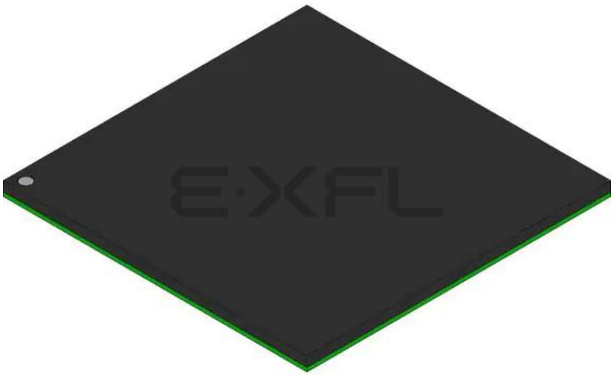


Welcome to E-XFL.COM

Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Active
Core Processor	MPC8xx
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	133MHz
Co-Processors/DSP	Communications; CPM
RAM Controllers	DRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10/100Mbps (2)
SATA	-
USB	USB 2.0 (1)
Voltage - I/O	3.3V
Operating Temperature	-40°C ~ 100°C (TA)
Security Features	-
Package / Case	256-BBGA
Supplier Device Package	256-PBGA (23x23)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mpc870czt133

- HDLC bus (implements an HDLC-based local area network (LAN))
- Asynchronous HDLC to support point-to-point protocol (PPP)
- AppleTalk
- Universal asynchronous receiver transmitter (UART)
- Synchronous UART
- Serial infrared (IrDA)
- Binary synchronous communication (BISYNC)
- Totally transparent (bit streams)
- Totally transparent (frame based with optional cyclic redundancy check (CRC))

• SMC (serial management channel)
  - UART (low-speed operation)
  - Transparent
• Universal serial bus (USB)—Supports operation as a USB function endpoint, a USB host controller, or both for testing purposes (loopback diagnostics)
  - USB 2.0 full-/low-speed compatible
  - The USB function mode has the following features:
    - Four independent endpoints support control, bulk, interrupt, and isochronous data transfers
    - CRC16 generation and checking
    - CRC5 checking
    - NRZI encoding/decoding with bit stuffing
    - 12- or 1.5-Mbps data rate
    - Flexible data buffers with multiple buffers per frame
    - Automatic retransmission upon transmit error
  - The USB host controller has the following features:
    - Supports control, bulk, interrupt, and isochronous data transfers
    - CRC16 generation and checking
    - NRZI encoding/decoding with bit stuffing
    - Supports both 12- and 1.5-Mbps data rates (automatic generation of preamble token and data rate configuration). Note that low-speed operation requires an external hub.
    - Flexible data buffers with multiple buffers per frame
    - Supports local loopback mode for diagnostics (12 Mbps only)
• Serial peripheral interface (SPI)
  - Supports master and slave modes
  - Supports multiple-master operation on the same bus
• Inter-integrated circuit (I^2C) port
  - Supports master and slave modes
  - Supports a multiple-master environment

One consequence of multiple power supplies is that when power is initially applied, the voltage rails ramp up at different rates. The rates depend on the nature of the power supply, the type of load on each power supply, and the manner in which different voltages are derived. The following restrictions apply:

- V_{DDL} must not exceed V_{DDH} during power up and power down
- V_{DDL} must not exceed 1.9 V, and V_{DDH} must not exceed 3.465 V

These cautions are necessary for the long-term reliability of the part. If they are violated, the electrostatic discharge (ESD) protection diodes are forward-biased, and excessive current can flow through these diodes. If the system power supply design does not control the voltage sequencing, the circuit shown in Figure 4 can be added to meet these requirements. The MUR420 Schottky diodes control the maximum potential difference between the external bus and core power supplies on power up, and the 1N5820 diodes regulate the maximum potential difference on power down.

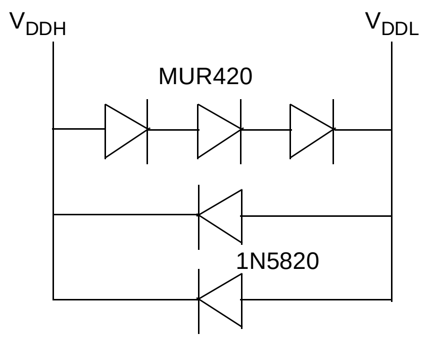


Figure 4. Example Voltage Sequencing Circuit

9 Mandatory Reset Configurations

The MPC875/MPC870 requires a mandatory configuration during reset.

If hardware reset configuration word (HRCW) is enabled, the HRCW[DBGC] value needs to be set to binary X1 in the HRCW and the SIUMCR[DBGC] should be programmed with the same value in the boot code after reset. This can be done by asserting the $\overline{\text{RSTCONF}}$ during $\overline{\text{HRESET}}$ assertion.

If HRCW is disabled, the SIUMCR[DBGC] should be programmed with binary X1 in the boot code after reset by negating the $\overline{\text{RSTCONF}}$ during the $\overline{\text{HRESET}}$ assertion.

The MBMR[GPLB4DIS], PAPAR, PADIR, PBPAR, PBDIR, PCPAR, and PCDIR need to be configured with the mandatory values in Table 7 in the boot code after the reset is negated.

Table 7. Mandatory Reset Configuration of MPC875/MPC870

Register/Configuration	Field	Value (Binary)
HRCW (Hardware reset configuration word)	HRCW[DBGC]	X1
SIUMCR (SIU module configuration register)	SIUMCR[DBGC]	X1
MBMR (Machine B mode register)	MBMR[GPLB4DIS}	0
PAPAR (Port A pin assignment register)	PAPAR[5:9] PAPAR[12:13]	0

Table 7. Mandatory Reset Configuration of MPC875/MPC870 (continued)

Register/Configuration	Field	Value (Binary)
PADIR (Port A data direction register)	PADIR[5:9] PADIR[12:13]	0
PBPAR (Port B pin assignment register)	PBPAR[14:18] PBPAR[20:22]	0
PBDIR (Port B data direction register)	PBDIR[14:8] PBDIR[20:22]	0
PCPAR (Port C pin assignment register)	PCPAR[4:5] PCPAR[8:9] PCPAR[14]	0
PCDIR (Port C data direction register)	PCDIR[4:5] PCDIR[8:9] PCDIR[14]	0
PDPAR (Port D pin assignment register)	PDPAR[3:7] PDPAR[9:5]	0
PDDIR (Port D data direction register)	PDDIR[3:7] PDDIR[9:15]	0

10 Layout Practices

Each V_{DD} pin on the MPC875/MPC870 should be provided with a low-impedance path to the board's supply. Each GND pin should likewise be provided with a low-impedance path to ground. The power supply pins drive distinct groups of logic on chip. The V_{DD} power supply should be bypassed to ground using at least four 0.1-µF bypass capacitors located as close as possible to the four sides of the package. Each board designed should be characterized and additional appropriate decoupling capacitors should be used if required. The capacitor leads and associated printed-circuit traces connecting to chip V_{DD} and GND should be kept to less than half an inch per capacitor lead. At a minimum, a four-layer board employing two inner layers as V_{DD} and GND planes should be used.

All output pins on the MPC875/MPC870 have fast rise and fall times. Printed circuit (PC) trace interconnection length should be minimized in order to minimize undershoot and reflections caused by these fast output switching times. This recommendation particularly applies to the address and data buses. Maximum PC trace lengths of 6 inches are recommended. Capacitance calculations should consider all device loads as well as parasitic capacitances due to the PC traces. Attention to proper PCB layout and bypassing becomes especially critical in systems with higher capacitive loads because these loads create higher transient currents in the V_{DD} and GND circuits. Pull up all unused inputs or signals that will be inputs during reset. Special care should be taken to minimize the noise levels on the PLL supply pins. For more information, refer to Section 14.4.3, "Clock Synthesizer Power (V_{DDSYN}, V_{SSSYN}, V_{SSSYN1})," in the *MPC885 PowerQUICC™ Family Reference Manual*.

