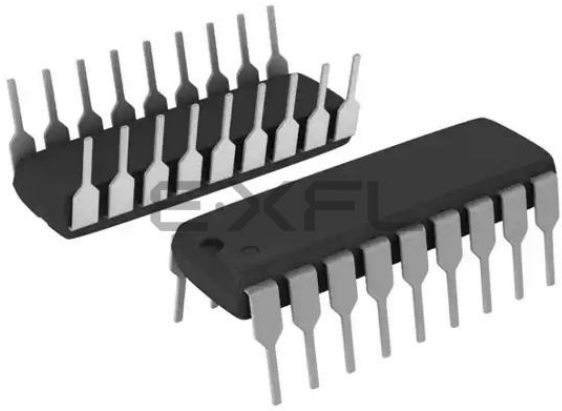

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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	4MHz
Connectivity	-
Peripherals	Brown-out Detect/Reset, POR, WDT
Number of I/O	13
Program Memory Size	3.5KB (2K x 14)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 6V
Data Converters	-
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Through Hole
Package / Case	18-DIP (0.300", 7.62mm)
Supplier Device Package	18-PDIP
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16c622-04e-p

NOTES:

4.0 MEMORY ORGANIZATION

4.1 Program Memory Organization

The PIC16C62X has a 13-bit program counter capable of addressing an 8K x 14 program memory space. Only the first 512 x 14 (0000h - 01FFh) for the PIC16C620(A) and PIC16CR620, 1K x 14 (0000h - 03FFh) for the PIC16C621(A) and 2K x 14 (0000h - 07FFh) for the PIC16C622(A) are physically implemented. Accessing a location above these boundaries will cause a wrap-around within the first 512 x 14 space (PIC16C(R)620(A)) or 1K x 14 space (PIC16C621(A)) or 2K x 14 space (PIC16C622(A)). The RESET vector is at 0000h and the interrupt vector is at 0004h (Figure 4-1, Figure 4-2, Figure 4-3).

FIGURE 4-1: PROGRAM MEMORY MAP AND STACK FOR THE PIC16C620/PIC16C620A/ PIC16CR620A

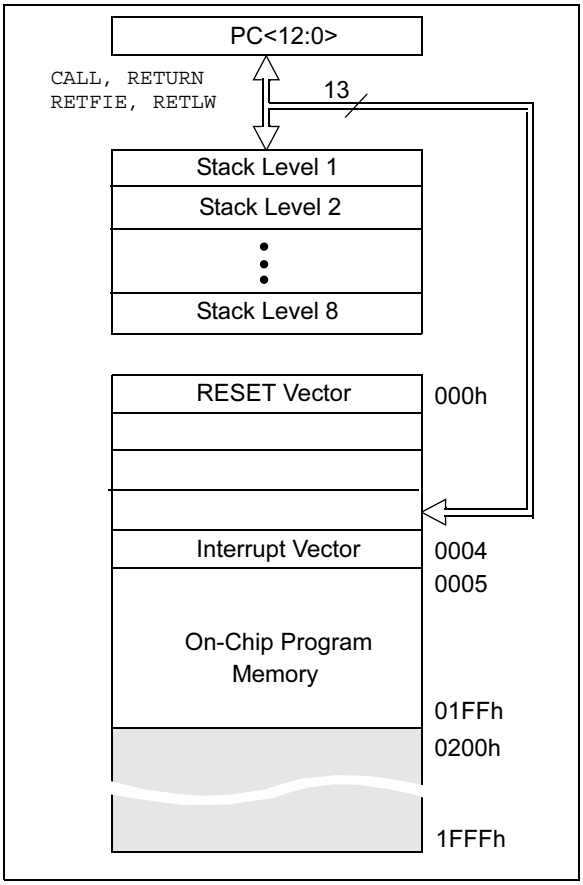


FIGURE 4-2: PROGRAM MEMORY MAP AND STACK FOR THE PIC16C621/PIC16C621A

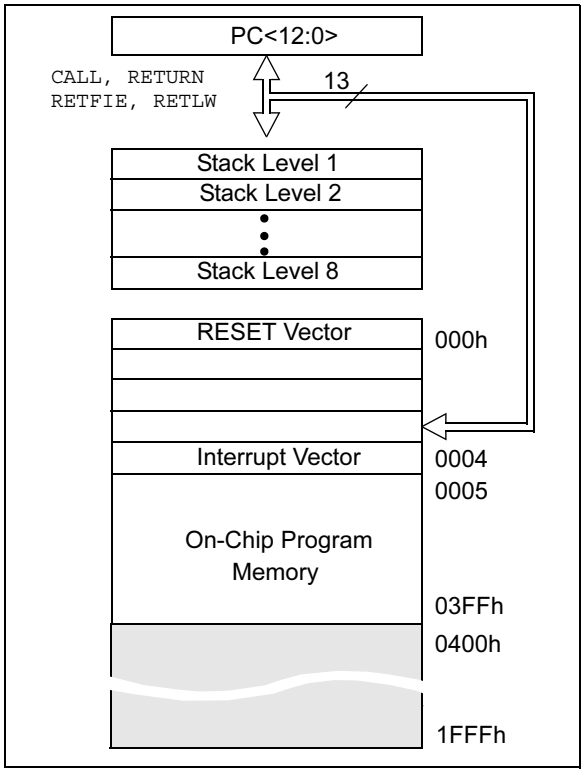
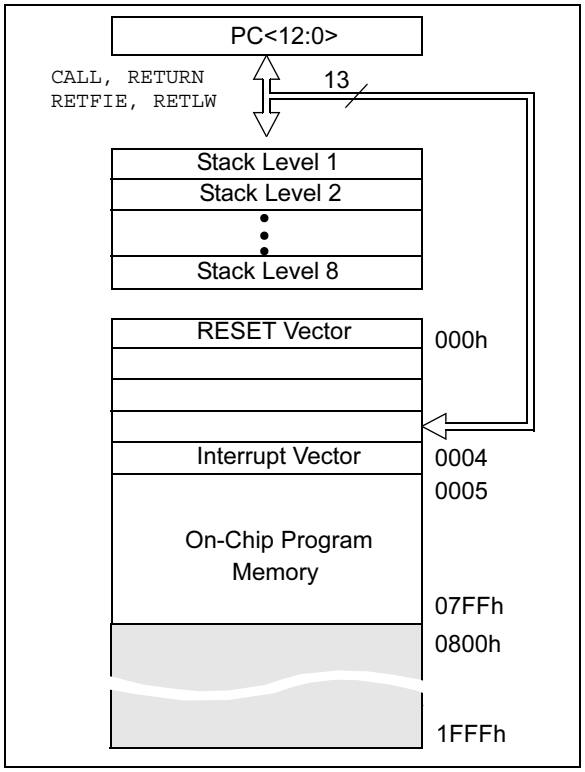


FIGURE 4-3: PROGRAM MEMORY MAP AND STACK FOR THE PIC16C622/PIC16C622A



4.2.2.1 STATUS Register

The STATUS register, shown in Register 4-1, contains the arithmetic status of the ALU, the RESET status and the bank select bits for data memory.

The STATUS register can be the destination for any instruction, like any other register. If the STATUS register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the $\overline{TO}$ and $\overline{PD}$ bits are not writable. Therefore, the result of an instruction with the STATUS register as destination may be different than intended.

For example, `CLRF STATUS` will clear the upper-three bits and set the Z bit. This leaves the STATUS register as `000uu1uu` (where `u` = unchanged).

It is recommended, therefore, that only `BCF`, `BSF`, `SWAPF` and `MOVWF` instructions are used to alter the STATUS register, because these instructions do not affect any STATUS bit. For other instructions not affecting any STATUS bits, see the "Instruction Set Summary".

Note 1: The IRP and RP1 bits (STATUS<7:6>) are not used by the PIC16C62X and should be programmed as '0'. Use of these bits as general purpose R/W bits is NOT recommended, since this may affect upward compatibility with future products.

2: The C and DC bits operate as a $\overline{\text{Borrow}}$ and Digit $\overline{\text{Borrow}}$ out bit, respectively, in subtraction. See the `SUBLW` and `SUBWF` instructions for examples.

REGISTER 4-1: STATUS REGISTER (ADDRESS 03H OR 83H)

Reserved	Reserved	R/W-0	R-1	R-1	R/W-x	R/W-x	R/W-x
IRP	RP1	RP0	$\overline{TO}$	$\overline{PD}$	Z	DC	C
bit 7							bit 0

bit 7 **IRP:** Register Bank Select bit (used for indirect addressing)
`1` = Bank 2, 3 (100h - 1FFh)
`0` = Bank 0, 1 (00h - FFh)
The IRP bit is reserved on the PIC16C62X; always maintain this bit clear.

bit 6-5 **RP<1:0>**: Register Bank Select bits (used for direct addressing)
`01` = Bank 1 (80h - FFh)
`00` = Bank 0 (00h - 7Fh)
Each bank is 128 bytes. The RP1 bit is reserved on the PIC16C62X; always maintain this bit clear.

bit 4 **$\overline{TO}$**: Time-out bit
`1` = After power-up, `CLRWDT` instruction, or `SLEEP` instruction
`0` = A WDT time-out occurred

bit 3 **$\overline{PD}$**: Power-down bit
`1` = After power-up or by the `CLRWDT` instruction
`0` = By execution of the `SLEEP` instruction

bit 2 **Z**: Zero bit
`1` = The result of an arithmetic or logic operation is zero
`0` = The result of an arithmetic or logic operation is not zero

bit 1 **DC**: Digit carry/$\overline{\text{borrow}}$ bit (`ADDWF, ADDLW,SUBLW,SUBWF` instructions)(for $\overline{\text{borrow}}$ the polarity is reversed)
`1` = A carry-out from the 4th low order bit of the result occurred
`0` = No carry-out from the 4th low order bit of the result

bit 0 **C**: Carry/$\overline{\text{borrow}}$ bit (`ADDWF, ADDLW,SUBLW,SUBWF` instructions)
`1` = A carry-out from the Most Significant bit of the result occurred
`0` = No carry-out from the Most Significant bit of the result occurred

Note: For $\overline{\text{borrow}}$ the polarity is reversed. A subtraction is executed by adding the two's complement of the second operand. For rotate (`RRF, RLF`) instructions, this bit is loaded with either the high or low order bit of the source register.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'	
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

4.4 Indirect Addressing, INDF and FSR Registers

The INDF register is not a physical register. Addressing the INDF register will cause indirect addressing.

