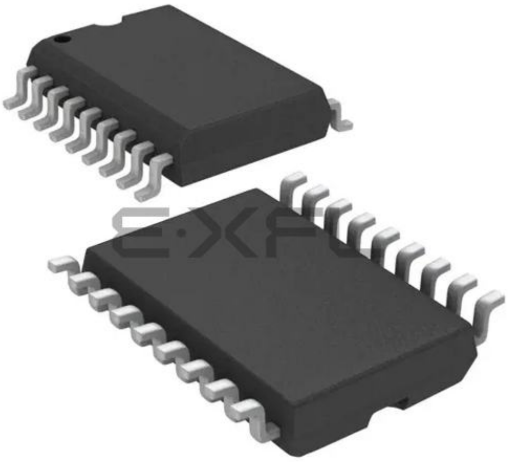
Microchip Technology - PIC16C622T-04I/SO Datasheet



Details	
Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	4MHz
Connectivity	-
Peripherals	Brown-out Detect/Reset, POR, WDT
Number of I/O	13
Program Memory Size	3.5KB (2K x 14)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 6V
Data Converters	-
Oscillator Type	External
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	18-SOIC (0.295", 7.50mm Width)
Supplier Device Package	18-SOIC
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic16c622t-04i-so

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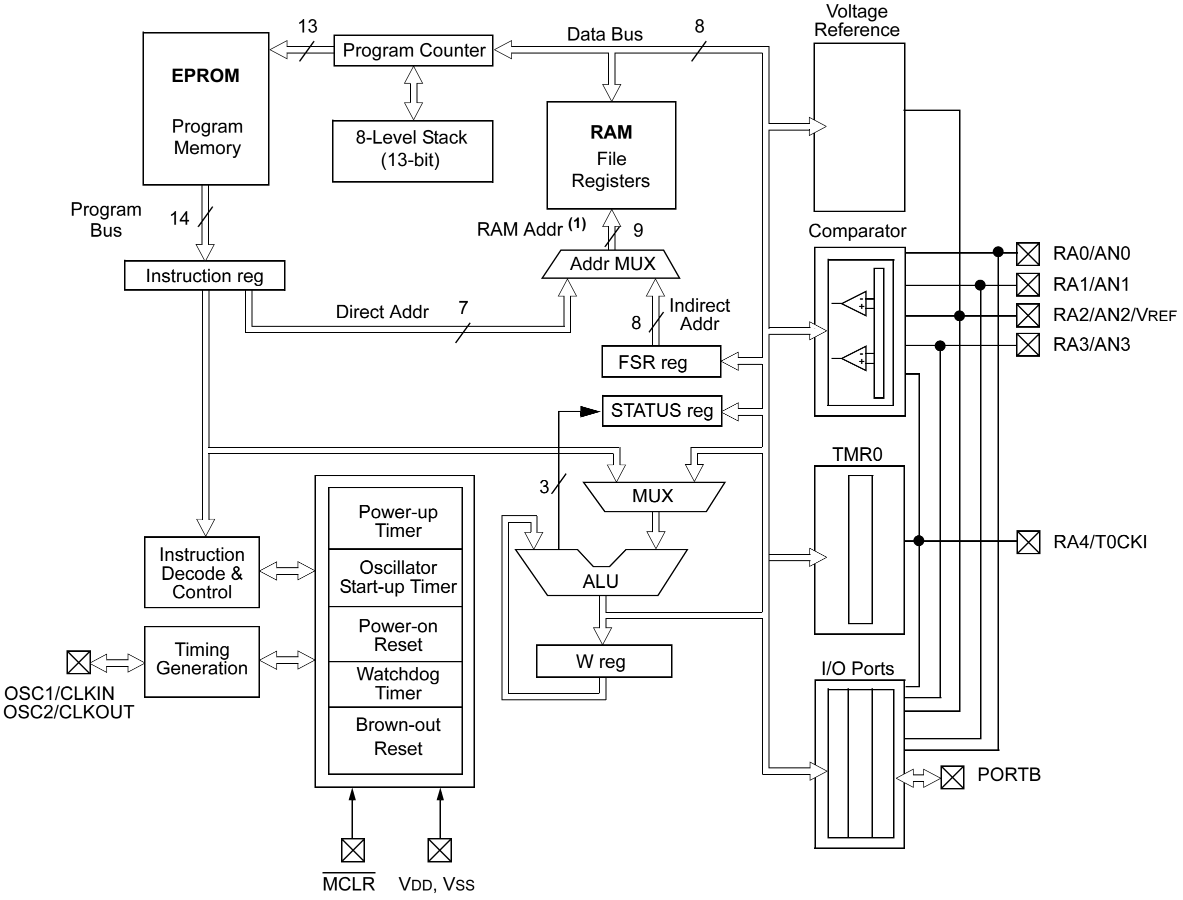
FIGURE 3-1: BLOCK DIAGRAM

Device	Program Memory	Data Memory (RAM)
PIC16C620	512 x 14	80 x 8
PIC16C620A	512 x 14	96 x 8
PIC16CR620A	512 x 14	96 x 8
PIC16C621	1K x 14	80 x 8
PIC16C621A	1K x 14	96 x 8
PIC16C622	2K x 14	128 x 8
PIC16C622A	2K x 14	128 x 8

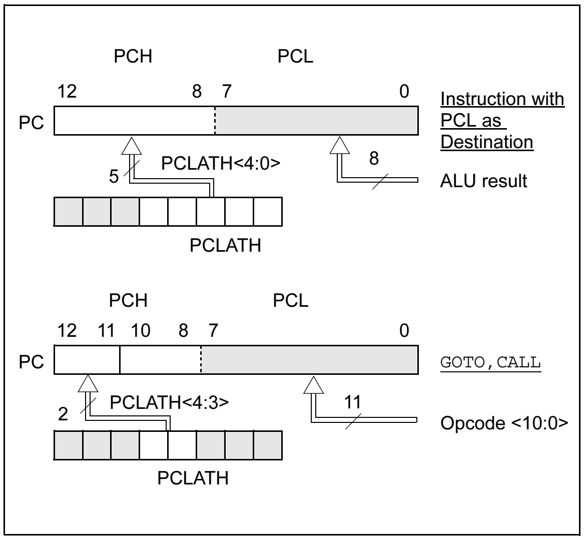


Note 1: Higher order bits are from the STATUS register.

4.3 PCL and PCLATH

The program counter (PC) is 13-bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<12:8>) is not directly readable or writable and comes from PCLATH. On any RESET, the PC is cleared. Figure 4-8 shows the two situations for the loading of the PC. The upper example in the figure shows how the PC is loaded on a write to PCL (PCLATH<4:0> → PCH). The lower example in the figure shows how the PC is loaded during a `CALL` or `GOTO` instruction (PCLATH<4:3> → PCH).

FIGURE 4-8: LOADING OF PC IN DIFFERENT SITUATIONS



4.3.1 COMPUTED GOTO

A computed GOTO is accomplished by adding an offset to the program counter (`ADDWF PCL`). When doing a table read using a computed GOTO method, care should be exercised if the table location crosses a PCL memory boundary (each 256 byte block). Refer to the application note, *“Implementing a Table Read"* (AN556).