Table 10. Bus Operation Timings (continued)

Num	Characteristic	33 MHz		40 MHz		66 MHz		80 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B2	CLKOUT pulse width low (MIN = 0.4 × B1, MAX = 0.6 × B1)	12.1	18.2	10.0	15.0	6.1	9.1	5.0	7.5	ns
B3	CLKOUT pulse width high (MIN = 0.4 × B1, MAX = 0.6 × B1)	12.1	18.2	10.0	15.0	6.1	9.1	5.0	7.5	ns
B4	CLKOUT rise time	—	4.00	—	4.00	—	4.00	—	4.00	ns
B5	CLKOUT fall time	—	4.00	—	4.00	—	4.00	—	4.00	ns
B7	CLKOUT to A(0:31), BADDR(28:30), RD/$\overline{WR}$, $\overline{BURST}$, D(0:31) output hold (MIN = 0.25 × B1)	7.60	—	6.30	—	3.80	—	3.13	—	ns
B7a	CLKOUT to TSIZ(0:1), $\overline{REG}$, $\overline{RSV}$, $\overline{BDIP}$, PTR output hold (MIN = 0.25 × B1)	7.60	—	6.30	—	3.80	—	3.13	—	ns
B7b	CLKOUT to $\overline{BR}$, $\overline{BG}$, FRZ, VFLS(0:1), VF(0:2) IWP(0:2), LWP(0:1), $\overline{STS}$ output hold (MIN = 0.25 × B1)	7.60	—	6.30	—	3.80	—	3.13	—	ns
B8	CLKOUT to A(0:31), BADDR(28:30), RD/$\overline{WR}$, $\overline{BURST}$, D(0:31) valid (MAX = 0.25 × B1 + 6.3)	—	13.80	—	12.50	—	10.00	—	9.43	ns
B8a	CLKOUT to TSIZ(0:1), $\overline{REG}$, $\overline{RSV}$, $\overline{BDIP}$, PTR valid (MAX = 0.25 × B1 + 6.3)	—	13.80	—	12.50	—	10.00	—	9.43	ns
B8b	CLKOUT to $\overline{BR}$, $\overline{BG}$, VFLS(0:1), VF(0:2), IWP(0:2), FRZ, LWP(0:1), $\overline{STS}$ valid [2] (MAX = 0.25 × B1 + 6.3)	—	13.80	—	12.50	—	10.00	—	9.43	ns
B9	CLKOUT to A(0:31), BADDR(28:30), RD/$\overline{WR}$, $\overline{BURST}$, D(0:31), TSIZ(0:1), $\overline{REG}$, $\overline{RSV}$, PTR High-Z (MAX = 0.25 × B1 + 6.3)	7.60	13.80	6.30	12.50	3.80	10.00	3.13	9.43	ns
B11	CLKOUT to $\overline{TS}$, $\overline{BB}$ assertion (MAX = 0.25 × B1 + 6.0)	7.60	13.60	6.30	12.30	3.80	9.80	3.13	9.13	ns
B11a	CLKOUT to $\overline{TA}$, $\overline{BI}$ assertion (when driven by the memory controller or PCMCIA interface) (MAX = 0.00 × B1 + 9.30[1])	2.50	9.30	2.50	9.30	2.50	9.80	2.5	9.3	ns
B12	CLKOUT to $\overline{TS}$, $\overline{BB}$ negation (MAX = 0.25 × B1 + 4.8)	7.60	12.30	6.30	11.00	3.80	8.50	3.13	7.92	ns
B12a	CLKOUT to $\overline{TA}$, $\overline{BI}$ negation (when driven by the memory controller or PCMCIA interface) (MAX = 0.00 × B1 + 9.00)	2.50	9.00	2.50	9.00	2.50	9.00	2.5	9.00	ns
B13	CLKOUT to $\overline{TS}$, $\overline{BB}$ High-Z (MIN = 0.25 × B1)	7.60	21.60	6.30	20.30	3.80	14.00	3.13	12.93	ns
B13a	CLKOUT to $\overline{TA}$, $\overline{BI}$ High-Z (when driven by the memory controller or PCMCIA interface) (MIN = 0.00 × B1 + 2.5)	2.50	15.00	2.50	15.00	2.50	15.00	2.5	15.00	ns
B14	CLKOUT to $\overline{TEA}$ assertion (MAX = 0.00 × B1 + 9.00)	2.50	9.00	2.50	9.00	2.50	9.00	2.50	9.00	ns

Figure 20 provides the timing for the asynchronous asserted UPWAIT signal controlled by the UPM.

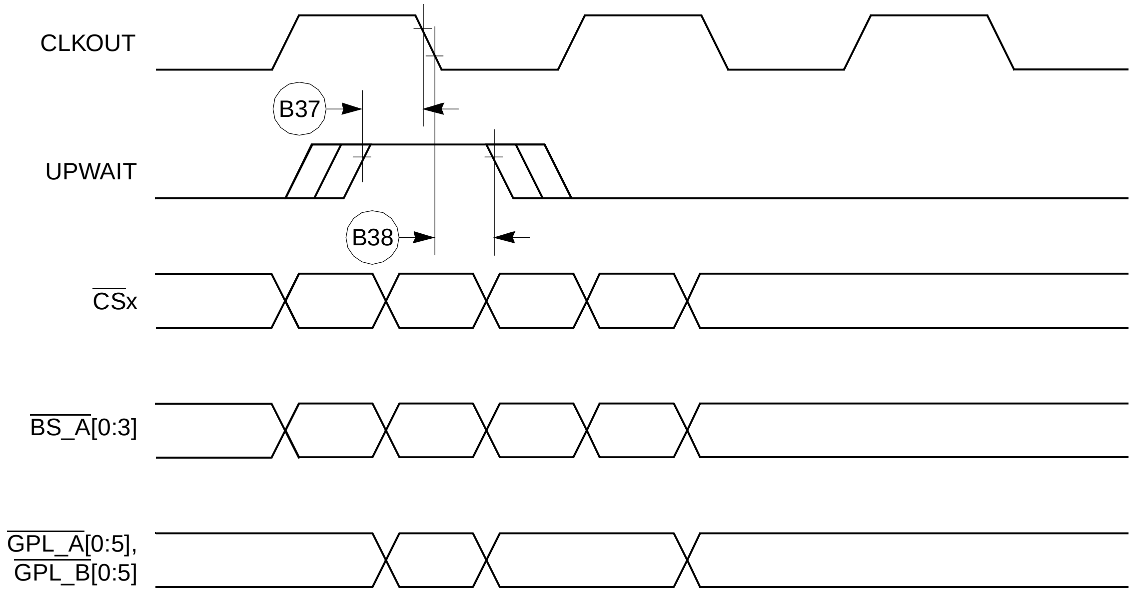


Figure 20. Asynchronous UPWAIT Asserted Detection in UPM Handled Cycles Timing

Figure 21 provides the timing for the asynchronous negated UPWAIT signal controlled by the UPM.

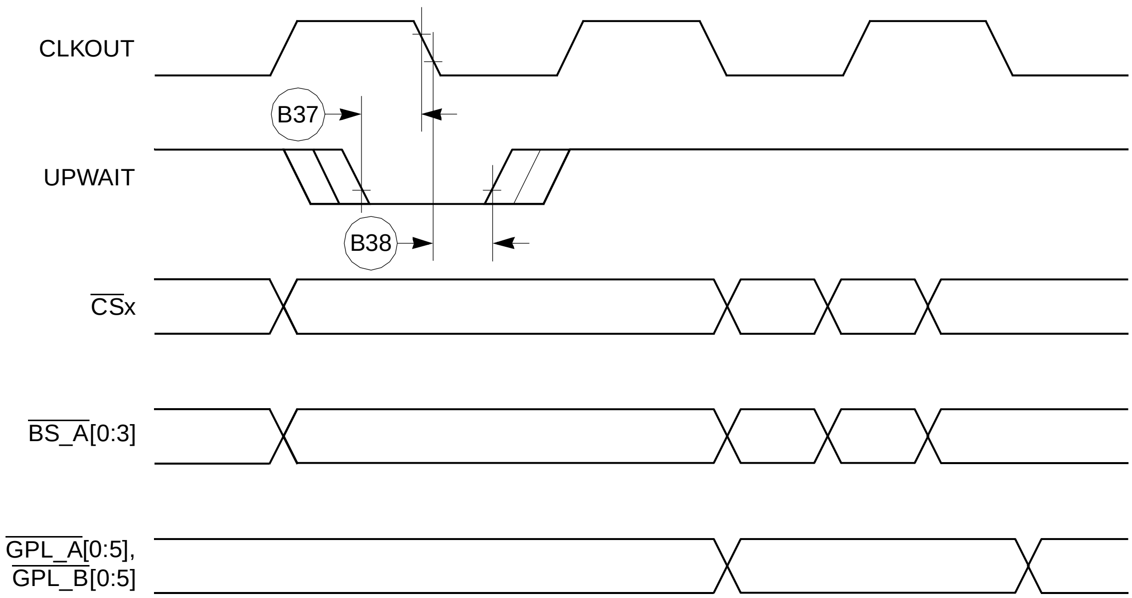


Figure 21. Asynchronous UPWAIT Negated Detection in UPM Handled Cycles Timing

Figure 22 provides the timing for the synchronous external master access controlled by the GPCM.

