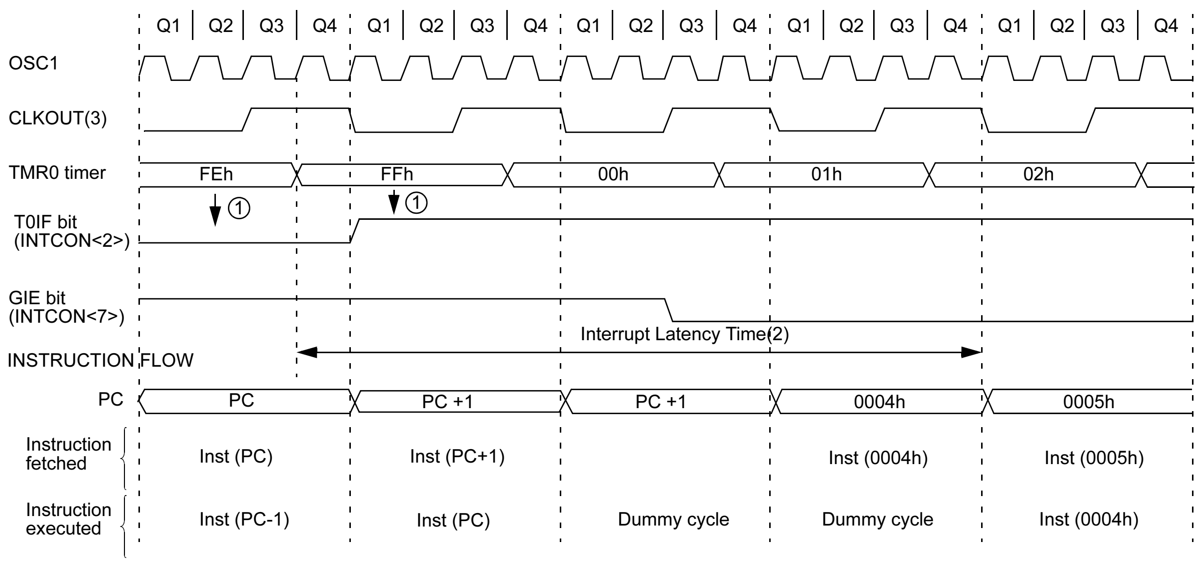


FIGURE 6-4: TIMER0 INTERRUPT TIMING



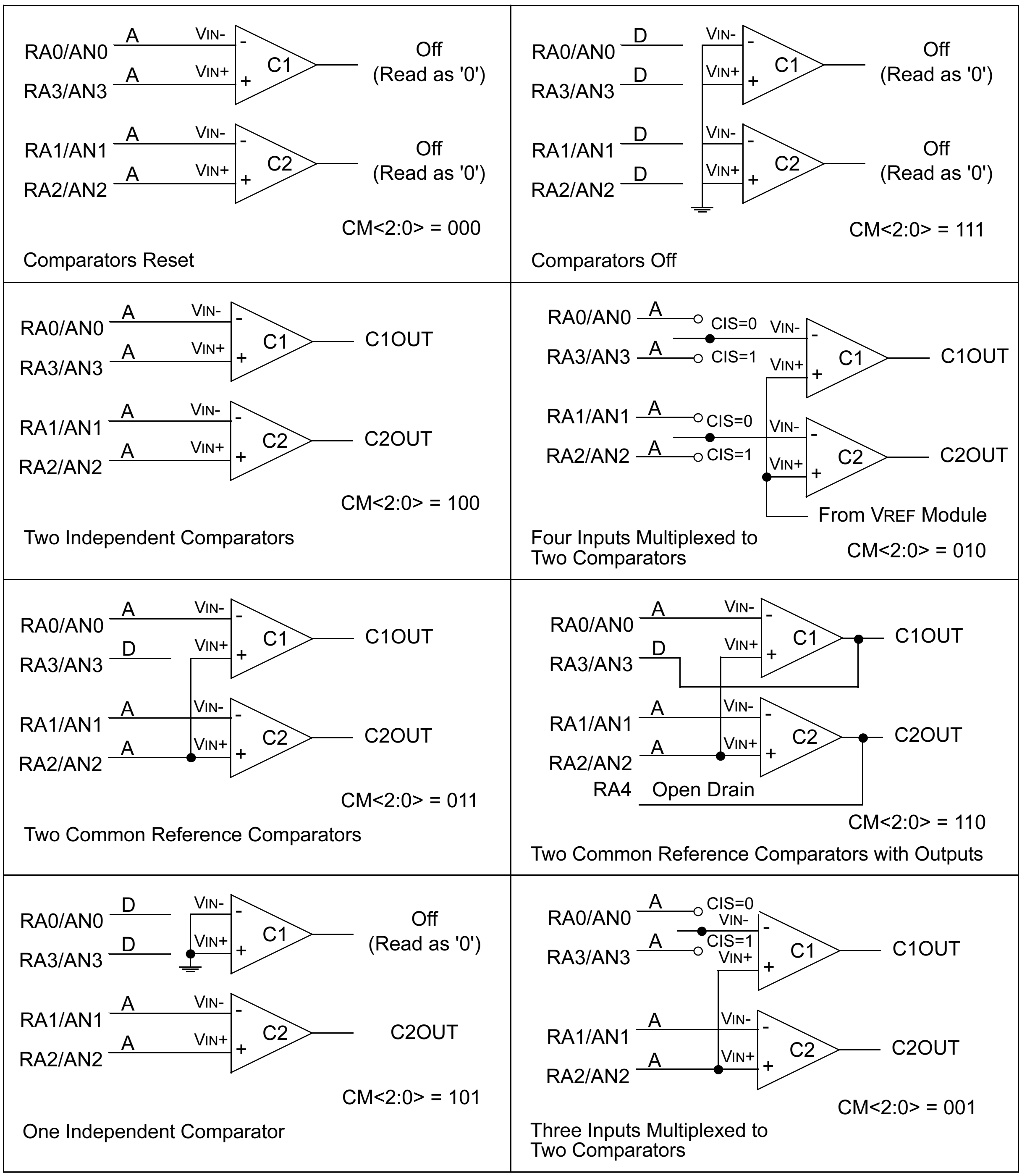
Note 1: T0IF interrupt flag is sampled here (every Q1).
2: Interrupt latency = 3TCY, where TCY = instruction cycle time.
3: CLKOUT is available only in RC Oscillator mode.

7.1 Comparator Configuration

There are eight modes of operation for the comparators. The CMCON register is used to select the mode. Figure 7-1 shows the eight possible modes. The TRISA register controls the data direction of the comparator pins for each mode. If the Comparator mode is changed, the comparator output level may not be valid for the specified mode change delay shown in Table 12-2.

Note:	Comparator interrupts should be disabled during a Comparator mode change otherwise a false interrupt may occur.

FIGURE 7-1: COMPARATOR I/O OPERATING MODES



A = Analog Input, Port Reads Zeros Always
D = Digital Input
CIS = CMCON<3>, Comparator Input Switch

The code example in Example 7-1 depicts the steps required to configure the comparator module. RA3 and RA4 are configured as digital output. RA0 and RA1 are configured as the V- inputs and RA2 as the V+ input to both comparators.