4.3.2 STACK

The PIC16C62X family has an 8-level deep x 13-bit wide hardware stack (Figure 4-2 and Figure 4-3). The stack space is not part of either program or data space and the stack pointer is not readable or writable. The PC is PUSHed onto the stack when a `CALL` instruction is executed or an interrupt causes a branch. The stack is POPed in the event of a `RETURN`, `RETLW` or a `RETFIE` instruction execution. PCLATH is not affected by a `PUSH` or `POP` operation.

The stack operates as a circular buffer. This means that after the stack has been PUSHed eight times, the ninth push overwrites the value that was stored from the first push. The tenth push overwrites the second push (and so on).

Note 1: There are no STATUS bits to indicate stack overflow or stack underflow conditions.

2: There are no instructions/mnemonics called `PUSH` or `POP`. These are actions that occur from the execution of the `CALL`, `RETURN`, `RETLW` and `RETFIE` instructions, or the vectoring to an interrupt address.

TABLE 5-1: PORTA FUNCTIONS

Name	Bit #	Buffer Type	Function
RA0/AN0	bit0	ST	Input/output or comparator input
RA1/AN1	bit1	ST	Input/output or comparator input
RA2/AN2/VREF	bit2	ST	Input/output or comparator input or VREF output
RA3/AN3	bit3	ST	Input/output or comparator input/output
RA4/T0CKI	bit4	ST	Input/output or external clock input for TMR0 or comparator output. Output is open drain type.

Legend: ST = Schmitt Trigger input

TABLE 5-2: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR	Value on All Other RESETS
05h	PORTA	—	—	—	RA4	RA3	RA2	RA1	RA0	`---x 0000`	`---u 0000`
85h	TRISA	—	—	—	TRISA 4	TRISA 3	TRISA 2	TRISA 1	TRISA 0	`---1 1111`	`---1 1111`
1Fh	CMCON	C2OUT	C1OUT	—	—	CIS	CM2	CM1	CM0	`00-- 0000`	`00-- 0000`
9Fh	VRCON	VREN	VROE	VRR	—	VR3	VR2	VR1	VR0	`000- 0000`	`000- 0000`

Legend: — = Unimplemented locations, read as ‘0’, u = unchanged, x = unknown

Note: Shaded bits are not used by PORTA.

5.2 PORTB and TRISB Registers

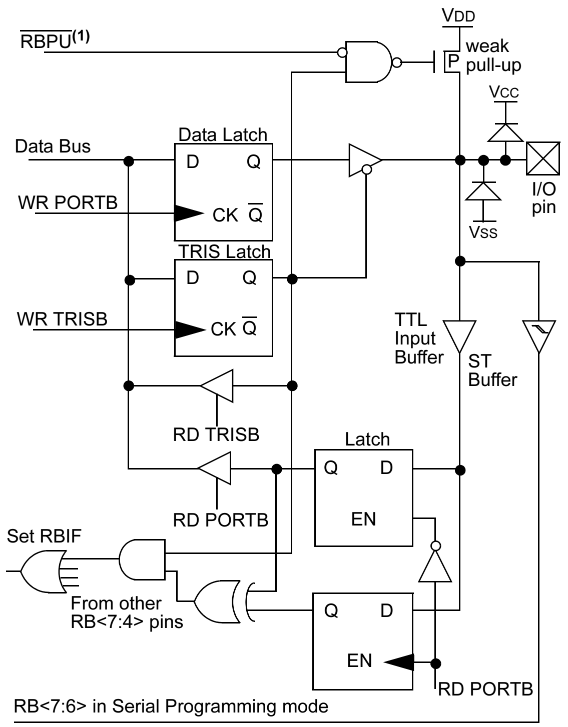
PORTB is an 8-bit wide, bi-directional port. The corresponding data direction register is TRISB. A '1' in the TRISB register puts the corresponding output driver in a High Impedance mode. A '0' in the TRISB register puts the contents of the output latch on the selected pin(s).

Reading PORTB register reads the status of the pins, whereas writing to it will write to the port latch. All write operations are read-modify-write operations. So a write to a port implies that the port pins are first read, then this value is modified and written to the port data latch.

Each of the PORTB pins has a weak internal pull-up (≈200 μA typical). A single control bit can turn on all the pull-ups. This is done by clearing the $\overline{\text{RBPU}}$ (OPTION<7>) bit. The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on Power-on Reset.

Four of PORTB's pins, RB<7:4>, have an interrupt on change feature. Only pins configured as inputs can cause this interrupt to occur (e.g., any RB<7:4> pin configured as an output is excluded from the interrupt on change comparison). The input pins (of RB<7:4>) are compared with the old value latched on the last read of PORTB. The "mismatch" outputs of RB<7:4> are OR'ed together to generate the RBIF interrupt (flag latched in INTCON<0>).

FIGURE 5-5: BLOCK DIAGRAM OF RB<7:4> PINS



Note 1: TRISB = 1 enables weak pull-up if $\overline{\text{RBPU}}$ = '0' (OPTION<7>).

This interrupt can wake the device from SLEEP. The user, in the interrupt service routine, can clear the interrupt in the following manner:

a) Any read or write of PORTB. This will end the mismatch condition.

b) Clear flag bit RBIF.

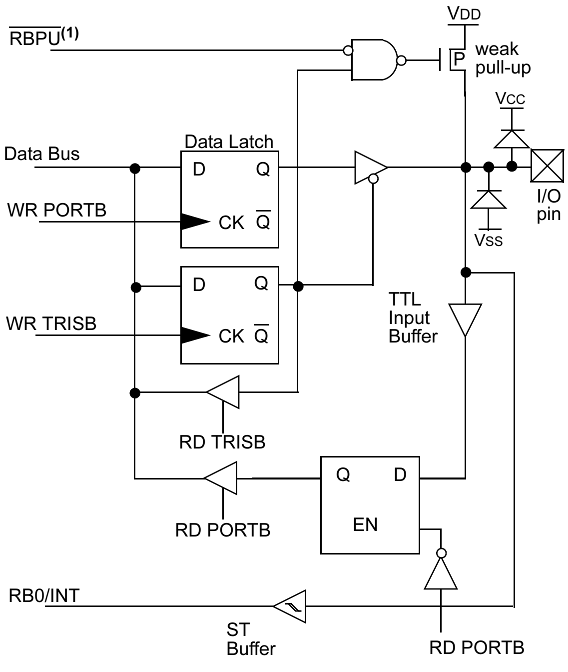
A mismatch condition will continue to set flag bit RBIF. Reading PORTB will end the mismatch condition and allow flag bit RBIF to be cleared.

This interrupt on mismatch feature, together with software configurable pull-ups on these four pins allow easy interface to a key pad and make it possible for wake-up on key-depression. (See AN552, "Implementing Wake-Up on Key Strokes.)

> **Note:** If a change on the I/O pin should occur when the read operation is being executed (start of the Q2 cycle), then the RBIF interrupt flag may not get set.