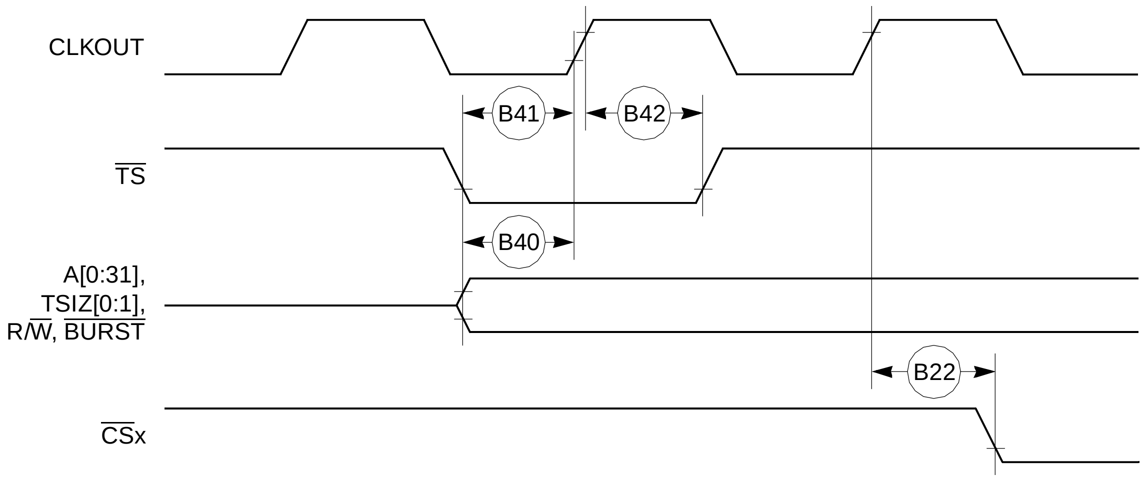


Figure 22. Synchronous External Master Access Timing (GPCM Handled ACS = 00)

Figure 23 provides the timing for the asynchronous external master memory access controlled by the GPCM.

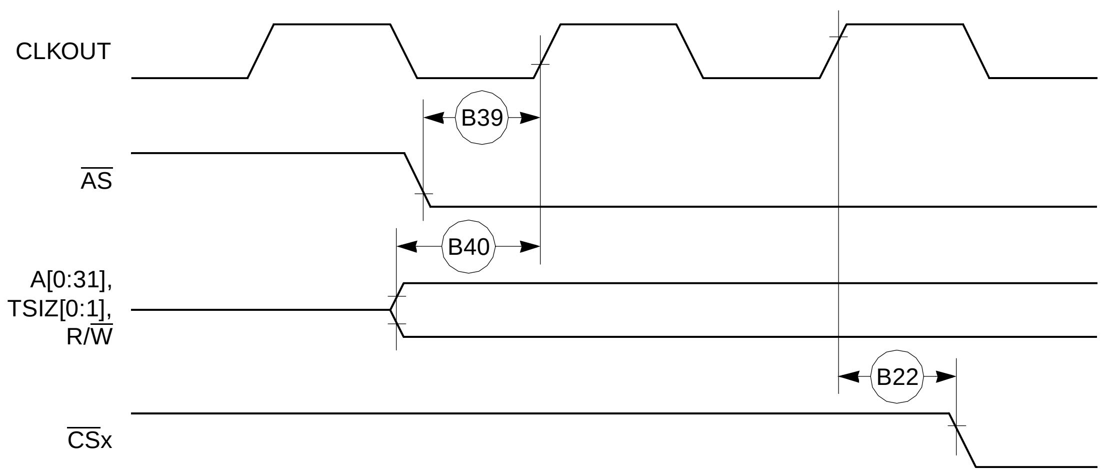


Figure 23. Asynchronous External Master Memory Access Timing (GPCM Controlled—ACS = 00)

Figure 24 provides the timing for the asynchronous external master control signals negation.

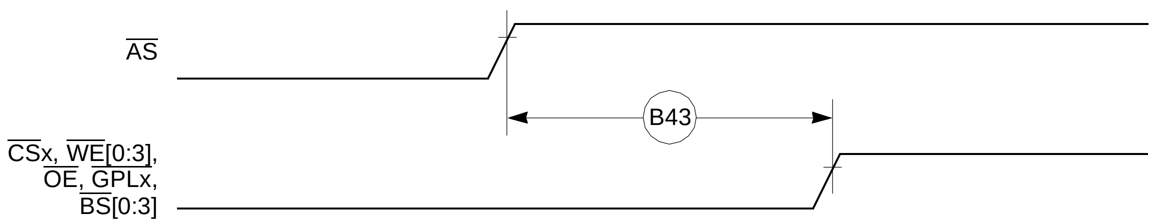


Figure 24. Asynchronous External Master—Control Signals Negation Timing

Table 12 shows the PCMCIA timing for the MPC875/MPC870.

Table 12. PCMCIA Timing

Num	Characteristic	33 MHz		40 MHz		66 MHz		80 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
P44	A(0:31), $\overline{\text{REG}}$ valid to PCMCIA strobe asserted[1] (MIN = 0.75 × B1 – 2.00)	20.70	—	16.70	—	9.40	—	7.40	—	ns
P45	A(0:31), $\overline{\text{REG}}$ valid to ALE negation[1] (MIN = 1.00 × B1 – 2.00)	28.30	—	23.00	—	13.20	—	10.50	—	ns
P46	CLKOUT to $\overline{\text{REG}}$ valid (MAX = 0.25 × B1 + 8.00)	7.60	15.60	6.30	14.30	3.80	11.80	3.13	11.13	ns
P47	CLKOUT to $\overline{\text{REG}}$ invalid (MIN = 0.25 × B1 + 1.00)	8.60	—	7.30	—	4.80	—	4.125	—	ns
P48	CLKOUT to $\overline{\text{CE1}}$, $\overline{\text{CE2}}$ asserted (MAX = 0.25 × B1 + 8.00)	7.60	15.60	6.30	14.30	3.80	11.80	3.13	11.13	ns
P49	CLKOUT to $\overline{\text{CE1}}$, $\overline{\text{CE2}}$ negated (MAX = 0.25 × B1 + 8.00)	7.60	15.60	6.30	14.30	3.80	11.80	3.13	11.13	ns
P50	CLKOUT to $\overline{\text{PCOE}}$, $\overline{\text{IORD}}$, $\overline{\text{PCWE}}$, $\overline{\text{IOWR}}$ assert time (MAX = 0.00 × B1 + 11.00)	—	11.00	—	11.00	—	11.00	—	11.00	ns
P51	CLKOUT to $\overline{\text{PCOE}}$, $\overline{\text{IORD}}$, $\overline{\text{PCWE}}$, $\overline{\text{IOWR}}$ negate time (MAX = 0.00 × B1 + 11.00)	2.00	11.00	2.00	11.00	2.00	11.00	2.00	11.00	ns
P52	CLKOUT to ALE assert time (MAX = 0.25 × B1 + 6.30)	7.60	13.80	6.30	12.50	3.80	10.00	3.13	9.40	ns
P53	CLKOUT to ALE negate time (MAX = 0.25 × B1 + 8.00)	—	15.60	—	14.30	—	11.80	—	11.13	ns
P54	$\overline{\text{PCWE}}$, $\overline{\text{IOWR}}$ negated to D(0:31) invalid[1] (MIN = 0.25 × B1 – 2.00)	5.60	—	4.30	—	1.80	—	1.125	—	ns
P55	$\overline{\text{WAITA}}$ and $\overline{\text{WAITB}}$ valid to CLKOUT rising edge[1] (MIN = 0.00 × B1 + 8.00)	8.00	—	8.00	—	8.00	—	8.00	—	ns
P56	CLKOUT rising edge to $\overline{\text{WAITA}}$ and $\overline{\text{WAITB}}$ invalid[1] (MIN = 0.00 × B1 + 2.00)	2.00	—	2.00	—	2.00	—	2.00	—	ns

[1] PSST = 1. Otherwise add PSST times cycle time.
PSHT = 0. Otherwise add PSHT times cycle time.
These synchronous timings define when the $\overline{\text{WAITA}}$ signals are detected in order to freeze (or relieve) the PCMCIA current cycle. The $\overline{\text{WAITA}}$ assertion will be effective only if it is detected 2 cycles before the PSL timer expiration. See Chapter 16, "PCMCIA Interface," in the *MPC885 PowerQUICC™ Family Reference Manual*.

Figure 27 provides the PCMCIA access cycle timing for the external bus read.