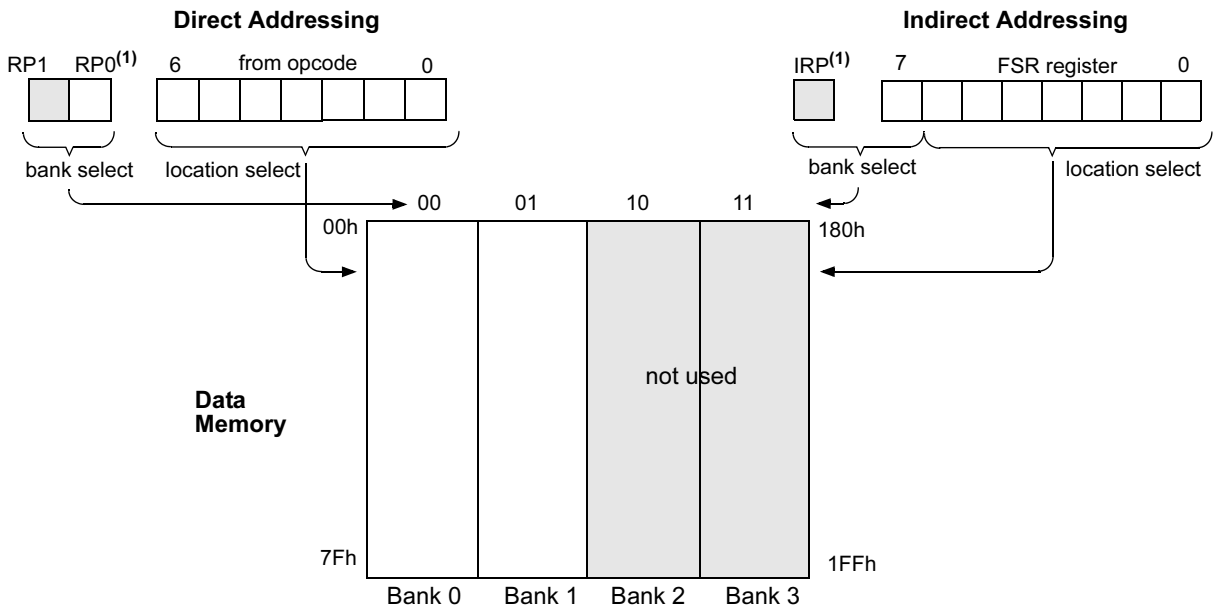
Indirect addressing is possible by using the INDF register. Any instruction using the INDF register actually accesses data pointed to by the File Select Register (FSR). Reading INDF itself indirectly will produce 00h. Writing to the INDF register indirectly results in a no-operation (although STATUS bits may be affected). An effective 9-bit address is obtained by concatenating the 8-bit FSR register and the IRP bit (STATUS<7>), as shown in Figure 4-9. However, IRP is not used in the PIC16C62X.

A simple program to clear RAM location 20h-7Fh using indirect addressing is shown in Example 4-1.

EXAMPLE 4-1: INDIRECT ADDRESSING

```
      movlw   0x20       ;initialize pointer
      movwf   FSR        ;to RAM
NEXT  clrf    INDF       ;clear INDF register
      incf    FSR        ;inc pointer
      btfss   FSR,7      ;all done?
      goto    NEXT       ;no clear next
                         ;yes continue
CONTINUE:
```

FIGURE 4-9: DIRECT/INDIRECT ADDRESSING PIC16C62X



For memory map detail see (Figure 4-4, Figure 4-5, Figure 4-6 and Figure 4-7).

Note 1: The RP1 and IRP bits are reserved; always maintain these bits clear.

5.0 I/O PORTS

The PIC16C62X have two ports, PORTA and PORTB. Some pins for these I/O ports are multiplexed with an alternate function for the peripheral features on the device. In general, when a peripheral is enabled, that pin may not be used as a general purpose I/O pin.

5.1 PORTA and TRISA Registers

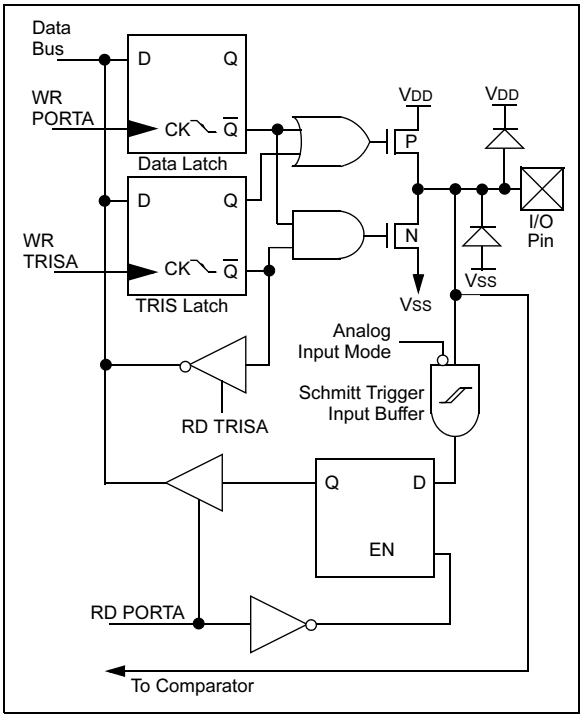
PORTA is a 5-bit wide latch. RA4 is a Schmitt Trigger input and an open drain output. Port RA4 is multiplexed with the T0CKI clock input. All other RA port pins have Schmitt Trigger input levels and full CMOS output drivers. All pins have data direction bits (TRIS registers), which can configure these pins as input or output.

A '1' in the TRISA register puts the corresponding output driver in a Hi-impedance mode. A '0' in the TRISA register puts the contents of the output latch on the selected pin(s).

Reading the PORTA register reads the status of the pins, whereas writing to it will write to the port latch. All write operations are read-modify-write operations. So a write to a port implies that the port pins are first read, then this value is modified and written to the port data latch.

The PORTA pins are multiplexed with comparator and voltage reference functions. The operation of these pins are selected by control bits in the CMCON (comparator control register) register and the VRCON (voltage reference control register) register. When selected as a comparator input, these pins will read as '0's.

FIGURE 5-1: BLOCK DIAGRAM OF RA1:RA0 PINS



Note:	On RESET, the TRISA register is set to all inputs. The digital inputs are disabled and the comparator inputs are forced to ground to reduce excess current consumption.

TRISA controls the direction of the RA pins, even when they are being used as comparator inputs. The user must make sure to keep the pins configured as inputs when using them as comparator inputs.

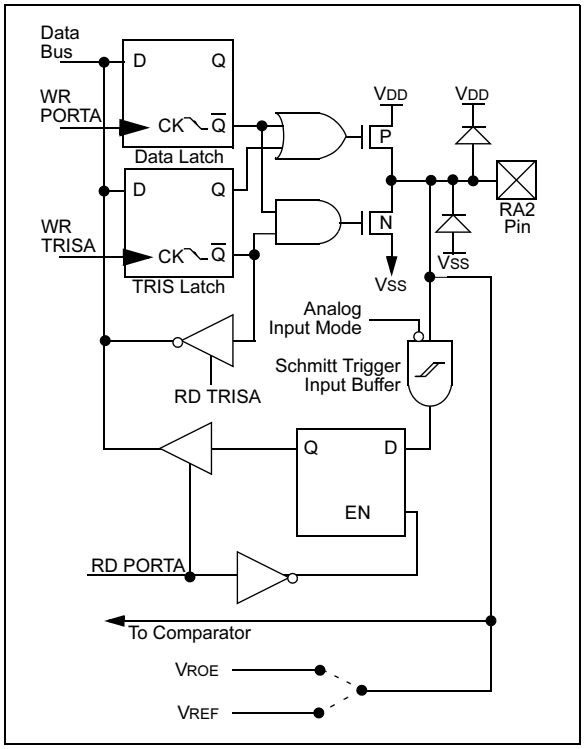
The RA2 pin will also function as the output for the voltage reference. When in this mode, the VREF pin is a very high impedance output and must be buffered prior to any external load. The user must configure TRISA<2> bit as an input and use high impedance loads.

In one of the Comparator modes defined by the CMCON register, pins RA3 and RA4 become outputs of the comparators. The TRISA<4:3> bits must be cleared to enable outputs to use this function.