The interrupt-on-change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt on change feature. Polling of PORTB is not recommended while using the interrupt-on-change feature.

FIGURE 5-6: BLOCK DIAGRAM OF RB<3:0> PINS



Note 1: TRISB = 1 enables weak pull-up if $\overline{\text{RBPU}}$ = '0' (OPTION<7>).

6.2 Using Timer0 with External Clock

When an external clock input is used for Timer0, it must meet certain requirements. The external clock requirement is due to internal phase clock (TOSC) synchronization. Also, there is a delay in the actual incrementing of Timer0 after synchronization.

6.2.1 EXTERNAL CLOCK SYNCHRONIZATION

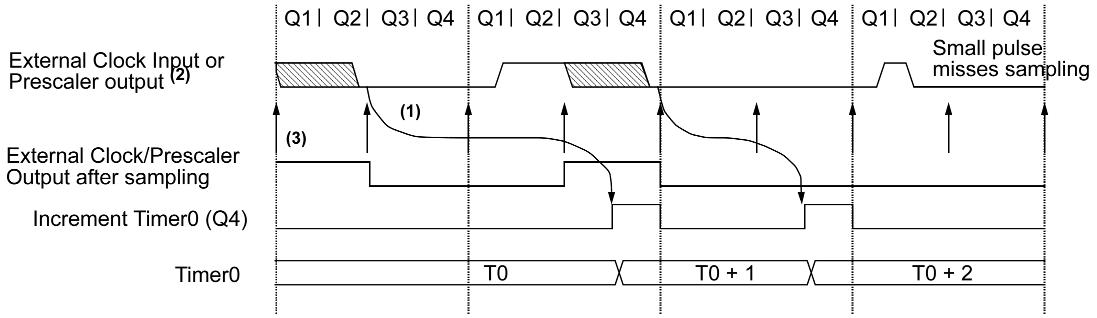
When no prescaler is used, the external clock input is the same as the prescaler output. The synchronization of T0CKI with the internal phase clocks is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks (Figure 6-5). Therefore, it is necessary for T0CKI to be high for at least 2TOSC (and a small RC delay of 20 ns) and low for at least 2TOSC (and a small RC delay of 20 ns). Refer to the electrical specification of the desired device.

When a prescaler is used, the external clock input is divided by the asynchronous ripple-counter type prescaler, so that the prescaler output is symmetrical. For the external clock to meet the sampling requirement, the ripple-counter must be taken into account. Therefore, it is necessary for T0CKI to have a period of at least 4TOSC (and a small RC delay of 40 ns) divided by the prescaler value. The only requirement on T0CKI high and low time is that they do not violate the minimum pulse width requirement of 10 ns. Refer to parameters 40, 41 and 42 in the electrical specification of the desired device.

6.2.2 TIMER0 INCREMENT DELAY

Since the prescaler output is synchronized with the internal clocks, there is a small delay from the time the external clock edge occurs to the time the TMR0 is actually incremented. Figure 6-5 shows the delay from the external clock edge to the timer incrementing.

FIGURE 6-5: TIMER0 TIMING WITH EXTERNAL CLOCK



Note 1: Delay from clock input change to Timer0 increment is 3Tosc to 7Tosc. (Duration of Q = Tosc). Therefore, the error in measuring the interval between two edges on Timer0 input = ±4Tosc max.

2: External clock if no prescaler selected, Prescaler output otherwise.

3: The arrows indicate the points in time where sampling occurs.

The code example in Example 7-1 depicts the steps required to configure the comparator module. RA3 and RA4 are configured as digital output. RA0 and RA1 are configured as the V- inputs and RA2 as the V+ input to both comparators.

EXAMPLE 7-1: INITIALIZING COMPARATOR MODULE

```
MOVLW  0x03        ;Init comparator mode
MOVWF  CMCON       ;CM<2:0> = 011
CLRF   PORTA       ;Init PORTA
BSF    STATUS,RP0  ;Select Bank1
MOVLW  0x07        ;Initialize data direction
MOVWF  TRISA       ;Set RA<2:0> as inputs
                   ;RA<4:3> as outputs
                   ;TRISA<7:5> always read ‘0’
BCF    STATUS,RP0  ;Select Bank 0
CALL   DELAY 10    ;10µs delay
MOVF   CMCON,F     ;Read CMCON to end change condition
BCF    PIR1,CMIF   ;Clear pending interrupts
BSF    STATUS,RP0  ;Select Bank 1
BSF    PIE1,CMIE   ;Enable comparator interrupts
BCF    STATUS,RP0  ;Select Bank 0
BSF    INTCON,PEIE ;Enable peripheral interrupts
BSF    INTCON,GIE  ;Global interrupt enable
```

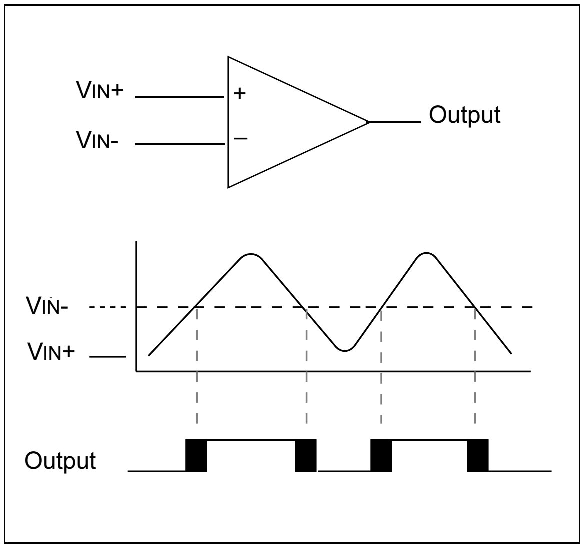
7.2 Comparator Operation

A single comparator is shown in Figure 7-2 along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input VIN-, the output of the comparator is a digital low level. When the analog input at VIN+ is greater than the analog input VIN-, the output of the comparator is a digital high level. The shaded areas of the output of the comparator in Figure 7-2 represent the uncertainty due to input offsets and response time.

7.3 Comparator Reference

An external or internal reference signal may be used depending on the comparator Operating mode. The analog signal that is present at VIN- is compared to the signal at VIN+, and the digital output of the comparator is adjusted accordingly (Figure 7-2).

FIGURE 7-2: SINGLE COMPARATOR



7.3.1 EXTERNAL REFERENCE SIGNAL

When external voltage references are used, the comparator module can be configured to have the comparators operate from the same or different reference sources. However, threshold detector applications may require the same reference. The reference signal must be between VSS and VDD, and can be applied to either pin of the comparator(s).

7.3.2 INTERNAL REFERENCE SIGNAL

The comparator module also allows the selection of an internally generated voltage reference for the comparators. Section 10, Instruction Sets, contains a detailed description of the Voltage Reference Module that provides this signal. The internal reference signal is used when the comparators are in mode CM<2:0>=010 (Figure 7-1). In this mode, the internal voltage reference is applied to the VIN+ pin of both comparators.