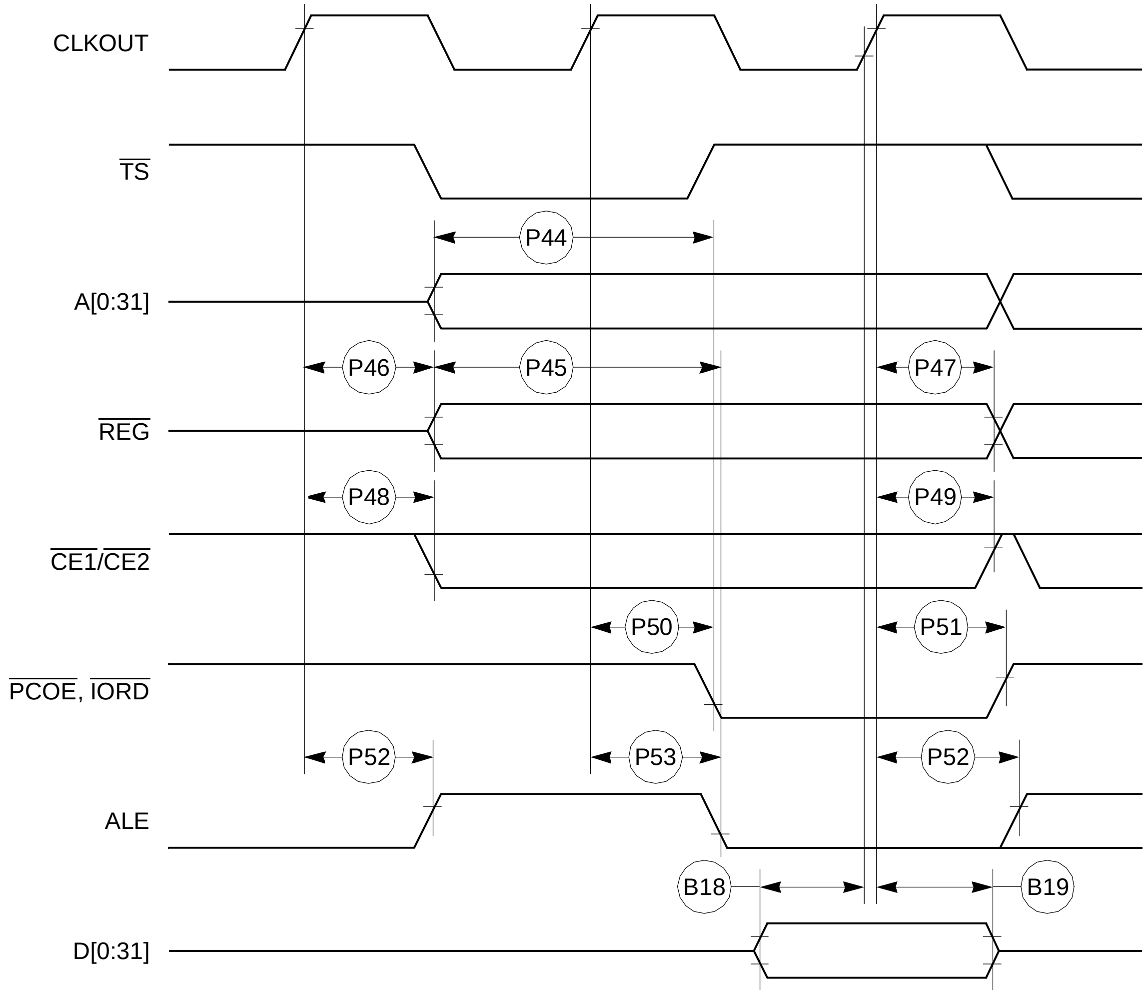


Figure 27. PCMCIA Access Cycles Timing External Bus Read

Figure 28 provides the PCMCIA access cycle timing for the external bus write.

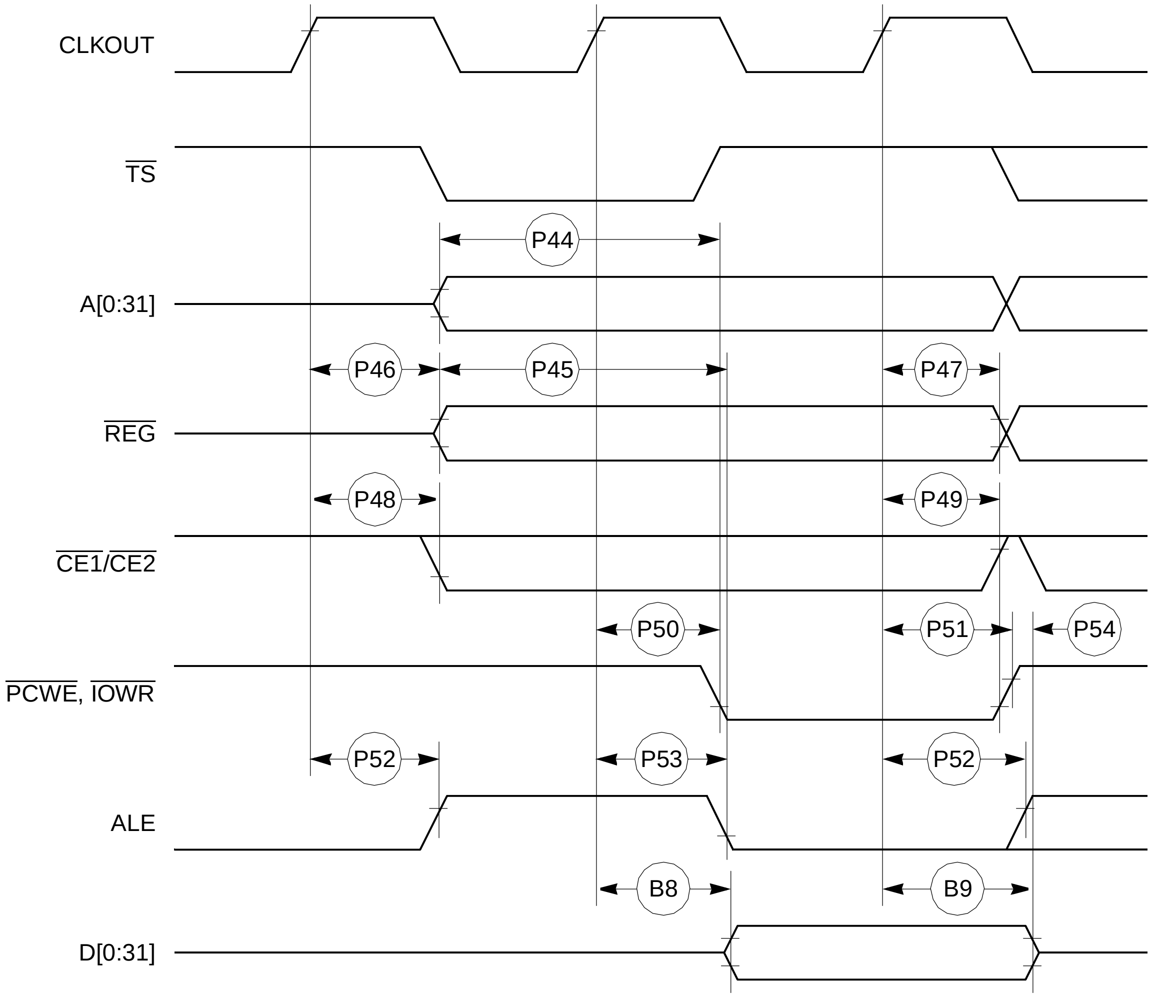


Figure 28. PCMCIA Access Cycles Timing External Bus Write

Figure 29 provides the PCMCIA $\overline{\text{WAIT}}$ signals detection timing.

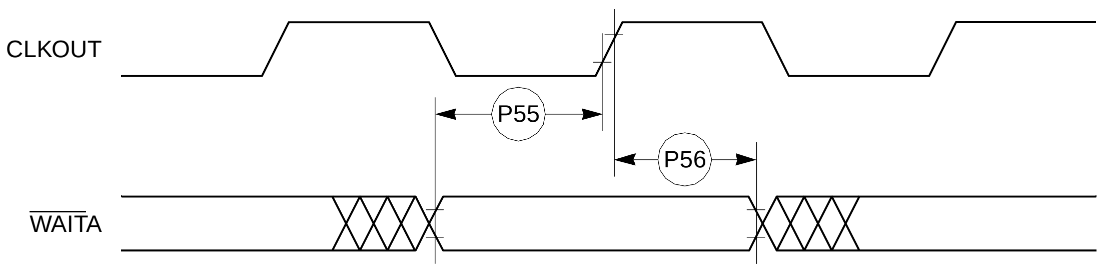


Figure 29. PCMCIA $\overline{\text{WAIT}}$ Signals Detection Timing

Table 15 shows the reset timing for the MPC875/MPC870.