EXAMPLE 5-1: INITIALIZING PORTA

```
CLRF    PORTA           ;Initialize PORTA by setting
                        ;output data latches
MOVLW   0X07            ;Turn comparators off and
MOVWF   CMCON           ;enable pins for I/O
                        ;functions
BSF     STATUS, RP0     ;Select Bank1
MOVLW   0x1F            ;Value used to initialize
                        ;data direction
MOVWF   TRISA           ;Set RA<4:0> as inputs
                        ;TRISA<7:5> are always
                        ;read as '0'.
```

FIGURE 5-2: BLOCK DIAGRAM OF RA2 PIN



6.0 TIMER0 MODULE

The Timer0 module timer/counter has the following features:

- 8-bit timer/counter
- Readable and writable
- 8-bit software programmable prescaler
- Internal or external clock select
- Interrupt on overflow from FFh to 00h
- Edge select for external clock

Figure 6-1 is a simplified block diagram of the Timer0 module.

Timer mode is selected by clearing the T0CS bit (OPTION<5>). In Timer mode, the TMR0 will increment every instruction cycle (without prescaler). If Timer0 is written, the increment is inhibited for the following two cycles (Figure 6-2 and Figure 6-3). The user can work around this by writing an adjusted value to TMR0.

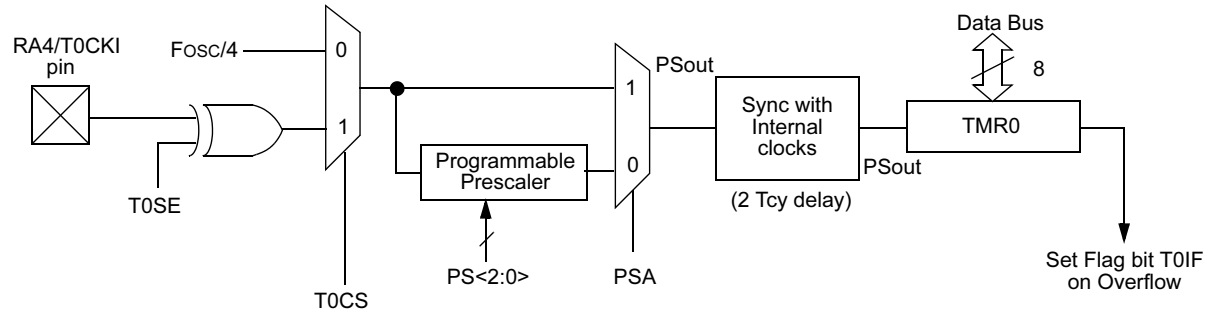
Counter mode is selected by setting the T0CS bit. In this mode, Timer0 will increment either on every rising or falling edge of pin RA4/T0CKI. The incrementing edge is determined by the source edge (T0SE) control bit (OPTION<4>). Clearing the T0SE bit selects the rising edge. Restrictions on the external clock input are discussed in detail in Section 6.2.

The prescaler is shared between the Timer0 module and the Watchdog Timer. The prescaler assignment is controlled in software by the control bit PSA (OPTION<3>). Clearing the PSA bit will assign the prescaler to Timer0. The prescaler is not readable or writable. When the prescaler is assigned to the Timer0 module, prescale value of 1:2, 1:4, ..., 1:256 are selectable. Section 6.3 details the operation of the prescaler.

6.1 TIMER0 Interrupt

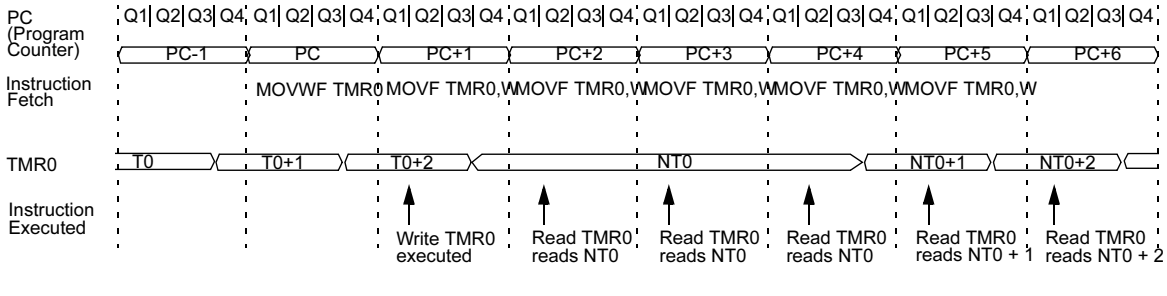
Timer0 interrupt is generated when the TMR0 register timer/counter overflows from FFh to 00h. This overflow sets the T0IF bit. The interrupt can be masked by clearing the T0IE bit (INTCON<5>). The T0IF bit (INTCON<2>) must be cleared in software by the Timer0 module interrupt service routine before re-enabling this interrupt. The Timer0 interrupt cannot wake the processor from SLEEP, since the timer is shut off during SLEEP. See Figure 6-4 for Timer0 interrupt timing.

FIGURE 6-1: TIMER0 BLOCK DIAGRAM



Note 1: Bits T0SE, T0CS, PS2, PS1, PS0 and PSA are located in the OPTION register.
2: The prescaler is shared with Watchdog Timer (Figure 6-6).

FIGURE 6-2: TIMER0 (TMR0) TIMING: INTERNAL CLOCK/NO PRESCALER



6.2 Using Timer0 with External Clock

When an external clock input is used for Timer0, it must meet certain requirements. The external clock requirement is due to internal phase clock (TOSC) synchronization. Also, there is a delay in the actual incrementing of Timer0 after synchronization.

6.2.1 EXTERNAL CLOCK SYNCHRONIZATION

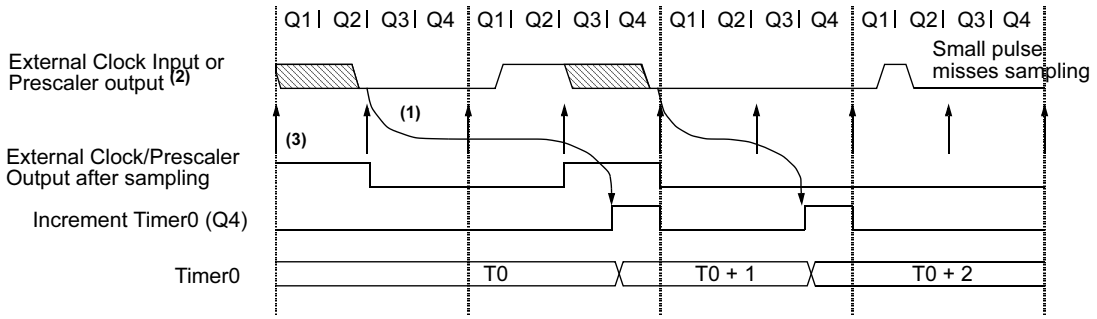
When no prescaler is used, the external clock input is the same as the prescaler output. The synchronization of T0CKI with the internal phase clocks is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks (Figure 6-5). Therefore, it is necessary for T0CKI to be high for at least 2TOSC (and a small RC delay of 20 ns) and low for at least 2TOSC (and a small RC delay of 20 ns). Refer to the electrical specification of the desired device.

When a prescaler is used, the external clock input is divided by the asynchronous ripple-counter type prescaler, so that the prescaler output is symmetrical. For the external clock to meet the sampling requirement, the ripple-counter must be taken into account. Therefore, it is necessary for T0CKI to have a period of at least 4TOSC (and a small RC delay of 40 ns) divided by the prescaler value. The only requirement on T0CKI high and low time is that they do not violate the minimum pulse width requirement of 10 ns. Refer to parameters 40, 41 and 42 in the electrical specification of the desired device.

6.2.2 TIMER0 INCREMENT DELAY

Since the prescaler output is synchronized with the internal clocks, there is a small delay from the time the external clock edge occurs to the time the TMR0 is actually incremented. Figure 6-5 shows the delay from the external clock edge to the timer incrementing.

FIGURE 6-5: TIMER0 TIMING WITH EXTERNAL CLOCK



Note 1: Delay from clock input change to Timer0 increment is 3Tosc to 7Tosc. (Duration of Q = Tosc). Therefore, the error in measuring the interval between two edges on Timer0 input = ±4Tosc max.
2: External clock if no prescaler selected, Prescaler output otherwise.
3: The arrows indicate the points in time where sampling occurs.

7.4 Comparator Response Time

Response time is the minimum time, after selecting a new reference voltage or input source, before the comparator output has a valid level. If the internal reference is changed, the maximum delay of the internal voltage reference must be considered when using the comparator outputs. Otherwise the maximum delay of the comparators should be used (Table 12-2).