7.6 Comparator Interrupts

The comparator interrupt flag is set whenever there is a change in the output value of either comparator. Software will need to maintain information about the status of the output bits, as read from CMCON<7:6>, to determine the actual change that has occurred. The CMIF bit, PIR1<6>, is the comparator interrupt flag. The CMIF bit must be RESET by clearing ‘0’. Since it is also possible to write a '1' to this register, a simulated interrupt may be initiated.

The CMIE bit (PIE1<6>) and the PEIE bit (INTCON<6>) must be set to enable the interrupt. In addition, the GIE bit must also be set. If any of these bits are clear, the interrupt is not enabled, though the CMIF bit will still be set if an interrupt condition occurs.

Note:	If a change in the CMCON register (C1OUT or C2OUT) should occur when a read operation is being executed (start of the Q2 cycle), then the CMIF (PIR1<6>) interrupt flag may not get set.

The user, in the interrupt service routine, can clear the interrupt in the following manner:

a) Any read or write of CMCON. This will end the mismatch condition.

b) Clear flag bit CMIF.

A mismatch condition will continue to set flag bit CMIF. Reading CMCON will end the mismatch condition and allow flag bit CMIF to be cleared.

7.7 Comparator Operation During SLEEP

When a comparator is active and the device is placed in SLEEP mode, the comparator remains active and the interrupt is functional if enabled. This interrupt will wake up the device from SLEEP mode when enabled. While the comparator is powered-up, higher SLEEP currents than shown in the power-down current specification will occur. Each comparator that is operational will consume additional current as shown in the comparator specifications. To minimize power consumption while in SLEEP mode, turn off the comparators, CM<2:0> = 111, before entering SLEEP. If the device wakes up from SLEEP, the contents of the CMCON register are not affected.

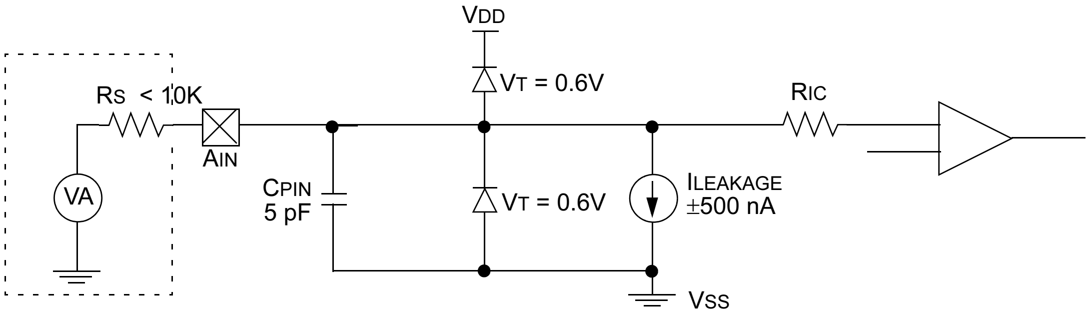
7.8 Effects of a RESET

A device RESET forces the CMCON register to its RESET state. This forces the comparator module to be in the comparator RESET mode, CM<2:0> = 000. This ensures that all potential inputs are analog inputs. Device current is minimized when analog inputs are present at RESET time. The comparators will be powered-down during the RESET interval.

7.9 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 7-4. Since the analog pins are connected to a digital output, they have reverse biased diodes to V_{DD} and V_{SS}. The analog input therefore, must be between V_{SS} and V_{DD}. If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latchup may occur. A maximum source impedance of 10 kΩ is recommended for the analog sources. Any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current.

FIGURE 7-4: ANALOG INPUT MODEL



Legend
C_{PIN} = Input Capacitance
V_T = Threshold Voltage
$I_{LEAKAGE}$ = Leakage Current at the pin due to various junctions
R_{IC} = Interconnect Resistance
R_S = Source Impedance
VA = Analog Voltage

9.2 Oscillator Configurations

9.2.1 OSCILLATOR TYPES

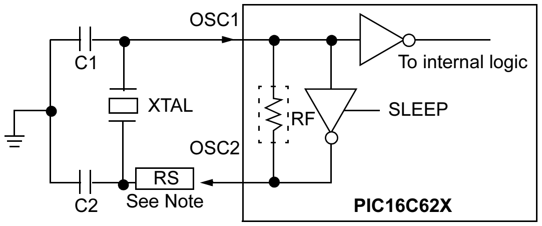
The PIC16C62X devices can be operated in four different oscillator options. The user can program two configuration bits (FOSC1 and FOSC0) to select one of these four modes:

- LP Low Power Crystal
- XT Crystal/Resonator
- HS High Speed Crystal/Resonator
- RC Resistor/Capacitor

9.2.2 CRYSTAL OSCILLATOR / CERAMIC RESONATORS

In XT, LP or HS modes, a crystal or ceramic resonator is connected to the OSC1 and OSC2 pins to establish oscillation (Figure 9-1). The PIC16C62X oscillator design requires the use of a parallel cut crystal. Use of a series cut crystal may give a frequency out of the crystal manufacturers specifications. When in XT, LP or HS modes, the device can have an external clock source to drive the OSC1 pin (Figure 9-2).

FIGURE 9-1: CRYSTAL OPERATION (OR CERAMIC RESONATOR) (HS, XT OR LP OSC CONFIGURATION)



See Table 9-1 and Table 9-2 for recommended values of C1 and C2.

Note: A series resistor may be required for AT strip cut crystals.

FIGURE 9-2: EXTERNAL CLOCK INPUT OPERATION (HS, XT OR LP OSC CONFIGURATION)

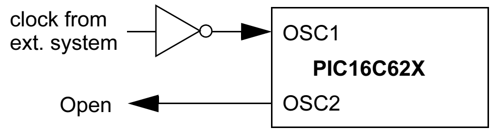


TABLE 9-1: CAPACITOR SELECTION FOR CERAMIC RESONATORS

Ranges Characterized:			
Mode	**Freq**	**OSC1(C1)**	**OSC2(C2)**
XT	455 kHz 2.0 MHz 4.0 MHz	22 - 100 pF 15 - 68 pF 15 - 68 pF	22 - 100 pF 15 - 68 pF 15 - 68 pF
HS	8.0 MHz 16.0 MHz	10 - 68 pF 10 - 22 pF	10 - 68 pF 10 - 22 pF

Higher capacitance increases the stability of the oscillator but also increases the start-up time. These values are for design guidance only. Since each resonator has its own characteristics, the user should consult the resonator manufacturer for appropriate values of external components.

TABLE 9-2: CAPACITOR SELECTION FOR CRYSTAL OSCILLATOR

Mode	Freq	OSC1(C1)	OSC2(C2)
LP	32 kHz 200 kHz	68 - 100 pF 15 - 30 pF	68 - 100 pF 15 - 30 pF
XT	100 kHz 2 MHz 4 MHz	68 - 150 pF 15 - 30 pF 15 - 30 pF	150 - 200 pF 15 - 30 pF 15 - 30 pF
HS	8 MHz 10 MHz 20 MHz	15 - 30 pF 15 - 30 pF 15 - 30 pF	15 - 30 pF 15 - 30 pF 15 - 30 pF

Higher capacitance increases the stability of the oscillator but also increases the start-up time. These values are for design guidance only. Rs may be required in HS mode as well as XT mode to avoid overdriving crystals with low drive level specification. Since each crystal has its own characteristics, the user should consult the crystal manufacturer for appropriate values of external components.