Table 15. Reset Timing

Num	Characteristic	33 MHz		40 MHz		66 MHz		80 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
R69	CLKOUT to HRESET high impedance (MAX = 0.00 × B1 + 20.00)	—	20.00	—	20.00	—	20.00	—	20.00	ns
R70	CLKOUT to SRESET high impedance (MAX = 0.00 × B1 + 20.00)	—	20.00	—	20.00	—	20.00	—	20.00	ns
R71	RSTCONF pulse width (MIN = 17.00 × B1)	515.20	—	425.00	—	257.60	—	212.50	—	ns
R72	—	—	—	—	—	—	—	—	—	—
R73	Configuration data to HRESET rising edge setup time (MIN = 15.00 × B1 + 50.00)	504.50	—	425.00	—	277.30	—	237.50	—	ns
R74	Configuration data to RSTCONF rising edge setup time (MIN = 0.00 × B1 + 350.00)	350.00	—	350.00	—	350.00	—	350.00	—	ns
R75	Configuration data hold time after RSTCONF negation (MIN = 0.00 × B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R76	Configuration data hold time after HRESET negation (MIN = 0.00 × B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R77	HRESET and RSTCONF asserted to data out drive (MAX = 0.00 × B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R78	RSTCONF negated to data out high impedance (MAX = 0.00 × B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R79	CLKOUT of last rising edge before chip three-states HRESET to data out high impedance (MAX = 0.00 × B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R80	DSDI, DSCK setup (MIN = 3.00 × B1)	90.90	—	75.00	—	45.50	—	37.50	—	ns
R81	DSDI, DSCK hold time (MIN = 0.00 × B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R82	SRESET negated to CLKOUT rising edge for DSDI and DSCK sample (MIN = 8.00 × B1)	242.40	—	200.00	—	121.20	—	100.00	—	ns

Figure 34 shows the reset timing for the data bus configuration.

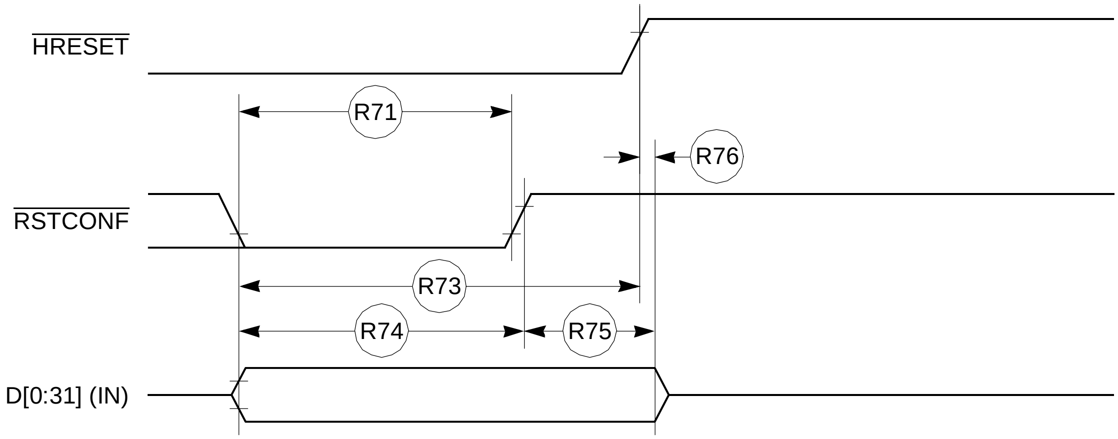


Figure 34. Reset Timing—Configuration from Data Bus

Figure 35 provides the reset timing for the data bus weak drive during configuration.

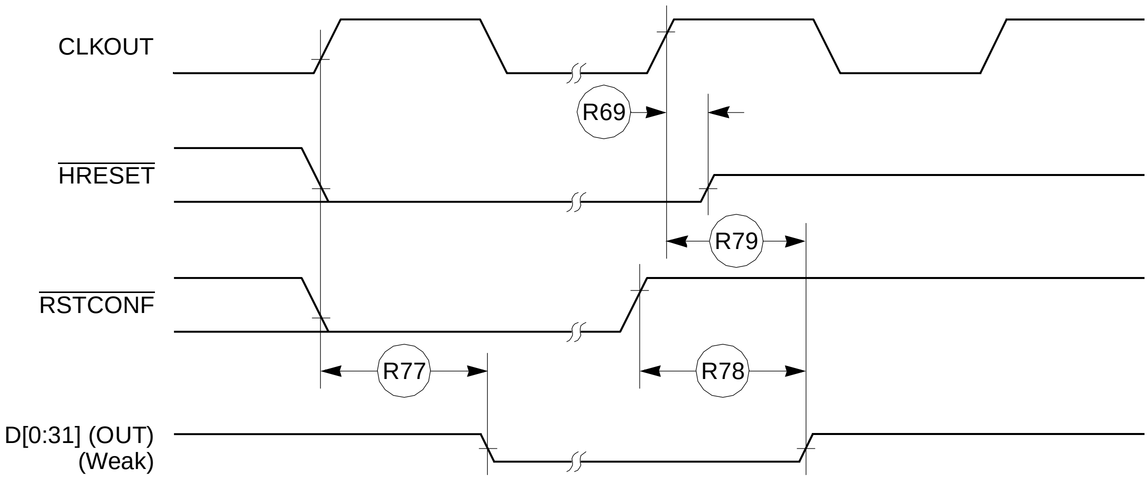


Figure 35. Reset Timing—Data Bus Weak Drive During Configuration

Figure 36 provides the reset timing for the debug port configuration.

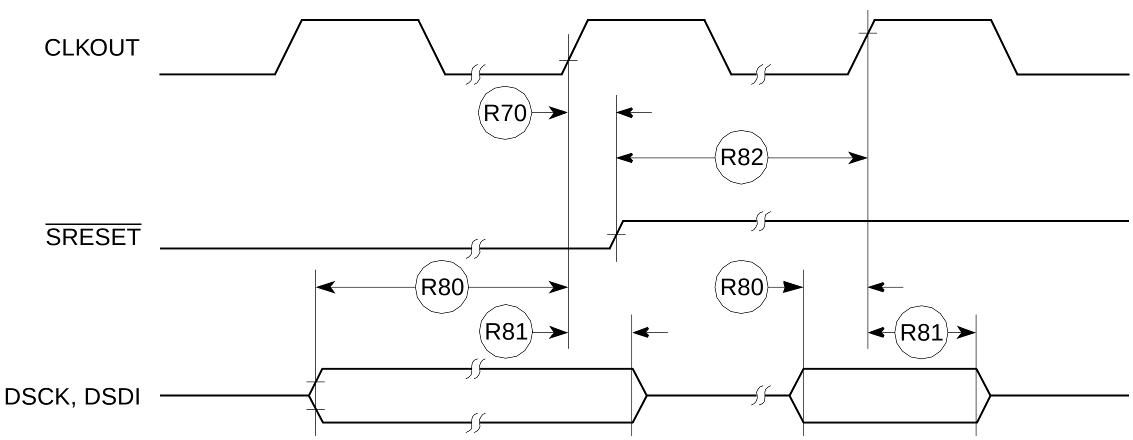


Figure 36. Reset Timing—Debug Port Configuration

13.3 Baud Rate Generator AC Electrical Specifications

Table 19 provides the baud rate generator timings as shown in Figure 46.

Table 19. Baud Rate Generator Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
50	BRGO rise and fall time	—	10	ns
51	BRGO duty cycle	40	60	%
52	BRGO cycle	40	—	ns

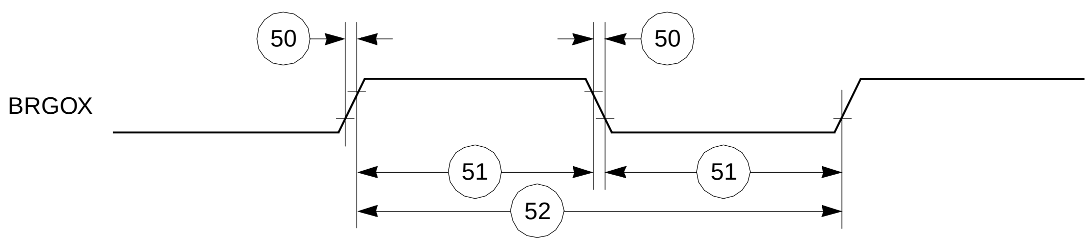


Figure 46. Baud Rate Generator Timing Diagram

13.4 Timer AC Electrical Specifications

Table 20 provides the general-purpose timer timings as shown in Figure 47.