7.5 Comparator Outputs

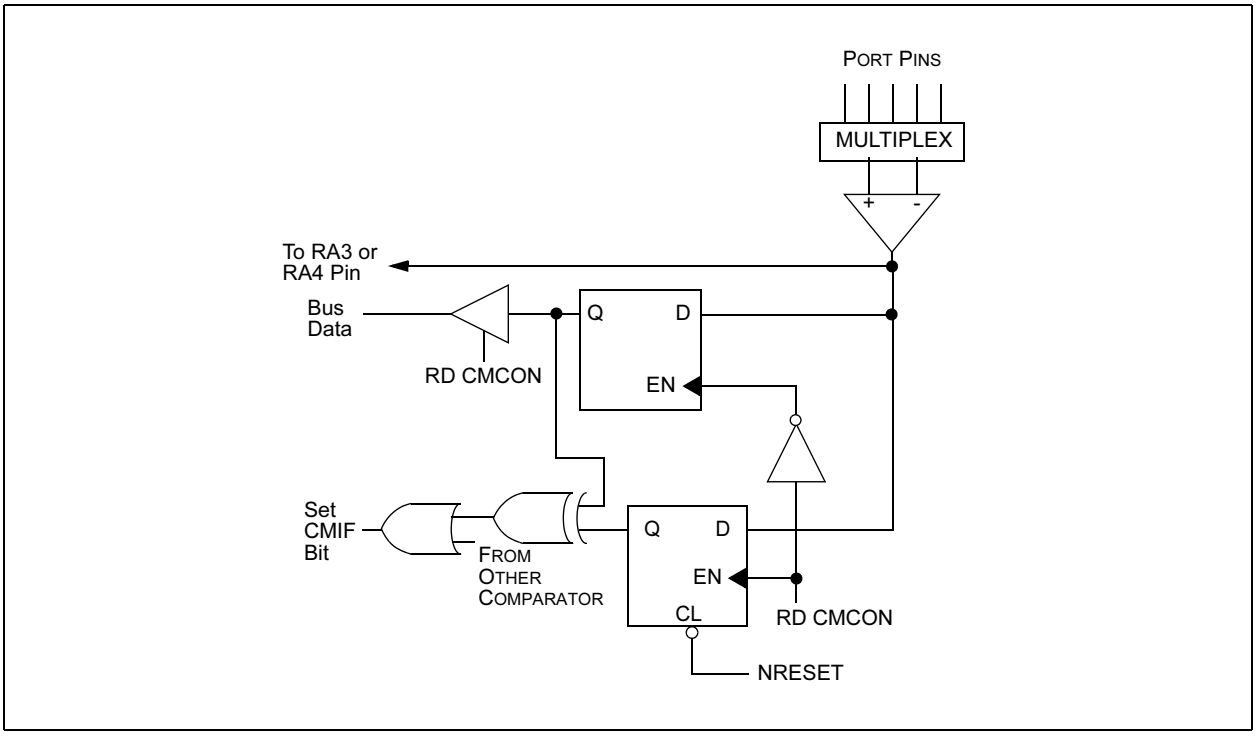
The comparator outputs are read through the CMCON register. These bits are read only. The comparator outputs may also be directly output to the RA3 and RA4 I/O pins. When the CM<2:0> = 110, multiplexors in the output path of the RA3 and RA4 pins will switch and the output of each pin will be the unsynchronized output of the comparator. The uncertainty of each of the comparators is related to the input offset voltage and the response time given in the specifications. Figure 7-3 shows the comparator output block diagram.

The TRISA bits will still function as an output enable/ disable for the RA3 and RA4 pins while in this mode.

Note 1: When reading the PORT register, all pins configured as analog inputs will read as a '0'. Pins configured as digital inputs will convert an analog input according to the Schmitt Trigger input specification.

2: Analog levels on any pin that is defined as a digital input may cause the input buffer to consume more current than is specified.

FIGURE 7-3: COMPARATOR OUTPUT BLOCK DIAGRAM



EXAMPLE 8-1: VOLTAGE REFERENCE CONFIGURATION

```
MOVLW     0x02          ; 4 Inputs Muxed
MOVWF     CMCON         ; to 2 comps.
BSF       STATUS,RP0    ; go to Bank 1
MOVLW     0x0F          ; RA3-RA0 are
MOVWF     TRISA         ; inputs
MOVLW     0xA6          ; enable VREF
MOVWF     VRCON         ; low range
                        ; set VR<3:0>=6
BCF       STATUS,RP0    ; go to Bank 0
CALL      DELAY10       ; 10µs delay
```

8.2 Voltage Reference Accuracy/Error

The full range of VSS to VDD cannot be realized due to the construction of the module. The transistors on the top and bottom of the resistor ladder network (Figure 8-1) keep VREF from approaching VSS or VDD. The voltage reference is VDD derived and therefore, the VREF output changes with fluctuations in VDD. The tested absolute accuracy of the voltage reference can be found in Table 12-2.

8.3 Operation During SLEEP

When the device wakes up from SLEEP through an interrupt or a Watchdog Timer time-out, the contents of the VRCON register are not affected. To minimize current consumption in SLEEP mode, the voltage reference should be disabled.

8.4 Effects of a RESET

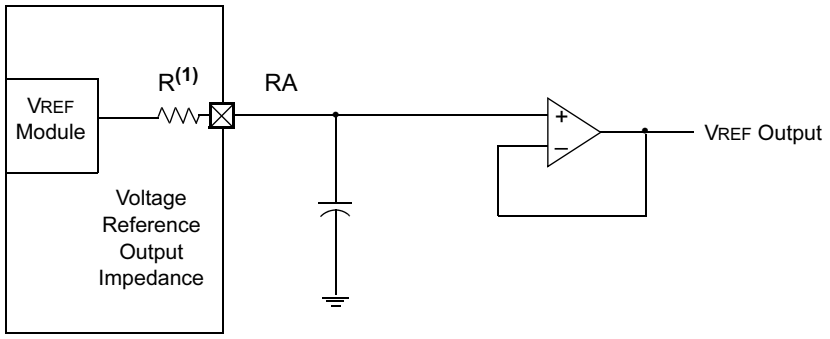
A device RESET disables the voltage reference by clearing bit VREN (VRCON<7>). This reset also disconnects the reference from the RA2 pin by clearing bit VROE (VRCON<6>) and selects the high voltage range by clearing bit VRR (VRCON<5>). The VREF value select bits, VRCON<3:0>, are also cleared.

8.5 Connection Considerations

The voltage reference module operates independently of the comparator module. The output of the reference generator may be connected to the RA2 pin if the TRISA<2> bit is set and the VROE bit, VRCON<6>, is set. Enabling the voltage reference output onto the RA2 pin with an input signal present will increase current consumption. Connecting RA2 as a digital output with VREF enabled will also increase current consumption.

The RA2 pin can be used as a simple D/A output with limited drive capability. Due to the limited drive capability, a buffer must be used in conjunction with the voltage reference output for external connections to VREF. Figure 8-2 shows an example buffering technique.

FIGURE 8-2: VOLTAGE REFERENCE OUTPUT BUFFER EXAMPLE



Note 1: R is dependent upon the Voltage Reference Configuration VRCON<3:0> and VRCON<5>.

TABLE 8-1: REGISTERS ASSOCIATED WITH VOLTAGE REFERENCE

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value On POR	Value On All Other RESETS
9Fh	VRCON	VREN	VROE	VRR	—	VR3	VR2	VR1	VR0	000- 0000	000- 0000
1Fh	CMCON	C2OUT	C1OUT	—	—	CIS	CM2	CM1	CM0	00-- 0000	00-- 0000
85h	TRISA	—	—	—	TRISA4	TRISA3	TRISA2	TRISA1	TRISA0	---1 1111	---1 1111

Note: - = Unimplemented, read as "0"

9.4 Power-on Reset (POR), Power-up Timer (PWRT), Oscillator Start-up Timer (OST) and Brown-out Reset (BOR)

9.4.1 POWER-ON RESET (POR)

The on-chip POR circuit holds the chip in RESET until VDD has reached a high enough level for proper operation. To take advantage of the POR, just tie the MCLR pin through a resistor to VDD. This will eliminate external RC components usually needed to create Power-on Reset. A maximum rise time for VDD is required. See Electrical Specifications for details.

The POR circuit does not produce an internal RESET when VDD declines.

When the device starts normal operation (exits the RESET condition), device operating parameters (voltage, frequency, temperature, etc.) must be met to ensure operation. If these conditions are not met, the device must be held in RESET until the operating conditions are met.

For additional information, refer to Application Note AN607, “Power-up Trouble Shooting”.

9.4.2 POWER-UP TIMER (PWRT)

The Power-up Timer provides a fixed 72 ms (nominal) time-out on power-up only, from POR or Brown-out Reset. The Power-up Timer operates on an internal RC oscillator. The chip is kept in RESET as long as PWRT is active. The PWRT delay allows the VDD to rise to an acceptable level. A configuration bit, PWRTE can disable (if set) or enable (if cleared or programmed) the Power-up Timer. The Power-up Timer should always be enabled when Brown-out Reset is enabled.

The Power-up Time delay will vary from chip-to-chip and due to VDD, temperature and process variation. See DC parameters for details.

9.4.3 OSCILLATOR START-UP TIMER (OST)

The Oscillator Start-Up Timer (OST) provides a 1024 oscillator cycle (from OSC1 input) delay after the PWRT delay is over. This ensures that the crystal oscillator or resonator has started and stabilized.

The OST time-out is invoked only for XT, LP and HS modes and only on Power-on Reset or wake-up from SLEEP.