TABLE 9-4: INITIALIZATION CONDITION FOR SPECIAL REGISTERS

Condition	Program Counter	STATUS Register	PCON Register
Power-on Reset	000h	0001 1xxx	---- --0x
$\overline{\text{MCLR}}$ Reset during normal operation	000h	000u uuuu	---- --uu
$\overline{\text{MCLR}}$ Reset during SLEEP	000h	0001 0uuu	---- --uu
WDT Reset	000h	0000 uuuu	---- --uu
WDT Wake-up	PC + 1	uuu0 0uuu	---- --uu
Brown-out Reset	000h	000x xuuu	---- --u0
Interrupt Wake-up from SLEEP	PC + 1(1)	uuu1 0uuu	---- --uu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0'.

Note 1: When the wake-up is due to an interrupt and global enable bit, GIE is set, the PC is loaded with the interrupt vector (0004h) after execution of PC+1.

TABLE 9-5: INITIALIZATION CONDITION FOR REGISTERS

Register	Address	Power-on Reset	• $\overline{\text{MCLR}}$ Reset during normal operation • $\overline{\text{MCLR}}$ Reset during SLEEP • WDT Reset • Brown-out Reset (1)	• Wake-up from SLEEP through interrupt • Wake-up from SLEEP through WDT time-out
W	—	xxxx xxxx	uuuu uuuu	uuuu uuuu
INDF	00h	—	—	—
TMR0	01h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PCL	02h	0000 0000	0000 0000	PC + 1(3)
STATUS	03h	0001 1xxx	000q quuu(4)	uuuq quuu(4)
FSR	04h	xxxx xxxx	uuuu uuuu	uuuu uuuu
PORTA	05h	---x xxxx	---u uuuu	---u uuuu
PORTB	06h	xxxx xxxx	uuuu uuuu	uuuu uuuu
CMCON	1Fh	00-- 0000	00-- 0000	uu-- uuuu
PCLATH	0Ah	---0 0000	---0 0000	---u uuuu
INTCON	0Bh	0000 000x	0000 000u	uuuu uqqq(2)
PIR1	0Ch	-0-- ----	-0-- ----	-q-- ----(2,5)
OPTION	81h	1111 1111	1111 1111	uuuu uuuu
TRISA	85h	---1 1111	---1 1111	---u uuuu
TRISB	86h	1111 1111	1111 1111	uuuu uuuu
PIE1	8Ch	-0-- ----	-0-- ----	-u-- ----
PCON	8Eh	---- --0x	---- --uq(1,6)	---- --uu
VRCON	9Fh	000- 0000	000- 0000	uuu- uuuu

Legend: u = unchanged, x = unknown, - = unimplemented bit, reads as '0',q = value depends on condition.

Note 1: If VDD goes too low, Power-on Reset will be activated and registers will be affected differently.

2: One or more bits in INTCON, PIR1 and/or PIR2 will be affected (to cause wake-up).

3: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

4: See Table 9-4 for RESET value for specific condition.

5: If wake-up was due to comparator input changing, then bit 6 = 1. All other interrupts generating a wake-up will cause bit 6 = u.

6: If RESET was due to brown-out, then bit 0 = 0. All other RESETS will cause bit 0 = u.

CLRW	**Clear W**
Syntax:	[*label*] CLRW
Operands:	None
Operation:	00h → (W) 1 → Z
Status Affected:	Z
Encoding:	`00` `0001` `0000` `0011`
Description:	W register is cleared. Zero bit (Z) is set.
Words:	1
Cycles:	1
Example	`CLRW` Before Instruction W = 0x5A After Instruction W = 0x00 Z = 1

CLRWDT	**Clear Watchdog Timer**
Syntax:	[*label*] CLRWDT
Operands:	None
Operation:	00h → WDT 0 → WDT prescaler, 1 → $\overline{\text{TO}}$ 1 → $\overline{\text{PD}}$
Status Affected:	$\overline{\text{TO}}$, $\overline{\text{PD}}$
Encoding:	`00` `0000` `0110` `0100`
Description:	`CLRWDT` instruction resets the Watchdog Timer. It also resets the prescaler of the WDT. STATUS bits $\overline{\text{TO}}$ and $\overline{\text{PD}}$ are set.
Words:	1
Cycles:	1
Example	`CLRWDT` Before Instruction WDT counter = ? After Instruction WDT counter = 0x00 WDT prescaler = 0 $\overline{\text{TO}}$ = 1 $\overline{\text{PD}}$ = 1

COMF	**Complement f**
Syntax:	[*label*] COMF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	($\bar{f}$) → (dest)
Status Affected:	Z
Encoding:	`00` `1001` `dfff` `ffff`
Description:	The contents of register 'f' are complemented. If 'd' is 0, the result is stored in W. If 'd' is 1, the result is stored back in register 'f'.
Words:	1
Cycles:	1
Example	`COMF REG1,0` Before Instruction REG1 = 0x13 After Instruction REG1 = 0x13 W = 0xEC

DECF	**Decrement f**
Syntax:	[*label*] DECF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(f) - 1 → (dest)
Status Affected:	Z
Encoding:	`00` `0011` `dfff` `ffff`
Description:	Decrement register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is stored back in register 'f'.
Words:	1
Cycles:	1
Example	`DECF CNT,` 1 Before Instruction CNT = 0x01 Z = 0 After Instruction CNT = 0x00 Z = 1

SUBLW Subtract W from Literal

Syntax: [*label*] SUBLW k

Operands: $0 \leq k \leq 255$

Operation: k - (W) → (W)

Status Affected: C, DC, Z

Encoding:

`11`	`110x`	`kkkk`	`kkkk`

Description: The W register is subtracted (2’s complement method) from the eight bit literal 'k'. The result is placed in the W register.

Words: 1

Cycles: 1

Example 1: `SUBLW 0x02`

Before Instruction

W = 1
C = ?

After Instruction

W = 1
C = 1; result is positive

Example 2: Before Instruction

W = 2
C = ?

After Instruction

W = 0
C = 1; result is zero

Example 3: Before Instruction

W = 3
C = ?

After Instruction

W = 0xFF
C = 0; result is negative

SUBWF Subtract W from f

Syntax: [*label*] SUBWF f,d

Operands: $0 \leq f \leq 127$
$d \in [0,1]$

Operation: (f) - (W) → (dest)

Status Affected: C, DC, Z

Encoding:

`00`	`0010`	`dfff`	`ffff`

Description: Subtract (2’s complement method) W register from register 'f'. If 'd' is 0, the result is stored in the W register. If 'd' is 1, the result is stored back in register 'f'.

Words: 1

Cycles: 1

Example 1: `SUBWF REG1,1`

Before Instruction

REG1= 3
W = 2
C = ?