Table 20. Timer Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
61	TIN/$\overline{\text{TGATE}}$ rise and fall time	10	—	ns
62	TIN/$\overline{\text{TGATE}}$ low time	1	—	clk
63	TIN/$\overline{\text{TGATE}}$ high time	2	—	clk
64	TIN/$\overline{\text{TGATE}}$ cycle time	3	—	clk
65	CLKO low to $\overline{\text{TOUT}}$ valid	3	25	ns

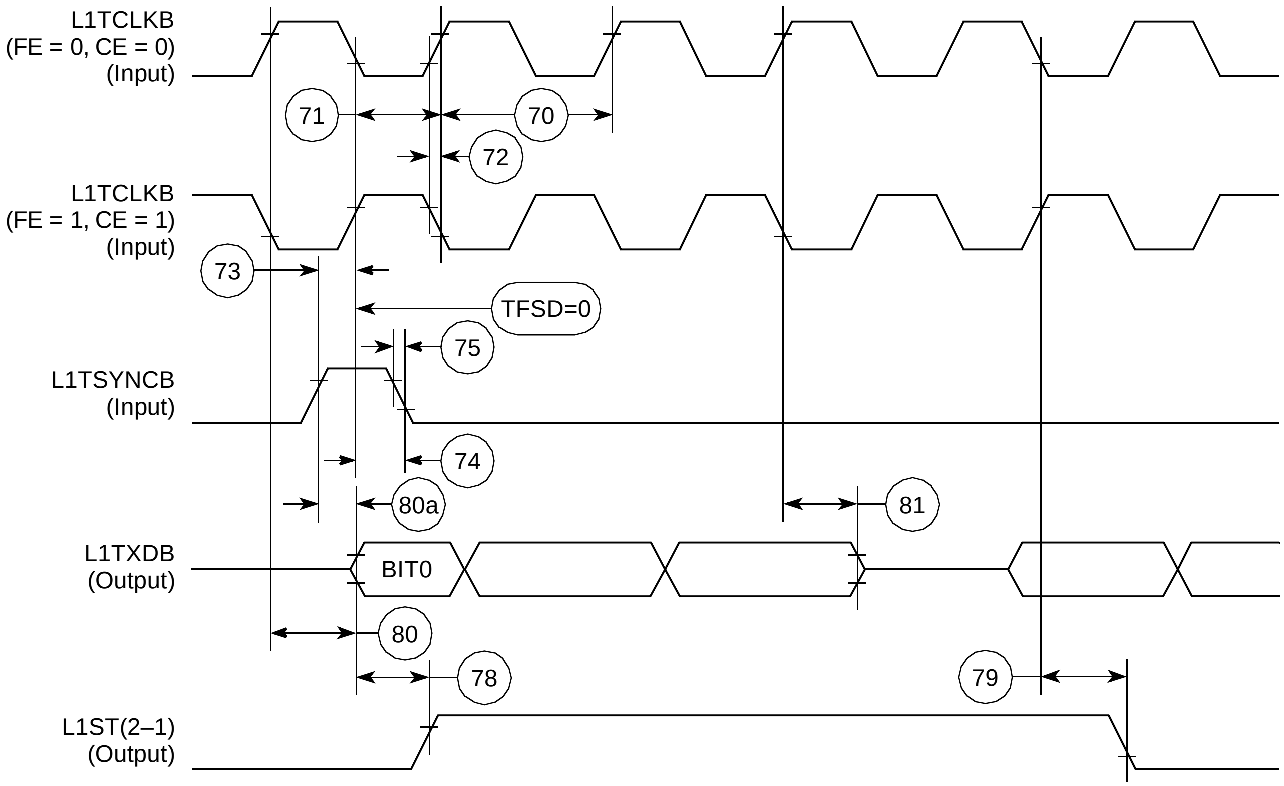


Figure 50. SI Transmit Timing Diagram (DSC = 0)

Figure 53 through Figure 55 show the NMSI timings.

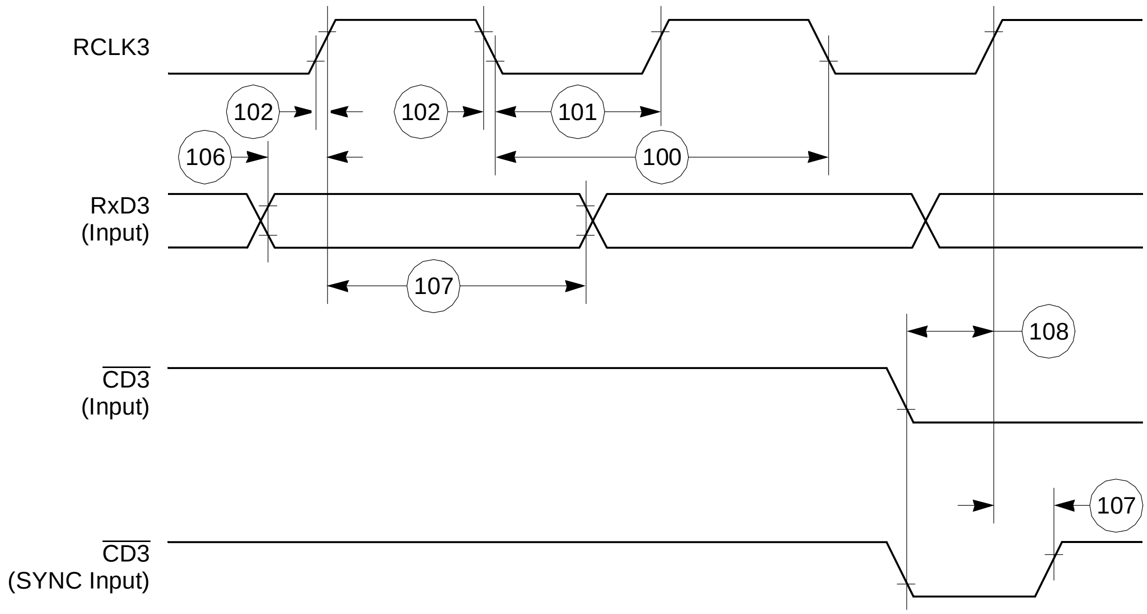


Figure 53. SCC NMSI Receive Timing Diagram

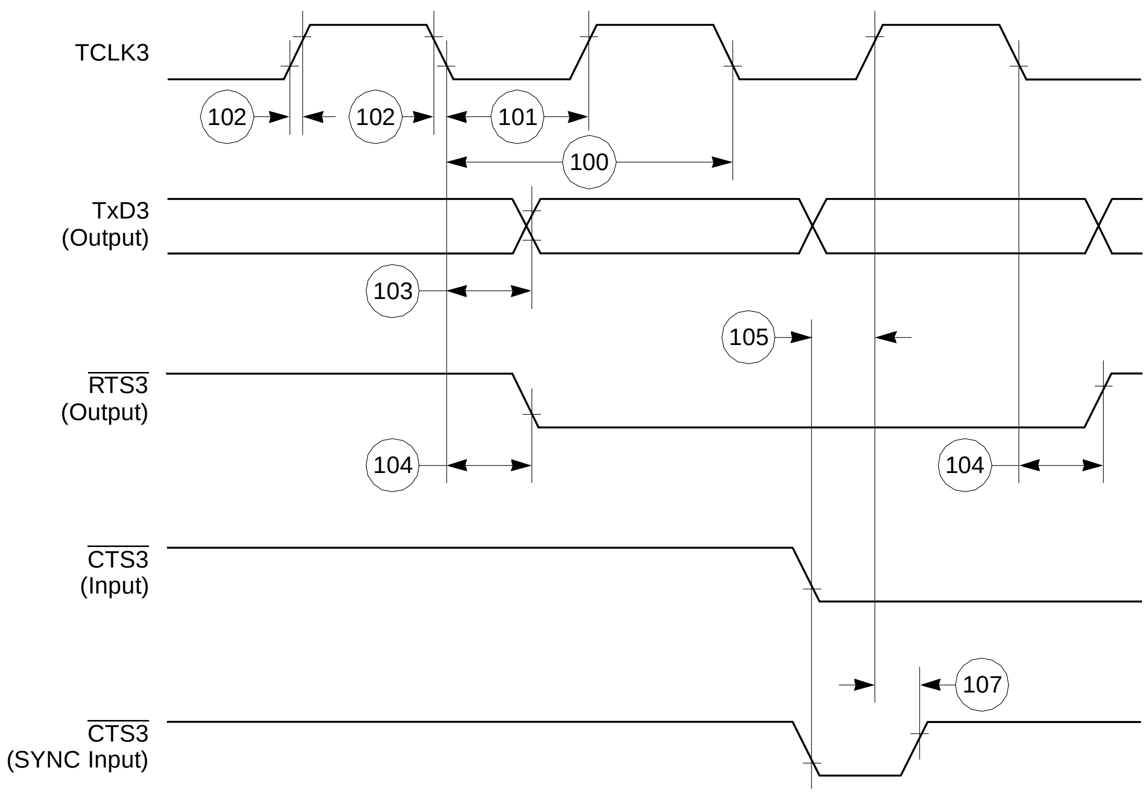
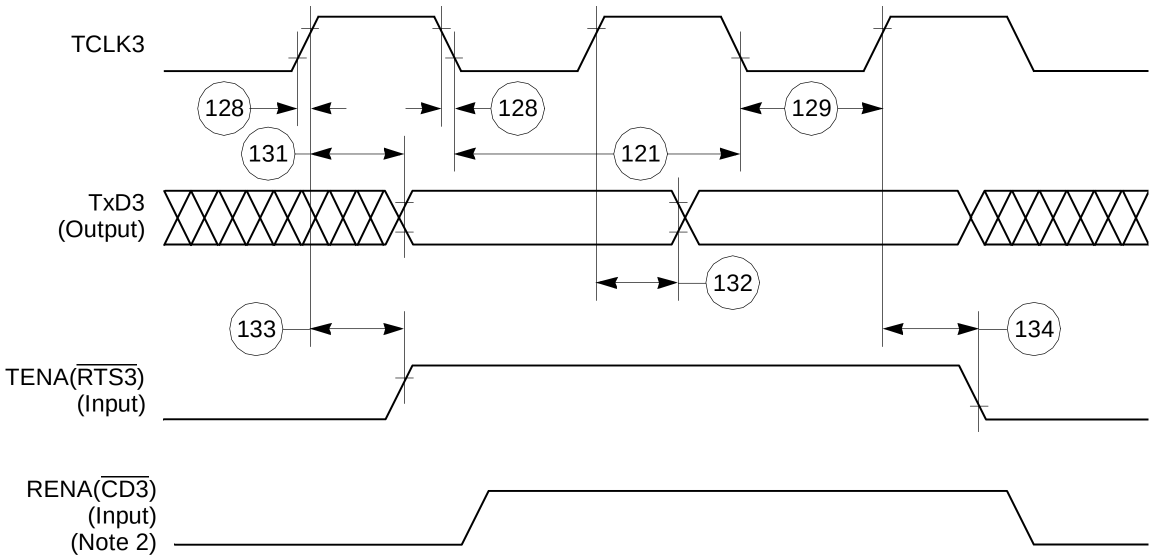


Figure 54. SCC NMSI Transmit Timing Diagram



Notes:
1. Transmit clock invert (TCI) bit in GSMR is set.
2. If RENA is negated before TENA or RENA is not asserted at all during transmit, then the CSL bit is set in the buffer descriptor at the end of the frame transmission.

Figure 58. Ethernet Transmit Timing Diagram

13.8 SMC Transparent AC Electrical Specifications

Table 25 provides the SMC transparent timings as shown in Figure 59.

Table 25. SMC Transparent Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
150	SMCLK clock period[1]	100	—	ns
151	SMCLK width low	50	—	ns
151A	SMCLK width high	50	—	ns
152	SMCLK rise/fall time	—	15	ns
153	SMTXD active delay (from SMCLK falling edge)	10	50	ns
154	SMRXD/SMSYNC setup time	20	—	ns
155	RXD1/SMSYNC hold time	5	—	ns

[1] SYNCCLK must be at least twice as fast as SMCLK.

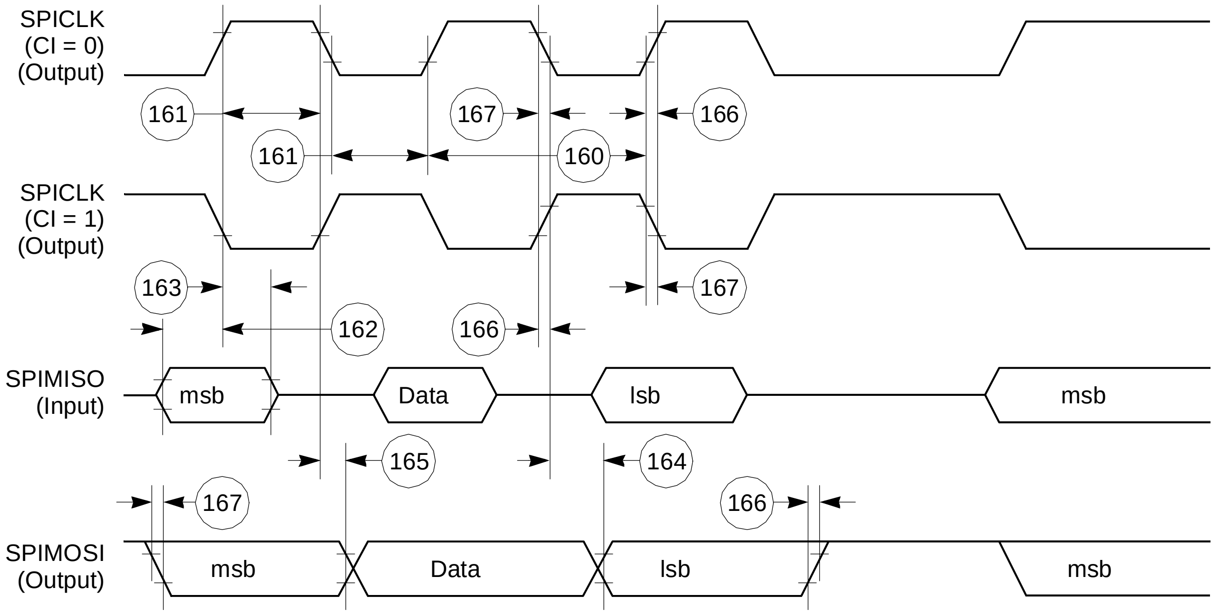


Figure 60. SPI Master (CP = 0) Timing Diagram

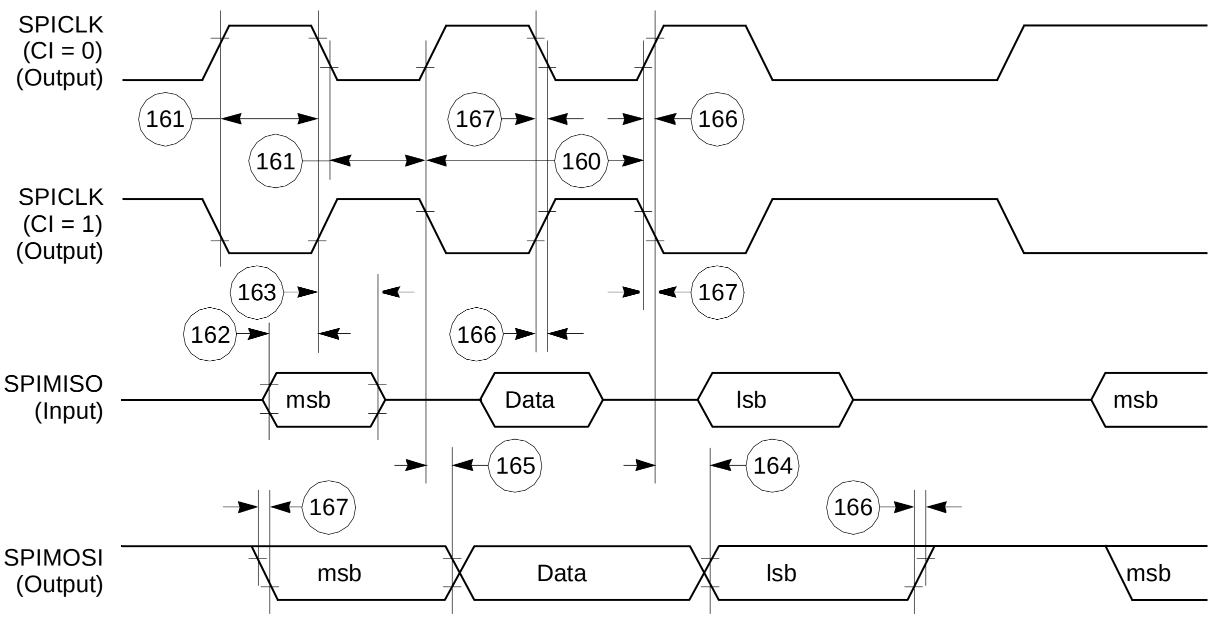


Figure 61. SPI Master (CP = 1) Timing Diagram

13.10 SPI Slave AC Electrical Specifications

Table 27 provides the SPI slave timings as shown in Figure 62 and Figure 63.

Table 27. SPI Slave Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
170	Slave cycle time	2	—	t_{cyc}
171	Slave enable lead time	15	—	ns
172	Slave enable lag time	15	—	ns
173	Slave clock (SPICLK) high or low time	1	—	t_{cyc}
174	Slave sequential transfer delay (does not require deselect)	1	—	t_{cyc}
175	Slave data setup time (inputs)	20	—	ns
176	Slave data hold time (inputs)	20	—	ns
177	Slave access time	—	50	ns

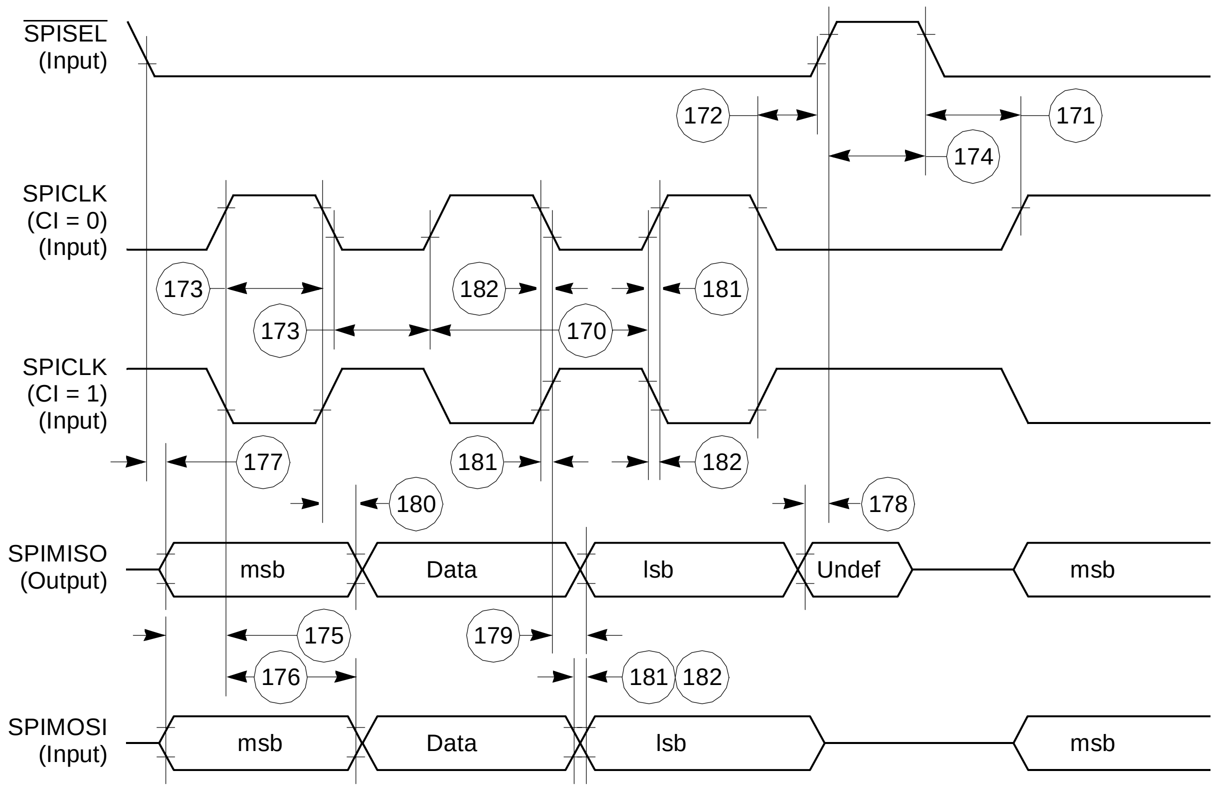


Figure 62. SPI Slave (CP = 0) Timing Diagram

14 USB Electrical Characteristics

This section provides the AC timings for the USB interface.

14.1 USB Interface AC Timing Specifications

The USB Port uses the transmit clock on SCC1. Table 30 lists the USB interface timings.

Table 30. USB Interface AC Timing Specifications

Name	Characteristic	All Frequencies		Unit
		Min	Max	
US1	USBCLK frequency of operation[1] Low speed Full speed	6 48		MHz
US4	USBCLK duty cycle (measured at 1.5 V)	45	55	%

[1] USBCLK accuracy should be ±500 ppm or better. USBCLK may be stopped to conserve power.

15 FEC Electrical Characteristics

This section provides the AC electrical specifications for the Fast Ethernet controller (FEC). Note that the timing specifications for the MII signals are independent of system clock frequency (part speed designation). Also, MII signals use TTL signal levels compatible with devices operating at either 5.0 or 3.3 V.

15.1 MII and Reduced MII Receive Signal Timing

The receiver functions correctly up to a MII_RX_CLK maximum frequency of 25 MHz + 1%. The reduced MII (RMII) receiver functions correctly up to a RMII_REFCLK maximum frequency of 50 MHz + 1%. There is no minimum frequency requirement. In addition, the processor clock frequency must exceed the MII_RX_CLK frequency – 1%.

Table 31 provides information on the MII receive signal timing.

Table 31. MII Receive Signal Timing

Num	Characteristic	Min	Max	Unit
M1	MII_RXD[3:0], MII_RX_DV, MII_RX_ER to MII_RX_CLK setup	5	—	ns
M2	MII_RX_CLK to MII_RXD[3:0], MII_RX_DV, MII_RX_ER hold	5	—	ns
M3	MII_RX_CLK pulse width high	35%	65%	MII_RX_CLK period
M4	MII_RX_CLK pulse width low	35%	65%	MII_RX_CLK period
M1_RMII	RMII_RXD[1:0], RMII_CRS_DV, RMII_RX_ERR to RMII_REFCLK setup	4	—	ns
M2_RMII	RMII_REFCLK to RMII_RXD[1:0], RMII_CRS_DV, RMII_RX_ERR hold	2	—	ns

Figure 68 shows the MII serial management channel timing diagram.

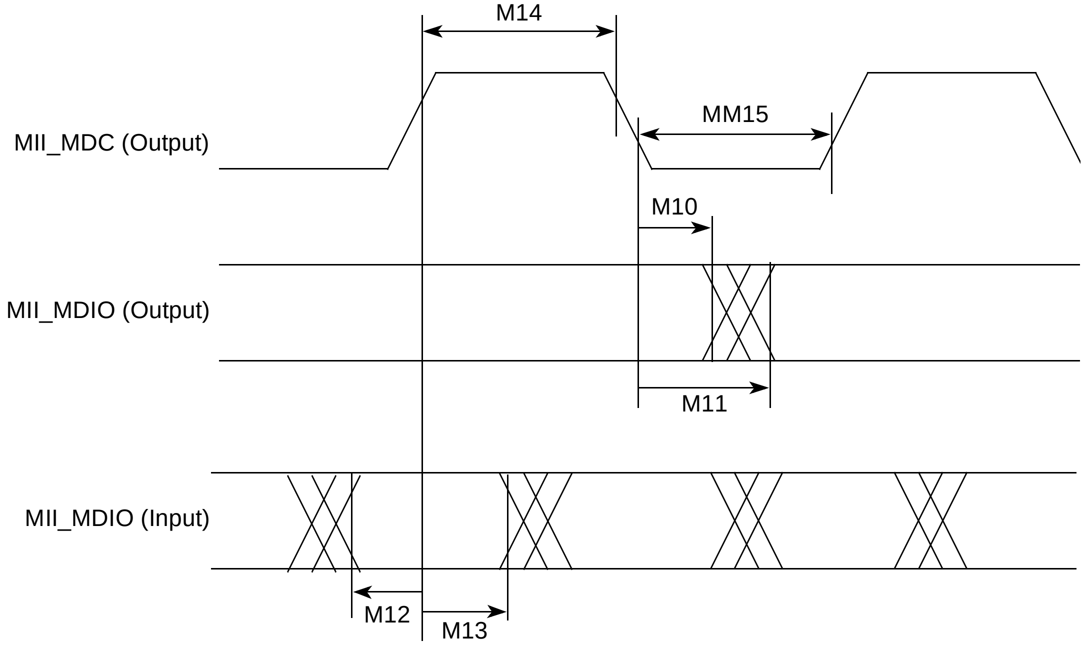


Figure 68. MII Serial Management Channel Timing Diagram

Table 36. Pin Assignments—JEDEC Standard (continued)

Name	Pin Number	Type
PE29, MII2-CRS	U7	Bidirectional (Optional: open-drain)
PE28, TOUT3, MII2-COL	R7	Bidirectional (Optional: open-drain)
PE27, L1RQB, MII2-RXERR, RMII2-RXERR	T6	Bidirectional (Optional: open-drain)
PE26, L1CLKOB, MII2-RXDV, RMII2-CRS_DV	T2	Bidirectional (Optional: open-drain)
PE25, RXD4, MII2-RXD3, L1ST2	R4	Bidirectional (Optional: open-drain)
PE24, SMRXD1, BRGO1, MII2-RXD2	U8	Bidirectional (Optional: open-drain)
PE23, TXD4, MII2-RXCLK, L1ST1	U4	Bidirectional (Optional: open-drain)
PE22, TOUT2, MII2-RXD1, RMII2-RXD1, SDACK1	P4	Bidirectional (Optional: open-drain)
PE21, TOUT1, MII2-RXD0, RMII2-RXD0	T9	Bidirectional (Optional: open-drain)
PE20, MII2-TXER	U3	Bidirectional (Optional: open-drain)
PE19, L1TXDB, MII2-TXEN, RMII2-TXEN	R6	Bidirectional (Optional: open-drain)
PE18, SMTXD1, MII2-TXD3	M5	Bidirectional (Optional: open-drain)
PE17, TIN3, CLK5, BRGO3, SMSYN1, MII2-TXD2	T8	Bidirectional (Optional: open-drain)
PE16, L1RCLKB, CLK6, MII2-TXCLK, RMII2-REFCLK	U6	Bidirectional (Optional: open-drain)
PE15, TGATE1, MII2-TXD1, RMII2-TXD1	T7	Bidirectional
PE14, MII2-TXD0, RMII2-TXD0	P8	Bidirectional
TMS	T14	Input (5-V tolerant)
TDI, DSDI	T13	Input (5-V tolerant)
TCK, DSCK	R13	Input (5-V tolerant)
TRST	U14	Input (5-V tolerant)