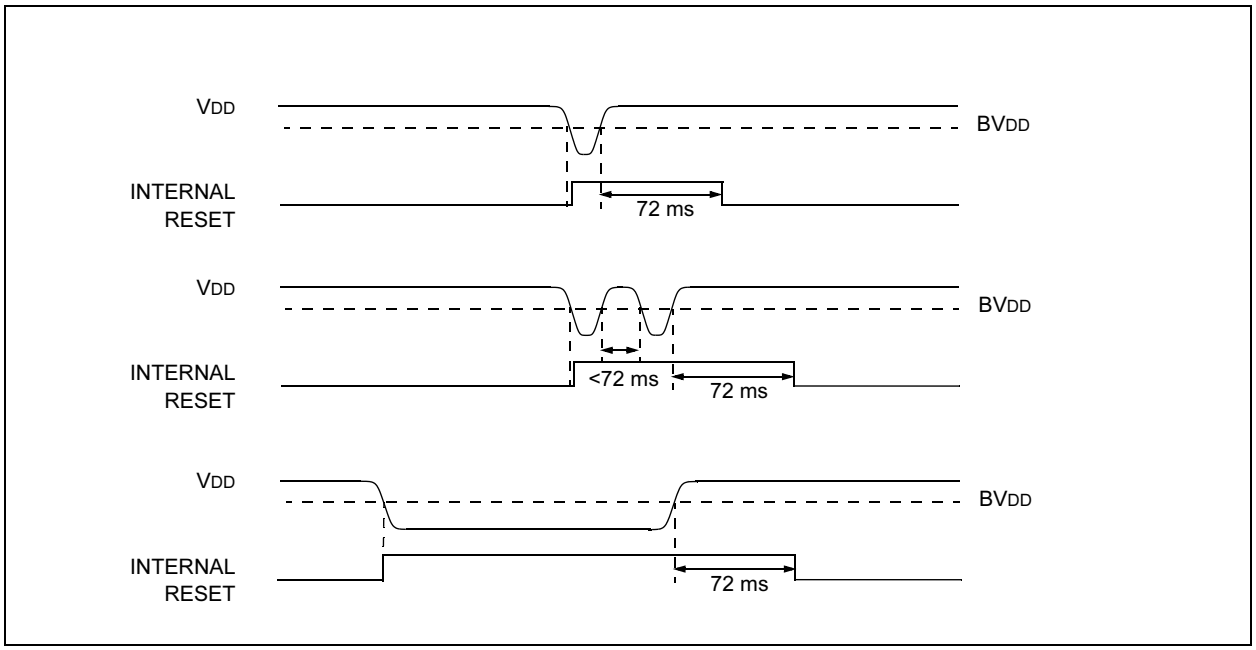
9.4.4 BROWN-OUT RESET (BOR)

The PIC16C62X members have on-chip Brown-out Reset circuitry. A configuration bit, BODEN, can disable (if clear/programmed) or enable (if set) the Brown-out Reset circuitry. If VDD falls below 4.0V refer to VBOR parameter D005 (VBOR) for greater than parameter (TBOR) in Table 12-5. The brown-out situation will RESET the chip. A RESET won’t occur if VDD falls below 4.0V for less than parameter (TBOR).

On any RESET (Power-on, Brown-out, Watchdog, etc.) the chip will remain in RESET until VDD rises above BVDD. The Power-up Timer will now be invoked and will keep the chip in RESET an additional 72 ms.

If VDD drops below BVDD while the Power-up Timer is running, the chip will go back into a Brown-out Reset and the Power-up Timer will be re-initialized. Once VDD rises above BVDD, the Power-Up Timer will execute a 72 ms RESET. The Power-up Timer should always be enabled when Brown-out Reset is enabled. Figure 9-7 shows typical Brown-out situations.

FIGURE 9-7: BROWN-OUT SITUATIONS



9.8 Power-Down Mode (SLEEP)

The Power-down mode is entered by executing a SLEEP instruction.

If enabled, the Watchdog Timer will be cleared but keeps running, the $\overline{PD}$ bit in the STATUS register is cleared, the $\overline{TO}$ bit is set, and the oscillator driver is turned off. The I/O ports maintain the status they had, before SLEEP was executed (driving high, low, or hi-impedance).

For lowest current consumption in this mode, all I/O pins should be either at VDD or VSS with no external circuitry drawing current from the I/O pin and the comparators and VREF should be disabled. I/O pins that are hi-impedance inputs should be pulled high or low externally to avoid switching currents caused by floating inputs. The T0CKI input should also be at VDD or VSS for lowest current consumption. The contribution from on chip pull-ups on PORTB should be considered.

The $\overline{MCLR}$ pin must be at a logic high level (VIHMC).

Note:	It should be noted that a RESET generated by a WDT time-out does not drive $\overline{MCLR}$ pin low.

9.8.1 WAKE-UP FROM SLEEP

The device can wake-up from SLEEP through one of the following events:

1. External RESET input on $\overline{MCLR}$ pin
2. Watchdog Timer Wake-up (if WDT was enabled)
3. Interrupt from RB0/INT pin, RB Port change, or the Peripheral Interrupt (Comparator).

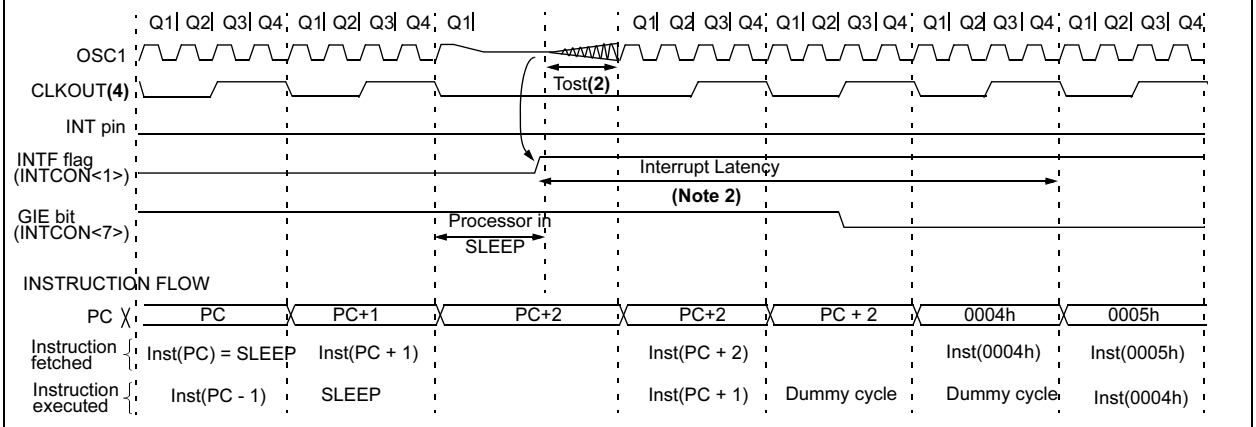
The first event will cause a device RESET. The two latter events are considered a continuation of program execution. The $\overline{TO}$ and $\overline{PD}$ bits in the STATUS register can be used to determine the cause of device RESET. $\overline{PD}$ bit, which is set on power-up, is cleared when SLEEP is invoked. $\overline{TO}$ bit is cleared if WDT wake-up occurred.

When the SLEEP instruction is being executed, the next instruction (PC + 1) is pre-fetched. For the device to wake-up through an interrupt event, the corresponding interrupt enable bit must be set (enabled). Wake-up is regardless of the state of the GIE bit. If the GIE bit is clear (disabled), the device continues execution at the instruction after the SLEEP instruction. If the GIE bit is set (enabled), the device executes the instruction after the SLEEP instruction and then branches to the interrupt address (0004h). In cases where the execution of the instruction following SLEEP is not desirable, the user should have an NOP after the SLEEP instruction.

Note:	If the global interrupts are disabled (GIE is cleared), but any interrupt source has both its interrupt enable bit and the corresponding interrupt flag bits set, the device will immediately wake-up from SLEEP. The SLEEP instruction is completely executed.

The WDT is cleared when the device wakes up from SLEEP, regardless of the source of wake-up.

FIGURE 9-18: WAKE-UP FROM SLEEP THROUGH INTERRUPT



Note 1: XT, HS or LP Oscillator mode assumed.
2: TOST = 1024TOSC (drawing not to scale) This delay will not be there for RC Osc mode.
3: GIE = '1' assumed. In this case, after wake-up, the processor jumps to the interrupt routine. If GIE = '0', execution will continue in-line.
4: CLKOUT is not available in these Osc modes, but shown here for timing reference.

TABLE 10-2: PIC16C62X INSTRUCTION SET

Mnemonic, Operands		Description	Cycles	14-Bit Opcode MSb			LSb	Status Affected	Notes
BYTE-ORIENTED FILE REGISTER OPERATIONS									
ADDWF	**f, d**	Add W and f	1	00	0111	dfff	ffff	C,DC,Z	1,2
ANDWF	**f, d**	AND W with f	1	00	0101	dfff	ffff	Z	1,2
CLRF	**f**	Clear f	1	00	0001	lfff	ffff	Z	2
CLRW	**-**	Clear W	1	00	0001	0000	0011	Z	
COMF	**f, d**	Complement f	1	00	1001	dfff	ffff	Z	1,2
DECF	**f, d**	Decrement f	1	00	0011	dfff	ffff	Z	1,2
DECFSZ	**f, d**	Decrement f, Skip if 0	1(2)	00	1011	dfff	ffff		1,2,3
INCF	**f, d**	Increment f	1	00	1010	dfff	ffff	Z	1,2
INCFSZ	**f, d**	Increment f, Skip if 0	1(2)	00	1111	dfff	ffff		1,2,3
IORWF	**f, d**	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	1,2
MOVF	**f, d**	Move f	1	00	1000	dfff	ffff	Z	1,2
MOVWF	**f**	Move W to f	1	00	0000	lfff	ffff		
NOP	**-**	No Operation	1	00	0000	0xx0	0000		
RLF	**f, d**	Rotate Left f through Carry	1	00	1101	dfff	ffff	C	1,2
RRF	**f, d**	Rotate Right f through Carry	1	00	1100	dfff	ffff	C	1,2
SUBWF	**f, d**	Subtract W from f	1	00	0010	dfff	ffff	C,DC,Z	1,2
SWAPF	**f, d**	Swap nibbles in f	1	00	1110	dfff	ffff		1,2
XORWF	**f, d**	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	1,2
BIT-ORIENTED FILE REGISTER OPERATIONS									
BCF	**f, b**	Bit Clear f	1	01	00bb	bfff	ffff		1,2
BSF	**f, b**	Bit Set f	1	01	01bb	bfff	ffff		1,2
BTFSC	**f, b**	Bit Test f, Skip if Clear	1 (2)	01	10bb	bfff	ffff		3
BTFSS	**f, b**	Bit Test f, Skip if Set	1 (2)	01	11bb	bfff	ffff		3
LITERAL AND CONTROL OPERATIONS									
ADDLW	**k**	Add literal and W	1	11	111x	kkkk	kkkk	C,DC,Z	
ANDLW	**k**	AND literal with W	1	11	1001	kkkk	kkkk	Z	
CALL	**k**	Call subroutine	2	10	0kkk	kkkk	kkkk		
CLRWDT	**-**	Clear Watchdog Timer	1	00	0000	0110	0100	$\overline{TO}$,$\overline{PD}$	
GOTO	**k**	Go to address	2	10	1kkk	kkkk	kkkk		
IORLW	**k**	Inclusive OR literal with W	1	11	1000	kkkk	kkkk	Z	
MOVLW	**k**	Move literal to W	1	11	00xx	kkkk	kkkk		
RETFIE	**-**	Return from interrupt	2	00	0000	0000	1001		
RETLW	**k**	Return with literal in W	2	11	01xx	kkkk	kkkk		
RETURN	**-**	Return from Subroutine	2	00	0000	0000	1000		
SLEEP	**-**	Go into Standby mode	1	00	0000	0110	0011	$\overline{TO}$,$\overline{PD}$	
SUBLW	**k**	Subtract W from literal	1	11	110x	kkkk	kkkk	C,DC,Z	
XORLW	**k**	Exclusive OR literal with W	1	11	1010	kkkk	kkkk	Z	

Note 1: When an I/O register is modified as a function of itself (e.g., `MOVF PORTB, 1`), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and, where applicable, d = 1), the prescaler will be cleared if assigned to the Timer0 Module.

3: If Program Counter (PC) is modified or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a `NOP`.

INCFSZ **Increment f, Skip if 0**

Syntax: [*label*] INCFSZ f,d

Operands: $0 \le f \le 127$
$d \in [0,1]$

Operation: (f) + 1 → (dest), skip if result = 0

Status Affected: None

Encoding:

00	1111	dfff	ffff

Description: The contents of register 'f' are incremented. If 'd' is 0 the result is placed in the W register. If 'd' is 1, the result is placed back in register 'f'.
If the result is 0, the next instruction, which is already fetched, is discarded. A NOP is executed instead making it a two-cycle instruction.

Words: 1

Cycles: 1(2)

Example
```
HERE     INCFSZ     CNT,  1
         GOTO       LOOP
CONTINUE •
         •
         •
```
Before Instruction
PC = address HERE
After Instruction
CNT = CNT + 1
if CNT= 0,
PC = address CONTINUE
if CNT≠ 0,
PC = address HERE +1

IORLW **Inclusive OR Literal with W**

Syntax: [*label*] IORLW k

Operands: $0 \le k \le 255$

Operation: (W) .OR. k → (W)

Status Affected: Z

Encoding:

11	1000	kkkk	kkkk

Description: The contents of the W register is OR'ed with the eight bit literal 'k'. The result is placed in the W register.

Words: 1

Cycles: 1

Example
```
IORLW   0x35
```
Before Instruction
W = 0x9A
After Instruction
W = 0xBF
Z = 1

IORWF **Inclusive OR W with f**

Syntax: [*label*] IORWF f,d

Operands: $0 \le f \le 127$
$d \in [0,1]$

Operation: (W) .OR. (f) → (dest)

Status Affected: $\overline{Z}$

Encoding:

00	0100	dfff	ffff

Description: Inclusive OR the W register with register 'f'. If 'd' is 0 the result is placed in the W register. If 'd' is 1 the result is placed back in register 'f'.

Words: 1

Cycles: 1

Example
```
IORWF          RESULT, 0
```
Before Instruction
RESULT = 0x13
W = 0x91
After Instruction
RESULT = 0x13
W = 0x93
Z = 1

MOVLW **Move Literal to W**

Syntax: [*label*] MOVLW k

Operands: $0 \le k \le 255$

Operation: k → (W)

Status Affected: None

Encoding:

11	00xx	kkkk	kkkk

Description: The eight bit literal 'k' is loaded into W register. The don't cares will assemble as 0's.

Words: 1

Cycles: 1

Example
```
MOVLW   0x5A
```
After Instruction
W = 0x5A

11.9 MPLAB ICE 2000 High Performance Universal In-Circuit Emulator

The MPLAB ICE 2000 universal in-circuit emulator is intended to provide the product development engineer with a complete microcontroller design tool set for PICmicro microcontrollers. Software control of the MPLAB ICE 2000 in-circuit emulator is advanced by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICE 2000 is a full-featured emulator system with enhanced trace, trigger and data monitoring features. Interchangeable processor modules allow the system to be easily reconfigured for emulation of different processors. The universal architecture of the MPLAB ICE in-circuit emulator allows expansion to support new PICmicro microcontrollers.

The MPLAB ICE 2000 in-circuit emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft® Windows 32-bit operating system were chosen to best make these features available in a simple, unified application.

11.10 MPLAB ICE 4000 High Performance Universal In-Circuit Emulator

The MPLAB ICE 4000 universal in-circuit emulator is intended to provide the product development engineer with a complete microcontroller design tool set for high-end PICmicro microcontrollers. Software control of the MPLAB ICE in-circuit emulator is provided by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICD 4000 is a premium emulator system, providing the features of MPLAB ICE 2000, but with increased emulation memory and high speed performance for dsPIC30F and PIC18XXXX devices. Its advanced emulator features include complex triggering and timing, up to 2 Mb of emulation memory, and the ability to view variables in real-time.

The MPLAB ICE 4000 in-circuit emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft Windows 32-bit operating system were chosen to best make these features available in a simple, unified application.

11.11 MPLAB ICD 2 In-Circuit Debugger

Microchip's In-Circuit Debugger, MPLAB ICD 2, is a powerful, low cost, run-time development tool, connecting to the host PC via an RS-232 or high speed USB interface. This tool is based on the FLASH PICmicro MCUs and can be used to develop for these and other PICmicro microcontrollers. The MPLAB ICD 2 utilizes the in-circuit debugging capability built into the FLASH devices. This feature, along with Microchip's In-Circuit Serial Programming™ (ICSP™) protocol, offers cost effective in-circuit FLASH debugging from the graphical user interface of the MPLAB Integrated Development Environment. This enables a designer to develop and debug source code by setting breakpoints, single-stepping and watching variables, CPU status and peripheral registers. Running at full speed enables testing hardware and applications in real-time. MPLAB ICD 2 also serves as a development programmer for selected PICmicro devices.

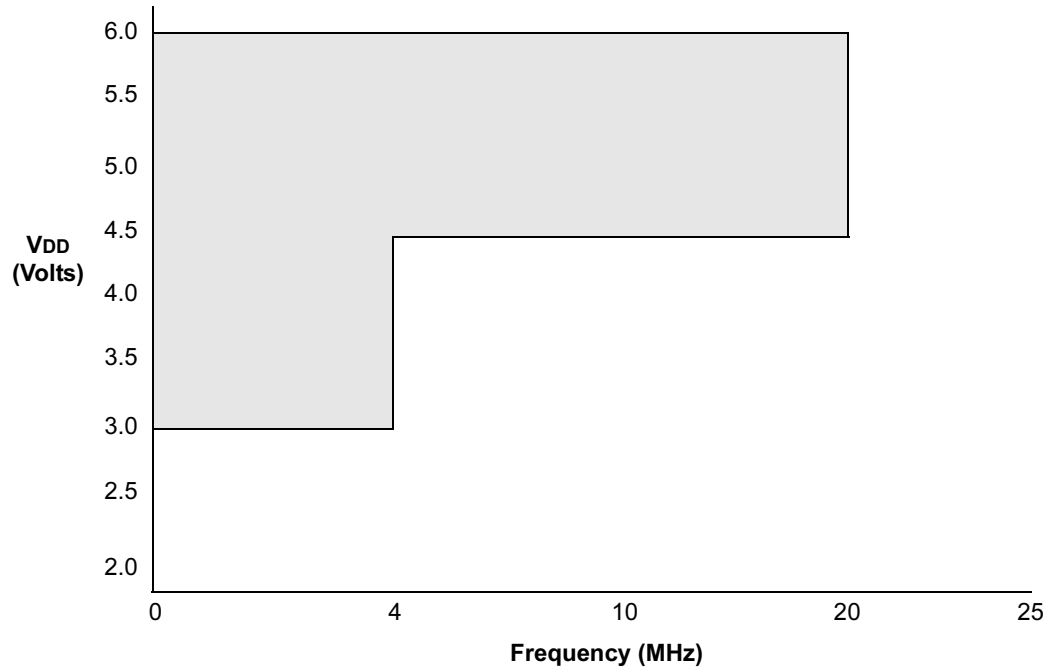
11.12 PRO MATE II Universal Device Programmer

The PRO MATE II is a universal, CE compliant device programmer with programmable voltage verification at V_{DDMIN} and V_{DDMAX} for maximum reliability. It features an LCD display for instructions and error messages and a modular detachable socket assembly to support various package types. In Stand-Alone mode, the PRO MATE II device programmer can read, verify, and program PICmicro devices without a PC connection. It can also set code protection in this mode.

11.13 PICSTART Plus Development Programmer

The PICSTART Plus development programmer is an easy-to-use, low cost, prototype programmer. It connects to the PC via a COM (RS-232) port. MPLAB Integrated Development Environment software makes using the programmer simple and efficient. The PICSTART Plus development programmer supports most PICmicro devices up to 40 pins. Larger pin count devices, such as the PIC16C92X and PIC17C76X, may be supported with an adapter socket. The PICSTART Plus development programmer is CE compliant.

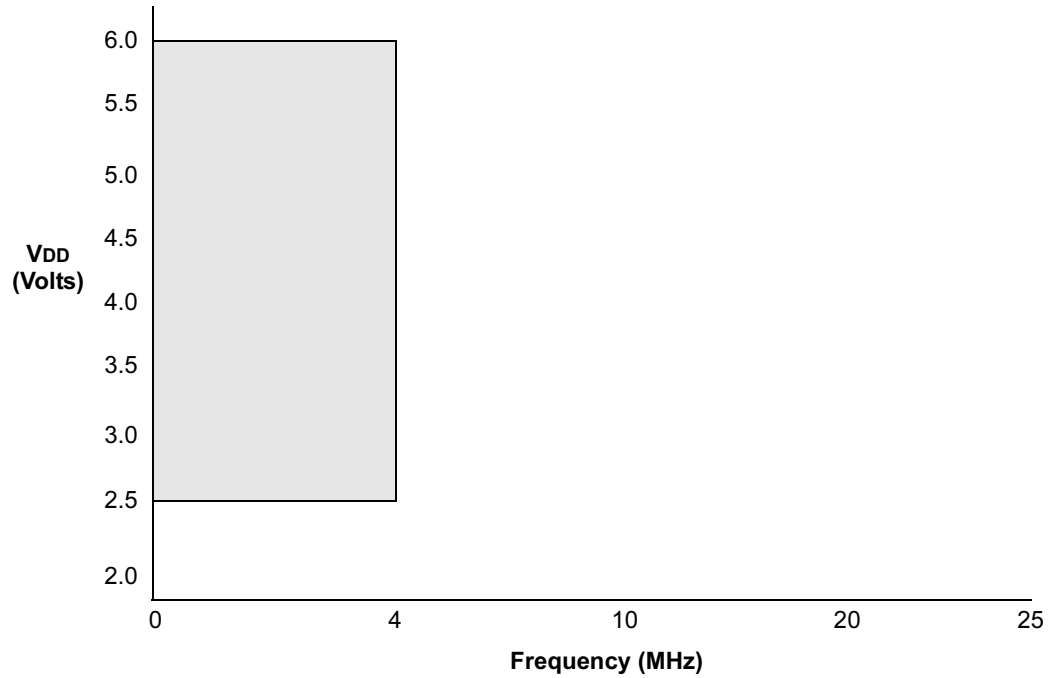
FIGURE 12-1: PIC16C62X VOLTAGE-FREQUENCY GRAPH, -40°C ≤ TA ≤ +125°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

2: The maximum rated speed of the part limits the permissible combinations of voltage and frequency. Please reference the Product Identification System section for the maximum rated speed of the parts.

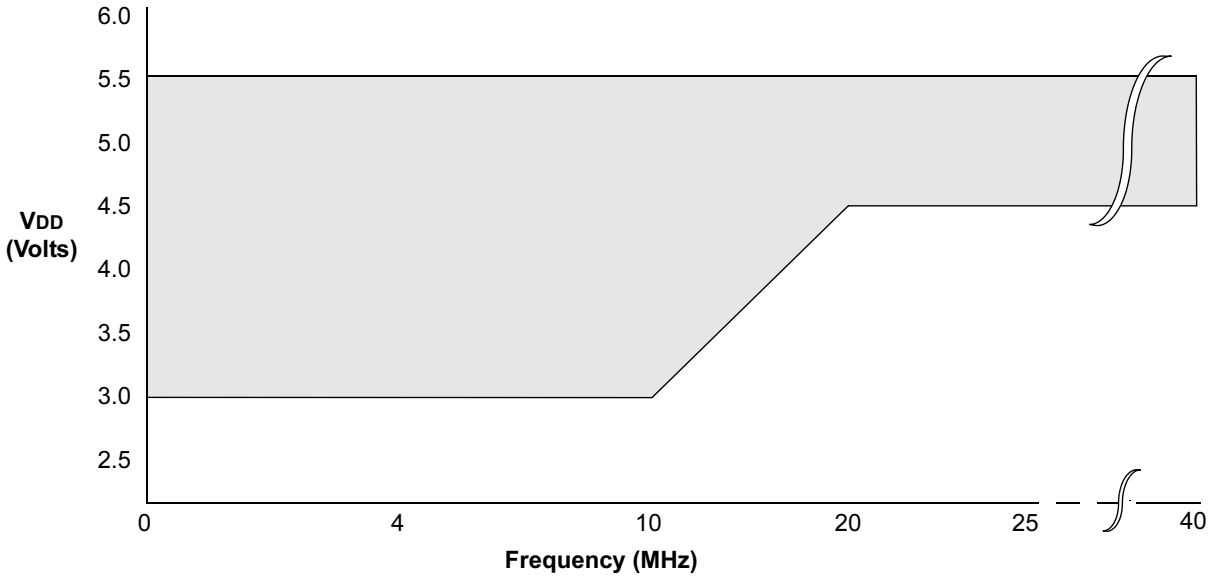
FIGURE 12-2: PIC16LC62X VOLTAGE-FREQUENCY GRAPH, -40°C ≤ TA ≤ +125°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

2: The maximum rated speed of the part limits the permissible combinations of voltage and frequency. Please reference the Product Identification System section for the maximum rated speed of the parts.

FIGURE 12-10: PIC16C620A/C621A/C622A/CR620A - 40 VOLTAGE-FREQUENCY GRAPH, 0°C ≤ TA ≤ +70°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

2: The maximum rated speed of the part limits the permissible combinations of voltage and frequency. Please reference the Product Identification System section for the maximum rated speed of the parts.

3: Operation between 20 to 40 MHz requires the following:

- VDD between 4.5V. and 5.5V
- OSC1 externally driven
- OSC2 not connected
- HS mode
- Commercial temperatures

Devices qualified for 40 MHz operation have -40 designation (ex: PIC16C620A-40/P).

12.2 DC Characteristics: PIC16C62XA-04 (Commercial, Industrial, Extended) PIC16C62XA-20 (Commercial, Industrial, Extended) PIC16LC62XA-04 (Commercial, Industrial, Extended) (CONT.)

PIC16C62XA			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial and -40°C ≤ TA ≤ +125°C for extended				
PIC16LC62XA			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial and -40°C ≤ TA ≤ +125°C for extended				
Param. No.	**Sym**	**Characteristic**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
D010	IDD	Supply Current(2, 4)	—	1.2	2.0	mA	FOSC = 4 MHz, VDD = 5.5V, WDT disabled, XT mode, **(Note 4)***
			—	0.4	1.2	mA	FOSC = 4 MHz, VDD = 3.0V, WDT disabled, XT mode, **(Note 4)***
			—	1.0	2.0	mA	FOSC = 10 MHz, VDD = 3.0V, WDT disabled, HS mode, **(Note 6)**
			—	4.0	6.0	mA	FOSC = 20 MHz, VDD = 4.5V, WDT disabled, HS mode
			—	4.0	7.0	mA	FOSC = 20 MHz, VDD = 5.5V, WDT disabled*, HS mode
			—	35	70	µA	FOSC = 32 kHz, VDD = 3.0V, WDT disabled, LP mode
D010	IDD	Supply Current(2)	—	1.2	2.0	mA	FOSC = 4 MHz, VDD = 5.5V, WDT disabled, XT mode, **(Note 4)***
			—	—	1.1	mA	FOSC = 4 MHz, VDD = 2.5V, WDT disabled, XT mode, **(Note 4)**
			—	35	70	µA	FOSC = 32 kHz, VDD = 2.5V, WDT disabled, LP mode
D020	IPD	Power-down Current(3)	—	—	2.2	µA	VDD = 3.0V
			—	—	5.0	µA	VDD = 4.5V*
			—	—	9.0	µA	VDD = 5.5V
			—	—	15	µA	VDD = 5.5V Extended Temp.
D020	IPD	Power-down Current(3)	—	—	2.0	µA	VDD = 2.5V
			—	—	2.2	µA	VDD = 3.0V*
			—	—	9.0	µA	VDD = 5.5V
			—	—	15	µA	VDD = 5.5V Extended Temp.

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.
The test conditions for all IDD measurements in Active Operation mode are:
OSC1 = external square wave, from rail to rail; all I/O pins tri-stated, pulled to VDD,
$\overline{MCLR}$ = VDD; WDT enabled/disabled as specified.

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD or VSS.

4: For RC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula: Ir = VDD/2REXT (mA) with REXT in kΩ.

5: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.

6: Commercial temperature range only.

12.3 DC CHARACTERISTICS: PIC16CR62XA-04 (Commercial, Industrial, Extended) PIC16CR62XA-20 (Commercial, Industrial, Extended) PIC16LCR62XA-04 (Commercial, Industrial, Extended) (CONT.)