After Instruction

REG1= 1
W = 2
C = 1; result is positive

Example 2: Before Instruction

REG1= 2
W = 2
C = ?

After Instruction

REG1= 0
W = 2
C = 1; result is zero

Example 3: Before Instruction

REG1= 1
W = 2
C = ?

After Instruction

REG1= 0xFF
W = 2
C = 0; result is negative

NOTES:

11.0 DEVELOPMENT SUPPORT

The PICmicro® microcontrollers are supported with a full range of hardware and software development tools:

- Integrated Development Environment
    - MPLAB® IDE Software
- Assemblers/Compilers/Linkers
    - MPASM™ Assembler
    - MPLAB C17 and MPLAB C18 C Compilers
    - MPLINK™ Object Linker/
      MPLIB™ Object Librarian
    - MPLAB C30 C Compiler
    - MPLAB ASM30 Assembler/Linker/Library
- Simulators
    - MPLAB SIM Software Simulator
    - MPLAB dsPIC30 Software Simulator
- Emulators
    - MPLAB ICE 2000 In-Circuit Emulator
    - MPLAB ICE 4000 In-Circuit Emulator
- In-Circuit Debugger
    - MPLAB ICD 2
- Device Programmers
    - PRO MATE® II Universal Device Programmer
    - PICSTART® Plus Development Programmer
- Low Cost Demonstration Boards
    - PICDEM™ 1 Demonstration Board
    - PICDEM.net™ Demonstration Board
    - PICDEM 2 Plus Demonstration Board
    - PICDEM 3 Demonstration Board
    - PICDEM 4 Demonstration Board
    - PICDEM 17 Demonstration Board
    - PICDEM 18R Demonstration Board
    - PICDEM LIN Demonstration Board
    - PICDEM USB Demonstration Board
- Evaluation Kits
    - KEELOQ®
    - PICDEM MSC
    - microID®
    - CAN
    - PowerSmart®
    - Analog

11.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16-bit microcontroller market. The MPLAB IDE is a Windows® based application that contains:

- An interface to debugging tools
    - simulator
    - programmer (sold separately)
    - emulator (sold separately)
    - in-circuit debugger (sold separately)
- A full-featured editor with color coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High level source code debugging
- Mouse over variable inspection
- Extensive on-line help

The MPLAB IDE allows you to:

- Edit your source files (either assembly or C)
- One touch assemble (or compile) and download to PICmicro emulator and simulator tools (automatically updates all project information)
- Debug using:
    - source files (assembly or C)
    - absolute listing file (mixed assembly and C)
    - machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost effective simulators, through low cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increasing flexibility and power.

11.2 MPASM Assembler

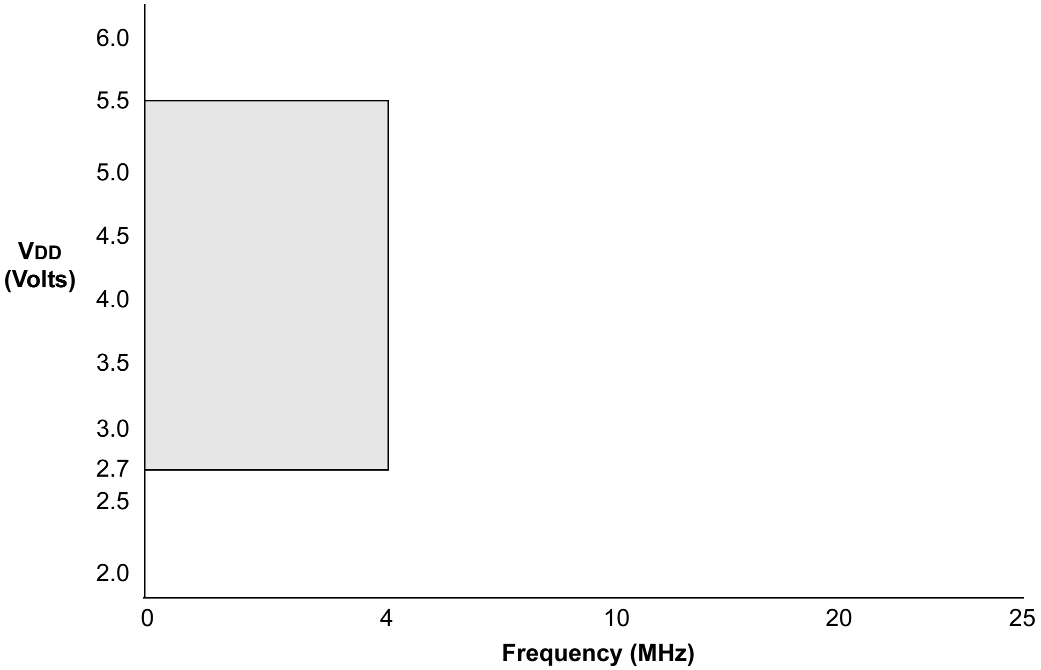
The MPASM assembler is a full-featured, universal macro assembler for all PICmicro MCUs.

The MPASM assembler generates relocatable object files for the MPLINK object linker, Intel® standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code and COFF files for debugging.

The MPASM assembler features include:

- Integration into MPLAB IDE projects
- User defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process

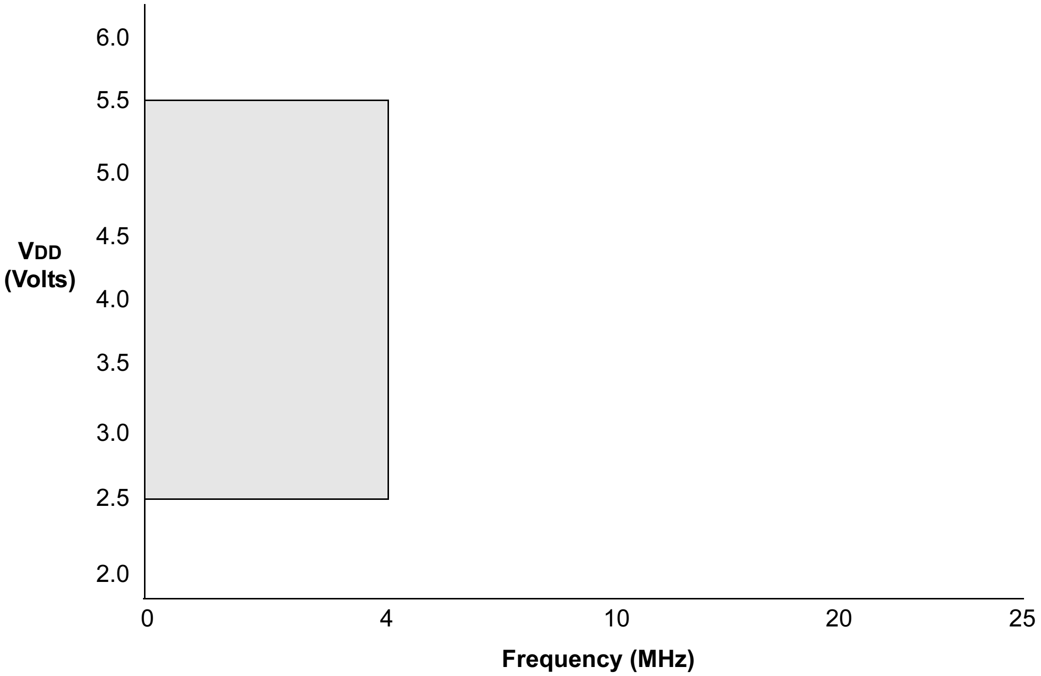
FIGURE 12-5: PIC16LC620A/LC621A/LC622A VOLTAGE-FREQUENCY GRAPH, -40°C ≤ TA ≤ 0°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

2: The maximum rated speed of the part limits the permissible combinations of voltage and frequency. Please reference the Product Identification System section for the maximum rated speed of the parts.