EXAMPLE 7-1: INITIALIZING COMPARATOR MODULE

```
MOVLW  0x03          ;Init comparator mode
MOVWF  CMCON         ;CM<2:0> = 011
CLRF   PORTA         ;Init PORTA
BSF    STATUS,RP0    ;Select Bank1
MOVLW  0x07          ;Initialize data direction
MOVWF  TRISA         ;Set RA<2:0> as inputs
                     ;RA<4:3> as outputs
                     ;TRISA<7:5> always read '0'
BCF    STATUS,RP0    ;Select Bank 0
CALL   DELAY 10      ;10µs delay
MOVF   CMCON,F       ;Read CMCON to end change condition
BCF    PIR1,CMIF     ;Clear pending interrupts
BSF    STATUS,RP0    ;Select Bank 1
BSF    PIE1,CMIE     ;Enable comparator interrupts
BCF    STATUS,RP0    ;Select Bank 0
BSF    INTCON,PEIE   ;Enable peripheral interrupts
BSF    INTCON,GIE    ;Global interrupt enable
```

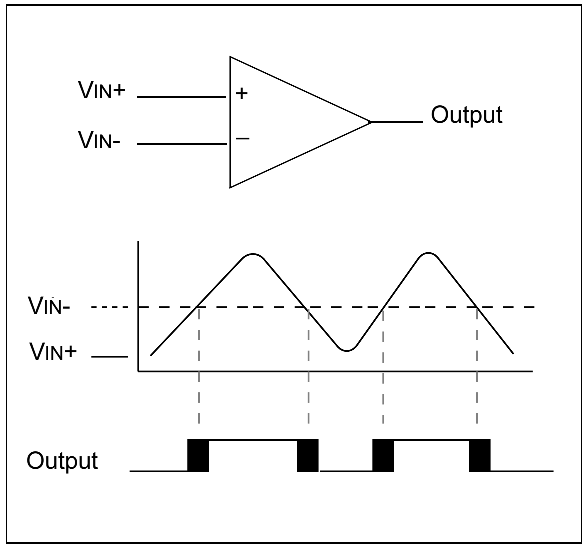
7.2 Comparator Operation

A single comparator is shown in Figure 7-2 along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input VIN-, the output of the comparator is a digital low level. When the analog input at VIN+ is greater than the analog input VIN-, the output of the comparator is a digital high level. The shaded areas of the output of the comparator in Figure 7-2 represent the uncertainty due to input offsets and response time.

7.3 Comparator Reference

An external or internal reference signal may be used depending on the comparator Operating mode. The analog signal that is present at VIN- is compared to the signal at VIN+, and the digital output of the comparator is adjusted accordingly (Figure 7-2).

FIGURE 7-2: SINGLE COMPARATOR



7.3.1 EXTERNAL REFERENCE SIGNAL

When external voltage references are used, the comparator module can be configured to have the comparators operate from the same or different reference sources. However, threshold detector applications may require the same reference. The reference signal must be between VSS and VDD, and can be applied to either pin of the comparator(s).

7.3.2 INTERNAL REFERENCE SIGNAL

The comparator module also allows the selection of an internally generated voltage reference for the comparators. Section 10, Instruction Sets, contains a detailed description of the Voltage Reference Module that provides this signal. The internal reference signal is used when the comparators are in mode CM<2:0>=010 (Figure 7-1). In this mode, the internal voltage reference is applied to the VIN+ pin of both comparators.

EXAMPLE 8-1: VOLTAGE REFERENCE CONFIGURATION

```
MOVLW    0x02          ; 4 Inputs Muxed
MOVWF    CMCON         ; to 2 comps.
BSF      STATUS,RP0    ; go to Bank 1
MOVLW    0x0F          ; RA3-RA0 are
MOVWF    TRISA         ; inputs
MOVLW    0xA6          ; enable VREF
MOVWF    VRCON         ; low range
                       ; set VR<3:0>=6
BCF      STATUS,RP0    ; go to Bank 0
CALL     DELAY10       ; 10µs delay
```

8.2 Voltage Reference Accuracy/Error

The full range of VSS to VDD cannot be realized due to the construction of the module. The transistors on the top and bottom of the resistor ladder network (Figure 8-1) keep VREF from approaching VSS or VDD. The voltage reference is VDD derived and therefore, the VREF output changes with fluctuations in VDD. The tested absolute accuracy of the voltage reference can be found in Table 12-2.

8.3 Operation During SLEEP

When the device wakes up from SLEEP through an interrupt or a Watchdog Timer time-out, the contents of the VRCON register are not affected. To minimize current consumption in SLEEP mode, the voltage reference should be disabled.

8.4 Effects of a RESET

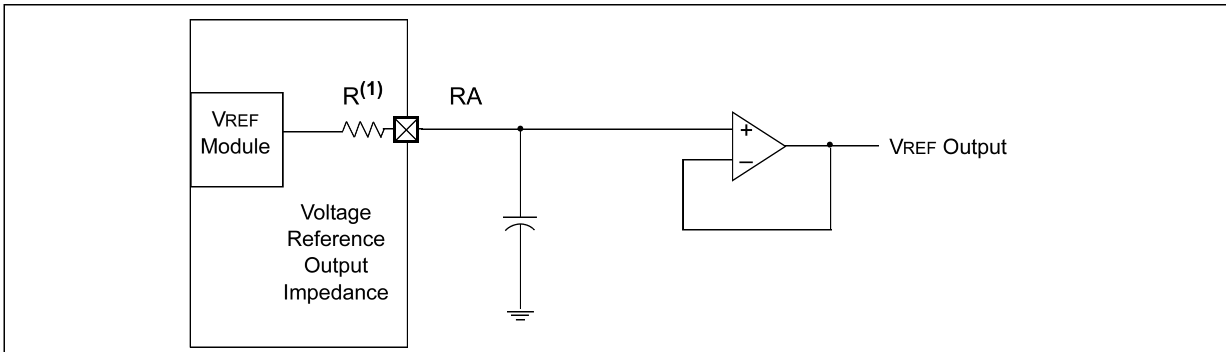
A device RESET disables the voltage reference by clearing bit VREN (VRCON<7>). This reset also disconnects the reference from the RA2 pin by clearing bit VROE (VRCON<6>) and selects the high voltage range by clearing bit VRR (VRCON<5>). The VREF value select bits, VRCON<3:0>, are also cleared.

8.5 Connection Considerations

The voltage reference module operates independently of the comparator module. The output of the reference generator may be connected to the RA2 pin if the TRISA<2> bit is set and the VROE bit, VRCON<6>, is set. Enabling the voltage reference output onto the RA2 pin with an input signal present will increase current consumption. Connecting RA2 as a digital output with VREF enabled will also increase current consumption.

The RA2 pin can be used as a simple D/A output with limited drive capability. Due to the limited drive capability, a buffer must be used in conjunction with the voltage reference output for external connections to VREF. Figure 8-2 shows an example buffering technique.

FIGURE 8-2: VOLTAGE REFERENCE OUTPUT BUFFER EXAMPLE



Note 1: R is dependent upon the Voltage Reference Configuration VRCON<3:0> and VRCON<5>.

TABLE 8-1: REGISTERS ASSOCIATED WITH VOLTAGE REFERENCE

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value On POR	Value On All Other RESETS
9Fh	VRCON	VREN	VROE	VRR	—	VR3	VR2	VR1	VR0	000- 0000	000- 0000
1Fh	CMCON	C2OUT	C1OUT	—	—	CIS	CM2	CM1	CM0	00-- 0000	00-- 0000
85h	TRISA	—	—	—	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	---1 1111	---1 1111

Note: - = Unimplemented, read as "0"

9.1 Configuration Bits

The configuration bits can be programmed (read as '0') or left unprogrammed (read as '1') to select various device configurations. These bits are mapped in program memory location 2007h.

The user will note that address 2007h is beyond the user program memory space. In fact, it belongs to the special test/configuration memory space (2000h – 3FFFh), which can be accessed only during programming.

REGISTER 9-1: CONFIGURATION WORD (ADDRESS 2007h)

CP1	CP0 (2)	CP1	$\overline{\text{CP0}}$ (2)	$\overline{\text{CP1}}$	CP0 (2)		BODEN	CP1	CP0 (2)	$\overline{\text{PWRTE}}$	WDTE	F0SC1	F0SC0

bit 13 bit 0

bit 13-8, 5-4: **CP<1:0>:** Code protection bit pairs (2)

Code protection for 2K program memory
`11` = Program memory code protection off
`10` = 0400h-07FFh code protected
`01` = 0200h-07FFh code protected
`00` = 0000h-07FFh code protected

Code protection for 1K program memory
`11` = Program memory code protection off
`10` = Program memory code protection off
`01` = 0200h-03FFh code protected
`00` = 0000h-03FFh code protected

Code protection for 0.5K program memory
`11` = Program memory code protection off
`10` = Program memory code protection off
`01` = Program memory code protection off
`00` = 0000h-01FFh code protected

bit 7 **Unimplemented**: Read as ‘0’

bit 6 **BODEN**: Brown-out Reset Enable bit (1)
`1` = BOR enabled
`0` = BOR disabled

bit 3 **$\overline{\text{PWRTE}}$**: Power-up Timer Enable bit (1, 3)
`1` = PWRT disabled
`0` = PWRT enabled

bit 2 **WDTE**: Watchdog Timer Enable bit
`1` = WDT enabled
`0` = WDT disabled

bit 1-0 **FOSC1:FOSC0**: Oscillator Selection bits
`11` = RC oscillator
`10` = HS oscillator
`01` = XT oscillator
`00` = LP oscillator

Note 1: Enabling Brown-out Reset automatically enables Power-up Timer (PWRT) regardless of the value of bit $\overline{\text{PWRTE}}$. Ensure the Power-up Timer is enabled anytime Brown-out Detect Reset is enabled.
2: All of the CP<1:0> pairs have to be given the same value to enable the code protection scheme listed.
3: Unprogrammed parts default the Power-up Timer disabled.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’	
-n = Value at POR	1 = bit is set	0 = bit is cleared	x = bit is unknown

9.4.5 TIME-OUT SEQUENCE

On power-up the time-out sequence is as follows: First PWRT time-out is invoked after POR has expired. Then OST is activated. The total time-out will vary based on oscillator configuration and $\overline{\text{PWRTE}}$ bit status. For example, in RC mode with $\overline{\text{PWRTE}}$ bit erased (PWRT disabled), there will be no time-out at all. Figure 9-8, Figure 9-9 and Figure 9-10 depict time-out sequences.

Since the time-outs occur from the POR pulse, if $\overline{\text{MCLR}}$ is kept low long enough, the time-outs will expire. Then bringing $\overline{\text{MCLR}}$ high will begin execution immediately (see Figure 9-9). This is useful for testing purposes or to synchronize more than one PIC16C62X device operating in parallel.

Table 9-4 shows the RESET conditions for some special registers, while Table 9-5 shows the RESET conditions for all the registers.

9.4.6 POWER CONTROL (PCON)/ STATUS REGISTER

The power control/STATUS register, PCON (address 8Eh), has two bits.

Bit0 is $\overline{\text{BOR}}$ (Brown-out). $\overline{\text{BOR}}$ is unknown on Power-on Reset. It must then be set by the user and checked on subsequent RESETS to see if $\overline{\text{BOR}}$ = 0, indicating that a brown-out has occurred. The $\overline{\text{BOR}}$ STATUS bit is a don't care and is not necessarily predictable if the brown-out circuit is disabled (by setting BODEN bit = 0 in the Configuration word).

Bit1 is $\overline{\text{POR}}$ (Power-on Reset). It is a '0' on Power-on Reset and unaffected otherwise. The user must write a '1' to this bit following a Power-on Reset. On a subsequent RESET, if $\overline{\text{POR}}$ is '0', it will indicate that a Power-on Reset must have occurred (VDD may have gone too low).

TABLE 9-1: TIME-OUT IN VARIOUS SITUATIONS

Oscillator Configuration	Power-up		Brown-out Reset	Wake-up from SLEEP
	$\overline{\text{PWRTE}}$ = 0	$\overline{\text{PWRTE}}$ = 1		
XT, HS, LP	72 ms + 1024 TOSC	1024 TOSC	72 ms + 1024 TOSC	1024 TOSC
RC	72 ms	—	72 ms	—

TABLE 9-2: STATUS/PCON BITS AND THEIR SIGNIFICANCE

POR	BOR	TO	PD	
0	X	1	1	Power-on Reset
0	X	0	X	Illegal, $\overline{\text{TO}}$ is set on $\overline{\text{POR}}$
0	X	X	0	Illegal, $\overline{\text{PD}}$ is set on $\overline{\text{POR}}$
1	0	X	X	Brown-out Reset
1	1	0	u	WDT Reset
1	1	0	0	WDT Wake-up
1	1	u	u	$\overline{\text{MCLR}}$ Reset during normal operation
1	1	1	0	$\overline{\text{MCLR}}$ Reset during SLEEP

Legend: u = unchanged, x = unknown

TABLE 9-3: SUMMARY OF REGISTERS ASSOCIATED WITH BROWN-OUT

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR Reset	Value on all other RESETS(1)
83h	STATUS				$\overline{\text{TO}}$	$\overline{\text{PD}}$				0001 1xxx	000q quuu
8Eh	PCON	—	—	—	—	—	—	$\overline{\text{POR}}$	$\overline{\text{BOR}}$	---- --0x	---- --uq

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0', q = value depends on condition.

Note 1: Other (non Power-up) Resets include $\overline{\text{MCLR}}$ Reset, Brown-out Reset and Watchdog Timer Reset during normal operation.

FIGURE 9-8: TIME-OUT SEQUENCE ON POWER-UP ($\overline{MCLR}$ NOT TIED TO V_{DD}): CASE 1

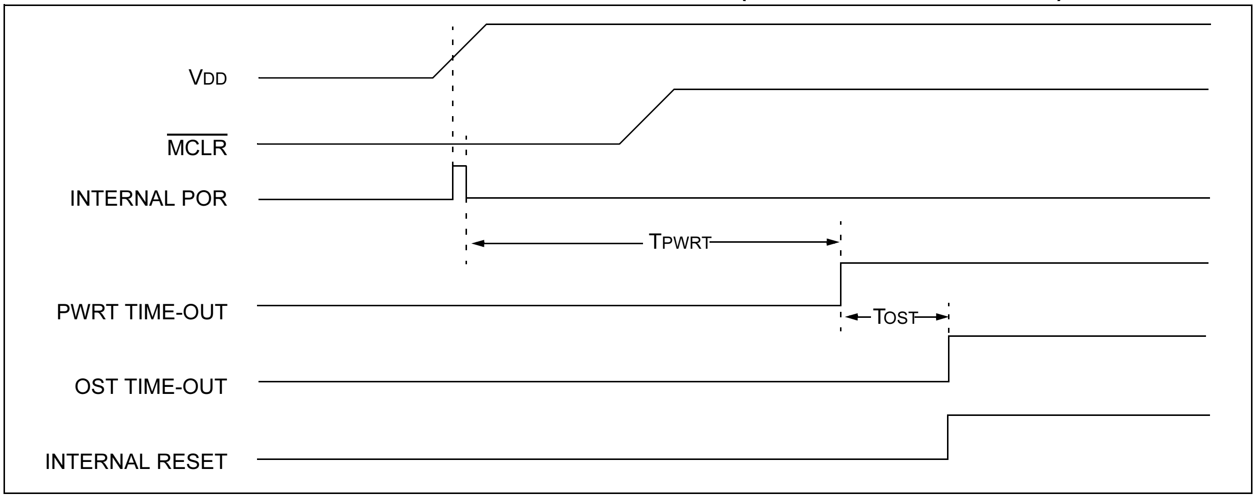


FIGURE 9-9: TIME-OUT SEQUENCE ON POWER-UP ($\overline{MCLR}$ NOT TIED TO V_{DD}): CASE 2

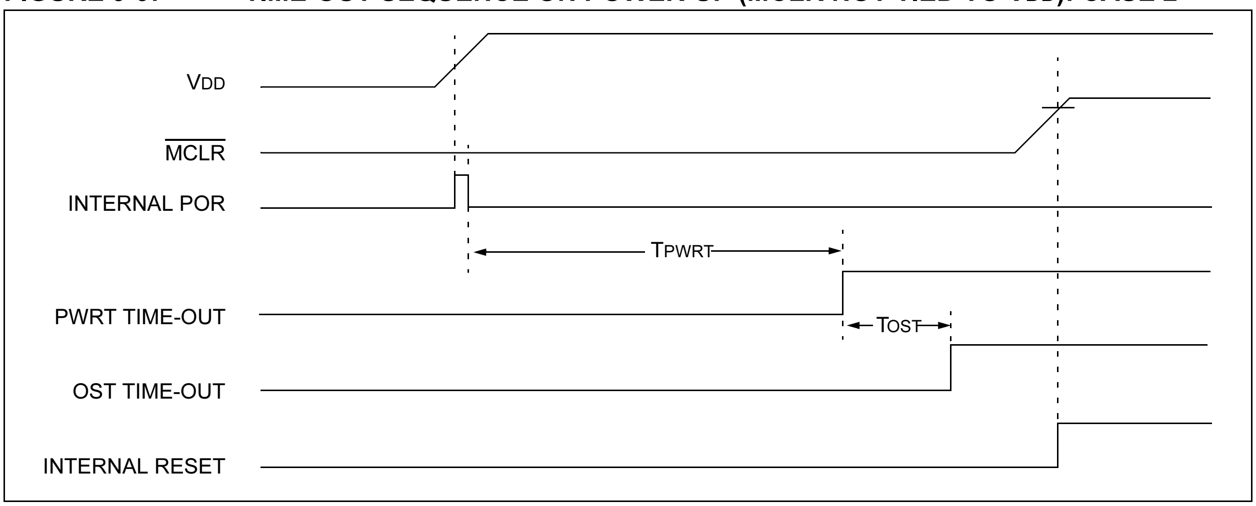
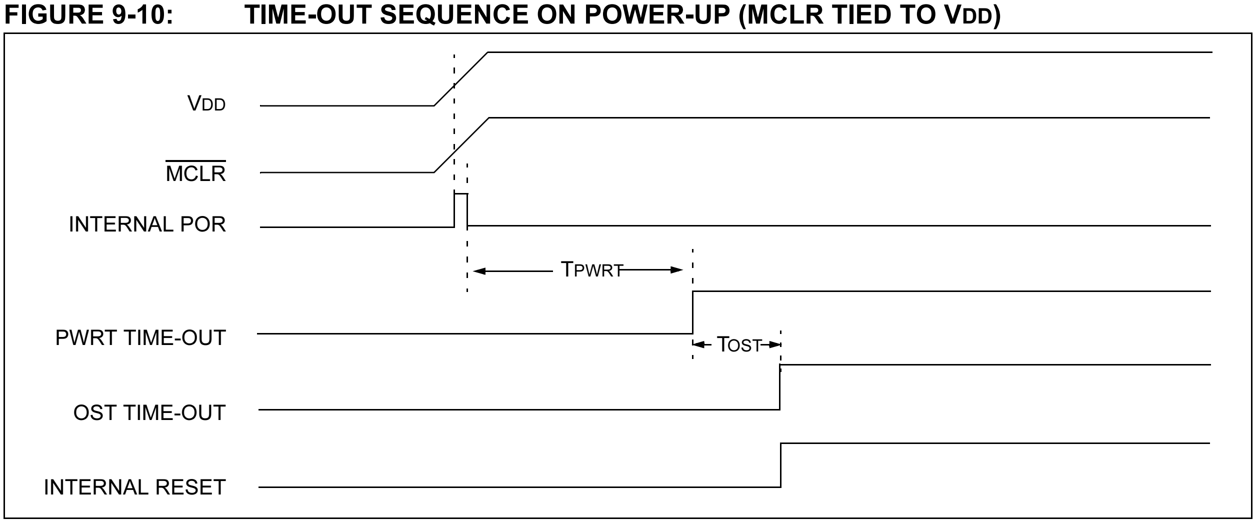


FIGURE 9-10: TIME-OUT SEQUENCE ON POWER-UP ($\overline{MCLR}$ TIED TO V_{DD})



9.5 Interrupts

The PIC16C62X has 4 sources of interrupt:

- External interrupt RB0/INT
- TMR0 overflow interrupt
- PORTB change interrupts (pins RB<7:4>)
- Comparator interrupt

The interrupt control register (INTCON) records individual interrupt requests in flag bits. It also has individual and global interrupt enable bits.

A global interrupt enable bit, GIE (INTCON<7>) enables (if set) all un-masked interrupts or disables (if cleared) all interrupts. Individual interrupts can be disabled through their corresponding enable bits in INTCON register. GIE is cleared on RESET.

The “return from interrupt” instruction, `RETFIE`, exits interrupt routine, as well as sets the GIE bit, which re-enable RB0/INT interrupts.

The INT pin interrupt, the RB port change interrupt and the TMR0 overflow interrupt flags are contained in the INTCON register.

The peripheral interrupt flag is contained in the special register PIR1. The corresponding interrupt enable bit is contained in special registers PIE1.

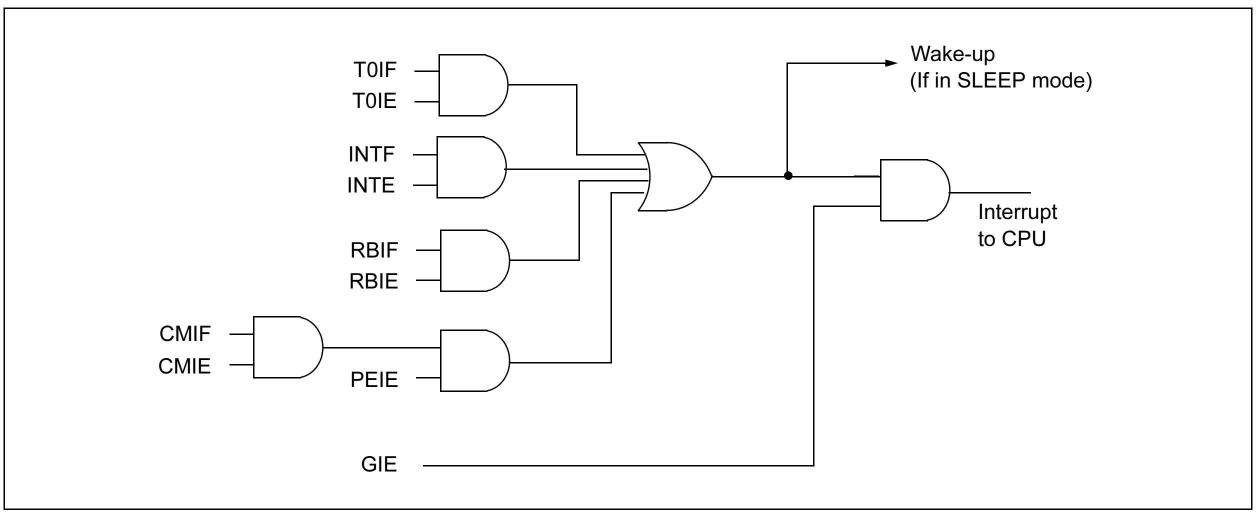
When an interrupt is responded to, the GIE is cleared to disable any further interrupt, the return address is pushed into the stack and the PC is loaded with 0004h.

Once in the interrupt service routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid RB0/INT recursive interrupts.

For external interrupt events, such as the INT pin or PORTB change interrupt, the interrupt latency will be three or four instruction cycles. The exact latency depends when the interrupt event occurs (Figure 9-16). The latency is the same for one or two cycle instructions. Once in the interrupt service routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid multiple interrupt requests.

> **Note 1:** Individual interrupt flag bits are set regardless of the status of their corresponding mask bit or the GIE bit.
>
> **2:** When an instruction that clears the GIE bit is executed, any interrupts that were pending for execution in the next cycle are ignored. The CPU will execute a `NOP` in the cycle immediately following the instruction which clears the GIE bit. The interrupts which were ignored are still pending to be serviced when the GIE bit is set again.

FIGURE 9-15: INTERRUPT LOGIC



RLF — Rotate Left f through Carry

Syntax: [*label*] RLF f,d

Operands: $0 \leq f \leq 127$
$d \in [0,1]$

Operation: See description below

Status Affected: C

Encoding:

`00`	`1101`	`dfff`	`ffff`

Description: The contents of register 'f' are rotated one bit to the left through the Carry Flag. If 'd' is 0, the result is placed in the W register. If 'd' is 1, the result is stored back in register 'f'.

Words: 1

Cycles: 1

Example

```
RLF     REG1,0
```

Before Instruction

```
REG1   =   1110 0110
C      =   0
```

After Instruction

```
REG1   =   1110 0110
W      =   1100 1100
C      =   1
```

RRF — Rotate Right f through Carry

Syntax: [*label*] RRF f,d

Operands: $0 \leq f \leq 127$
$d \in [0,1]$

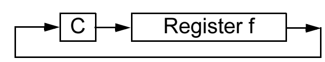
Operation: See description below

Status Affected: C

Encoding:

`00`	`1100`	`dfff`	`ffff`

Description: The contents of register 'f' are rotated one bit to the right through the Carry Flag. If 'd' is 0, the result is placed in the W register. If 'd' is 1, the result is placed back in register 'f'.



Words: 1

Cycles: 1

Example `RRF` REG1, 0

Before Instruction

```
REG1   =   1110 0110
C      =   0
```

After Instruction

```
REG1   =   1110 0110
W      =   0111 0011
C      =   0
```

SLEEP

Syntax: [*label*] SLEEP

Operands: None

Operation: 00h → WDT,
0 → WDT prescaler,
1 → $\overline{\text{TO}}$,
0 → $\overline{\text{PD}}$

Status Affected: $\overline{\text{TO}}$, $\overline{\text{PD}}$

Encoding:

`00`	`0000`	`0110`	`0011`

Description: The power-down STATUS bit, $\overline{\text{PD}}$ is cleared. Time-out STATUS bit, $\overline{\text{TO}}$ is set. Watchdog Timer and its prescaler are cleared.
The processor is put into SLEEP mode with the oscillator stopped. See Section 9.8 for more details.

Words: 1

Cycles: 1

Example: `SLEEP`

NOTES:

11.14 PICDEM 1 PICmicro Demonstration Board

The PICDEM 1 demonstration board demonstrates the capabilities of the PIC16C5X (PIC16C54 to PIC16C58A), PIC16C61, PIC16C62X, PIC16C71, PIC16C8X, PIC17C42, PIC17C43 and PIC17C44. All necessary hardware and software is included to run basic demo programs. The sample microcontrollers provided with the PICDEM 1 demonstration board can be programmed with a PRO MATE II device programmer, or a PICSTART Plus development programmer. The PICDEM 1 demonstration board can be connected to the MPLAB ICE in-circuit emulator for testing. A prototype area extends the circuitry for additional application components. Features include an RS-232 interface, a potentiometer for simulated analog input, push button switches and eight LEDs.

11.15 PICDEM.net Internet/Ethernet Demonstration Board

The PICDEM.net demonstration board is an Internet/Ethernet demonstration board using the PIC18F452 microcontroller and TCP/IP firmware. The board supports any 40-pin DIP device that conforms to the standard pinout used by the PIC16F877 or PIC18C452. This kit features a user friendly TCP/IP stack, web server with HTML, a 24L256 Serial EEPROM for Xmodem download to web pages into Serial EEPROM, ICSP/MPLAB ICD 2 interface connector, an Ethernet interface, RS-232 interface, and a 16 x 2 LCD display. Also included is the book and CD-ROM *"TCP/IP Lean, Web Servers for Embedded Systems,"* by Jeremy Bentham

11.16 PICDEM 2 Plus Demonstration Board

The PICDEM 2 Plus demonstration board supports many 18-, 28-, and 40-pin microcontrollers, including PIC16F87X and PIC18FXX2 devices. All the necessary hardware and software is included to run the demonstration programs. The sample microcontrollers provided with the PICDEM 2 demonstration board can be programmed with a PRO MATE II device programmer, PICSTART Plus development programmer, or MPLAB ICD 2 with a Universal Programmer Adapter. The MPLAB ICD 2 and MPLAB ICE in-circuit emulators may also be used with the PICDEM 2 demonstration board to test firmware. A prototype area extends the circuitry for additional application components. Some of the features include an RS-232 interface, a 2 x 16 LCD display, a piezo speaker, an on-board temperature sensor, four LEDs, and sample PIC18F452 and PIC16F877 FLASH microcontrollers.

11.17 PICDEM 3 PIC16C92X Demonstration Board

The PICDEM 3 demonstration board supports the PIC16C923 and PIC16C924 in the PLCC package. All the necessary hardware and software is included to run the demonstration programs.

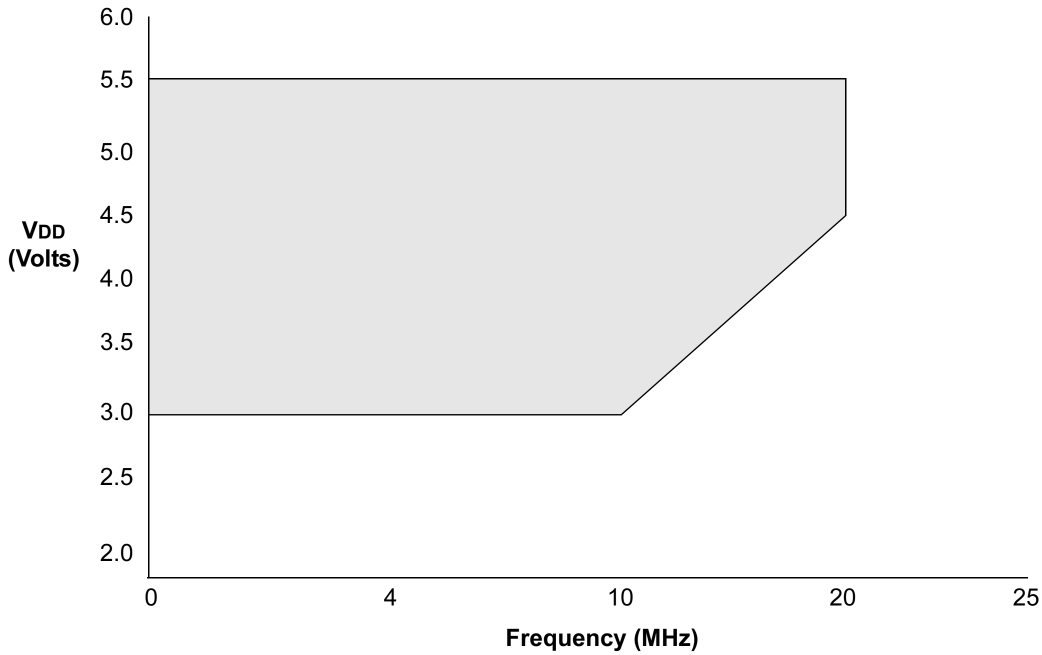
11.18 PICDEM 4 8/14/18-Pin Demonstration Board

The PICDEM 4 can be used to demonstrate the capabilities of the 8-, 14-, and 18-pin PIC16XXXX and PIC18XXXX MCUs, including the PIC16F818/819, PIC16F87/88, PIC16F62XA and the PIC18F1320 family of microcontrollers. PICDEM 4 is intended to showcase the many features of these low pin count parts, including LIN and Motor Control using ECCP. Special provisions are made for low power operation with the supercapacitor circuit, and jumpers allow on-board hardware to be disabled to eliminate current draw in this mode. Included on the demo board are provisions for Crystal, RC or Canned Oscillator modes, a five volt regulator for use with a nine volt wall adapter or battery, DB-9 RS-232 interface, ICD connector for programming via ICSP and development with MPLAB ICD 2, 2x16 liquid crystal display, PCB footprints for H-Bridge motor driver, LIN transceiver and EEPROM. Also included are: header for expansion, eight LEDs, four potentiometers, three push buttons and a prototyping area. Included with the kit is a PIC16F627A and a PIC18F1320. Tutorial firmware is included along with the User's Guide.

11.19 PICDEM 17 Demonstration Board

The PICDEM 17 demonstration board is an evaluation board that demonstrates the capabilities of several Microchip microcontrollers, including PIC17C752, PIC17C756A, PIC17C762 and PIC17C766. A programmed sample is included. The PRO MATE II device programmer, or the PICSTART Plus development programmer, can be used to reprogram the device for user tailored application development. The PICDEM 17 demonstration board supports program download and execution from external on-board FLASH memory. A generous prototype area is available for user hardware expansion.

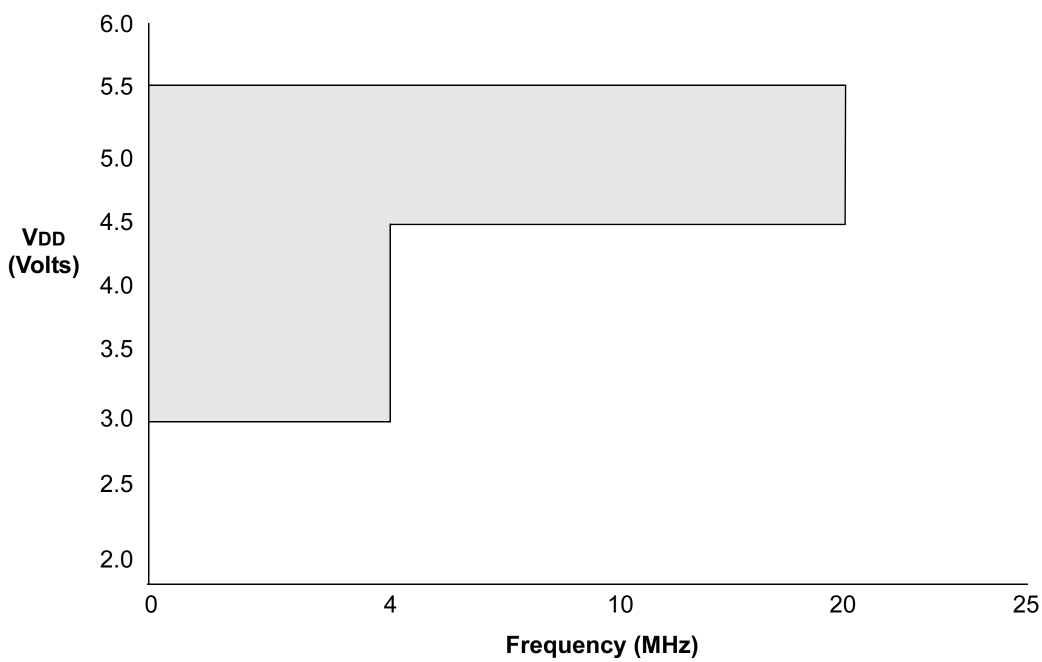
FIGURE 12-3: PIC16C62XA VOLTAGE-FREQUENCY GRAPH, 0°C ≤ TA ≤ +70°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

2: The maximum rated speed of the part limits the permissible combinations of voltage and frequency. Please reference the Product Identification System section for the maximum rated speed of the parts.

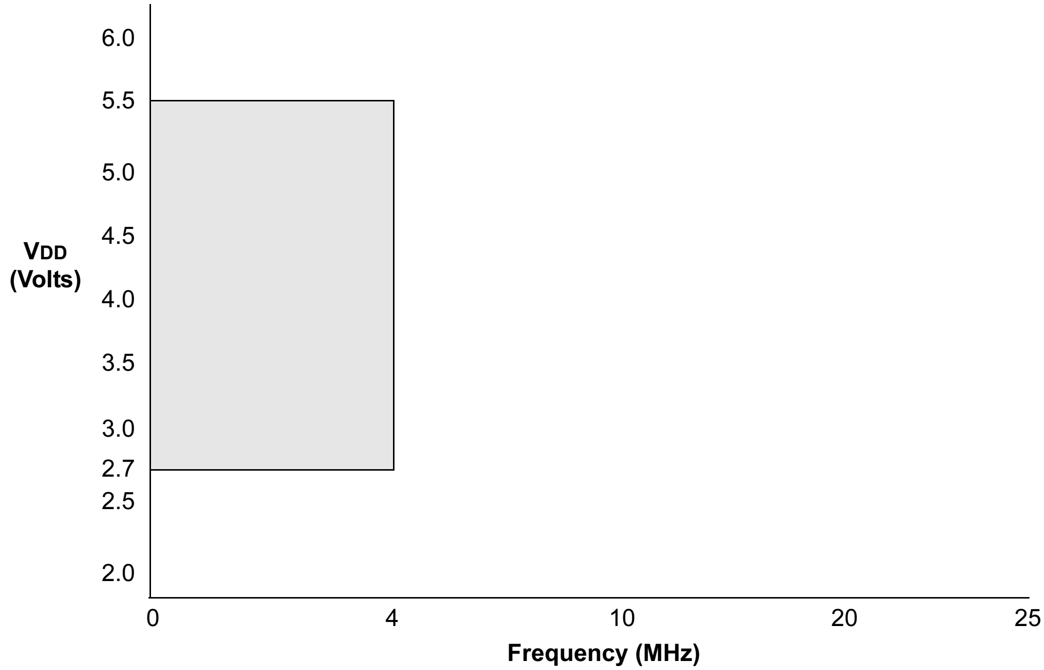
FIGURE 12-4: PIC16C62XA VOLTAGE-FREQUENCY GRAPH, -40°C ≤ TA ≤ 0°C, +70°C ≤ TA ≤ +125°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

2: The maximum rated speed of the part limits the permissible combinations of voltage and frequency. Please reference the Product Identification System section for the maximum rated speed of the parts.

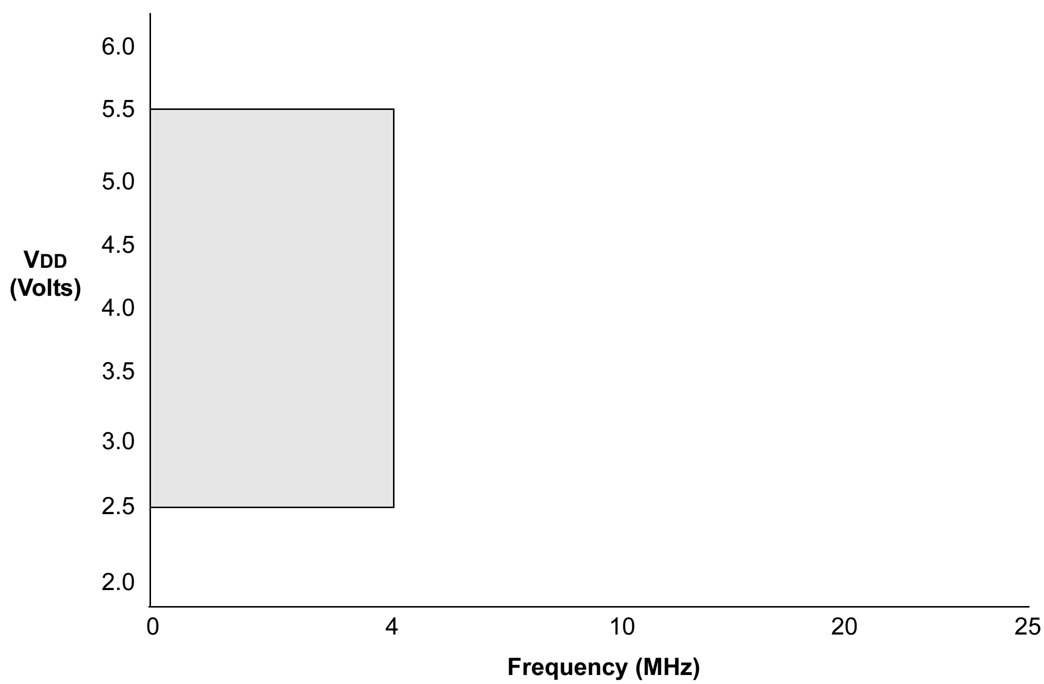
FIGURE 12-5: PIC16LC620A/LC621A/LC622A VOLTAGE-FREQUENCY GRAPH, -40°C ≤ TA ≤ 0°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

2: The maximum rated speed of the part limits the permissible combinations of voltage and frequency. Please reference the Product Identification System section for the maximum rated speed of the parts.