PIC16CR62XA-04 **PIC16CR62XA-20**			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial and -40°C ≤ TA ≤ +125°C for extended				
PIC16LCR62XA-04			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial and -40°C ≤ TA ≤ +125°C for extended				
Param. No.	**Sym**	**Characteristic**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
D020	IPD	Power-down Current(3)	—	200	950	nA	VDD = 3.0V
			—	0.400	1.8	µA	VDD = 4.5V*
			—	0.600	2.2	µA	VDD = 5.5V
			—	5.0	9.0	µA	VDD = 5.5V Extended Temp.
D020	IPD	Power-down Current(3)	—	200	850	nA	VDD = 2.5V
			—	200	950	nA	VDD = 3.0V*
			—	0.600	2.2	µA	VDD = 5.5V
			—	5.0	9.0	µA	VDD = 5.5V Extended
D022	ΔIWDT	WDT Current(5)	—	6.0	10	µA	VDD=4.0V
					12	µA	(125°C)
D022A	ΔIBOR	Brown-out Reset Current(5)	—	75	125	µA	BOD enabled, VDD = 5.0V
D023	ΔICOMP	Comparator Current for each Comparator(5)	—	30	60	µA	VDD = 4.0V
D023A	ΔIVREF	VREF Current(5)	—	80	135	µA	VDD = 4.0V
D022	ΔIWDT	WDT Current(5)	—	6.0	10	µA	VDD=4.0V
					12	µA	(125°C)
D022A	ΔIBOR	Brown-out Reset Current(5)	—	75	125	µA	BOD enabled, VDD = 5.0V
D023	ΔICOMP	Comparator Current for each Comparator(5)	—	30	60	µA	VDD = 4.0V
D023A	ΔIVREF	VREF Current(5)	—	80	135	µA	VDD = 4.0V
1A	FOSC	LP Oscillator Operating Frequency	0	—	200	kHz	All temperatures
		RC Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		XT Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		HS Oscillator Operating Frequency	0	—	20	MHz	All temperatures
1A	FOSC	LP Oscillator Operating Frequency	0	—	200	kHz	All temperatures
		RC Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		XT Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		HS Oscillator Operating Frequency	0	—	20	MHz	All temperatures

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.
The test conditions for all IDD measurements in Active Operation mode are:
OSC1 = external square wave, from rail to rail; all I/O pins tri-stated, pulled to VDD,
MCLR = VDD; WDT enabled/disabled as specified.

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD or VSS.

4: For RC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula: Ir = VDD/2REXT (mA) with REXT in kΩ.

5: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.

6: Commercial temperature range only.

12.5 DC CHARACTERISTICS: PIC16C620A/C621A/C622A-40(7) (Commercial) PIC16CR620A-40(7) (Commercial)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature 0°C ≤ TA ≤ +70°C for commercial				
Param No.	**Sym**	**Characteristic**	**Min**	**Typ†**	**Max**	**Unit**	**Conditions**
	VIL	**Input Low Voltage**					
		I/O ports					
D030		with TTL buffer	VSS	—	0.8V 0.15VDD	V	VDD = 4.5V to 5.5V, otherwise
D031		with Schmitt Trigger input	VSS		0.2VDD	V	
D032		$\overline{MCLR}$, RA4/T0CKI, OSC1 (in RC mode)	VSS	—	0.2VDD	V	**(Note 1)**
D033		OSC1 (in XT and HS)	VSS	—	0.3VDD	V	
		OSC1 (in LP)	VSS	—	0.6VDD - 1.0	V	
	VIH	**Input High Voltage**					
		I/O ports					
D040		with TTL buffer	2.0V 0.25 VDD + 0.8	—	VDD VDD	V	VDD = 4.5V to 5.5V, otherwise
D041		with Schmitt Trigger input	0.8 VDD		VDD		
D042		$\overline{MCLR}$ RA4/T0CKI	0.8 VDD	—	VDD	V	
D043		OSC1 (XT, HS and LP)	0.7 VDD	—	VDD	V	
D043A		OSC1 (in RC mode)	0.9 VDD				**(Note 1)**
D070	IPURB	**PORTB Weak Pull-up Current**	50	200	400	µA	VDD = 5.0V, VPIN = VSS
	IIL	**Input Leakage Current(2, 3)**					
		I/O ports (except PORTA)			±1.0	µA	VSS ≤ VPIN ≤ VDD, pin at hi-impedance
D060		PORTA	—	—	±0.5	µA	Vss ≤ VPIN ≤ VDD, pin at hi-impedance
D061		RA4/T0CKI	—	—	±1.0	µA	Vss ≤ VPIN ≤ VDD
D063		OSC1, $\overline{MCLR}$	—	—	±5.0	µA	Vss ≤ VPIN ≤ VDD, XT, HS and LP OSC configuration
	VOL	**Output Low Voltage**					
D080		I/O ports	—	—	0.6	V	IOL = 8.5 mA, VDD = 4.5V, -40° to +85°C
			—	—	0.6	V	IOL = 7.0 mA, VDD = 4.5V, +125°C
D083		OSC2/CLKOUT (RC only)	—	—	0.6	V	IOL = 1.6 mA, VDD = 4.5V, -40° to +85°C
			—	—	0.6	V	IOL = 1.2 mA, VDD = 4.5V, +125°C
	VOH	**Output High Voltage(3)**					
D090		I/O ports (except RA4)	VDD-0.7	—	—	V	IOH = -3.0 mA, VDD = 4.5V, -40° to +85°C
			VDD-0.7	—	—	V	IOH = -2.5 mA, VDD = 4.5V, +125°C
D092		OSC2/CLKOUT (RC only)	VDD-0.7	—	—	V	IOH = -1.3 mA, VDD = 4.5V, -40° to +85°C
			VDD-0.7	—	—	V	IOH = -1.0 mA, VDD = 4.5V, +125°C
*D150	VOD	**Open Drain High Voltage**			8.5	V	RA4 pin
		Capacitive Loading Specs on Output Pins					
D100	COSC2	OSC2 pin			15	pF	In XT, HS and LP modes when external clock used to drive OSC1.
D101	CIO	All I/O pins/OSC2 (in RC mode)			50	pF	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C, unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in SLEEP mode without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.
The test conditions for all IDD measurements in Active Operation mode are:
OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD, $\overline{MCLR}$ = VDD; WDT enabled/disabled as specified.

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD or VSS.

4: For RC OSC configuration, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/2REXT (mA) with REXT in kΩ.

5: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.

6: Commercial temperature range only.

7: See Section 12.1 and Section 12.3 for 16C62X and 16CR62X devices for operation between 20 MHz and 40 MHz for valid modified characteristics.

FIGURE 13-5: IOH vs. VOH, VDD = 3.0V)

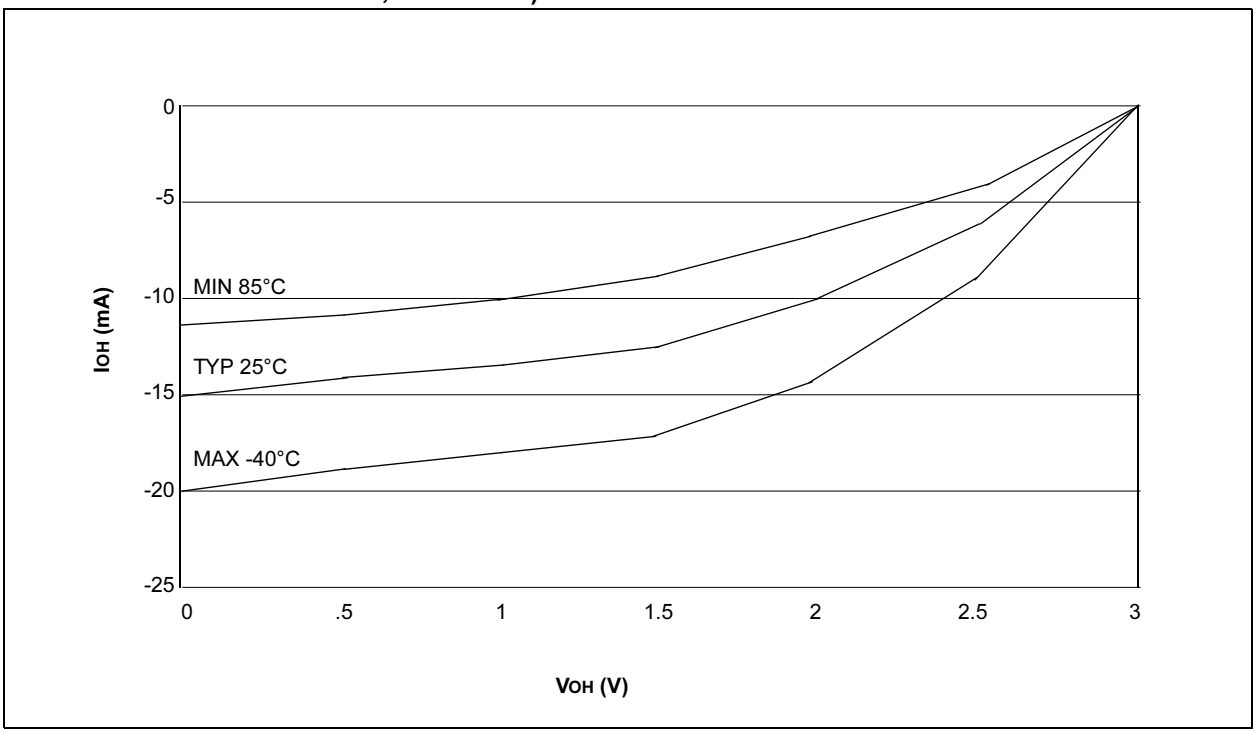


FIGURE 13-6: IOl vs. VOL, VDD = 5.5V)