FIGURE 12-6: PIC16LC620A/LC621A/LC622A VOLTAGE-FREQUENCY GRAPH, 0°C ≤ TA ≤ +125°C



Note 1: The shaded region indicates the permissible combinations of voltage and frequency.

2: The maximum rated speed of the part limits the permissible combinations of voltage and frequency. Please reference the Product Identification System section for the maximum rated speed of the parts.

12.1 DC Characteristics: PIC16C62X-04 (Commercial, Industrial, Extended) PIC16C62X-20 (Commercial, Industrial, Extended) PIC16LC62X-04 (Commercial, Industrial, Extended) (CONT.)

PIC16C62X			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial and -40°C ≤ TA ≤ +125°C for extended				
PIC16LC62X			**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial and -40°C ≤ TA ≤ +125°C for extended Operating voltage VDD range is the PIC16C62X range.				
Param . No.	**Sym**	**Characteristic**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
D022	ΔIWDT	WDT Current$^{(5)}$	—	6.0	20 25	μA μA	VDD=4.0V (125°C)
D022A	ΔIBOR	Brown-out Reset Current$^{(5)}$	—	350	425	μA	$\overline{\text{BOD}}$ enabled, VDD = 5.0V
D023	ΔICOMP	Comparator Current for each Comparator$^{(5)}$	—	—	100	μA	VDD = 4.0V
D023A	ΔIVREF	VREF Current$^{(5)}$	—	—	300	μA	VDD = 4.0V
D022	ΔIWDT	WDT Current$^{(5)}$	—	6.0	15	μA	VDD=3.0V
D022A	ΔIBOR	Brown-out Reset Current$^{(5)}$	—	350	425	μA	$\overline{\text{BOD}}$ enabled, VDD = 5.0V
D023	ΔICOMP	Comparator Current for each Comparator$^{(5)}$	—	—	100	μA	VDD = 3.0V
D023A	ΔIVREF	VREF Current$^{(5)}$	—	—	300	μA	VDD = 3.0V
1A	FOSC	LP Oscillator Operating Frequency	0	—	200	kHz	All temperatures
		RC Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		XT Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		HS Oscillator Operating Frequency	0	—	20	MHz	All temperatures
1A	FOSC	LP Oscillator Operating Frequency	0	—	200	kHz	All temperatures
		RC Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		XT Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		HS Oscillator Operating Frequency	0	—	20	MHz	All temperatures

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.
The test conditions for all IDD measurements in Active Operation mode are:
$\overline{\text{OSC1}}$ = external square wave, from rail to rail; all I/O pins tri-stated, pulled to VDD,
$\overline{\text{MCLR}}$ = VDD; WDT enabled/disabled as specified.

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD or VSS.

4: For RC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula: Ir = VDD/2REXT (mA) with REXT in kΩ.

5: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.

12.2 DC Characteristics: PIC16C62XA-04 (Commercial, Industrial, Extended)
PIC16C62XA-20 (Commercial, Industrial, Extended)
PIC16LC62XA-04 (Commercial, Industrial, Extended) (CONT.)

PIC16C62XA	**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial and -40°C ≤ TA ≤ +125°C for extended
PIC16LC62XA	**Standard Operating Conditions (unless otherwise stated)** Operating temperature -40°C ≤ TA ≤ +85°C for industrial and 0°C ≤ TA ≤ +70°C for commercial and -40°C ≤ TA ≤ +125°C for extended

Param. No.	Sym	Characteristic	Min	Typ†	Max	Units	Conditions
D010	IDD	Supply Current(2, 4)	—	1.2	2.0	mA	FOSC = 4 MHz, VDD = 5.5V, WDT disabled, XT mode, **(Note 4)***
			—	0.4	1.2	mA	FOSC = 4 MHz, VDD = 3.0V, WDT disabled, XT mode, **(Note 4)***
			—	1.0	2.0	mA	FOSC = 10 MHz, VDD = 3.0V, WDT disabled, HS mode, **(Note 6)**
			—	4.0	6.0	mA	FOSC = 20 MHz, VDD = 4.5V, WDT disabled, HS mode
			—	4.0	7.0	mA	FOSC = 20 MHz, VDD = 5.5V, WDT disabled*, HS mode
			—	35	70	μA	FOSC = 32 kHz, VDD = 3.0V, WDT disabled, LP mode
D010	IDD	Supply Current(2)	—	1.2	2.0	mA	FOSC = 4 MHz, VDD = 5.5V, WDT disabled, XT mode, **(Note 4)***
			—	—	1.1	mA	FOSC = 4 MHz, VDD = 2.5V, WDT disabled, XT mode, **(Note 4)**
			—	35	70	μA	FOSC = 32 kHz, VDD = 2.5V, WDT disabled, LP mode
D020	IPD	Power-down Current(3)	—	—	2.2	μA	VDD = 3.0V
			—	—	5.0	μA	VDD = 4.5V*
			—	—	9.0	μA	VDD = 5.5V
			—	—	15	μA	VDD = 5.5V Extended Temp.
D020	IPD	Power-down Current(3)	—	—	2.0	μA	VDD = 2.5V
			—	—	2.2	μA	VDD = 3.0V*
			—	—	9.0	μA	VDD = 5.5V
			—	—	15	μA	VDD = 5.5V Extended Temp.

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.
The test conditions for all IDD measurements in Active Operation mode are:
OSC1 = external square wave, from rail to rail; all I/O pins tri-stated, pulled to VDD,
$\overline{MCLR}$ = VDD; WDT enabled/disabled as specified.

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD or VSS.

4: For RC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula: Ir = VDD/2REXT (mA) with REXT in kΩ.

5: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.

6: Commercial temperature range only.