FIGURE 12-6: PIC16LC620A/LC621A/LC622A VOLTAGE-FREQUENCY GRAPH, 0°C ≤ TA ≤ +125°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

2: The maximum rated speed of the part limits the permissible combinations of voltage and frequency. Please reference the Product Identification System section for the maximum rated speed of the parts.

12.1 DC Characteristics: PIC16C62X-04 (Commercial, Industrial, Extended) PIC16C62X-20 (Commercial, Industrial, Extended) PIC16LC62X-04 (Commercial, Industrial, Extended)

PIC16C62X	**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial and -40°C ≤ TA ≤ +125°C for extended
PIC16LC62X	**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial and -40°C ≤ TA ≤ +125°C for extended Operating voltage VDD range is the PIC16C62X range.

Param. No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
D001	VDD	Supply Voltage	3.0	—	6.0	V	See Figures 12-1, 12-2, 12-3, 12-4, and 12-5
D001	VDD	Supply Voltage	2.5	—	6.0	V	See Figures 12-1, 12-2, 12-3, 12-4, and 12-5
D002	VDR	RAM Data Retention Voltage(1)	—	1.5*	—	V	Device in SLEEP mode
D002	VDR	RAM Data Retention Voltage(1)	—	1.5*	—	V	Device in SLEEP mode
D003	VPOR	VDD start voltage to ensure Power-on Reset	—	Vss	—	V	See section on Power-on Reset for details
D003	VPOR	VDD start voltage to ensure Power-on Reset	—	VSS	—	V	See section on Power-on Reset for details
D004	SVDD	VDD rise rate to ensure Power-on Reset	0.05*	—	—	V/ms	See section on Power-on Reset for details
D004	SVDD	VDD rise rate to ensure Power-on Reset	0.05*	—	—	V/ms	See section on Power-on Reset for details
D005	VBOR	Brown-out Detect Voltage	3.7	4.0	4.3	V	BOREN configuration bit is cleared
D005	VBOR	Brown-out Detect Voltage	3.7	4.0	4.3	V	BOREN configuration bit is cleared
D010	IDD	Supply Current(2)	— — —	1.8 35 9.0	3.3 70 20	mA μA mA	FOSC = 4 MHz, VDD = 5.5V, WDT disabled, XT mode, (**Note 4**)* FOSC = 32 kHz, VDD = 4.0V, WDT disabled, LP mode FOSC = 20 MHz, VDD = 5.5V, WDT disabled, HS mode
D010	IDD	Supply Current(2)	— —	1.4 26	2.5 53	mA μA	FOSC = 2.0 MHz, VDD = 3.0V, WDT disabled, XT mode, (**Note 4**) FOSC = 32 kHz, VDD = 3.0V, WDT disabled, LP mode
D020	IPD	Power-down Current(3)	—	1.0	2.5 15	μA μA	VDD=4.0V, WDT disabled (125°C)
D020	IPD	Power-down Current(3)	—	0.7	2	μA	VDD=3.0V, WDT disabled

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.
The test conditions for all IDD measurements in Active Operation mode are:
$\overline{\text{OSC1}}$ = external square wave, from rail to rail; all I/O pins tri-stated, pulled to VDD,
$\overline{\text{MCLR}}$ = VDD; WDT enabled/disabled as specified.

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD or VSS.

4: For RC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula: Ir = VDD/2REXT (mA) with REXT in kΩ.

5: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.

12.5 DC CHARACTERISTICS: PIC16C620A/C621A/C622A-40(7) (Commercial) PIC16CR620A-40(7) (Commercial)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature 0°C ≤ TA ≤ +70°C for commercial				
Param No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
D001	VDD	Supply Voltage	3.0	—	5.5	V	FOSC = DC to 20 MHz
D002	VDR	RAM Data Retention Voltage(1)	—	1.5*	—	V	Device in SLEEP mode
D003	VPOR	VDD start voltage to ensure Power-on Reset	—	VSS	—	V	See section on Power-on Reset for details
D004	SVDD	VDD rise rate to ensure Power-on Reset	0.05 *	—	—	V/ms	See section on Power-on Reset for details
D005	VBOR	Brown-out Detect Voltage	3.65	4.0	4.35	V	BOREN configuration bit is cleared
D010	IDD	Supply Current(2,4)	—	1.2	2.0	mA	FOSC = 4 MHz, VDD = 5.5V, WDT disabled, XT OSC mode, **(Note 4)***
			—	0.4	1.2	mA	FOSC = 4 MHz, VDD = 3.0V, WDT disabled, XT OSC mode, **(Note 4)**
			—	1.0	2.0	mA	FOSC = 10 MHz, VDD = 3.0V, WDT disabled, HS OSC mode, **(Note 6)**
			—	4.0	6.0	mA	FOSC = 20 MHz, VDD = 4.5V, WDT disabled, HS OSC mode
			—	4.0	7.0	mA	FOSC = 20 MHz, VDD = 5.5V, WDT disabled*, HS OSC mode
			—	35	70	µA	FOSC = 32 kHz, VDD = 3.0V, WDT disabled, LP OSC mode
D020	IPD	Power Down Current(3)	—	—	2.2	µA	VDD = 3.0V
			—	—	5.0	µA	VDD = 4.5V*
			—	—	9.0	µA	VDD = 5.5V
			—	—	15	µA	VDD = 5.5V Extended
D022	ΔIWDT	WDT Current(5)	—	6.0	10	µA	VDD = 4.0V
					12	µA	(125°C)
D022A	ΔIBOR	Brown-out Reset Current(5)	—	75	125	µA	$\overline{\text{BOD}}$ enabled, VDD = 5.0V
D023	ΔICOMP	Comparator Current for each Comparator(5)	—	30	60	µA	VDD = 4.0V
D023A	ΔIVREF	VREF Current(5)	—	80	135	µA	VDD = 4.0V
	ΔIEE Write	Operating Current	—		3	mA	VCC = 5.5V, SCL = 400 kHz
	ΔIEE Read	Operating Current	—		1	mA	
	ΔIEE	Standby Current	—		30	µA	VCC = 3.0V, EE VDD = VCC
	ΔIEE	Standby Current	—		100	µA	VCC = 3.0V, EE VDD = VCC
1A	FOSC	LP Oscillator Operating Frequency	0	—	200	kHz	All temperatures
		RC Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		XT Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		HS Oscillator Operating Frequency	0	—	20	MHz	All temperatures

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C, unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in SLEEP mode without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.
The test conditions for all IDD measurements in Active Operation mode are:
OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD, $\overline{\text{MCLR}}$ = VDD; WDT enabled/disabled as specified.

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD or VSS.

4: For RC OSC configuration, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/2REXT (mA) with REXT in kΩ.

5: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.

6: Commercial temperature range only.

7: See Section 12.1 and Section 12.3 for 16C62X and 16CR62X devices for operation between 20 MHz and 40 MHz for valid modified characteristics.

12.9 Timing Diagrams and Specifications

FIGURE 12-12: EXTERNAL CLOCK TIMING

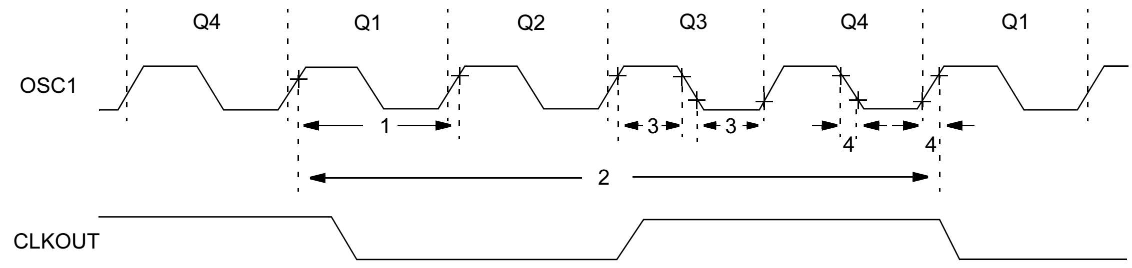


TABLE 12-3: EXTERNAL CLOCK TIMING REQUIREMENTS

Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
1A	Fosc	External CLKIN Frequency(1)	DC	—	4	MHz	XT and RC Osc mode, VDD=5.0V
			DC	—	20	MHz	HS Osc mode
			DC	—	200	kHz	LP Osc mode
		Oscillator Frequency(1)	DC	—	4	MHz	RC Osc mode, VDD=5.0V
			0.1	—	4	MHz	XT Osc mode
			1	—	20	MHz	HS Osc mode
			DC	—	200	kHz	LP Osc mode
1	Tosc	External CLKIN Period(1)	250	—	—	ns	XT and RC Osc mode
			50	—	—	ns	HS Osc mode
			5	—	—	μs	LP Osc mode
		Oscillator Period(1)	250	—	—	ns	RC Osc mode
			250	—	10,000	ns	XT Osc mode
			50	—	1,000	ns	HS Osc mode
			5	—	—	μs	LP Osc mode
2	TCY	Instruction Cycle Time(1)	1.0	FOSC/4	DC	μs	TCYS=FOSC/4
3*	TosL, TosH	External Clock in (OSC1) High or Low Time	100*	—	—	ns	XT oscillator, TOSC L/H duty cycle
			2*	—	—	μs	LP oscillator, TOSC L/H duty cycle
			20*	—	—	ns	HS oscillator, TOSC L/H duty cycle
4*	TosR, TosF	External Clock in (OSC1) Rise or Fall Time	25*	—	—	ns	XT oscillator
			50*	—	—	ns	LP oscillator
			15*	—	—	ns	HS oscillator

2: * These parameters are characterized but not tested.

3: † Data in "Typ" column is at 5.0V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (TCY) equals four times the input oscillator time-base period. All specified values are based on characterization data for that particular oscillator type under standard operating conditions with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min." values with an external clock applied to the OSC1 pin. When an external clock input is used, the "Max." cycle time limit is "DC" (no clock) for all devices.

FIGURE 12-14: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

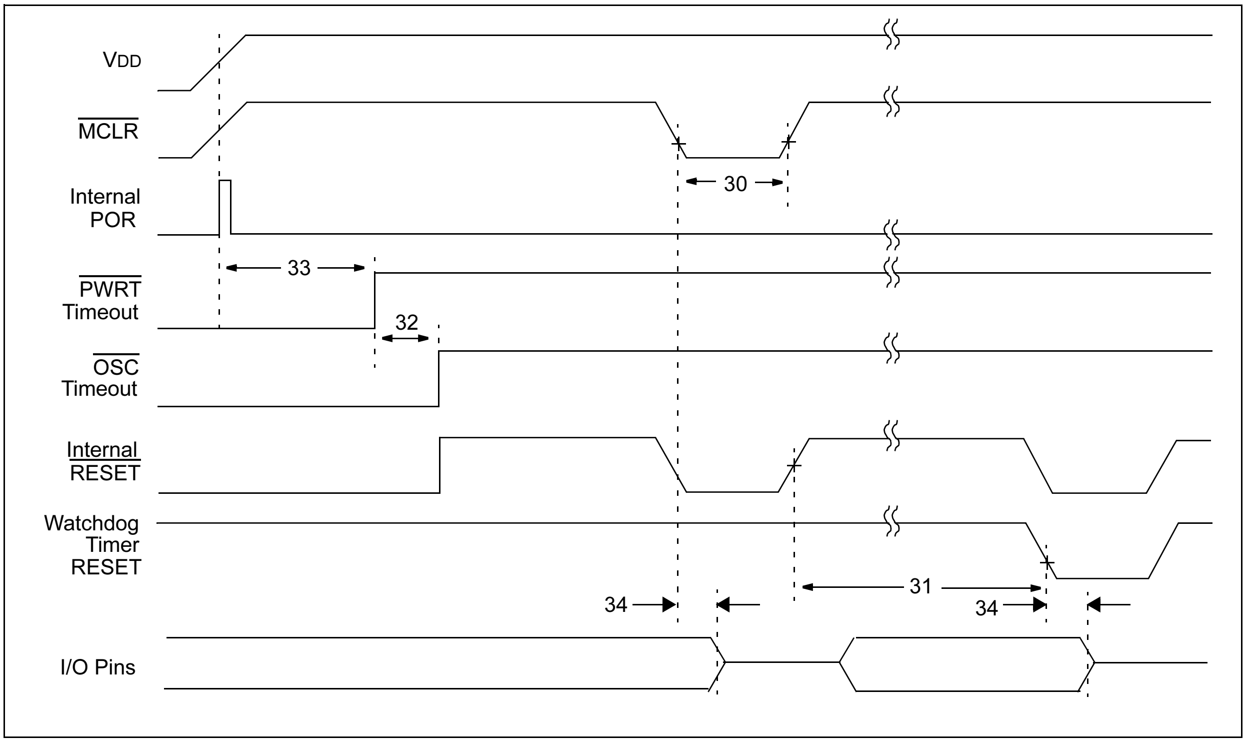


FIGURE 12-15: BROWN-OUT RESET TIMING

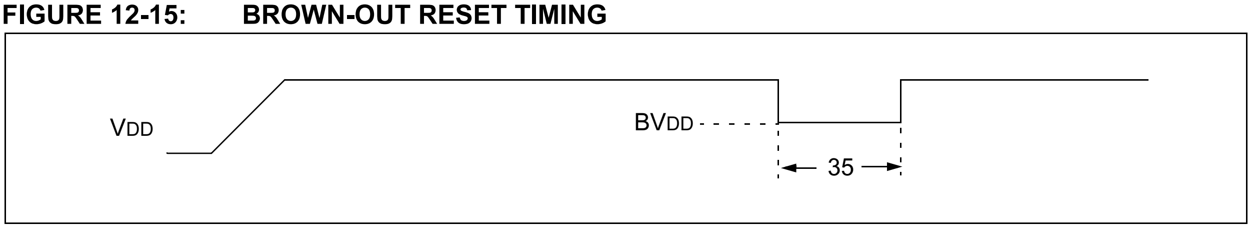


TABLE 12-5: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER REQUIREMENTS

Parameter No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
30	TmcL	MCLR Pulse Width (low)	2000	—	—	ns	-40° to +85°C
31	Twdt	Watchdog Timer Time-out Period (No Prescaler)	7*	18	33*	ms	VDD = 5.0V, -40° to +85°C
32	Tost	Oscillation Start-up Timer Period	—	1024 TOSC	—	—	TOSC = OSC1 period
33	Tpwrt	Power-up Timer Period	28*	72	132*	ms	VDD = 5.0V, -40° to +85°C
34	TIOZ	I/O hi-impedance from MCLR low		—	2.0	μs	
35	TBOR	Brown-out Reset Pulse Width	100*	—	—	μs	3.7V ≤ VDD ≤ 4.3V

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C, unless otherwise stated. These parameters are for design guidance only and are not tested.