12.5 DC CHARACTERISTICS: PIC16C620A/C621A/C622A-40(7) (Commercial) PIC16CR620A-40(7) (Commercial)

DC CHARACTERISTICS			Standard Operating Conditions (unless otherwise stated) Operating temperature 0°C ≤ TA ≤ +70°C for commercial				
Param No.	**Sym**	**Characteristic**	**Min**	**Typ†**	**Max**	**Units**	**Conditions**
D001	VDD	Supply Voltage	3.0	—	5.5	V	FOSC = DC to 20 MHz
D002	VDR	RAM Data Retention Voltage(1)	—	1.5*	—	V	Device in SLEEP mode
D003	VPOR	VDD start voltage to ensure Power-on Reset	—	VSS	—	V	See section on Power-on Reset for details
D004	SVDD	VDD rise rate to ensure Power-on Reset	0.05 *	—	—	V/ms	See section on Power-on Reset for details
D005	VBOR	Brown-out Detect Voltage	3.65	4.0	4.35	V	BOREN configuration bit is cleared
D010	IDD	Supply Current(2,4)	—	1.2	2.0	mA	FOSC = 4 MHz, VDD = 5.5V, WDT disabled, XT OSC mode, **(Note 4)***
			—	0.4	1.2	mA	FOSC = 4 MHz, VDD = 3.0V, WDT disabled, XT OSC mode, **(Note 4)**
			—	1.0	2.0	mA	FOSC = 10 MHz, VDD = 3.0V, WDT disabled, HS OSC mode, **(Note 6)**
			—	4.0	6.0	mA	FOSC = 20 MHz, VDD = 4.5V, WDT disabled, HS OSC mode
			—	4.0	7.0	mA	FOSC = 20 MHz, VDD = 5.5V, WDT disabled*, HS OSC mode
			—	35	70	μA	FOSC = 32 kHz, VDD = 3.0V, WDT disabled, LP OSC mode
D020	IPD	Power Down Current(3)	—	—	2.2	μA	VDD = 3.0V
			—	—	5.0	μA	VDD = 4.5V*
			—	—	9.0	μA	VDD = 5.5V
			—	—	15	μA	VDD = 5.5V Extended
D022	ΔIWDT	WDT Current(5)	—	6.0	10	μA	VDD = 4.0V
					12	μA	(125°C)
D022A	ΔIBOR	Brown-out Reset Current(5)	—	75	125	μA	$\overline{\text{BOD}}$ enabled, VDD = 5.0V
D023	ΔICOMP	Comparator Current for each Comparator(5)	—	30	60	μA	VDD = 4.0V
D023A	ΔIVREF	VREF Current(5)	—	80	135	μA	VDD = 4.0V
	ΔIEE Write	Operating Current	—		3	mA	VCC = 5.5V, SCL = 400 kHz
	ΔIEE Read	Operating Current	—		1	mA	
	ΔIEE	Standby Current	—		30	μA	VCC = 3.0V, EE VDD = VCC
	ΔIEE	Standby Current	—		100	μA	VCC = 3.0V, EE VDD = VCC
1A	FOSC	LP Oscillator Operating Frequency	0	—	200	kHz	All temperatures
		RC Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		XT Oscillator Operating Frequency	0	—	4	MHz	All temperatures
		HS Oscillator Operating Frequency	0	—	20	MHz	All temperatures

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5.0V, 25°C, unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered in SLEEP mode without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors such as I/O pin loading and switching rate, oscillator type, internal code execution pattern, and temperature also have an impact on the current consumption.
The test conditions for all IDD measurements in Active Operation mode are:
OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD, $\overline{\text{MCLR}}$ = VDD; WDT enabled/disabled as specified.

3: The power-down current in SLEEP mode does not depend on the oscillator type. Power-down current is measured with the part in SLEEP mode, with all I/O pins in hi-impedance state and tied to VDD or VSS.

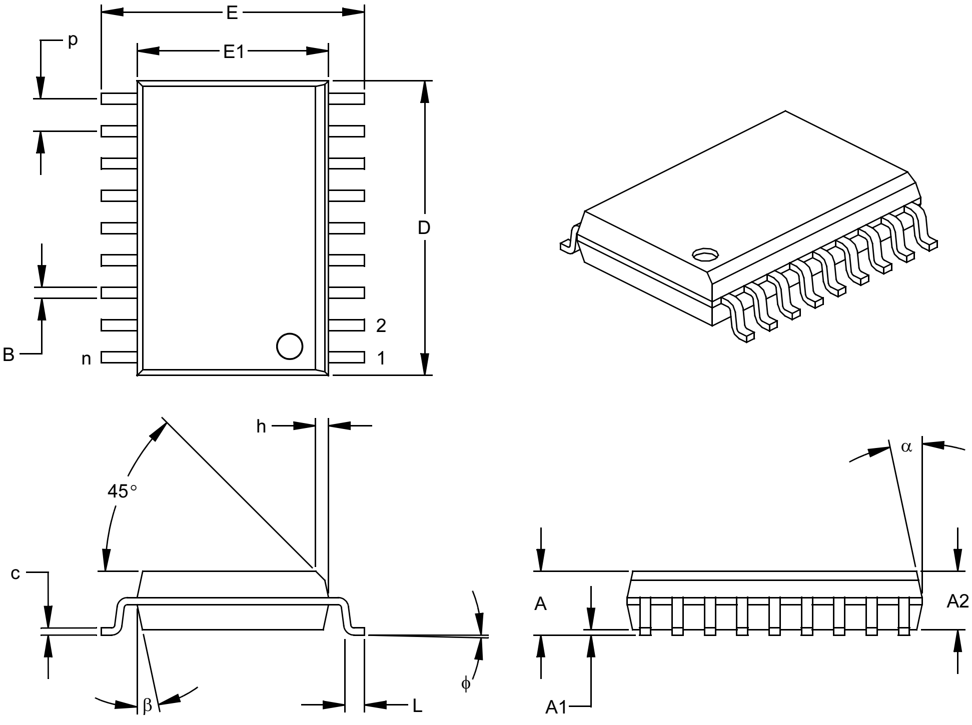
4: For RC OSC configuration, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/ 2REXT (mA) with REXT in kΩ.

5: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.

6: Commercial temperature range only.

7: See Section 12.1 and Section 12.3 for 16C62X and 16CR62X devices for operation between 20 MHz and 40 MHz for valid modified characteristics.

18-Lead Plastic Small Outline (SO) – Wide, 300 mil (SOIC)



	Units	INCHES*			MILLIMETERS		
	Dimension Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		18			18	
Pitch	p		.050			1.27	
Overall Height	A	.093	.099	.104	2.36	2.50	2.64
Molded Package Thickness	A2	.088	.091	.094	2.24	2.31	2.39
Standoff §	A1	.004	.008	.012	0.10	0.20	0.30
Overall Width	E	.394	.407	.420	10.01	10.34	10.67
Molded Package Width	E1	.291	.295	.299	7.39	7.49	7.59
Overall Length	D	.446	.454	.462	11.33	11.53	11.73
Chamfer Distance	h	.010	.020	.029	0.25	0.50	0.74
Foot Length	L	.016	.033	.050	0.41	0.84	1.27
Foot Angle	φ	0	4	8	0	4	8
Lead Thickness	c	.009	.011	.012	0.23	0.27	0.30
Lead Width	B	.014	.017	.020	0.36	0.42	0.51
Mold Draft Angle Top	α	0	12	15	0	12	15
Mold Draft Angle Bottom	β	0	12	15	0	12	15

* Controlling Parameter
§ Significant Characteristic

Notes:
Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010” (0.254mm) per side.
JEDEC Equivalent: MS-013
Drawing No. C04-051

NOTES: